

## **Operational Amplifiers**

# **Low Noise Operational Amplifiers**

BA4580Rxxx BA4584FV BA4584Rxx

#### **General Description**

BA4580Rxxx, BA4584FV, BA4584Rxx integrates two or four independent high voltage gain Op-Amps on a single chip. Especially, this series are suitable for any audio applications due to low noise and low distortion characteristics and are usable for other many applications by wide operating supply voltage range.

#### Features

- High Voltage Gain
- Low Input Referred Noise Voltage
- Low Distortion
- Wide Operating Supply Voltage Range
- Wide Temperature Range

### Application

- Audio Application
- Consumer Electronics

## **Simplified Schematic**

 Packages
 W(Typ) x D(Typ) x H(Max)

 SOP8
 5.00mm x 6.20mm x 1.71mm

 SOP-J8
 4.90mm x 6.00mm x 1.65mm

 TSSOP-B8
 3.00mm x 6.40mm x 1.20mm

 MSOP8
 2.90mm x 4.00mm x 0.90mm

 SOP14
 8.70mm x 6.20mm x 1.71mm

 SSOP-B14
 5.00mm x 6.40mm x 1.35mm

### **Key Specification**

- Operating Supply Voltage Range (Split Supply): BA4580Rxxx, BA4584FV ±2V to ±16V BA4584Rxx ±2V to ±9.5V
   Slew Rate: 5V/µs(Typ)
- Total Harmonic Distortion: 0.0005%(Typ)
   Input Referred Noise Voltage: 5 nV/√Hz (Typ)
- Operating Temperature Range: BA4584FV
   BA4580Rxxx.BA4584Rxx
   -40°C to +85°C
   -40°C to +105°C

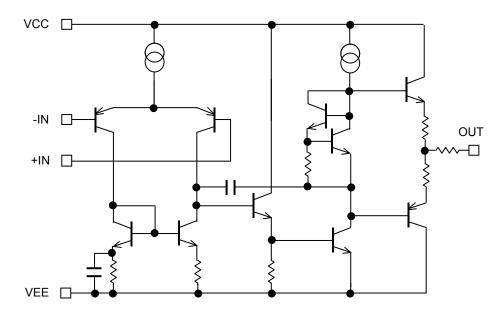
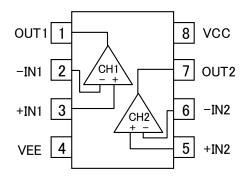


Figure 1. Simplified schematic

OProduct structure : Silicon monolithic integrated circuit OThis product is not designed protection against radioactive rays.

## **Pin Configuration**

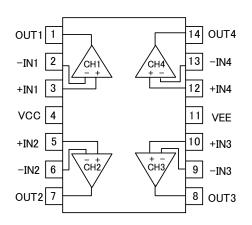
BA4580RF	: SOP8
BA4580RFJ	: SOP-J8
BA4580RFVT	: TSSOP-B8
BA4580RFVM	: MSOP8



Pin No.	Pin Name
1	OUT1
2	-IN1
3	+IN1
4	VEE
5	+IN2
6	-IN2
7	OUT2
8	VCC

BA4584RF	: :
BA4584FV, BA4584RFV	: :

: SOP14 : SSOP-B14



Pin No.	Pin Name
1	OUT1
2	-IN1
3	+IN1
4	VCC
5	+IN2
6	-IN2
7	OUT2
8	OUT3
9	-IN3
10	+IN3
11	VEE
12	+IN4
13	-IN4
14	OUT4
14	OUT4

Package												
SOP8	SOP-J8	TSSOP-B8	MSOP8	SOP14	SSOP-B14							
BA4580RF	BA4580RFJ	BA4580RFVT	BA4580RFVM	BA4584RF	BA4584FV BA4584RFV							

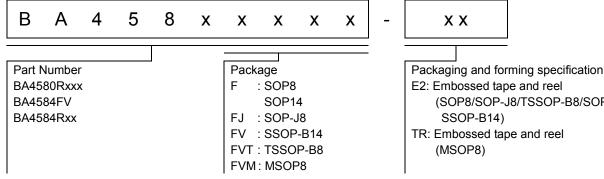
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SSOP-B14)

(MSOP8)

(SOP8/SOP-J8/TSSOP-B8/SOP14/

## **Ordering Information**



#### Line-up

Operating Temperature Range	Operating Supply Voltage Range (Split Supply)	Supply Current (Typ)	Slew Rate (Typ)	Pac	kage	Orderable Part Number
-40°C to +85°C		12mA		SSOP-B14	Reel of 2500	BA4584FV-E2
	±2.0V to ±16.0V		5V/µs	SOP8	Reel of 2500	BA4580RF-E2
		6mA		SOP-J8	Reel of 2500	BA4580RFJ-E2
-40°C to +105°C				TSSOP-B8	Reel of 3000	BA4580RFVT-E2
-40 C 10 + 105 C				MSOP8	Reel of 3000	BA4580RFVM-TR
	±2.0V to ±9.5V	11mA		SOP14	Reel of 2500	BA4584RF-E2
	12.0V 10 19.5V	TIMA		SSOP-B14	Reel of 2500	BA4584RFV-E2

## Absolute Maximum Ratings (T<sub>A</sub>=25°C)

Parameter		Sumbol		Ratings		Unit	
Falameter		Symbol	BA4580Rxxx	BA4584FV	Unit		
Supply Voltage		VCC-VEE		+36		V	
		SOP8	0.78 <sup>(Note1,7)</sup>		-		
		SOP-J8	0.67 <sup>(Note2,7)</sup>		-		
	-	TSSOP-B8	0.62 <sup>(Note3,7)</sup>		-		
Power Dissipation	$P_D$	MSOP8	0.59 <sup>(Note4,7)</sup>		-	W	
		SOP14	-	- 0.61 <sup>(Note5,7)</sup>			
		SSOP-B14	-	0.87 <sup>(Note6,7)</sup>		1	
Differential Input Voltage <sup>(Note 8)</sup>		V <sub>ID</sub>		V			
Input Common-mode Voltage Range		VICM	VEE to VEE+36			V	
Input Current <sup>(Note 9)</sup>		l <sub>l</sub>		±10		mA	
Operating Supply Voltage Range		V <sub>opr</sub>		o +32 o ±16)	+4 to +19 (±2 to ±9.5)	V	
Output Current		I <sub>OUT</sub>		±50			
Operating Temperature Range	T <sub>opr</sub>		-40 to +105	-40 to +85	-40 to +105	°C	
Storage Temperature Range	T <sub>stg</sub>		T <sub>stg</sub> -55 to +150				
Maximum Junction Temperature		T <sub>Jmax</sub>		+150		°C	

(Note 1) To use at temperature above  $T_A=25^{\circ}C$  reduce  $6.2mW/^{\circ}C$ .

(Note 2) To use at temperature above T<sub>A</sub>=25°C reduce 5.4mW/°C

(Note 3) To use at temperature above  $T_A=25^{\circ}C$  reduce 5.0mW/°C (Note 4) To use at temperature above T<sub>A</sub>=25°C reduce 4.8mW/°C

To use at temperature above  $T_A{=}25^\circ\!C$  reduce  $4.9mW/^\circ\!C$ (Note 5)

(Note 6) To use at temperature above T<sub>A</sub>=25°C reduce 7.0mW/°C

Mounted on a FR4 glass epoxy PCB(70mm×70mm×1.6mm). (Note 7)

The voltage difference between inverting input and non-inverting input is the differential input voltage. (Note 8)

Then input terminal voltage is set to more than VEE.

(Note 9) An excessive input current will flow when input voltages of more than VDD+0.6V or less than VSS-0.6V are applied.

The input current can be set to less than the rated current by adding a limiting resistor.

Caution: Operating the IC over the absolute maximum ratings may damage the IC. In addition, it is impossible to predict all destructive situations such as short-circuit modes, open circuit modes, etc. Therefore, it is important to consider circuit protection measures, like adding a fuse, in case the IC is operated in a special mode exceeding the absolute maximum ratings.

## **Electrical Characteristics**

OBA4580R (Unless otherwise specified VCC=+15V, VEE=-15V, T<sub>A</sub>=25°C)

Parameter	Symbol		Limits		Unit	Condition
Falameter	Symbol	Min	Тур	Max	Unit	Condition
Input Offset Voltage (Note 10)	V <sub>IO</sub>	-	0.3	3	mV	Rs≤ 10kΩ
Input Offset Current (Note 10)	I <sub>IO</sub>	-	5	200	nA	-
Input Bias Current (Note 11)	I <sub>B</sub>	-	100	500	nA	-
Large Signal Voltage Gain	Av	90	110	-	dB	R <sub>L</sub> ≥ 10kΩ, OUT=±10∨
Maximum Output Voltage	V <sub>OM</sub>	±12	±13.5	-	V	R <sub>L</sub> ≥ 2kΩ
Input Common-mode Voltage Range	VICM	±12	±13.5	-	V	-
Common-mode Rejection Ratio	CMRR	80	110	-	dB	R <sub>S</sub> ≤ 10kΩ
Power Supply Rejection Ratio	PSRR	80	110	-	dB	Rs≦ 10kΩ
Supply Current	Icc	-	6	9	mA	R <sub>L</sub> =∞, All Op-Amps, VIN+=0V
Slew Rate	SR	-	5	-	V/µs	R <sub>L</sub> ≥ 2kΩ
Gain Bandwidth Product	GBW	-	10	-	MHz	f=10kHz
Unity Gain Frequency	f <sub>T</sub>	-	5	-	MHz	$R_L=2k\Omega$
Total Harmonic Distortion+ Noise	THD+N	-	0.0005	-	%	$A_V$ =20dB, OUT=5Vrms R <sub>L</sub> =2kΩ f=1kHz, 20Hz~20kHz BPF
Input Deferred Noise Veltere	V	-	5	-	nV/√Hz	R <sub>S</sub> =100Ω, V <sub>I</sub> =0V, f=1kHz
Input Referred Noise Voltage	V <sub>N</sub>	-	0.8	-	μVrms	RIAA, R <sub>S</sub> =2.2 kΩ, 30kHz LPF
Channel Separation	CS	-	110	-	dB	R1=100Ω, f=1kHz

(Note 10) Absolute value

(Note 11) Current direction: Since first input stage is composed with PNP transistor, input bias current flows out of IC.

#### OBA4584 (Unless otherwise specified VCC=+15V, VEE=-15V, $T_A = 25^{\circ}C$ )

Parameter	Symbol		Limits		Unit	Condition		
Falameter	Symbol	Min.	Тур.	Max.	Unit	Condition		
Input Offset Voltage (Note 12)	V <sub>IO</sub>	-	0.3	3	mV	R <sub>S</sub> ≤ 10kΩ		
Input Offset Current (Note 12)	I <sub>IO</sub>	-	5	200	nA	-		
Input Bias Current (Note 13)	Ι <sub>Β</sub>	-	100	500	nA	-		
Large Signal Voltage Gain	Av	90	110	-	dB	R <sub>L</sub> ≥ 10kΩ, OUT=±10∨		
Maximum Output Voltage	V <sub>OM</sub>	±12	±13.5	-	V	R <sub>L</sub> ≥ 2kΩ		
Input Common-mode Voltage Range	V <sub>ICM</sub>	±12	±13.5	-	V	-		
Common-mode Rejection Ratio	CMRR	80	110	-	dB	R <sub>S</sub> ≤ 10kΩ		
Power Supply Rejection Ratio	PSRR	80	110	-	dB	R <sub>S</sub> ≤ 10kΩ		
Supply Current	I <sub>CC</sub>	-	12	18	mA	R <sub>L</sub> =∞, All Op-Amps, VIN+=0V		
Slew Rate	SR	-	5	-	V/µs	R <sub>L</sub> ≥ 2kΩ		
Gain Bandwidth Product	GBW	-	10	-	MHz	f=10kHz		
Unity Gain Frequency	f⊤	-	5	-	MHz	$R_L=2k\Omega$		
Total Harmonic Distortion+ Noise	THD+N	-	0.0005	-	%	$A_V$ =20dB, OUT=5Vrms R <sub>L</sub> =2k $\Omega$ f=1kHz, 20Hz~20kHz BPF		
Input Referred Noise Voltage	V <sub>N</sub>	-	5	-	nV/√Hz	R <sub>S</sub> =100Ω, V <sub>i</sub> =0V, f=1kHz		
input teleneu tiolse voltage	۷N	-	0.8	-	μVrms	RIAA, R <sub>S</sub> =2.2 k $\Omega$ , 30kHz LPF		
Channel Separation	CS	-	110	-	dB	R1=100Ω, f=1kHz		

(Note 12) Absolute value(Note 13) Current direction: Since first input stage is composed with PNP transistor, input bias current flows out of IC.

#### OBA4584R (Unless otherwise specified VCC=+9.5V, VEE=-9.5V, T<sub>A</sub> =25°C)

Parameter	Symbol	,	Limits		- Unit	Condition		
Parameter	Symbol	Min.	Тур.	Max.	Unit	Condition		
Input Offset Voltage (Note 14)	V <sub>IO</sub>	-	0.3	3	mV	R <sub>S</sub> ≤ 10kΩ		
Input Offset Current (Note 14)	I <sub>IO</sub>	-	5	200	nA	-		
Input Bias Current (Note 15)	Ι <sub>Β</sub>	-	100	500	nA	-		
Large Signal Voltage Gain	Av	90	110	-	dB	$R_{L} \ge 10 k\Omega$ , OUT=±10 V		
Maximum Output Voltage	V <sub>OM</sub>	±6.5	±8	-	V	R <sub>L</sub> ≥ 2kΩ		
Input Common-mode Voltage Range	V <sub>ICM</sub>	±6.5	±8	-	V	-		
Common-mode Rejection Ratio	CMRR	80	110	-	dB	R <sub>S</sub> ≤ 10kΩ		
Power Supply Rejection Ratio	PSRR	80	110	-	dB	R <sub>S</sub> ≤ 10kΩ		
Supply Current	I <sub>CC</sub>	-	11	17	mA	R <sub>L</sub> =∞, All Op-Amps, VIN+=0V		
Slew Rate	SR	-	5	-	V/µs	R <sub>L</sub> ≥ 2kΩ		
Gain Bandwidth Product	GBW	-	10	-	MHz	f=10kHz		
Unity Gain Frequency	f⊤	-	5	-	MHz	R <sub>L</sub> =2kΩ		
Total Harmonic Distortion+ Noise	THD+N	-	0.0005	-	%	$A_V$ =20dB, OUT=5Vrms R <sub>L</sub> =2k $\Omega$ f=1kHz, 20Hz~20kHz BPF		
Input Referred Noise Voltage	V <sub>N</sub>	-	5	-	nV/√Hz	R <sub>S</sub> =100Ω, V <sub>I</sub> =0V, f=1kHz		
input Neieneu Noise voitage	۷N	-	0.8	-	μVrms	RIAA, R <sub>S</sub> =2.2 kΩ, 30kHz LPF		
Channel Separation	CS	-	110	-	dB	R1=100Ω, f=1kHz		

(Note 14) Absolute value
 (Note 15) Current direction: Since first input stage is composed with PNP transistor, input bias current flows out of IC.

#### **Description of Electrical Characteristics**

Described below are descriptions of the relevant electrical terms used in this datasheet. Items and symbols used are also shown. Note that item name and symbol and their meaning may differ from those on another manufacturer's document or general document.

#### 1. Absolute Maximum Ratings

Absolute maximum rating items indicate the condition which must not be exceeded. Application of voltage in excess of absolute maximum rating or use out of absolute maximum rated temperature environment may cause deterioration of characteristics.

1.1 Power Supply Voltage (VCC-VEE)

Indicates the maximum voltage that can be applied between the positive power supply terminal and negative power supply terminal without deterioration or destruction of characteristics of internal circuit.

- 1.2 Differential Input Voltage (VID) Indicates the maximum voltage that can be applied between non-inverting and inverting terminals without damaging the IC.
- 1.3 Input Common-mode Voltage Range (VICM) Indicates the maximum voltage that can be applied to the non-inverting and inverting terminals without deterioration or destruction of electrical characteristics. Input common-mode voltage range of the maximum ratings does not assure normal operation of IC. For normal operation, use the IC within the input common-mode voltage range characteristics.
- 1.4 Power Dissipation (P<sub>D</sub>) Indicates the power that can be consumed by the IC when mounted on a specific board at the ambient temperature 25°C (normal temperature). As for package product, Pd is determined by the temperature that can be permitted by the IC in the package (maximum junction temperature) and the thermal resistance of the package.

#### 2. Electrical Characteristics Item

2.1 Input Offset Voltage (V<sub>IO</sub>) Indicates the voltage difference between non-inverting terminal and inverting terminals. It can be translated into the input voltage difference required for setting the output voltage at 0 V.

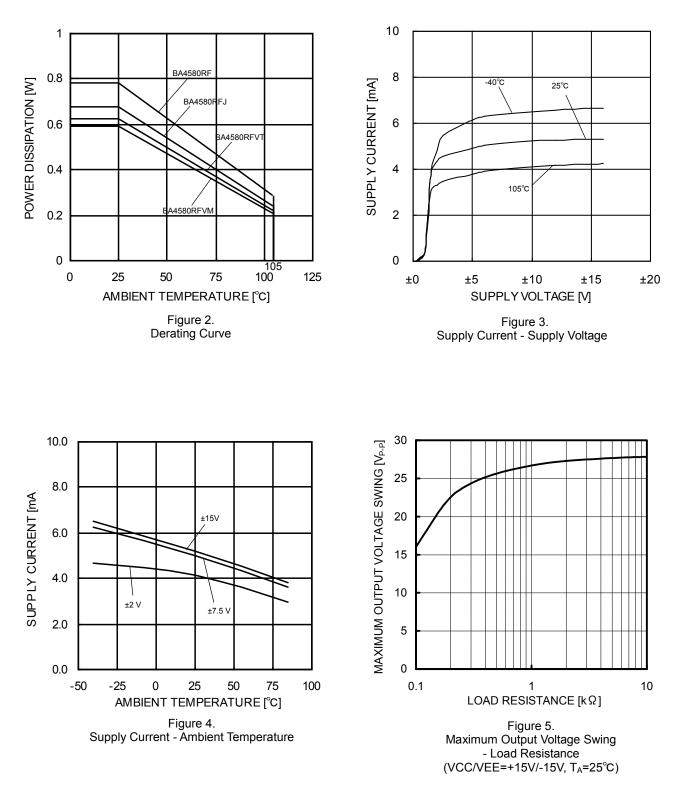
- 2.2 Input Offset Current (I<sub>IO</sub>) Indicates the difference of input bias current between the non-inverting and inverting terminals.
- 2.3 Input Bias Current (I<sub>B</sub>) Indicates the current that flows into or out of the input terminal. It is defined by the average of input bias currents at the non-inverting and inverting terminals.
- 2.4 Input Common-mode Voltage Range (VICM) Indicates the input voltage range where IC normally operates.
- 2.5 Large Signal Voltage Gain (A<sub>V</sub>) Indicates the amplifying rate (gain) of output voltage against the voltage difference between non-inverting terminal and inverting terminal. It is normally the amplifying rate (gain) with reference to DC voltage. Av = (Output voltage) / (Differential Input voltage)
- 2.6 Circuit Current (I<sub>CC</sub>) Indicates the current that flows within the IC under specified no-load conditions.
- 2.7 Output Saturation Voltage (V<sub>OM</sub>) Signifies the voltage range that can be output under specific output conditions.
- 2.8 Common-mode Rejection Ratio (CMRR) Indicates the ratio of fluctuation of input offset voltage when the input common mode voltage is changed. It is normally the fluctuation of DC. CMRR = (Change of Input common-mode voltage)/(Input offset fluctuation)
- 2.9 Power Supply Rejection Ratio (PSRR) Indicates the ratio of fluctuation of input offset voltage when supply voltage is changed. It is normally the fluctuation of DC.
  - PSRR= (Change of power supply voltage)/(Input offset fluctuation)
- 2.10 Channel Separation (CS) Indicates the fluctuation in the output voltage of the driven channel with reference to the change of output voltage of the channel which is not driven.
- 2.11 Slew Rate (SR)

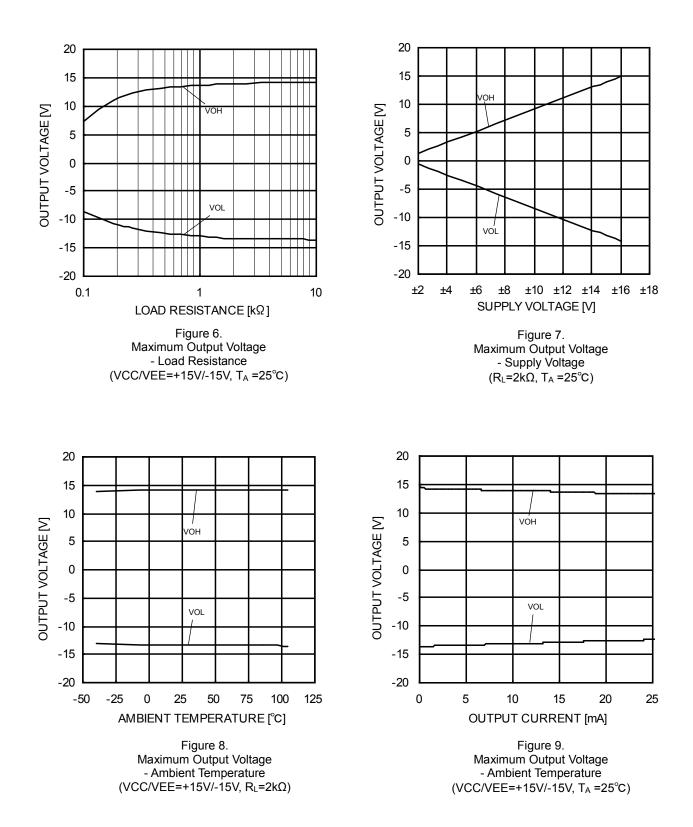
Indicates the ratio of the change in output voltage with time when a step input signal is applied.

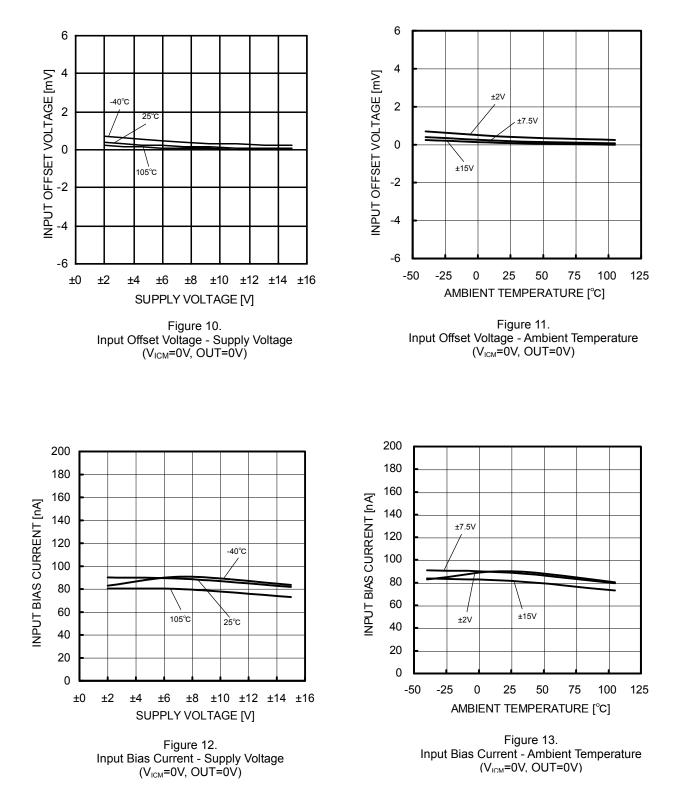
- 2.12 Gain Band Width (GBW) The product of the open-loop voltage gain and the frequency at which the voltage gain decreases 6dB/octave.
- 2.13 Unity Gain Frequency ( $f_T$ ) Indicates a frequency where the voltage gain of operational amplifier is 1.
- 2.14 Total Harmonic Distortion+ Noise (THD+N) Indicates the fluctuation of input offset voltage or that of output voltage with reference to the change of output voltage of driven channel.
- 2.15 Input Referred Noise Voltage ( $V_N$ ) Indicates a noise voltage generated inside the operational amplifier equivalent by ideal voltage source connected in series with input terminal.

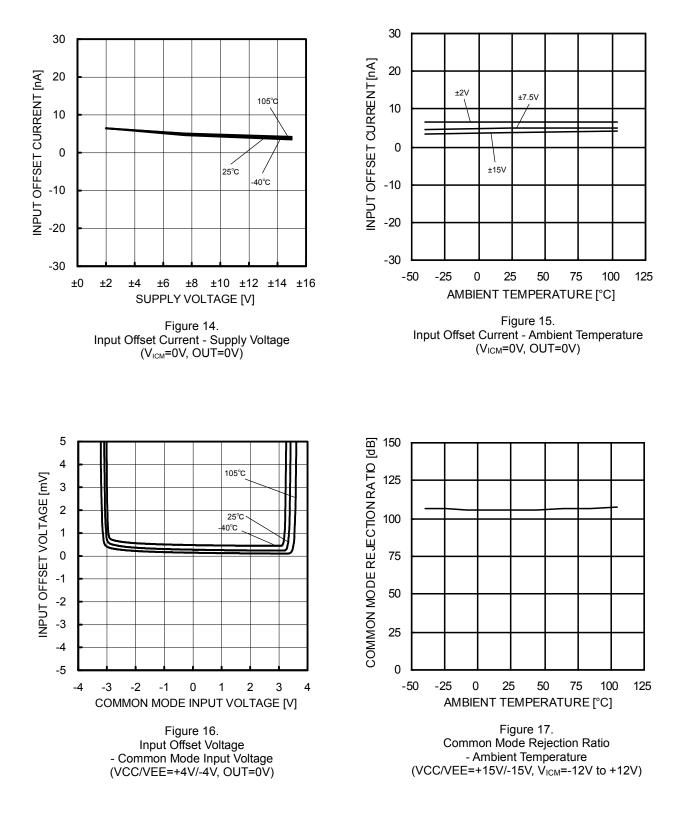
#### **Typical Performance Curves**

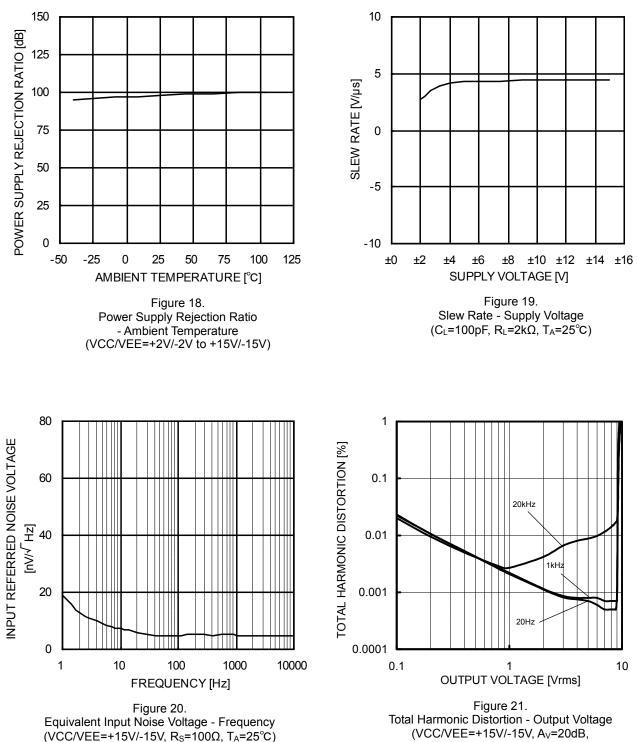
OBA4580Rxxx







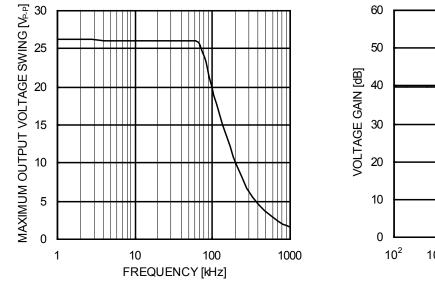


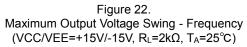


:C/VEE=+15V/-15V, R<sub>S</sub>=100Ω, T<sub>A</sub>=25°C)

(\*)The above data is measurement value of typical sample, it is not guaranteed.

 $R_L=2k\Omega$ , 80kHz-LPF,  $T_A=25^{\circ}C$ )





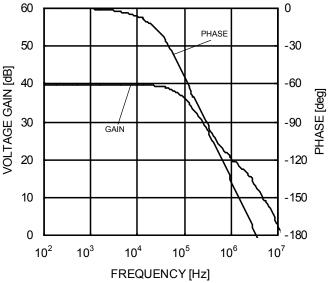
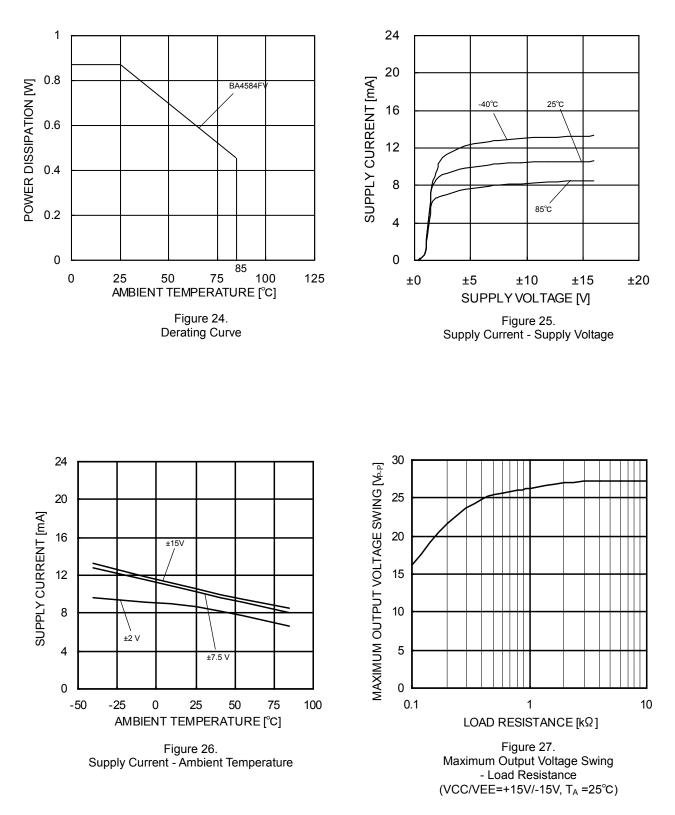
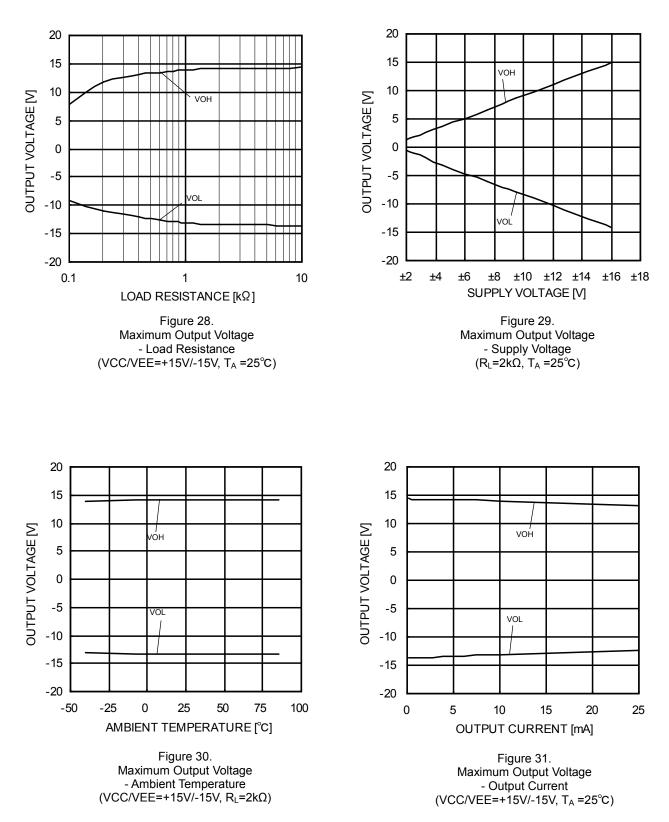
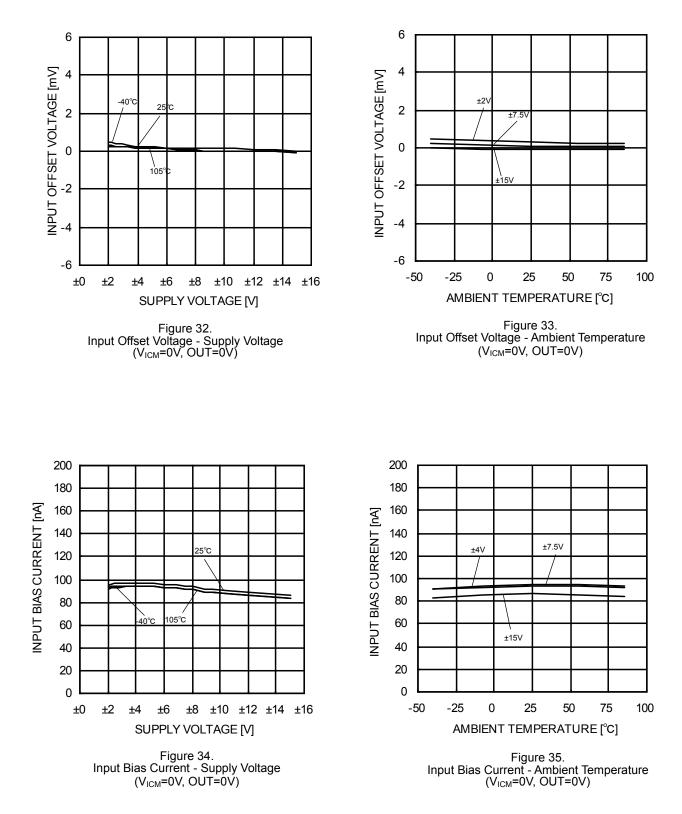
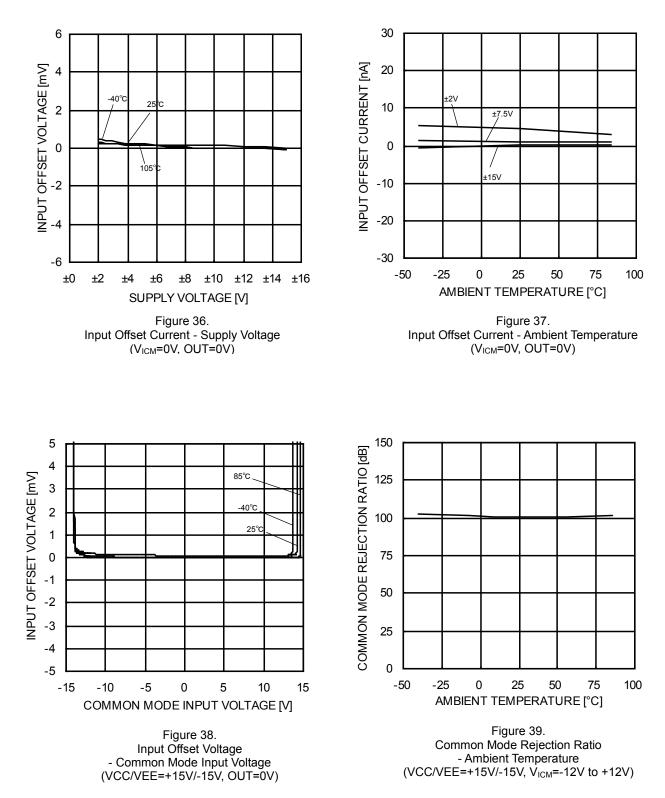


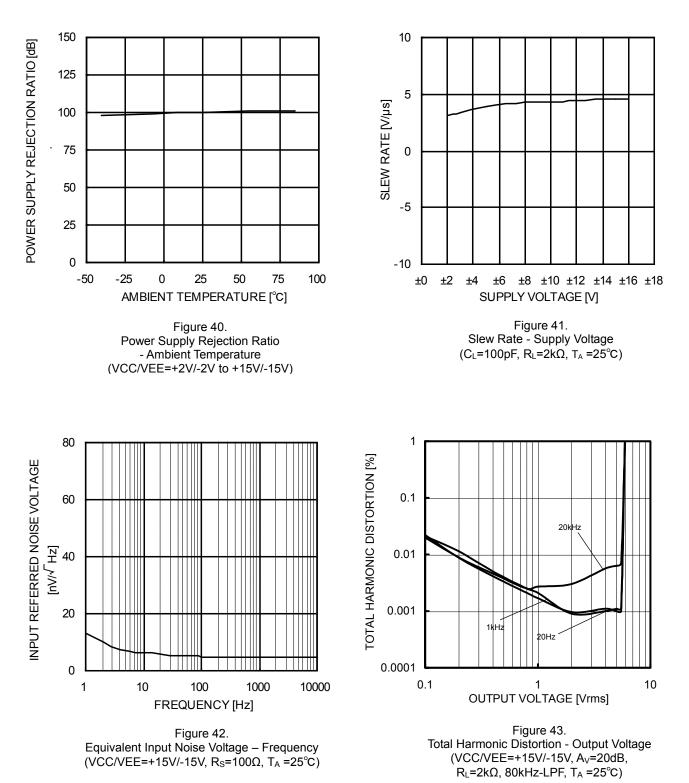
Figure 23. Voltage Gain • Phase - Frequency (VCC/VEE=+15V/-15V, A<sub>V</sub>=40dB, R<sub>L</sub>=2k $\Omega$ , T<sub>A</sub>=25°C)

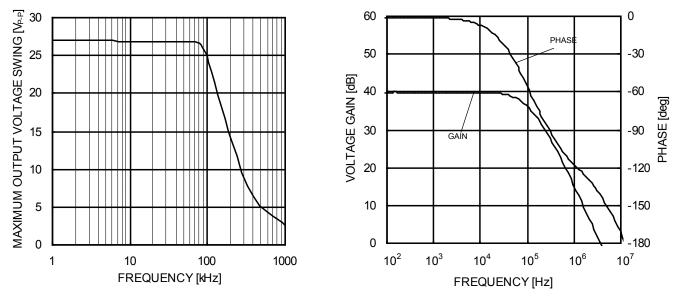












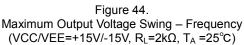
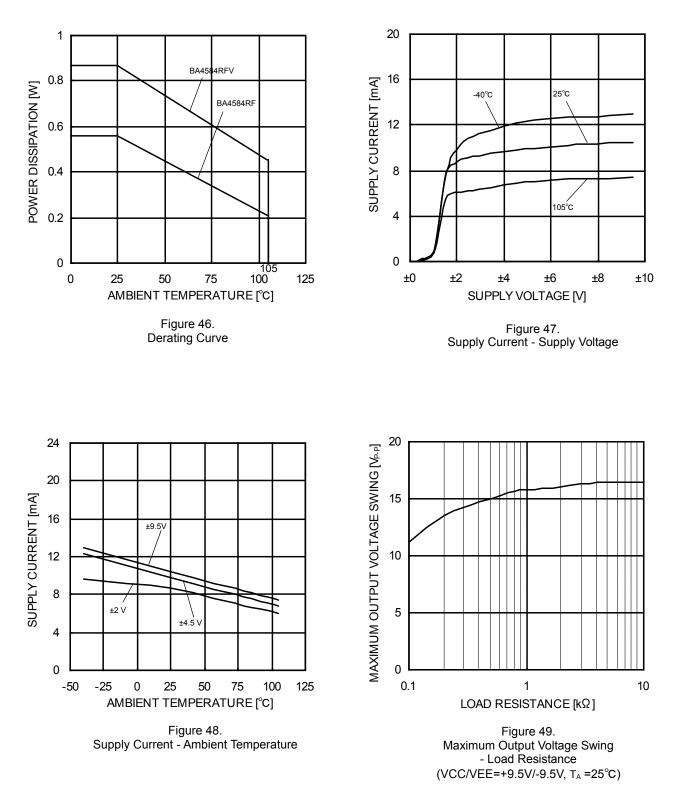
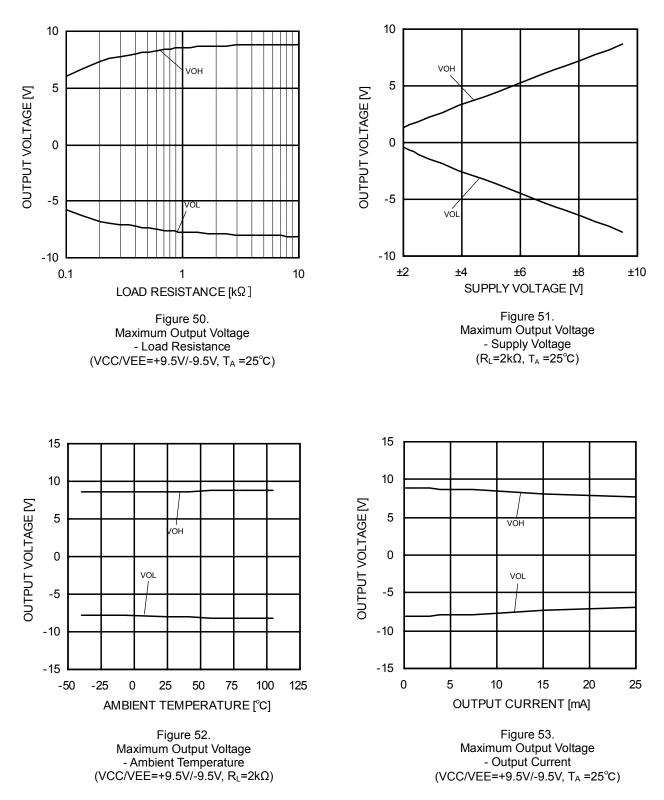
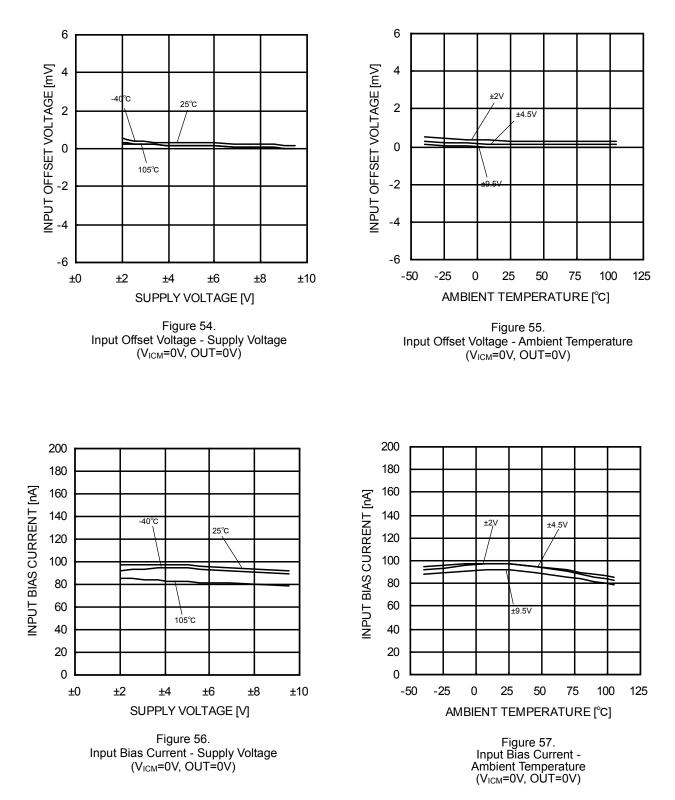
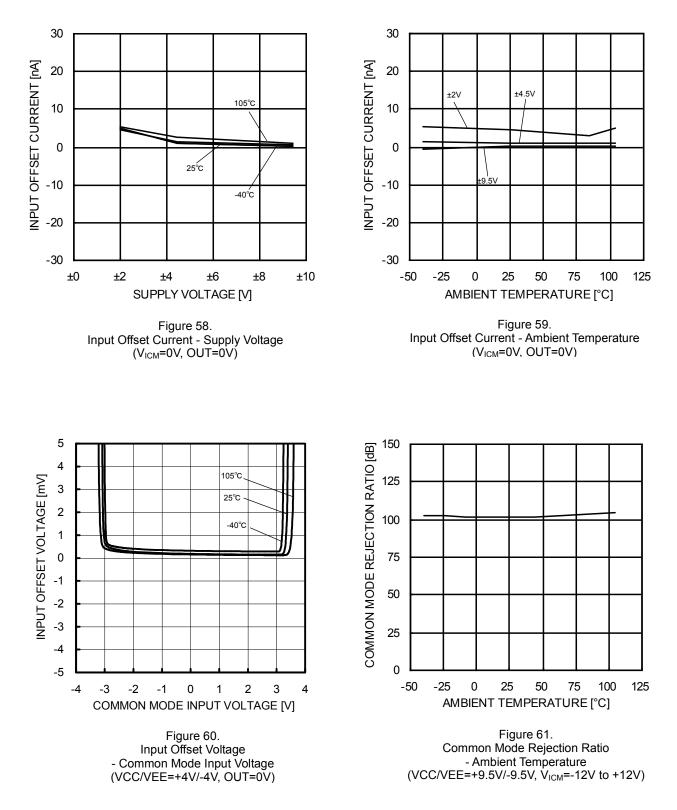


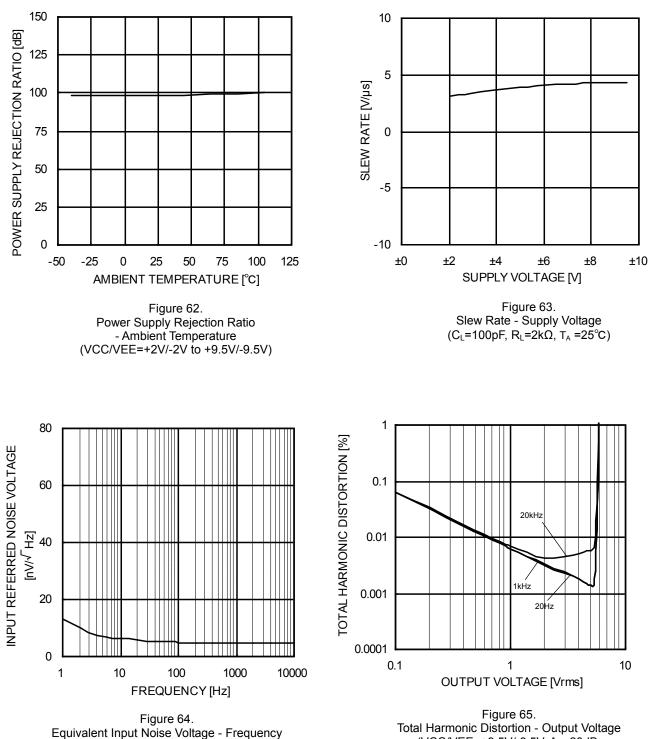
Figure 45. Voltage Gain • Phase - Frequency (VCC/VEE=+15V/-15V,  $A_V$ =40dB,  $R_L$ =2 $k\Omega$ ,  $T_A$ =25°C)





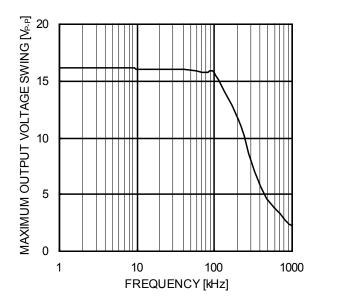


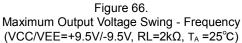


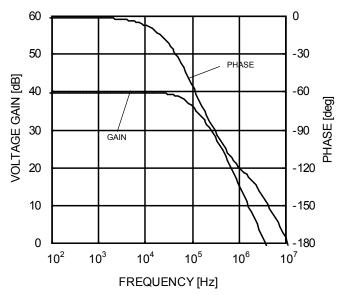


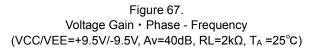
(VCC/VEE=+9.5V/-9.5V, R<sub>S</sub>=100Ω, T<sub>A</sub> =25°C)

(VCC/VEE=+9.5V/-9.5V, A<sub>V</sub>=20dB, R<sub>L</sub>=2kΩ, 80kHz-LPF, T<sub>A</sub> =25°C)









## Application Information

NULL method Condition for Test Circuit1

	Juiti								VCC, V	VEE, E <sub>κ</sub> ,	V <sub>ICM</sub> Unit: V	
Parameter	VF	S1	S2	S3	BA4580Rxxx, BA4584FV				Calculation			
					VCC	VEE	Eκ	VCC	VEE	Eκ		
Input Offset Voltage	V <sub>F1</sub>	ON	ON	OFF	15	-15	0	9.5	-9.5	0	1	
Input Offset Current	V <sub>F2</sub>	OFF	OFF	OFF	15	-15	0	9.5	-9.5	0	2	
Input Bias Current	$V_{F3}$	OFF	ON	OFF	15	-15	0	9.5	-9.5	0	3	
	$V_{F4}$	ON	OFF	OFF	15	-15	0	9.0	-9.0	0	5	
	$V_{F5}$	ON		ON	15	-15	-10	9.5	-9.5	-4.5	4	
Large Signal Voltage Gain	$V_{F6}$	ON	ON	ON	15	-15	10	9.5	-9.5	4.5	4	
Common-mode Rejection Ratio	$V_{F7}$			055	3	-27	12	3	-16	6.5	5	
(Input common-mode Voltage Range)	V <sub>F8</sub>	ON	ON	OFF	27	-3	-12	16	-3	-6.5	5	
Power Supply	$V_{F9}$				2	-2	0	2	-2	0	0	
Rejection Ratio	$V_{F10}$	ON	ON	OFF	15	-15	0	9.5	-9.5	0	6	

-Calculation-

2.

1. Input Offset Voltage (V<sub>IO</sub>)

$$V_{10} = \frac{|V_{F1}|}{1 + R_F/R_S}$$
 [V]

$$I_{IO} = \frac{|V_{F2}-V_{F1}|}{R_{I} \times (1+R_{F}/R_{S})}$$
 [A]

3. Input Bias Current (I<sub>B</sub>)

$$I_{B} = \frac{|V_{F4}-V_{F3}|}{2 \times R_{I} \times (1+R_{F}/R_{S})}$$
 [A]

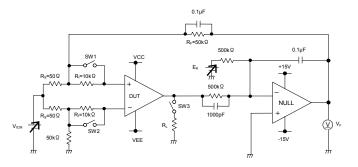
4. Large Signal Voltage Gain ( $A_V$ )

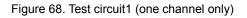
$$A_V = 20Log \qquad \frac{\Delta E_K \times (1+R_F/R_S)}{|V_{F5}-V_{F6}|} \qquad [dB]$$

- 5. Common-mode Rejection Ration (CMRR)  $CMRR = 20Log \quad \frac{\Delta V_{ICM} \times (1+R_F/R_S)}{|V_{F8}-V_{F7}|} \quad [dB]$
- 6. Power supply rejection ratio (PSRR)  $PSRR = 20Log \quad \frac{\Delta V_{CC} \times (1 + R_F/R_S)}{|V_{F10} - V_{F9}|} \quad [dB]$

## Switch Condition for Test Circuit 2

SW No.	SW1	SW2	SW3	SW4	SW5	SW6	SW7	SW8	SW9	SW10	SW11	SW12	SW13	SW14
Supply Current	OFF	OFF	OFF	ON	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF
High Level Output Voltage	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	ON	OFF
Low Level Output Voltage	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
Output Source Current	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON
Output Sink Current	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON
Slew Rate	OFF	OFF	OFF	ON	OFF	OFF	OFF	ON	ON	ON	OFF	OFF	OFF	OFF
Gain Bandwidth Product	OFF	ON	OFF	OFF	ON	ON	OFF	OFF	ON	ON	OFF	OFF	OFF	OFF
Equivalent Input Noise Voltage	ON	OFF	OFF	OFF	ON	ON	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF





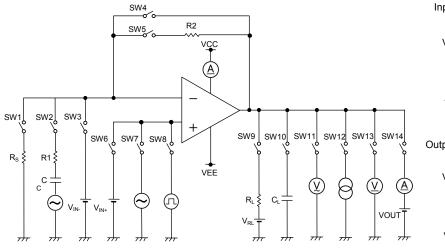


Figure 69. Test Circuit 2 (each Op-Amp)

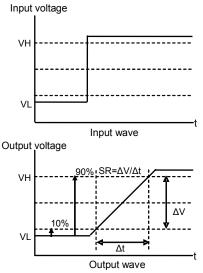


Figure 70. Slew Rate Input Waveform

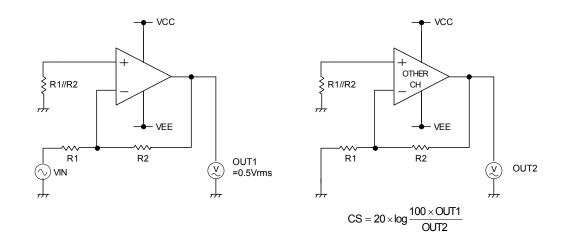


Figure 71. Test circuit 3 (Channel Separation) (VCC=+15V,VEE=-15V, R1=100Ω, R2=10kΩ)

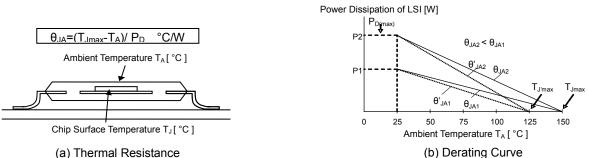
#### **Power Dissipation**

Power dissipation(total loss) indicates the power that can be consumed by IC at T<sub>A</sub> =25°C(normal temperature). IC is heated when it consumed power, and the temperature of IC chip becomes higher than ambient temperature. The temperature that can be accepted by IC chip depends on circuit configuration, manufacturing process, and consumable power is limited. Power dissipation is determined by the temperature allowed in IC chip(maximum junction temperature) and thermal resistance of package(heat dissipation capability). The maximum junction temperature is typically equal to the maximum value in the storage temperature range. Heat generated by consumed power of IC radiates from the mold resin or lead

frame of the package. The parameter which indicates this heat dissipation capability(hardness of heat release) is called thermal resistance, represented by the symbol  $\theta_{JA}^{\circ}C/W$ . The temperature of IC inside the package can be estimated by this thermal resistance. Figure 72. (a) shows the model of thermal resistance of the package. Thermal resistance  $\theta_{iA}$ , ambient temperature Ta, maximum junction temperature T<sub>JMAX</sub>, and power dissipation P<sub>D</sub> can be calculated by the equation below: :///

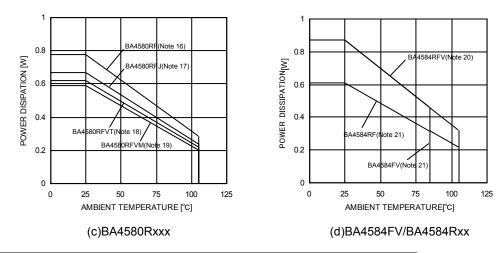
$$_{IA} = (T_{JMAX} - T_A) / P_D$$
 °C

Derating curve in Figure 72. (b) indicates power that can be consumed by IC with reference to ambient temperature. Power that can be consumed by IC with reference to ambient temperature. Power that can be consumed by IC begins to attenuate at certain ambient temperature. This gradient is determined by thermal resistance  $\theta_{JA}$ . Thermal resistance  $\theta_{JA}$  depends on chip size, power consumption, package, ambient temperature, package condition, wind velocity, etc even when the same of package is used. Thermal reduction curve indicates a reference value measured at a specified condition. Figure 73. (c),(d) show a derating curve for an example of BA4580Rxxx, BA4584FV, BA4584Rxx.



(a) Thermal Resistance

Figure 72. Thermal resistance and derating curve



(Note 16)	(Note 17)	(Note 18)	(Note 19)	(Note 20)	(Note 21)	Unit
6.2	5.4	5.0	4.8	7.0	4.9	mW/°C
When using the unit charge $T = 25\%$ , where the units charge and denote $\%$ . Demoissible distinction is the units						

When using the unit above  $T_A=25^{\circ}C$ , subtract the value above per degree  $^{\circ}C$ . Permissible dissipation is the value Permissible dissipation is the value when FR4 glass epoxy board 70mm ×70mm ×1.6mm (cooper foil area below 3%) is mounted.



## **Application Examples**

OVoltage Follower

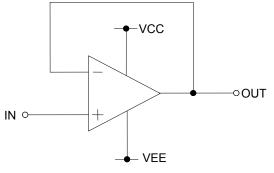


Figure 74. Voltage Follower Circuit

Voltage gain is 0dB.

Using this circuit, the output voltage (OUT) is configured to be equal to the input voltage (IN). This circuit also stabilizes the output voltage (OUT) due to high input impedance and low output impedance. Computation for output voltage (OUT) is shown below. OUT=IN

#### **OInverting Amplifier**

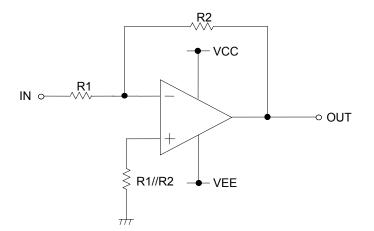


Figure 75. Inverting Amplifier Circuit

**ONon-inverting Amplifier** 

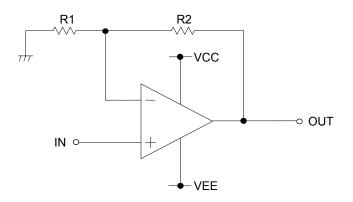


Figure 76. Non-inverting Amplifier Circuit

For inverting amplifier, input voltage (IN) is amplified by a voltage gain and depends on the ratio of R1 and R2. The out-of-phase output voltage is shown in the next expression

OUT=-(R2/R1) · IN

This circuit has input impedance equal to R1.

For non-inverting amplifier, input voltage (IN) is amplified by a voltage gain, which depends on the ratio of R1 and R2. The output voltage (OUT) is in-phase with the input voltage (IN) and is shown in the next expression.

OUT=(1 + R2/R1) · IN

Effectively, this circuit has high input impedance since its input side is the same as that of the operational amplifier.

### **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the  $P_D$  stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the  $P_D$  rating.

#### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

#### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

#### 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

## 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

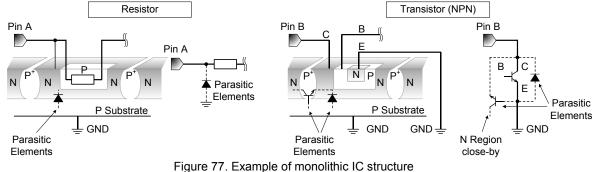
#### **Operational Notes – continued**

#### 11. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

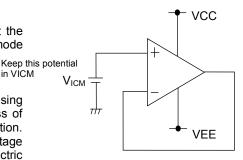


#### 12. Unused Circuits

It is recommended to apply the connection (see Figure 78.) and set the non-inverting input terminal at a potential within the Input Common-mode Voltage Range (VICM) for any unused circuit.

#### 13. Input Voltage

Applying VEE +36V to the input terminal is possible without causing deterioration of the electrical characteristics or destruction, regardless of the supply voltage. However, this does not ensure normal circuit operation. Please note that the circuit operates normally only when the input voltage is within the common mode input voltage range of the electric characteristics.



#### Figure 78. Example of Application Circuit for Unused Op-amp

#### 14. Power Supply(single/dual)

The operational amplifier operates when the voltage supplied is between VCC and VEE. Therefore, the single supply operational amplifier can be used as dual supply operational amplifier as well.

in VICM

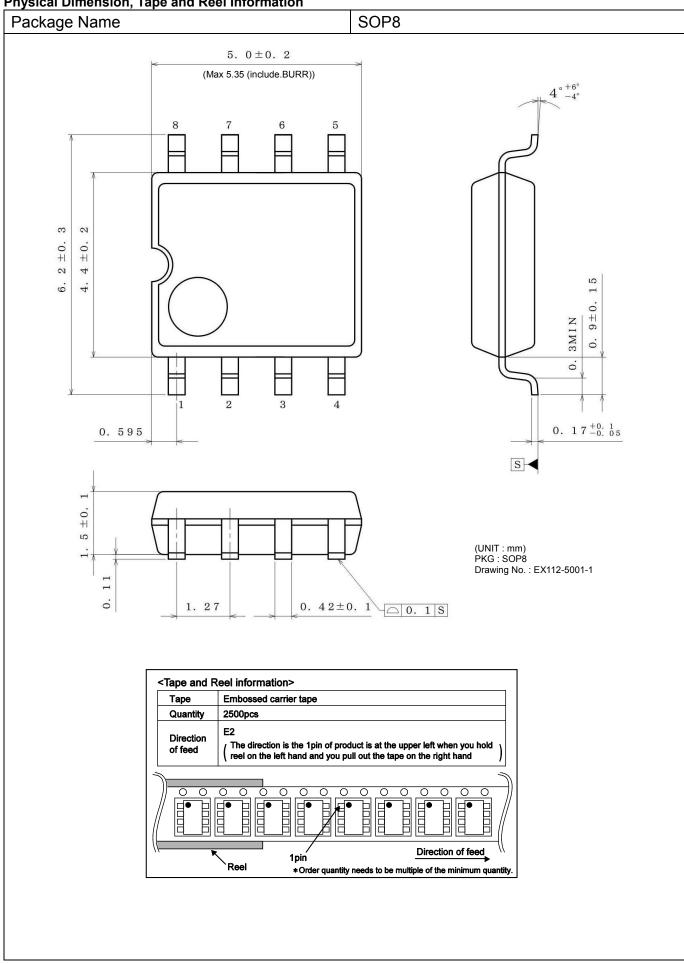
#### 15. IC Handling

When pressure is applied to the IC through warp on the printed circuit board, the characteristics may fluctuate due to the piezo effect. Be careful with the warp on the printed circuit board.

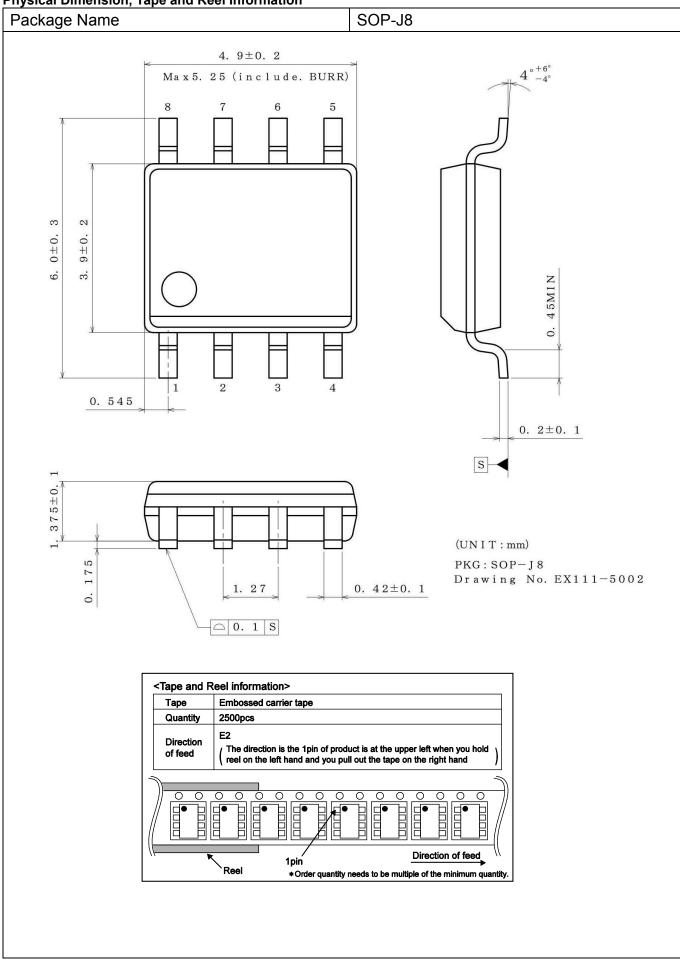
#### 16. The IC Destruction Caused by Capacitive Load

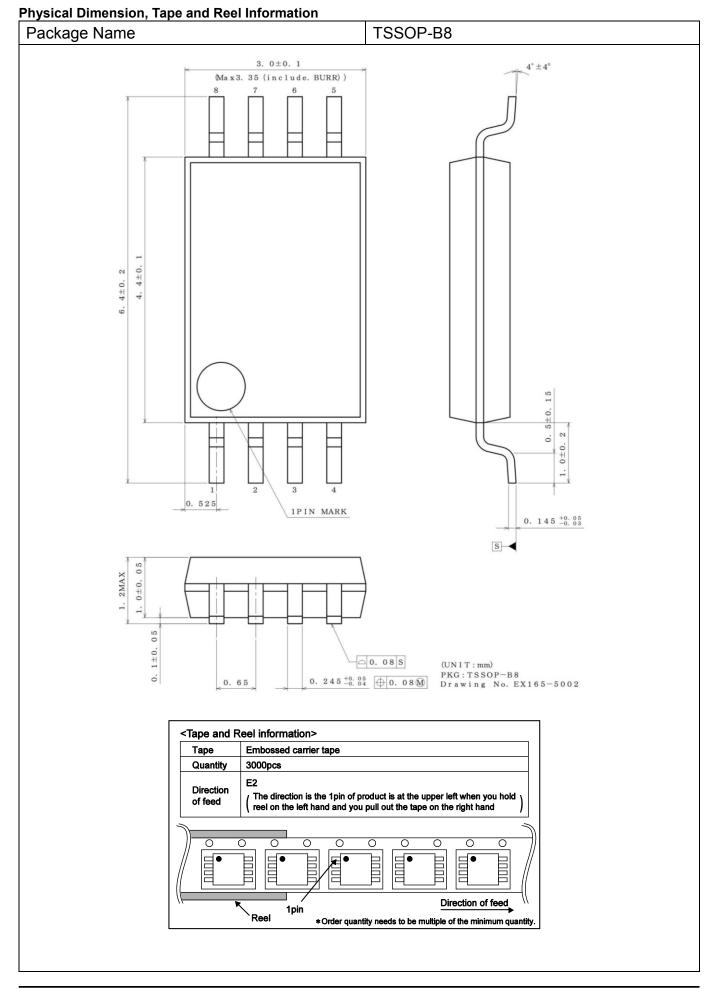
The IC may be damaged when VCC terminal and VEE terminal is shorted with the charged output terminal capacitor. When IC is used as an operational amplifier or as an application circuit where oscillation is not activated by an output capacitor, output capacitor must be kept below 0.1µF in order to prevent the damage mentioned above.

## Physical Dimension, Tape and Reel Information

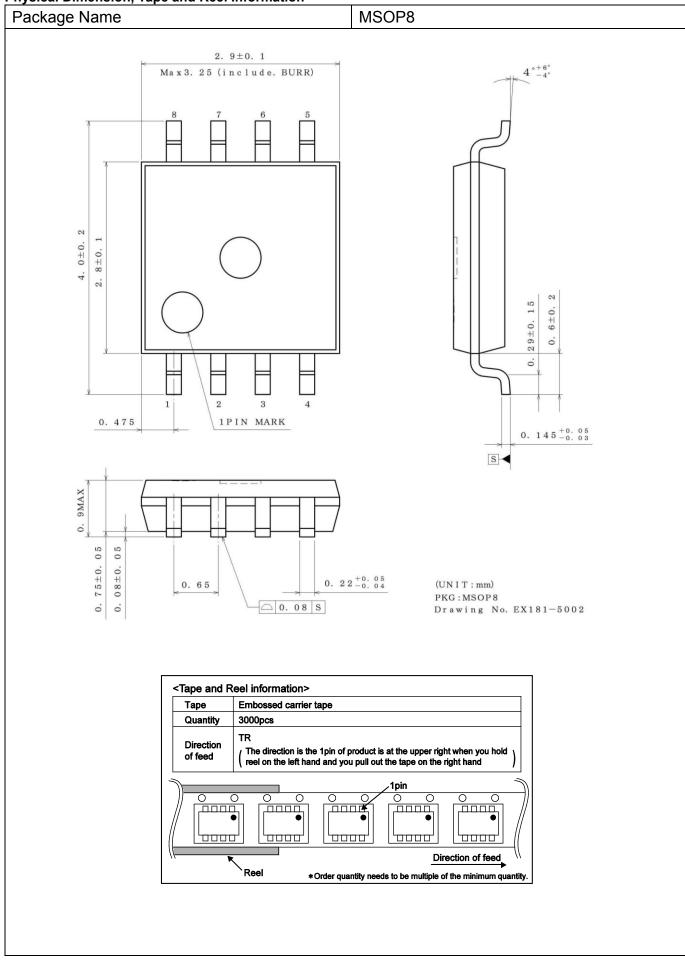


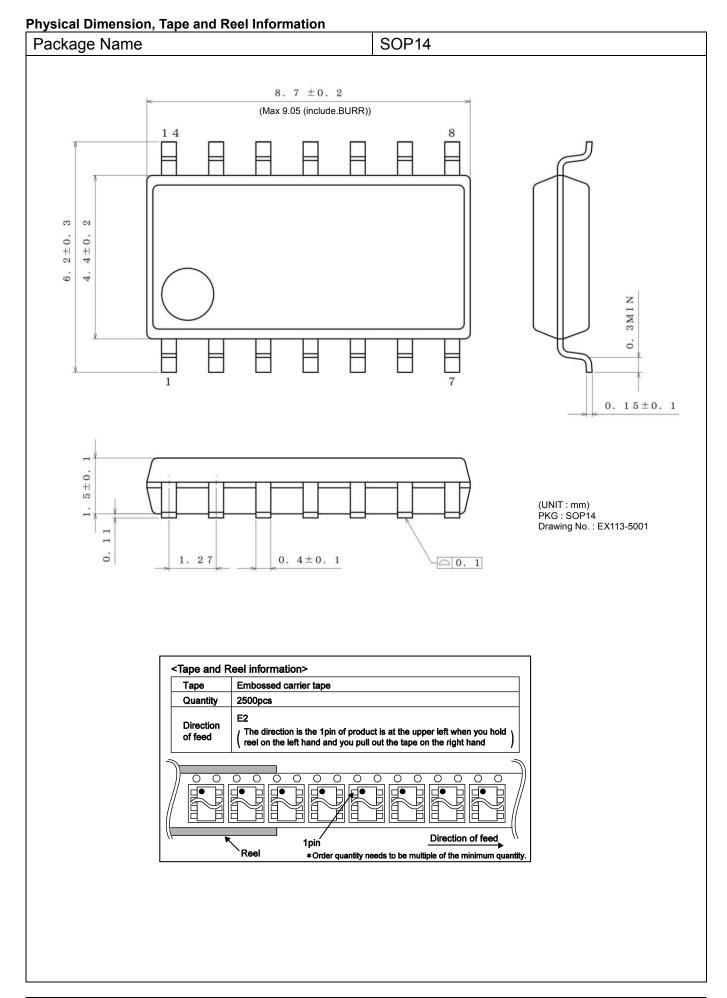




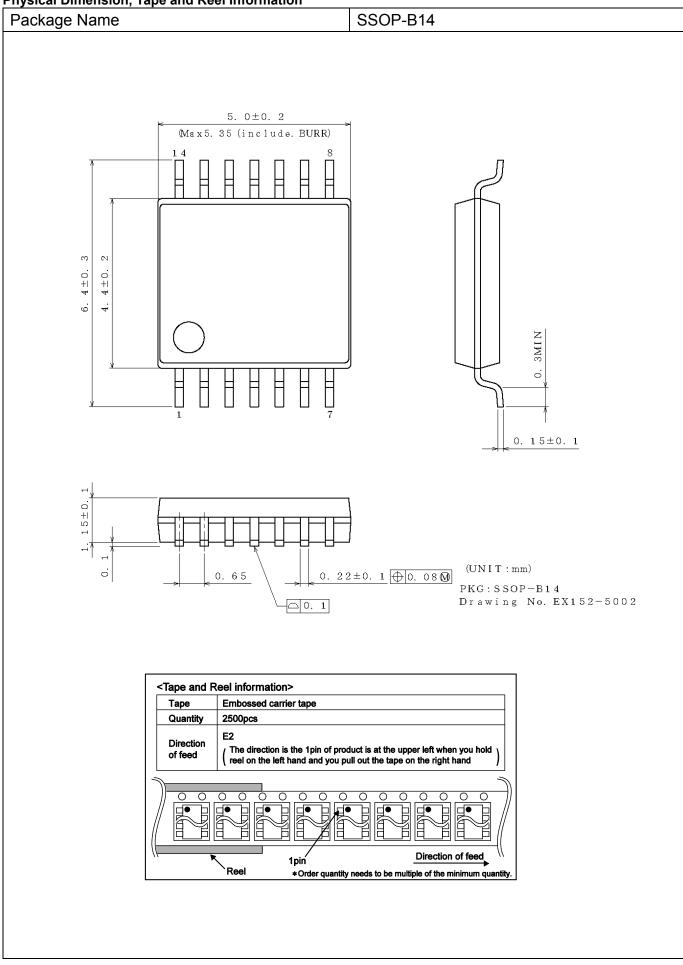


## Physical Dimension, Tape and Reel Information

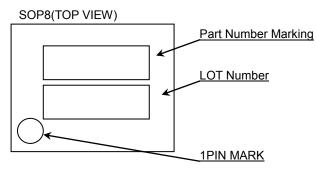


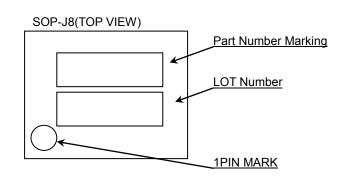


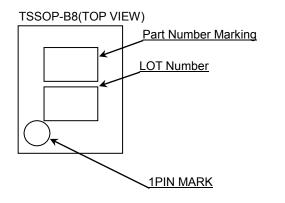
## Physical Dimension, Tape and Reel Information

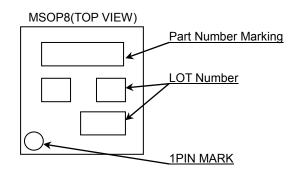


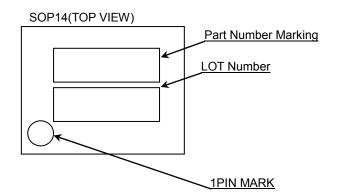
## **Marking Diagrams**

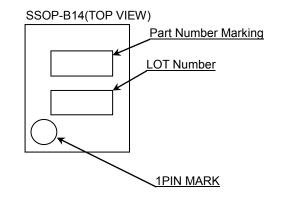












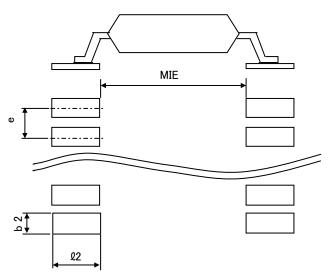
Product N	lame	Package Type	Marking	
	F	SOP8	4580R	
BA4580Rxxx	FJ	SOP-J8		
BA458URXXX	FVT	TSSOP-B8		
	FVM	MSOP8		
BA4584FV	BA4584FV FV		4584	
BA4584Rxx	F	SOP14	BA4584RF	
	FV	SSOP-B14	4584R	

## Land Pattern Data

			all ull	nensions in mm
PKG	Land pitch e	Land space MIE	Land length ≧ℓ 2	Land width b2
SOP8 SOP14	1.27	4.60	1.10	0.76
SOP-J8	1.27	3.90	1.35	0.76
SSOP-B14	0.65	4.60	1.20	0.35
MSOP8	0.65	2.62	0.99	0.35
TSSOP-B8	0.65	4.60	1.20	0.35

all dimensions in mm

## SOP8, SOP14, SOP-J8, SSOP-B14, MSOP8, TSSOP-B8



## **Revision History**

Date	Revision	Changes		
27.Feb.2012	001	New Release		
31.Oct.2014	002	Page.3 Absolute Maximum Ratings : Added Input Current		

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  - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl<sub>2</sub>, H<sub>2</sub>S, NH<sub>3</sub>, SO<sub>2</sub>, and NO<sub>2</sub>
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  - [c] the Products are exposed to direct sunshine or condensation
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- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



#### Как с нами связаться

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