

For air-conditioner fan motor **3-Phase Brushless Fan Motor** Driver



3.0W

BM6202FS

General Description

This motor driver IC adopts PrestoMOS[™] as the output transistor, and put in a small full molding package with the high voltage gate driver chip. The protection circuits for overcurrent, overheating, under voltage lock out and the high voltage bootstrap diode with current regulation are built-in. It provides optimum motor drive system for a wide variety of applications by the combination with controller BD6201X series and enables motor unit standardization.

Features

- 600V PrestoMOSTM built-in
- Output current 1.5A
- Bootstrap operation by floating high side driver (including diode)
- 3.3V logic input compatible
- Protection circuits provided: OCP, TSD and UVLO
- Fault output (open drain)

Applications

- Air conditioners; air cleaners; water pumps; dishwashers; washing machines
- General OA equipment

Key Specifications

Package

- Output MOSFET voltage: 600V
- Driver output current (DC): ±1.5A(Max.)
- Driver output current (Pulse): ±2.5A(Max.)
- Output MOSFET DC on resistance: 2.7Ω (Typ.)
- Operating case temperature: -20°C to +100°C
- Junction temperature: +150°C
- Power dissipation:

W (Typ.) x D (Typ.) x H (Max.) SSOP-A54 23 22.0 mm x 14.1 mm x 2.4 mm



VREG FG C ≶R9 R8≷ О C13 DTR C14 VSP (BD6201XFS C.7 C2~C4 C8 \$ Μ \$ -~~~ VREG ⊥ ___5 C, } |R3 C10 VCC O Ţ^{c6} Ŷ R6 BM6202FS GND O D1 4 ₩ R7 VDC O-

Figure 1. Application circuit example - BM6202FS & BD6201XFS

OProduct structure : Semiconductor IC OThis product is not designed protection against radioactive rays

Typical Application Circuit

Block Diagram and Pin Configuration



Figure 2. Block diagram

Figure 3. Pin configuration

Pin	Name	Function	Pin	Name	Function
1	VCC	Low voltage power supply	23	VDC	High voltage power supply
2	FOB	Fault signal output (open drain)	-	VDC	
3	UH	Phase U high side control input	22	BU	Phase U floating power supply
4	UL	Phase U low side control input	-	U	
5	NC		21	U	Phase U output
6	VH	Phase V high side control input	20	BV	Phase V floating power supply
7	VL	Phase V low side control input	-	V	
8	NC		19	V	Phase V output
9	NC		-	VDC	
10	WH	Phase W high side control input	18	VDC	High voltage power supply
11	WL	Phase W low side control input	17	BW	Phase W floating power supply
12	FOB	Fault signal output (open drain)	-	W	
13	VCC	Low voltage power supply	16	W	Phase W output
14	GND	Ground	15	PGND	Ground (current sense pin)

• Pin Descriptions (NC: No Connection)

Note) All pin cut surfaces visible from the side of package are no connected, except the pin number is expressed as a "-".

Functional Descriptions

1) Control input pins (UH, UL, VH, VL, WH, WL)

The input threshold voltages of the control pins are 2.5V and 0.8V, with a hysteresis voltage of approximately 0.4V. The IC will accept input voltages up to the VCC voltage. When the same phase control pins are input high at the same time, the high side and low side gate driver outputs become low. Dead time is installed in the control signals. The control input pins are connected internally to pull-down resistors (100k Ω nominal). However, the switching noise on the output stage may affect the input on these pins and cause undesired operation. In such cases, attaching an external pull-down resistor (10k Ω recommended) between each control pin and ground, or connecting each pin to an input voltage of 0.8V or less (preferably GND), is recommended.

2) Under voltage lock out (UVLO) circuit

To secure the lowest power supply voltage necessary to operate the driver, and to prevent under voltage malfunctions, the UVLO circuits are independently built into the upper side floating driver and the lower side driver. When the supply voltage falls to V_{UVL} or below, the controller forces driver outputs low. When the voltage rises to V_{UVH} or above, the UVLO circuit ends the lockout operation and returns the chip to normal operation. Even if the controller returns to normal operation, the output begins from the following control input signal.



Figure 4. Low voltage monitor - UVLO - timing chart

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Truth	table

I futri table						
HIN	LIN	HO	LO			
L	L	L	L			
Н	L	Н	L			
L	Н	L	Н			
Н	Н	Inhib	oition			

Note) HIN: UH,VH,WH, LIN: UL,VL,WL

3) Bootstrap operation



Figure 5. Charging period

Figure 6. Discharging period

The bootstrap is operated by the charge period and the discharge period being alternately repeated for bootstrap capacitor (CB) as shown in the figure above. In a word, this operation is repeated while the output of an external transistor is switching with synchronous rectification. Because the supply voltage of the floating driver is charged from the VCC power supply to CB through prevention of backflow diode DX, it is approximately (VCC-1V). The resistance series connection with DX has the impedance of approximately 200 Ω .

The capacitance value for the bootstrap is the following formula:

Example:

Floating driver power supply quiescence current I_{BBQ} : 150µA(max.) Bootstrap diode reverse bias current I_{LBD} : 10µA(max.) Carrier frequency F_{PWM} : 20kHz Output MOSFET total gate charge Qg : 25nC(max.) Floating driver transmission loss Q_{LOSS} : 1nC(max.) Drop voltage of the floating driver power supply dV_{DROP} : 3V

 C_{BOOT} » (($I_{BBQ} + I_{LBD}$) / $F_{PWM} + 2 \times Qg + Q_{LOSS}$) / $dV_{DROP} \approx 20nF$

The allowed drop voltage actually becomes smaller by the range of the used power supply voltage, the output MOSFET ON resistance, the forward voltages of the internal boot diode (the drop voltage to the capacitor by the charge current), and the power supply voltage monitor circuits etc. Please set the calculation value to the criterion about the capacitance value tenfold or more to secure the margin in consideration of temperature characteristics and the value change, etc. Moreover, the example of the mentioned above assumes the synchronous rectification switching. Because the total gate charge is needed only by the carrier frequency in the upper switching section, for example 150° commutation driving, it becomes a great capacity shortage in the above settings. Please set it after confirming actual application operation.

4) Thermal shutdown (TSD) circuit

The TSD circuit operates when the junction temperature of the gate driver exceeds the preset temperature (150°C nominal). At this time, the controller forces all driver outputs low. Since thermal hysteresis is provided in the TSD circuit, the chip returns to normal operation when the junction temperature falls below the preset temperature (125°C nominal). The TSD circuit is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation in the presence of extreme heat. Do not continue using the IC after the TSD circuit is activated, and do not use the IC in an environment where activation of the circuit is assumed. Moreover, it is not possible to follow the output MOSFET junction temperature rising rapidly because it is a gate driver chip that monitors the temperature and it is likely not to function effectively.

5) Overcurrent protection (OCP) circuit

The overcurrent protection circuit can be activated by connecting a low value resistor for current detection between the PGND pin and the GND pin. When the PGND pin voltage reaches or surpasses the threshold value (0.9V typical), the gate driver outputs low to the gate of all output MOSFETs, thus initiating the overcurrent protection operation.

6) Fault signal output

When the gate driver detects either state that should be protected (UVLO / TSD / OCP), the FOB pin outputs low (open drain) for at least 25µs nominal. The FOB pin has wired-OR connection with each phase gate driver chip internally, and into another phase also entering the protection operation. Even when this function is not used, the FOB pin is pull-up to the voltage of 3V or more and at least a resistor with a value 10k Ω or more. Moreover, the signal from the outside of the chip is not passed because of the built-in analog filter, but the internal control signals (UVLO / TSD / OCP) pass the filter (2.0µs Min.) for the malfunction prevention by the switching noise, etc.







Figure 8. Fault operation ~ OCP ~ timing chart

The release time from the protection operation can be changed by inserting an external capacitor. Refer to the formula below. Release time of 2ms or more is recommended.



Figure 9. Release time setting application circuit



Figure 10. Release time (reference data @R=100kΩ)

When using controller BD6201X series as a control IC, the FOB pin can be linked to the external fault signal input pin of the side of the control IC since it has the internal pull-up resistor. Refer to figure 11.



Figure 11. Interface equivalent circuit

7) Switching time



Figure 12. Switching time definition

Parameter	Symbol	Reference	Unit	Conditions
	t _{dH(on)}	930	ns	
	t _{rH}	170	ns	
High side switching time	t _{rrH}	250	ns	
	t _{dH(off)}	600	ns	
	t _{fH}	20	ns	VDC=300V, VCC=15V, I _D =0.75A
	t _{dL(on)}	990	ns	VIN= 0V↔5V, Inductive load
	t _{rL}	220	ns	
Low side switching time	t _{rrL}	250	ns	
	t _{dL(off)}	600	ns	
	t _{fL}	60	ns	

• Absolute Maximum Ratings (Ta=25°C)

Parameter	Symbol	Ratings	Unit
Parameter	Symbol	BM6202FS	Unit
Output MOSFET	V _{DSS}	600* ¹	V
Supply voltage	V _{DC}	-0.3 to 600*1	V
Output voltage	V_{U}, V_{V}, V_{W}	-0.3 to 600*1	V
High side supply pin voltage	V _{BU} , V _{BV} , V _{BW}	-0.3 to 600*1	V
High side floating supply voltage	V _{BU} -V _U , V _{BV} -V _V , V _{BW} -V _W	-0.3 to 20	V
Low side supply voltage	V _{CC}	-0.3 to 20	V
All others	V _{I/O}	-0.3 to V_{CC}	V
Driver outputs (DC)	I _{OMAX(DC)}	±1.5* ²	A
Driver outputs (Pulse)	I _{OMAX(PLS)}	±2.5* ²	А
Fault signal output	I _{OMAX(FOB)}	15* ¹	mA
Power dissipation	Pd	3.00* ³	W
Thermal resistance	R _{thj-c}	15	°C/W
Operating case temperature	T _c	-20 to 100	°C
Storage temperature	T _{STG}	-55 to 150	°C
Junction temperature	T _{jmax}	150	°C

Note) All voltages are with respect to ground.

*1 Do not, however, exceed Pd or ASO.

*2 Pw \leq 10µs, Duty cycle \leq 1%

*3 Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 24mW/°C above 25°C.

• Operating Conditions (Tc=25°C)

Parameter	Symbol		Unit		
Falameter	Symbol Min.		Тур.	Max.	Unit
Supply voltage	V _{DC}	-	310	400	V
High side floating supply voltage	V_{BU} - V_{U} , V_{BV} - V_{V} , V_{BW} - V_{W}	13.5	15	16.5	V
Low side supply voltage	V _{cc}	13.5	15	16.5	V
Minimum input pulse width	T _{MIN}	0.8	-	-	μs
Dead time	T _{DT}	1.5	-	-	μs
Shunt resistor (PGND)	R _S	0.4	-	-	Ω
Junction temperature	Tj	-	-	125	°C

Note) All voltages are with respect to ground.

• Electrical Characteristics (Unless otherwise specified, Ta=25°C and VCC=15V)

Parameter	Limits		– Unit	Que dition o			
Parameter	Symbol	Min. Typ.		Max.	Unit	Conditions	
Power supply					-		
HS quiescence current	I _{BBQ}	30	70	150	μA	XH=XL=L, each phase	
LS quiescence current	I _{CCQ}	0.4	0.9	1.5	mA	XH=XL=L	
Output MOSFET							
D-S breakdown voltage	V _{(BR)DSS}	600	-	-	V	I _D =1mA, XH=XL=L	
Leak current	I _{DSS}	-	-	100	μA	V _{DS} =600V, XH=XL=L	
DC on resistance	R _{DS(ON)}	-	2.7	3.5	Ω	I _D =0.75A	
Diode forward voltage	V _{SD}	-	1.1	1.5	V	I _D =0.75A	
Bootstrap diode	·					·	
Leak current	I _{LBD}	-	-	10	μA	V _{BX} =600V	
Forward voltage	V _{FBD}	1.5	1.8	2.1	V	I _{BD} =-5mA, including series-R	
Series resistance	R _{BD}	-	200	-	Ω		
Control inputs							
Input bias current	I _{XIN}	30	50	70	μA	V _{IN} =5V	
Input high voltage	V _{XINH}	2.5	-	VCC	V		
Input low voltage	V _{XINL}	0	-	0.8	V		
Under voltage lock out					1	•	
HS release voltage	V _{BUVH}	9.5	10.0	10.5	V	V _{BX} - V _X	
HS lockout voltage	V _{BUVL}	8.5	9.0	9.5	V	V _{BX} - V _X	
LS release voltage	V _{CCUVH}	11.0	11.5	12.0	V		
LS lockout voltage	V _{CCUVL}	10.0	10.5	11.0	V		
Overcurrent protection							
Threshold voltage	V _{SNS}	0.8	0.9	1.0	V		
Fault output							
Output low voltage	V _{FOL}	-	-	0.8	V	I ₀ =+10mA	
Input high voltage	V _{FINH}	2.5	-	VCC	V		
Input low voltage	V _{FINL}	0	-	0.8	V		
Noise masking time	T _{MASK}	2.0	-	-	μs		

• Typical Performance Curves (Reference data)



Figure 15. Low side drivers operating current (F_{PWM}: 20kHz, two phase switching)





Figure 17. High side driver operating current (F_{PWM}: 20kHz, each phase)

Figure 18. Input bias current (UH,UL,VH,VL,WH,WL)



Figure 19. Input threshold voltage (UH,UL,VH,VL,WH,WL,FOB)



Figure 20. Overcurrent detection voltage



Figure 21. Thermal shut down

Figure 22. Noise masking time



Figure 23. Release time (No external capacitor)



Figure 24. Fault output ON resistance



Figure 28. Input/Output propagation delay



Figure 29. Output MOSFET ON resistance

Figure 30. Output MOSFET body diode



Figure 31. Bootstrap diode forward voltage



Figure 32. Bootstrap series resistor









Figure 36. Low side recovery loss (VDC=300V)

Application Circuit Example



Figure 37. Application circuit example (150° commutation driver)

arts list		1		1			
Parts	Value	Manufacturer	Туре	Parts	Value	Ratings	Туре
IC1	-	ROHM	BM6202FS	C1	0.1µF	50V	Ceramic
IC2	-	ROHM	BD62012FS	C2~4	2200pF	50V	Ceramic
R1	1kΩ	ROHM	MCR18EZPF1001	C5	10 µF	50V	Ceramic
R2	150Ω	ROHM	MCR18EZPJ151	C6	10 µF	50V	Ceramic
R3	22kΩ	ROHM	MCR18EZPF2202	C7~9	1µF	50V	Ceramic
R4	100kΩ	ROHM	MCR18EZPF1003	C10	0.1µF	50V	Ceramic
R5	100kΩ	ROHM	MCR18EZPF1003	C11	1µF	50V	Ceramic
R6	0.5Ω	ROHM	MCR50JZHFL1R50 x 3	C12	100pF	50V	Ceramic
R7	10kΩ	ROHM	MCR18EZPF1002	C13	0.1µF	630V	Ceramic
R8	0Ω	ROHM	MCR18EZPJ000	C14	0.1µF	50V	Ceramic
R9	0Ω	ROHM	MCR18EZPJ000	HX	-	-	Hall elements
Q1	-	ROHM	DTC124EUA				
D1	-	ROHM	KDZ20B				

Interfaces







Figure 38. UH, UL, VH, VL, WH, WL





Figure 40. FOB

Figure 41. VCC, GND, VDC, BX(BU/BV/BW), X(U/V/W)

Notes for Use

1) Absolute maximum ratings

Devices may be destroyed when supply voltage or operating temperature exceeds the absolute maximum rating. Because the cause of this damage cannot be identified as, for example, a short circuit or an open circuit, it is important to consider circuit protection measures, such as adding fuses, if any value in excess of absolute maximum ratings is to be implemented.

2) Electrical potential at GND

Keep the GND terminal to the minimum potential under any operating condition. In addition, check to determine whether there is any terminal that provides voltage below GND, including the voltage during transient phenomena. However, note that even if the voltage does not fall below GND in any other operating condition, it can still swing below GND potential when the motor generates back electromotive force at the PGND pin. The chip layout in this product is designed to avoid this sort of electrical potential problem, but pulling excessive current may still result in malfunctions. Therefore, it is necessary to observe operation closely to conclusively confirm that there is no problem in actual operation. If there are a small signal GND and a high current GND, it is recommended to separate the patterns for the high current GND and the small signal GND and provide a proper grounding to the reference point of the set not to affect the voltage at the small signal GND with the change in voltage due to resistance component of pattern wiring and high current. Also for GND wiring pattern of the component externally connected, pay special attention not to cause undesirable change to it.

3) High voltage terminal – VDC, BU/U, BV/V and BW/W

When using this IC, the high voltage terminals VDC, BU/U, BV/V and BW/W need a resin coating between these pins. It is judged that the inter-pins distance is not enough. If any special mode in excess of absolute maximum ratings is to be implemented with this product or its application circuits, it is important to take physical safety measures, such as providing voltage-clamping diodes or fuses. And, set the output transistor so that it does not exceed absolute maximum ratings or ASO. In the event a large capacitor is connected between the output and ground, and if VCC and VDC are short-circuited with 0V or ground for any reason, the current charged in the capacitor flows into the output and may destroy the IC.

4) Power supply lines

Return current generated by the motor's Back-EMF requires countermeasures, such as providing a return current path by inserting capacitors across the power supply and GND (10μ F, ceramic capacitor is recommended). In this case, it is important to conclusively confirm that none of the negative effects sometimes seen with electrolytic capacitors including a capacitance drop at low temperatures occurs. Also, the connected power supply must have sufficient current absorbing capability. Otherwise, the regenerated current will increase voltage on the power supply line, which may in turn cause problems with the product, including peripheral circuits exceeding the absolute maximum rating. To help protect against damage or degradation, physical safety measures should be taken, such as providing a voltage-clamping diode across the power supply and GND.

5) Thermal design

Use a thermal design that allows sufficient margin in light of the power dissipation (Pd) in actual operating conditions.

6) Inter-pin shorts and mounting errors

Take caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error or if pins are shorted together. Also, connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply lines, such as establishing an external diode between the power supply and the IC power supply pin.

7) Operation in strong electromagnetic fields

Using this product in strong electromagnetic fields may cause IC malfunctions. Take extreme caution with electromagnetic fields.

8) Testing on application boards

When testing the IC on an application board, connecting a capacitor to a low impedance pin subjects the IC to stress. Always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

9) Regarding the input pin of the IC

Do not force voltage to the input pins when the power does not supply to the IC. Also, do not force voltage to the input pins that exceed the supply voltage or in the guaranteed the absolute maximum rating value even if the power is supplied to the IC.

Ordering Information



• Physical Dimension, Tape and Reel Information





Marking Diagram



Revision History

Date	Revision	Changes
22.FEB.2013	001	New release

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- Оперативные сроки поставки под заказ (от 5 рабочих дней);
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- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

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