

# BMA250

## Digital, triaxial acceleration sensor

### Data sheet

Bosch Sensortec




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#### **BMA250 Data sheet**

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## BMA250

Digital, triaxial  $\pm 2g$  to  $\pm 16g$  acceleration sensor  
with intelligent on-chip motion-triggered interrupt controller

### Key features


- Ultra-Small package LGA package (12 pins), footprint 2mm x 2mm, height 0.95mm
- Digital interface SPI (4-wire, 3-wire), I<sup>2</sup>C, 2 interrupt pins  
V<sub>DDIO</sub> voltage range: 1.2V to 3.6V
- Programmable functionality Acceleration ranges  $\pm 2g/\pm 4g/\pm 8g/\pm 16g$   
Low-pass filter bandwidths 1kHz - <8Hz
- On-chip interrupt controller Motion-triggered interrupt-signal generation for
  - new data
  - any-motion (slope) detection
  - tap sensing (single tap / double tap)
  - orientation recognition
  - flat detection
  - low-g/high-g detection
 Stand-alone capability (no microcontroller needed)
- Ultra-low power ASIC Low current consumption, short wake-up time,  
Advanced features for system power management
- RoHS compliant, halogen-free

### Typical applications

- Display profile switching
- Menu scrolling, tap / double tap sensing
- Gaming
- Pedometer / step counting
- Free-fall detection
- E-compass tilt compensation
- Drop detection for warranty logging
- Advanced system power management for mobile applications

### General description


The BMA250 is a triaxial, low-g acceleration sensor with digital output for consumer market applications. It allows measurements of acceleration in three perpendicular axes. An evaluation circuitry (ASIC) converts the output of a micromechanical acceleration-sensing structure (MEMS) that works according to the differential capacitance principle.

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Package and interfaces of the BMA250 have been defined to match a multitude of hardware requirements. Since the sensor features an ultra-small footprint and a flat package it is ingeniously suited for mobile applications.


The BMA250 offers a variable  $V_{DDIO}$  voltage range from 1.2V to 3.6V and can be programmed to optimize functionality, performance and power consumption in customer specific applications. In addition it features an on-chip interrupt controller enabling motion-based applications without use of a microcontroller.

The BMA250 senses tilt, motion and shock vibration in cell phones, handhelds, computer peripherals, man-machine interfaces, virtual reality features and game controllers.


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## 1. Specification

If not stated otherwise, the given values are over lifetime and full performance temperature and voltage ranges, minimum/maximum values are  $\pm 3 \sigma$ .


**Table 1: Parameter Specification**

OPERATING CONDITIONS						
Parameter	Symbol	Condition	Min	Typ	Max	Units
Acceleration Range	$g_{FS2g}$	Selectable via serial digital interface		$\pm 2$		g
	$g_{FS4g}$			$\pm 4$		g
	$g_{FS8g}$			$\pm 8$		g
	$g_{FS16g}$			$\pm 16$		g
Supply Voltage Internal Domains	$V_{DD}$		1.62	2.4	3.6	V
Supply Voltage I/O Domain	$V_{DDIO}$		1.2	2.4	3.6	V
Voltage Input Low Level	$V_{IL}$	SPI & I <sup>2</sup> C			$0.3V_{DDIO}$	-
Voltage Input High Level	$V_{IH}$	SPI & I <sup>2</sup> C	$0.7V_{DDIO}$			-
Voltage Output Low Level	$V_{OL}$	$V_{DDIO} = 1.62V$ $I_{OL} = 3mA$ , SPI & I <sup>2</sup> C			$0.2V_{DDIO}$	-
		$V_{DDIO} = 1.2V$ $I_{OL} = 3mA$ , SPI & I <sup>2</sup> C			$0.23 V_{DDIO}$	-
Voltage Output High Level	$V_{OH}$	$V_{DDIO} = 1.62V$ $I_{OL} = 2mA$ , SPI & I <sup>2</sup> C	$0.8V_{DDIO}$			-
		$V_{DDIO} = 1.2V$ $I_{OL} = 2mA$ , SPI & I <sup>2</sup> C	$0.62 V_{DDIO}$			-
Supply Current in Normal Mode	$I_{DD}$	Nominal $V_{DD}$ supplies $T_A = 25^\circ C$ , bw = 1kHz		139		$\mu A$
Supply Current in Low-Power Mode	$I_{DDlp}$	Nominal $V_{DD}$ supplies $T_A = 25^\circ C$ , bw = 1kHz sleep duration $\geq 25ms$		7		$\mu A$
Supply Current in Suspend Mode	$I_{DDsm}$	Nominal $V_{DD}$ supplies $T_A = 25^\circ C$		0.5		$\mu A$
Wake-Up Time	$t_{w\_up}$	from Low-Power Mode or Suspend Mode, bw = 1kHz		0.8		ms
Start-Up Time	$t_{s\_up}$	POR, bw = 1kHz		2		ms
Operating Temperature	$T_A$		-40		+85	$^\circ C$

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OUTPUT SIGNAL						
Parameter	Symbol	Condition	Min	Typ	Max	Units
Device Resolution	$D_{res}$	$g_{FS2g}$		3.91		mg
Sensitivity	$S_{2g}$	$g_{FS2g}, T_A=25^{\circ}C$		256		LSB/g
	$S_{4g}$	$g_{FS4g}, T_A=25^{\circ}C$		128		LSB/g
	$S_{8g}$	$g_{FS8g}, T_A=25^{\circ}C$		64		LSB/g
	$S_{16g}$	$g_{FS16g}, T_A=25^{\circ}C$		32		LSB/g
Sensitivity Temperature Drift	TCS	$g_{FS2g}, -40^{\circ}C \leq T_A \leq +85^{\circ}C$ Nominal $V_{DD}$ supplies		$\pm 0.02$		%/K
Zero-g Offset	Off	$g_{FS2g}, T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies		$\pm 80$		mg
Zero-g Offset Temperature Drift	TCO	$g_{FS2g}, -40^{\circ}C \leq T_A \leq +85^{\circ}C$ Nominal $V_{DD}$ supplies		$\pm 1$		mg/K
Bandwidth	$bw_8$	1 <sup>st</sup> order filter, selectable via serial digital interface		8		Hz
	$bw_{16}$			16		Hz
	$bw_{31}$			31		Hz
	$bw_{63}$			63		Hz
	$bw_{125}$			125		Hz
	$bw_{250}$			250		Hz
	$bw_{500}$			500		Hz
	$bw_{1000}$			1000		Hz
Nonlinearity	NL	best fit straight line		$\pm 0.5$		%FS
Output Noise	$n_{rms}$	$g_{FS2g}, T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies Normal mode		0.8		mg/ $\sqrt{Hz}$
Power Supply Rejection Rate	PSRR	$T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies			20	mg/V
Temperature Sensor Measurement Range	$T_S$	$T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies	-40		+87.5	$^{\circ}C$
Temperature Sensor Slope	$dT_S$	$T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies		0.5		LSB/K
Temperature Sensor Offset	$OT_S$	$T_A=25^{\circ}C$ Nominal $V_{DD}$ supplies		$\pm 5$		K
MECHANICAL CHARACTERISTICS						
Parameter	Symbol	Condition	Min	Typ	Max	Units
Cross Axis Sensitivity	S	relative contribution between any two of the three axes		1		%
Alignment Error	$E_A$	relative to package outline		$\pm 0.5$		$^{\circ}$



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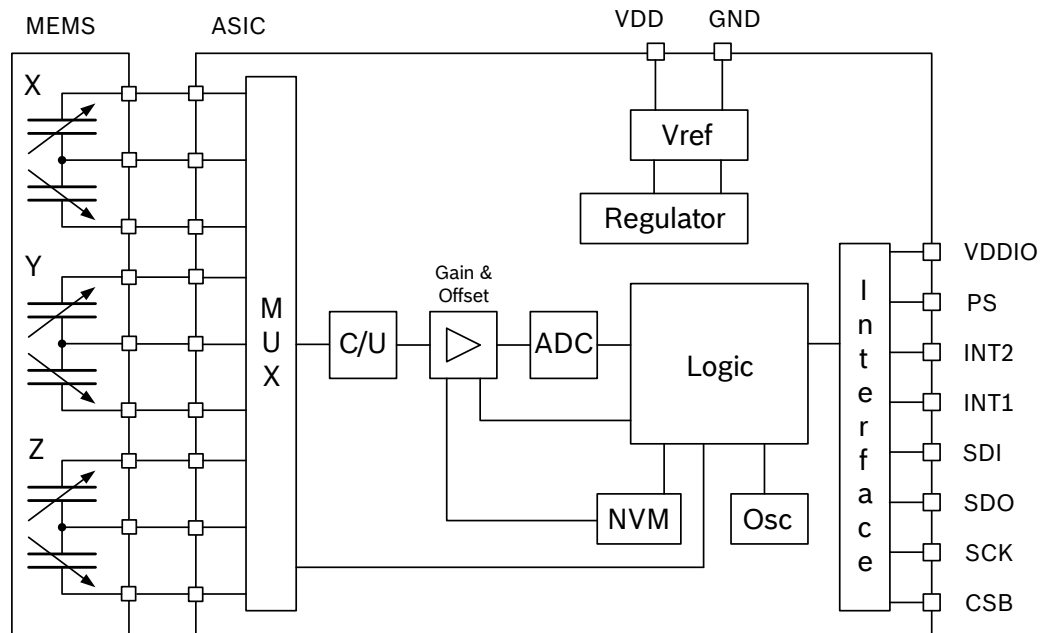
## 2. Absolute maximum ratings

**Table 2: Absolute maximum ratings**

Parameter	Condition	Min	Max	Units
Voltage at Supply Pin	V <sub>DD</sub> Pin	-0.3	4.25	V
	V <sub>DDIO</sub> Pin	-0.3	4.25	V
Voltage at any Logic Pad	Non-Supply Pin	-0.3	V <sub>DDIO</sub> +0.3	V
Passive Storage Temp. Range	≤ 65% rel. H.	-50	+150	°C
Mechanical Shock	Duration ≤ 200μs		10,000	g
	Duration ≤ 1.0ms		2,000	g
	Free fall onto hard surfaces		1.8	m
ESD	HBM, at any Pin		2	kV
	CDM		500	V

### 3. Block diagram

Figure 1 shows the basic building blocks of the BMA250:



**Figure 1: Block diagram of BMA250**

### 4. Functional description

Note: Default values for registers can be found in chapter 5.

#### 4.1 Power management

The BMA250 has two distinct power supply pins:

- $V_{DD}$  is the main power supply for all internal analog and digital functional blocks;
- $V_{DDIO}$  is a separate power supply pin, exclusively used for the supply of the digital interface.

There are no limitations on the voltage levels of both pins relative to each other, as long as each of them lies within its operating range. Furthermore, the device can be completely switched off ( $V_{DD} = 0V$ ) while keeping the  $V_{DDIO}$  supply on ( $V_{DDIO} > 0V$ ). To switch off the interface supply ( $V_{DDIO} = 0V$ ) and keep the internal supply on ( $V_{DD} > 0V$ ) is safe only in normal mode. If the device is in low-power mode or suspend mode while  $V_{DDIO} = 0V$ , there is a risk of excess current consumption on the  $V_{DD}$  supply (non-destructive).

It is absolutely prohibited to keep any interface at a logical high level when  $V_{DDIO}$  is switched off. Such a configuration will permanently damage the device (i.e. if  $V_{DDIO} = 0 \rightarrow [SDI \& SDO \& SCK \& CSB] \neq \text{high}$ ).

The device contains a power-on reset (POR) generator. It resets the logic part and the register values after powering-on  $V_{DD}$  and  $V_{DDIO}$ . There is no limitation on the sequence of switching on

both supply voltages. In case the I<sup>2</sup>C interface shall be used, a direct electrical connection between V<sub>DDIO</sub> supply and the PS pin is needed in order to ensure reliable protocol selection (see section 4.2 Operational modes).

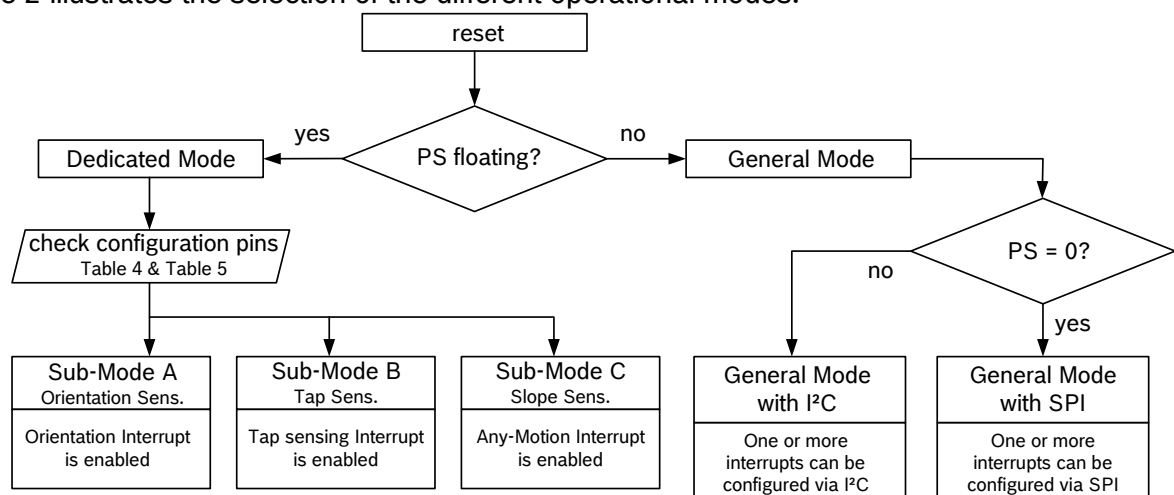
## 4.2 Operational modes

Depending on the configuration the BMA250 is able to operate in two different operational modes:


- **General mode:** The device is acting as a slave on a digital interface (SPI or I<sup>2</sup>C) and is controlled by the external bus master (e.g.  $\mu$ C). The master gets measurement data and status information from the device through the digital interface. In particular, the master can configure the interrupt controller and read out the interrupt status registers. Moreover, it can freely configure and use the two interrupt pins (INT1, INT2). Several interrupts may be enabled in parallel.
- **Dedicated mode:** The dedicated mode allows the sensor to be operated as a stand-alone device in a simple  $\mu$ C-less system without abandon of the interrupt functionality. No digital interface is needed and, as a consequence, no measurement data can be read from the device. Instead of the digital interface the internal interrupt engine with its default setting is used. The interrupt status is mapped onto dedicated output pins. One out of three different sub-modes can be chosen: A) orientation recognition, B) tap sensing or C) slope (any-motion) detection. Only one interrupt at a time can be assigned.

The selection of the operational mode is done during start-up or reset by the state of the PS pin. If PS is floating, the dedicated mode is selected. A defined digital state selects the general mode. All pads are in input mode (no output driver active) during the start-up sequence until the operational mode and, in case of the general mode, the interface type is selected. The start-up sequence is run after power-up and after reset.

Figure 2 illustrates the selection of the different operational modes:



**Figure 2: Operational mode selection**

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### 4.2.1 General mode

A defined digital state at the PS pin selects the general mode. Its polarity determines the kind of interface to be used:

- PS = GND enables the digital SPI interface
- PS =  $V_{DDIO}$  enables the digital I<sup>2</sup>C interface
- PS = float enables the dedicated mode

### 4.2.2 Dedicated mode (μC-less or stand-alone mode)

The dedicated mode operates with pre-defined settings of the interrupt engine in order to generate the motion-triggered interrupt-signals, i.e. bandwidth, sleep time, low-power mode, threshold, and hysteresis are use case optimized. Nevertheless some minor configurations can be selected by the user. The dedicated mode is entered if the device is connected according to table 3. During the start-up / power on sequence the PS pin (#11) must float.

**Table 3: Entering and operating dedicated mode**


<b>VDDIO</b> Pin#3	<b>NC</b> Pin#4	<b>VDD</b> Pin#7	<b>GNDIO</b> Pin#8	<b>GND</b> Pin#9	<b>PS</b> Pin#11
$V_{DDIO}$	NC	$V_{DD}$	GND	GND	float

Depending on the configuration of the other device pins according to table 4 the corresponding sub-mode of the dedicated mode is entered. In table 4 and table 5 the unshaded entries represent necessary input values for the corresponding sub-mode selection while the shaded entries represent corresponding output parameters of the events to be detected.

**Table 4: Sub-mode selection and specific outputs of the dedicated mode**

<b>Sub-Mode</b>	<b>SDO</b> Pin#1	<b>SDx</b> Pin#2	<b>INT1</b> Pin#5	<b>INT2</b> Pin#6	<b>CSB</b> Pin#10	<b>SCx</b> Pin#12
Orientation	output orient1-detect	output orient0-detect	output orient2-detect	output flat-detect	select orient sleep	GND
Tap	output double-detect	output single-detect	GND	select tap type	select tap sleep	$V_{DD}$
Slope	GND	output motion-detect	$V_{DD}$	GND	select slope sleep	$V_{DD}$

Table 5 contains state and description details of the parameters introduced in table 4. Unshaded entries represent input values to be set, shaded entries represent output parameters to be detected.

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**Table 5: Description of the parameters of table 4**

Sub-Mode	Parameter see Table 4	State	Description
Orientation BW = 62.5 Hz	output orient0-detect	low	“upright” for portrait / “left” for landscape
		high	“upside-down” for portrait / “right” for landscape
	output orient1-detect	low	portrait
		high	landscape
	output orient2-detect	low	z-axis upward looking i.e. $ \theta  < 90^\circ$ (Fig. 8)
		high	z-axis downward looking i.e. $ \theta  > 90^\circ$ (Fig. 8)
	output flat-detect	low	non flat i.e. $ \theta  > 19,5^\circ$ (Fig. 8)
		high	flat i.e. $ \theta  < 19,5^\circ$ (Fig. 8)
Tap BW = 1k Hz	select orient sleep	GND	Low-Power mode enabled, sleep time = 100ms
		V <sub>DD</sub>	Low-Power mode enabled, sleep time = 1s
	output double-detect	low	currently no Double-Tap event
		high	Double-Tap event detected
	output single-detect	low	currently no single-tap event
		high	Single-Tap event detected
	select tap type	GND	Single-Tap detection enabled
		V <sub>DD</sub>	Double-Tap detection enabled
Slope BW = 125 Hz	select tap sleep	GND	Low-Power Mode disabled
		V <sub>DD</sub>	Low-Power Mode enabled, sleep time = 10ms
	output motion-detect	low	currently no Any-Motion event
		high	Any-Motion event detected
	select slope sleep	GND	Low-Power mode enabled, sleep time = 50ms
		V <sub>DD</sub>	Low-Power mode enabled, sleep time = 1s

low = GND, high = V<sub>DDIO</sub>

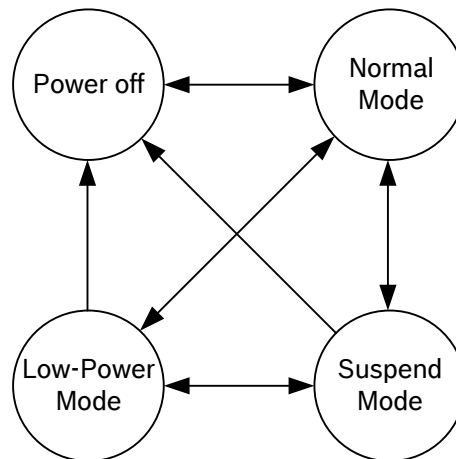
For more details, refer to chapter 4.3 Power modes and 4.8 Interrupt Controller

- Orientation recognition sub mode → refer to chapter 4.8.7
- Tap sensing sub mode → refer to chapter 4.8.6
- Any-motion (slope) detection) sub mode → refer to chapter 4.8.5

### 4.3 Power modes

The BMA250 has three different power modes. Besides normal mode, which represents the fully operational state of the device, there are two special energy saving modes: low-power mode and suspend mode.

The possible transitions between the power modes are illustrated in figure 3:



**Figure 3: Power mode transition diagram**

In normal mode, all parts of the electronic circuit are held powered-up and data acquisition is performed continuously.

In contrast to this, in suspend mode the whole analog part, oscillators included, is powered down. No data acquisition is performed, the only supported operations are reading registers (latest acceleration data are kept) and writing to the (0x11) *suspend* bit or (0x14) *softreset* register. Suspend mode is entered (left) by writing '1' ('0') to the (0x11) *suspend* bit.

In low-power mode, the device is periodically switching between a sleep phase and a wake-up phase. The wake-up phase essentially corresponds to operation in normal mode with complete power-up of the circuitry. During the sleep phase the analog part except the oscillators is powered down. Low-power mode is entered (left) by writing '1' ('0') to the (0x11) *lowpower\_en* bit.

During the wake-up phase the number of samples required by any enabled interrupt is processed. If an interrupt is detected, the device stays in the wake-up phase as long as the interrupt condition endures (non-latched interrupt), or until the latch time expires (temporary interrupt), or until the interrupt is reset (latched interrupt). If no interrupt is detected, the device enters the sleep phase.

The duration of the sleep phase is set by the (0x11) *sleep\_dur* bits as shown in the following table:

**Table 6: Sleep phase duration settings**

<b>(0x11) sleep_dur</b>	<b>Sleep Phase Duration <math>t_{sleep}</math></b>
0000b	0.5ms
0001b	0.5ms
0010b	0.5ms
0011b	0.5ms
0100b	0.5ms
0101b	0.5ms
0110b	1ms
0111b	2ms
1000b	4ms
1001b	6ms
1010b	10ms
1011b	25ms
1100b	50ms
1101b	100ms
1110b	500ms
1111b	1s

The current consumption of the BMA250 can be calculated according to this formula:

$$I_{DDIp} \approx \frac{t_{sleep} \cdot I_{DDsm} + t_{active} \cdot I_{DD}}{t_{sleep} + t_{active}}.$$

When making an estimation about the length of the wake-up phase  $t_{active}$ , the wake-up time,  $t_{w\_up}$ , has to be considered. Therefore,  $t_{active} = t_{ut} + t_{w\_up}$ , where  $t_{ut}$  is given in table 8. During the wake-up phase all analog modules are held powered-up, while during the sleep phase most analog modules are powered down. As a consequence, a wake-up time of less than 1ms (typ. value 0.8ms) is needed to settle the analog modules in order to get reliable acceleration data.

Table 7 gives an overview of the resulting average supply currents  $I_{DDIpe}$  for the different sleep phase durations and a selected bandwidth of 1000Hz, assuming no interrupt is active and thus only one sample per wake-up phase is taken:

**Table 7: Average current consumption in low-power mode**

Sleep phase duration	Average current consumption
0.5ms	100.5 $\mu$ A
1ms	78.8 $\mu$ A
2ms	55.0 $\mu$ A
4ms	34.5 $\mu$ A
6ms	25.2 $\mu$ A
10ms	16.4 $\mu$ A
25ms	7.4 $\mu$ A
50ms	4.0 $\mu$ A
100ms	2.3 $\mu$ A
500ms	0.9 $\mu$ A
1s	0.7 $\mu$ A


## 4.4 Sensor data

### 4.4.1 Acceleration data

The width of acceleration data is 10 bits given in two's complement representation. The 10 bits for each axis are split into an MSB upper part (one byte containing bits 9 to 2) and an LSB lower part (one byte containing bits 1 and 0 of acceleration and a (0x02, 0x04, 0x06) *new\_data* flag). Reading the acceleration data registers shall always start with the LSB part. The content of an MSB register is updated by reading the corresponding LSB register (shadowing procedure). The shadowing procedure can be disabled (enabled) by writing '1' ('0') to the bit *shadow\_dis*. With disabled shadowing, the content of both MSB and LSB registers is updated by a new value immediately. Unused bits of the LSB registers are fixed to 0. The (0x02, 0x04, 0x06) *new\_data* flag of each LSB register is set if the data registers are updated, it is reset if either the corresponding MSB or LSB part is read.

Two different streams of acceleration data are available, unfiltered and filtered. The unfiltered data is sampled with 2kHz. The sampling rate of the filtered data depends on the selected filter bandwidth; it is twice the bandwidth. Which kind of data is stored in the acceleration data registers depends on bit (0x13) *data\_high\_bw*. If (0x13) *data\_high\_bw* is '0' ('1'), then filtered (unfiltered) data is stored in the registers. Both data streams are separately offset-compensated. Both kinds of data can be processed by the interrupt controller.



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The bandwidth of filtered acceleration data is determined by setting the (0x10) *bw* bit as followed:

**Table 8: Bandwidth configuration**

<b>bw</b>	<b>Bandwidth</b>	<b>Update Time <math>t_{ut}</math></b>
00xxx	*)	-
01000	7.81Hz	64ms
01001	15.63Hz	32ms
01010	31.25Hz	16ms
01011	62.5Hz	8ms
01100	125Hz	4ms
01101	250Hz	2ms
01110	500Hz	1ms
01111	1000Hz	0.5ms
1xxxx	*)	-

\*) Note: Settings 00xxx result in a bandwidth of 7.81 Hz; settings 1xxxx result in a bandwidth of 1000 Hz. It is recommended to actively use the range from '01000b' to '01111b' only in order to be compatible with future products.

The BMA250 supports four different acceleration measurement ranges. A measurement range is selected by setting the (0x0F) range bits as follows:


**Table 9: Range selection**

<b>Range</b>	<b>Acceleration measurement range</b>	<b>Resolution</b>
0011	±2g	3.91mg/LSB
0101	±4g	7.81mg/LSB
1000	±8g	15.62mg/LSB
1100	±16g	31.25mg/LSB
others	reserved	-

#### 4.4.2 Temperature data

The width of temperature data is 8 bits given in two's complement representation. Temperature values are available in the (0x08) *temp* register.

The slope of the temperature sensor is 0.5K/LSB, its center temperature is 24°C [(0x08) *temp* = 0x00]. Therefore, the typical temperature measurement range is -40°C up to 87.5°C.

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#### 4.5 Self-test

This feature permits to check the sensor functionality by applying electrostatic forces to the sensor core instead of external accelerations. By actually deflecting the seismic mass, the entire signal path of the sensor can be tested. Activating the self-test results in a static offset of the acceleration data; any external acceleration or gravitational force applied to the sensor during active self-test will be observed in the output as a superposition of both acceleration and self-test signal.

The self-test is activated individually for each axis by writing the proper value to the (0x32) *self\_test\_axis* bits ('01b' for x-axis, '10b' for y-axis, '11b' for z-axis, '00b' to deactivate self-test). It is possible to control the direction of the deflection through bit (0x32) *self\_test\_sign*. The excitation occurs in positive (negative) direction if (0x32) *self\_test\_sign* = '0b' ('1b').

In order to ensure a proper interpretation of the self-test signal it is recommended to perform the self-test for both (positive and negative) directions and then to calculate the difference of the resulting acceleration values. Table 10 shows the minimum differences for each axis. The actually measured signal differences can be significantly larger.

**Table 10: Self-test difference values**

	x-axis signal	y-axis signal	z-axis signal
resulting minimum difference signal	+0.8 g	+0.8 g	+0.4 g

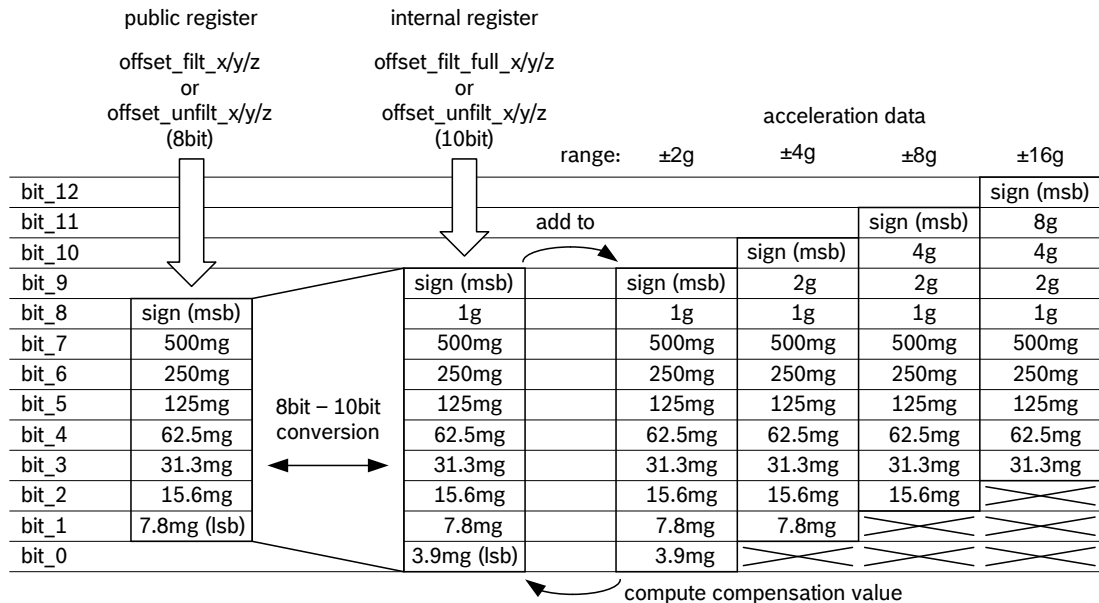
It is recommended to perform a reset of the device after self-test. If the reset cannot be performed, the following sequence must be kept to prevent unwanted interrupt generation: disable interrupts, change parameters of interrupts, wait for at least 600  $\mu$ s, enable desired interrupts.

#### 4.6 Offset compensation

Offsets in measured signals can have several causes but they are always unwanted and disturbing in many cases. Therefore, the BMA250 offers an advanced set of four digital offset compensation methods which are closely matched to each other. These are slow, fast, and manual compensation, and inline calibration.

The compensation is performed for unfiltered and filtered data independently. It is done by adding a compensation value to the acceleration data coming from the ADC. The result of this computation is saturated if necessary to prevent any overflow errors (the smallest or biggest possible value is set, depending on the sign). However, the public registers used to read and write compensation values have only a width of 8 bits.

An overview of the offset compensation principle is given in figure 4:



**Figure 4: Principle of offset compensation**

The meaning of both public and internal registers is the same for all acceleration measurement ranges. Therefore, with measurement ranges other than  $\pm 2g$ , one or more lower significant bits of the internal registers are lost when added to an acceleration value, or are set to zero when the internal compensation value is computed. If a compensation value is too small or too big to fit into the corresponding internal register, it is saturated to prevent an overflow error.

In a similar way the conversion of the internal register value to the public register value (10bit to 8bit) uses saturation.

Summarized, in dependence to the measurement range which has been set, the compensation value, which has been written into the public register will correct the data output according to figure 4.

e.g.  $\pm 2g$  range:

public register = 00000001b → add to acceleration data =  $\pm 7.8mg$  = +2LSB  
public register = 00000010b → add to acceleration data =  $\pm 15.6mg$  = +4LSB  
public register = 00000101b → add to acceleration data =  $\pm 39.1mg$  = +10LSB

The public registers are image registers of EEPROM registers. With each image update (see section 4.7 Non-volatile memory for details) the contents of the non-volatile EEPROM registers are written to the public registers. At any time the public register can be over-written by the user. After changing the contents of the public registers by either an image update or manually, all 8bit values are widened to 10bit values and stored in the corresponding internal registers. In the opposite direction, if the value of an internal register changes due to the computation performed by a compensation algorithm, it is converted to an 8bit value and stored in the public register.

For slow and fast offset compensation, the compensation target can be chosen by setting the bits (0x37) *offset\_target\_x*, (0x37) *offset\_target\_y*, and (0x37) *offset\_target\_z* according to table 11:

**Table 11: Offset target settings**

(0x37) <i>offset_target_x/y/z</i>	Target value
00b	0g
01b	+1g
10b	-1g
11b	0g

By writing '1' to the (0x36) *offset\_reset* bit, all offset compensation registers are reset to zero.

#### 4.6.1 Slow compensation

Slow compensation is a quasi-continuous process which regulates the acceleration value of each axis towards the target value by comparing the current value with the target and adding or subtracting a fixed value depending on the comparison.

The algorithm in detail: If an acceleration value is larger (smaller) than the target value (0x37) *offset\_target\_x/y/z* for a number of samples (given by the parameter Offset Period see table 12), the internal offset compensation value (0x38, 0x039, 0x3A) *offset\_filt\_x/y/z* or (0x3B, 0x03C, 0x3D) *offset\_unfilt\_x/y/z* is decremented (incremented) by 4 LSB.

The public registers (0x38, 0x039, 0x3A) *offset\_filt\_x/y/z* and (0x3B, 0x03C, 0x3D) *offset\_unfilt\_x/y/z* are not used for the computations but they are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

The compensation period *offset\_period* is set by the (0x37) *cut\_off* bit as represented in table 12:

**Table 12: Compensation period settings**

(0x37) <i>cut_off</i>	Offset Period
0b	8
1b	16

The slow compensation can be enabled (disabled) for each axis independently by setting the bits (0x36) *hp\_x\_en*, *hp\_y\_en*, *hp\_z\_en* to '1' ('0'), respectively.

Slow compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of slow compensation are not fulfilled.

#### 4.6.2 Fast compensation

Fast compensation is a one-shot process by which the compensation value is set in such a way that when added to the raw acceleration, the resulting acceleration value of each axis equals the target value.

The algorithm in detail: An average of 16 consecutive acceleration values is computed and the difference between target value and computed value is written to (0x38, 0x39, 0x3A) *offset\_filt\_x/y/z* or (0x3B, 0x3C, 0x3D) *offset\_unfilt\_x/y/z*. The public registers (0x38, 0x39, 0x3A) *offset\_filt\_x/y/z* and (0x3B, 0x3C, 0x3D) *offset\_unfilt\_x/y/z* are updated with the contents of the internal registers (using saturation if necessary) and can be read by the user.

Fast compensation is triggered for each axis individually by setting the (0x36) *cal\_trigger* bits as shown in table 13:

**Table 13: Fast compensation axis selection**

(0x36) <i>cal_trigger</i>	Selected Axis
00b	none
01b	x
10b	y
11b	z

The register (0x36) *cal\_trigger* keeps its non-zero value while the fast compensation procedure is running. Slow compensation is blocked as long as fast compensation endures. Bit (0x36) *cal\_rdy* is '0' when (0x36) *cal\_trigger* is not '00'.

Fast compensation should not be used in combination with low-power mode. In low-power mode the conditions (availability of necessary data) for proper function of fast compensation are not fulfilled.


#### 4.6.3 Manual compensation

As explained above, the contents of the public compensation registers (0x38, 0x39, 0x3A) *offset\_filt\_x/y/z* and (0x3B, 0x3C, 0x3D) *offset\_unfilt\_x/y/z* can be set manually via the digital interface. It is recommended to write into these registers immediately after a new data interrupt in order not to disturb running offset computations.

Writing to the offset compensation registers is not allowed if slow compensation is enabled or if the fast compensation procedure is running.

#### 4.6.4 Inline calibration

For a given application, it is often desirable to calibrate the offset once and to store the compensation values permanently. This can be achieved by using one of the aforementioned offset compensation methods to determine the proper compensation values and then storing

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these values permanently in the non-volatile memory (EEPROM). See section 4.7 Non-volatile memory for details of the storing procedure.

Each time the device is reset, the compensation values are loaded from the non-volatile memory into the image registers and used for offset compensation until they are possibly overwritten using one of the other compensation methods.

#### 4.7 Non-volatile memory

The entire memory of the BMA250 consists of three different kinds of registers: hard-wired, volatile, and non-volatile. Non-volatile memory is implemented as EEPROM. Part of it can be both read and written by the user. Access to non-volatile memory is only possible through (volatile) image registers.

Altogether, there are eight registers (bytes) of EEPROM which are accessible by the customer. The addresses of the image registers range from 0x38 to 0x3F. While the addresses up to 0x3D are used for offset compensation (see 4.6 Offset Compensation), addresses 0x3E and 0x3F are general purpose registers not linked to any sensor-specific functionality.

The content of the EEPROM is loaded to the image registers after a reset (either POR or softreset) or after a user request which is performed by writing '1' to bit (0x33) *nvm\_load*. As long as the image update is not yet complete, bit (0x33) *nvm\_load* is '1', otherwise it is '0'.

The image registers can be read and written like any other register.

Writing to the EEPROM is a three-step procedure:


1. Write the new contents to the image registers.
2. Write '1' to bit (0x33) *nvm\_prog\_mode* in order to unlock the EEPROM.
3. Write '1' to bit (0x33) *nvm\_prog\_trig* and keep '1' in bit (0x33) *nvm\_prog\_mode* in order to trigger the write process.

Writing to the EEPROM always renews the entire EEPROM contents. It is possible to check the write status by reading bit (0x33) *nvm\_rdy*. While (0x33) *nvm\_rdy* = '0', the write process is still enduring; if (0x33) *nvm\_rdy* = '1', then writing is completed. As long as the write process is ongoing, no power mode change and no image update is allowed. It is forbidden to write to the EEPROM while the image update is running, in low-power mode, and in suspend mode.

#### 4.8 Interrupt controller

Seven interrupt engines are integrated in the BMA250. Each interrupt can be independently enabled and configured. If the condition of an enabled interrupt is fulfilled, the corresponding status bit is set to '1' and the selected interrupt pin is activated. There are two interrupt pins, INT1 and INT2; interrupts can be freely mapped to any of these pins. The pin state is a logic 'or' combination of all mapped interrupts.

The interrupt status registers are updated together with writing new data into the acceleration data registers. If an interrupt is disabled, all active status bits and pins are immediately reset.

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All time constants are based upon the typical frequency of the internal oscillator. This is reflected by the bandwidths (bw) as specified in table 1.

#### 4.8.1 General features

An interrupt is cleared depending on the selected interrupt mode, which is common to all interrupts. There are three different interrupt modes: non-latched, latched, and temporary. The mode is selected by the (0x21) *latch\_int* bits according to table 14

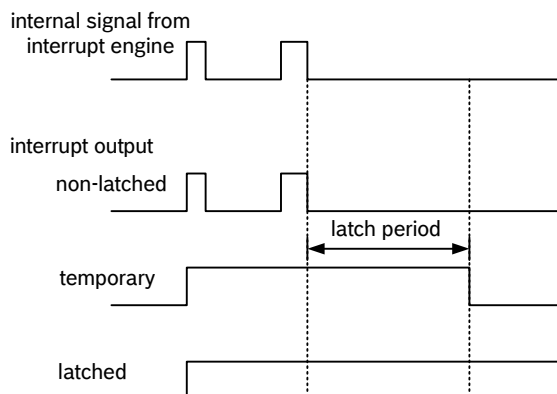
**Table 14: Interrupt mode selection**

(0x21) <i>latch_int</i>	Interrupt mode
0000b	non-latched
0001b	temporary, 250ms
0010b	temporary, 500ms
0011b	temporary, 1s
0100b	temporary, 2s
0101b	temporary, 4s
0110b	temporary, 8s
0111b	latched
1000b	non-latched
1001b	temporary, 500µs
1010b	temporary, 500µs
1011b	temporary, 1ms
1100b	temporary, 12.5ms
1101b	temporary, 25ms
1110b	temporary, 50ms
1111b	latched

An interrupt is generated if its activation condition is met. It can not be cleared as long as the activation condition is fulfilled. In the non-latched mode the interrupt status bit and the selected pin (the contribution to the 'or' condition for INT1 and/or INT2) are cleared as soon as the activation condition is no more valid. Exceptions to this behaviour are the new data, orientation, and flat interrupts, which are automatically reset after a fixed time.

In the latched mode an asserted interrupt status and the selected pin are cleared by writing '1' to bit (0x21) *reset\_int*. If the activation condition still holds when it is cleared, the interrupt status is asserted again with the next change of the acceleration registers.

In the temporary mode an asserted interrupt and selected pin are cleared after a defined period of time. The behaviour of the different interrupt modes is shown graphically in figure 5:



**Figure 5: Interrupt modes**

Several interrupt engines can use either unfiltered or filtered acceleration data as their input. For these interrupts, the source can be selected with the respective  $(0x1E)$   $int\_src\_...$  bits, in details these are  $(0x1E)$   $int\_src\_data$ ,  $(0x1E)$   $int\_src\_tap$ ,  $(0x1E)$   $int\_src\_slope$ ,  $(0x1E)$   $int\_src\_high$ , and  $(0x1E)$   $int\_src\_low$ . Setting the respective bits to '0' ('1') selects filtered (unfiltered) data as input. For the other interrupts, orientation recognition and flat detection, such a selection is not possible. They always use filtered input data.

It is strongly recommended to set interrupt parameters prior to enabling the interrupt. Changing parameters of an already enabled interrupt may cause unwanted interrupt generation and generation of a false interrupt history. A safe way to change parameters of an enabled interrupt is to keep the following sequence: disable the desired interrupt, change parameters, wait for at least 600  $\mu s$ , enable the desired interrupt.

#### 4.8.2 Mapping (inttype to INT Pin#)

The mapping of interrupts to the interrupt pins #05 or #06 is done by registers  $(0x19)$  to  $(0x1B)$ . Setting  $(0x19)$   $int1\_inttyp$  to '1' ('0') maps (unmaps) "inttyp" to pin #5 (INT1), correspondingly setting  $(0x1B)$   $int2\_inttyp$  to '1' ('0') maps (unmaps) "inttyp" to pin #6 (INT2).

Note: "inttyp" to be replaced with the precise notation, given in the memory map in chapter 5.


Example: For flat interrupt ( $int1\_flat$ ): Setting  $(0x19)$   $int1\_flat$  to '1' maps  $int1\_flat$  to pin #5 (INT1).

#### 4.8.3 Electrical behaviour (INT pin# to open-drive or push-pull)

Both interrupt pins can be configured to show desired electrical behaviour. The 'active' level of each pin is determined by the  $(0x20)$   $int1\_lvl$  and  $(0x20)$   $int2\_lvl$  bits.

If  $(0x20)$   $int1\_lvl$  = '1' ('0') /  $(0x20)$   $int2\_lvl$  = '1' ('0'), then pin #05 (INT1) / pin #06 (INT2) is active '1' ('0'). In addition to that, also the electric type of the interrupt pins can be selected. By



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setting bits  $(0x20) \text{ int1\_od}$  /  $(0x20) \text{ int2\_od}$  to '0', the interrupt pin output type gets push-pull, by setting the configuration bits to '1', the output type gets open-drive.

Remark: Due to their use for sub-mode selection in dedicated mode, the states of both INT pins are not defined during the first 2 ms after power-up.

#### 4.8.4 New data interrupt

This interrupt serves for synchronous reading of acceleration data. It is generated after storing a new value of z-axis acceleration data in the data register. The interrupt is cleared automatically when the next cycle of data acquisition starts. The interrupt status is '0' for at least 50µs.

The interrupt mode of the new data interrupt is fixed to non-latched.

It is enabled (disabled) by writing '1' ('0') to bit  $(0x17) \text{ data\_en}$ . The interrupt status is stored in bit  $(0x0A) \text{ data\_int}$ .

#### 4.8.5 Any-motion (slope) detection

Any-motion detection uses the slope between successive acceleration signals to detect changes in motion. An interrupt is generated when the slope (absolute value of acceleration difference) exceeds a preset threshold. It is cleared as soon as the slope falls below the threshold. The principle is made clear in figure 6.

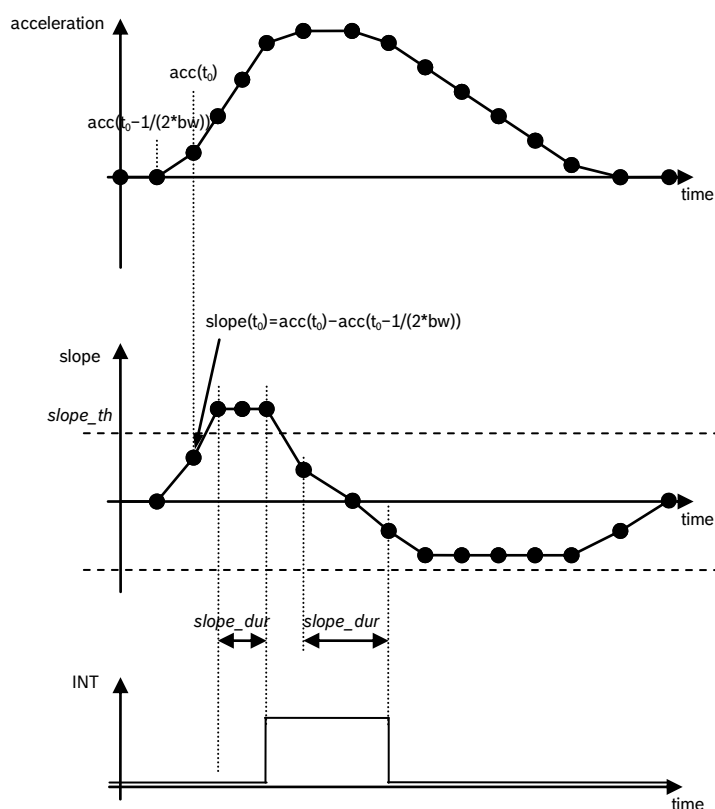



Figure 6: Principle of any-motion detection

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The threshold is set with the value of register (0x28) *slope\_th*. 1 LSB of (0x28) *slope\_th* corresponds to 1 LSB of acceleration data. Therefore, an increment of (0x28) *slope\_th* is 3.91 mg in 2g-range (7.81 mg in 4g-range, 15.6 mg in 8g-range and 31.3 mg in 16g-range). And the maximum value is 996 mg in 2g-range (1.99g in 4g-range, 3.98g in 8g-range and 7.97g in 16g-range).

The time difference between the successive acceleration signals depends on the selected bandwidth and equates to  $1/(2 \cdot \text{bandwidth})$  ( $\Delta t = 1/(2 \cdot \text{bw})$ ). In order to suppress failure signals, the interrupt is only generated (cleared) if a certain number *N* of consecutive slope data points is larger (smaller) than the slope threshold given by (0x28) *slope\_th*. This number is set by the (0x27) *slope\_dur* bits. It is  $N = (0x27) \text{ slope\_dur} + 1$  for (0x27).

Example: (0x27) *slope\_dur* = 00b, ..., 11b = 1decimal, ..., 4decimal

#### 4.8.5.1 Enabling (disabling) for each axis

Any-motion detection can be enabled (disabled) for each axis separately by writing '1' ('0') to bits (0x16) *slope\_en\_x*, (0x16) *slope\_en\_y*, (0x16) *slope\_en\_z*. The criteria for any-motion detection are fulfilled and the slope interrupt is generated if the slope of any of the enabled axes exceeds the threshold (0x28) *slope\_th* for [(0x27) *slope\_dur* + 1] consecutive times. As soon as the slopes of all enabled axes fall or stay below this threshold for [(0x27) *slope\_dur* + 1] consecutive times the interrupt is cleared unless interrupt signal is latched.

#### 4.8.5.2 Axis and sign information of any motion interrupt

The interrupt status is stored in bit (0x09) *slope\_int*. The any-motion interrupt supplies additional information about the detected slope. The axis which triggered the interrupt is given by that one of bits (0x0B) *slope\_first\_x*, (0x0B) *slope\_first\_y*, (0x0B) *slope\_first\_z* that contains a '1'. The sign of the triggering slope is held in bit (0x0B) *slope\_sign*. If (0x0B) *slope\_sign* = '0' ('1'), the sign is positive (negative).

#### 4.8.5.3 Serial interface and dedicated wake-up mode


When serial interface is active, any-motion detection logic is enabled if any of the axis specific (0x16) *slope\_en\_...* register bits are set. To disable the any-motion interrupt, clear all the axis specific (0x16) *slope\_en\_...* bits.

In the dedicated wake-up mode (see chapter 4.2.2), all three axes are enabled for any-motion detection whether the individual axis enable bits are set or not.

#### 4.8.6 Tap sensing

Tap sensing has a functional similarity with a common laptop touch-pad or clicking keys of a computer mouse. A tap event is detected if a pre-defined slope of the acceleration of at least one axis is exceeded. Two different tap events are distinguished: A 'single tap' is a single event within a certain time, followed by a certain quiet time. A 'double tap' consists of a first such event followed by a second event within a defined time frame.

Only one of the tap interrupts can be enabled at the same time. Single tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) *s\_tap\_en*. Double tap interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) *d\_tap\_en*. If one tries to enable both interrupts by writing '1' to

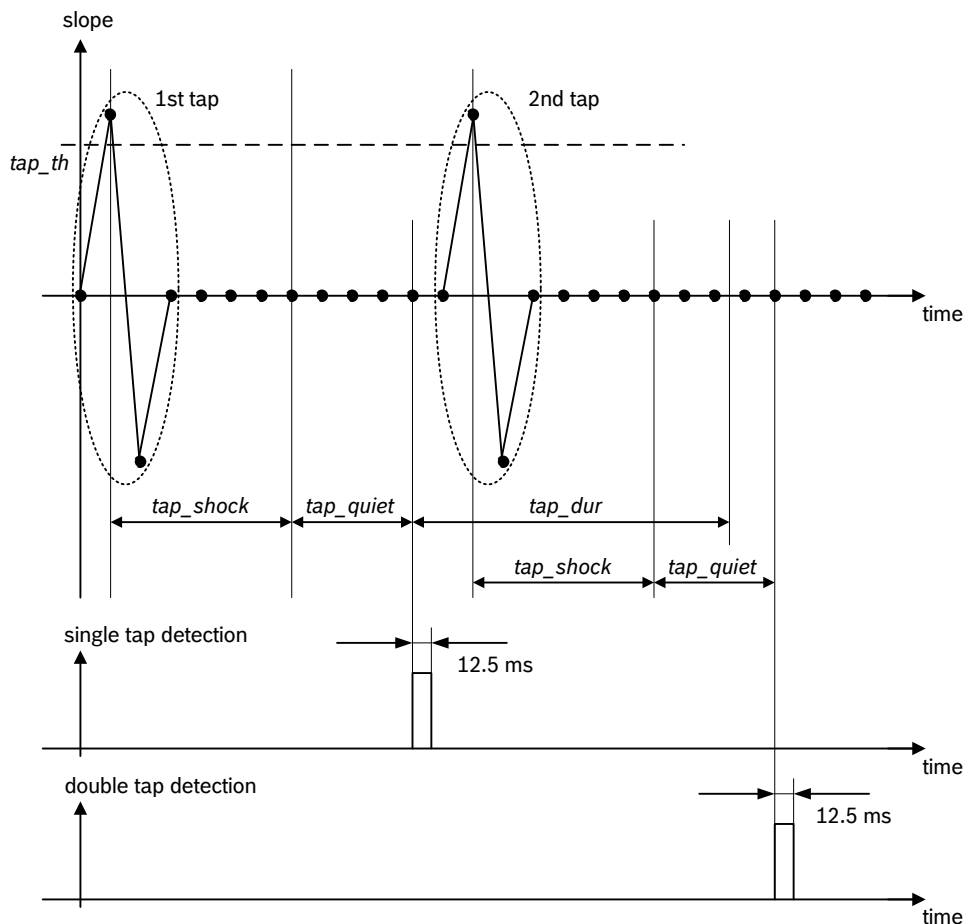
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(0x16) *s\_tap\_en* and (0x16) *d\_tap\_en*, then only (0x16) *d\_tap\_en* keeps the value '1' and the double tap interrupt is enabled.

The status of the single tap interrupt is stored in bit (0x09) *s\_tap\_int*, the status of the double tap interrupt is stored in bit (0x09) *d\_tap\_int*.

The slope threshold for detecting a tap event is set by bits (0x2B) *tap\_th*. The meaning of (0x2B) *tap\_th* depends on the range setting. 1 LSB of (0x2B) *tap\_th* corresponds to a slope of 62.5mg in 2g-range, 125mg in 4g-range, 250mg in 8g-range, and 500mg in 16g-range.

In figure 7 the meaning of the different timing parameters is visualized:



**Figure 7: Timing of tap detection**

The parameters (0x2A) *tap\_shock* and (0x2A) *tap\_quiet* apply to both single tap and double tap detection, while (0x2A) *tap\_dur* applies to double tap detection only. Within the duration of (0x2A) *tap\_shock* any slope exceeding (0x2B) *tap\_th* after the first event is ignored. Contrary to this, within the duration of (0x2A) *tap\_quiet* no slope exceeding (0x2B) *tap\_th* must occur, otherwise the first event will be cancelled.

#### 4.8.6.1 Single tap detection

A single tap is detected and the single tap interrupt is generated after the combined durations of (0x2A) *tap\_shock* and (0x2A) *tap\_quiet*, if the corresponding slope conditions are fulfilled. The interrupt is cleared after a delay of 12.5 ms.

#### 4.8.6.2 Double tap detection

A double tap is detected and the double tap interrupt is generated if an event fulfilling the conditions for a single tap occurs within the set duration in (0x2A) *tap\_dur* after the completion of the first tap event. The interrupt is cleared after a delay of 12.5 ms.

#### 4.8.6.3 Selecting the timing of tap detection

For each of parameters (0x2A) *tap\_shock* and (0x2A) *tap\_quiet* two values are selectable. By writing '0' ('1') to bit (0x2A) *tap\_shock* the duration of (0x2A) *tap\_shock* is set to 50 ms (75 ms). By writing '0' ('1') to bit (0x2A) *tap\_quiet* the duration of (0x2A) *tap\_quiet* is set to 30 ms (20 ms).

The length of (0x2A) *tap\_dur* can be selected by setting the (0x2A) *tap\_dur* bits according to table 15:

**Table 15: Selection of *tap\_dur***

(0x2A) <i>tap_dur</i>	length of <i>tap_dur</i>
000b	50 ms
001b	100 ms
010b	150 ms
011b	200 ms
100b	250 ms
101b	375 ms
110b	500 ms
111b	700 ms

#### 4.8.6.4 Axis and sign information of tap sensing

The sign of the slope of the first tap which triggered the interrupt is stored in bit (0x0B) *tap\_sign* ('0' means positive sign, '1' means negative sign). The value of this bit persists after clearing the interrupt.

The axis which triggered the interrupt is indicated by bits (0x0B) *tap\_first\_x*, (0x0B) *tap\_first\_y*, and (0x0B) *tap\_first\_z*.

The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status.

#### 4.8.6.5 Tap sensing in low power mode

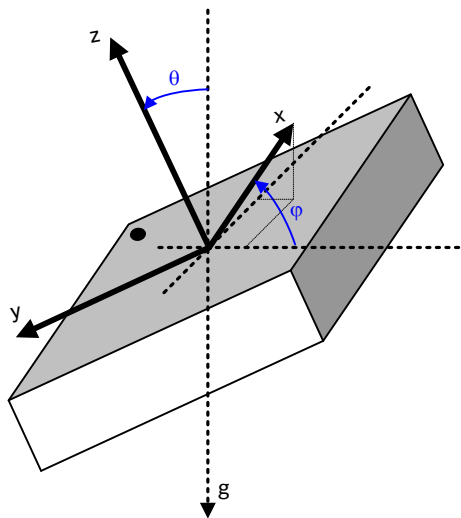
In low-power mode, a limited number of samples is processed after wake-up to decide whether an interrupt condition is fulfilled. The number of samples is selected by bits (0x2B) *tap\_samp* according to table 16.

**Table 16: Meaning of (0x2B) *tap\_samp***

(0x2B) <i>tap_samp</i>	Number of Samples
00b	2
01b	4
10b	8
11b	16

#### 4.8.7 Orientation recognition


The orientation recognition feature informs on an orientation change of the sensor with respect to the gravitational field vector 'g'. The measured acceleration vector components with respect to the gravitational field are defined as shown in figure 8.



**Figure 8: Definition of vector components**

Therefore, the magnitudes of the acceleration vectors are calculated as follows:

$$\begin{aligned}
 \text{acc\_x} &= 1g \cdot \sin\theta \cdot \cos\varphi \\
 \text{acc\_y} &= -1g \cdot \sin\theta \cdot \sin\varphi \\
 \text{acc\_z} &= 1g \cdot \cos\theta \\
 \rightarrow \text{acc\_y}/\text{acc\_x} &= -\tan\varphi
 \end{aligned}$$

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Depending on the magnitudes of the acceleration vectors the orientation of the device in the space is determined and stored in the three (0x0C) *orient* bits. These bits may not be reset in the sleep phase of low-power mode. There are three orientation calculation modes with different thresholds for switching between different orientations: symmetrical, high-asymmetrical, and low-asymmetrical. The mode is selected by setting the (0x2C) *orient\_mode* bits as given in table 17.

**Table 17: Orientation mode settings**

(0x2C) <i>orient_mode</i>	Orientation Mode
00b	symmetrical
01b	high-asymmetrical
10b	low-asymmetrical
11b	symmetrical

For each orientation mode the (0x0C) *orient* bits have a different meaning as shown in table 18 to table 20:

**Table 18: Meaning of the (0x0C) *orient* bits in symmetrical mode**

(0x0C) <i>orient</i>	Name	Angle	Condition
x00	portrait upright	$315^\circ < \varphi < 45^\circ$	$ acc\_y  <  acc\_x  - 'hyst'$ and $acc\_x - 'hyst' \geq 0$
x01	portrait upside down	$135^\circ < \varphi < 225^\circ$	$ acc\_y  <  acc\_x  - 'hyst'$ and $acc\_x + 'hyst' < 0$
x10	landscape left	$45^\circ < \varphi < 135^\circ$	$ acc\_y  \geq  acc\_x  + 'hyst'$ and $acc\_y < 0$
x11	landscape right	$225^\circ < \varphi < 315^\circ$	$ acc\_y  \geq  acc\_x  + 'hyst'$ and $acc\_y \geq 0$

**Table 19: Meaning of the (0x0C) *orient* bits in high-asymmetrical mode**

(0x0C) <i>orient</i>	Name	Angle	Condition
x00	portrait upright	$297^\circ < \varphi < 63^\circ$	$ acc\_y  < 2 \cdot  acc\_x  - 'hyst'$ and $acc\_x - 'hyst' \geq 0$
x01	portrait upside down	$117^\circ < \varphi < 243^\circ$	$ acc\_y  < 2 \cdot  acc\_x  - 'hyst'$ and $acc\_x + 'hyst' < 0$
x10	landscape left	$63^\circ < \varphi < 117^\circ$	$ acc\_y  \geq 2 \cdot  acc\_x  + 'hyst'$ and $acc\_y < 0$
x11	landscape right	$243^\circ < \varphi < 297^\circ$	$ acc\_y  \geq 2 \cdot  acc\_x  + 'hyst'$ and $acc\_y \geq 0$

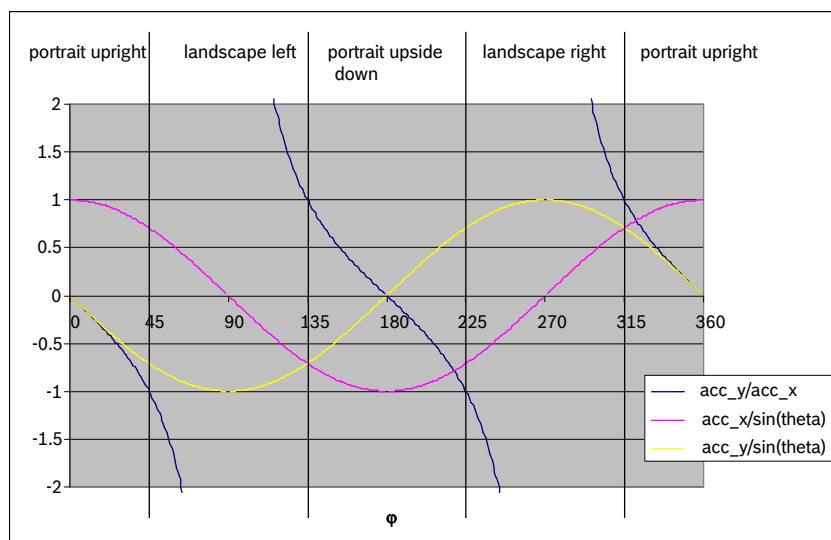
**Table 20: Meaning of the (0x0C) orient bits in low-asymmetrical mode**

(0x0C) orient	Name	Angle	Condition
x00	portrait upright	$333^{\circ} < \varphi < 27^{\circ}$	$ acc\_y  < 0.5 \cdot  acc\_x  - 'hyst'$ and $acc\_x - 'hyst' \geq 0$
x01	portrait upside down	$153^{\circ} < \varphi < 207^{\circ}$	$ acc\_y  < 0.5 \cdot  acc\_x  - 'hyst'$ and $acc\_x + 'hyst' < 0$
x10	landscape left	$27^{\circ} < \varphi < 153^{\circ}$	$ acc\_y  \geq 0.5 \cdot  acc\_x  + 'hyst'$ and $acc\_y < 0$
x11	landscape right	$207^{\circ} < \varphi < 333^{\circ}$	$ acc\_y  \geq 0.5 \cdot  acc\_x  + 'hyst'$ and $acc\_y \geq 0$

In the preceding tables, the parameter 'hyst' stands for a hysteresis, which can be selected by setting the (0x0C) orient\_hyst bits. 1 LSB of (0x0C) orient\_hyst always corresponds to 62.5 mg, in 2g-range, 125 mg in 4g-range, 250 mg in 8g-range and 500 mg in 16g-range.. It is important to note that by using a hysteresis  $\neq 0$  the actual switching angles become different from the angles given in the tables since there is an overlap between the different orientations.

The most significant bit of the (0x0C) orient bits (which is displayed as an 'x' in the above given tables) contains information about the direction of the z-axis. It is set to '0' ('1') if  $acc\_z \geq 0$  ( $acc\_z < 0$ ).

Figure 9 shows the typical switching conditions between the four different orientations for the symmetrical mode (i.e. without hysteresis):



**Figure 9: Typical orientation switching conditions w/o hysteresis**

The orientation interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) orient\_en. The interrupt is generated if the value of (0x0C) orient has changed. It is automatically cleared after

one stable period of the (0x0C) *orient* value. The interrupt status is stored in the (0x09) *orient\_int* bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed (0x0C) *orient* value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the (0x0C) *orient* is only updated with the next following value change (i.e. with the next occurring interrupt). In order to ensure the continuous availability of up-to-date orientation data it is therefore optimal to use the non-latched interrupt. It is strongly advised against using latched interrupt mode or temporary interrupt mode with latch times above 50 ms for orient recognition.

#### 4.8.7.1 Orientation blocking

The change of the (0x0C) *orient* value and – as a consequence – the generation of the interrupt can be blocked according to conditions selected by setting the value of the (0x2C) *orient\_blocking* bits as described by table 21.

**Table 21: Blocking conditions for orientation recognition**

(0x2C) <i>orient_blocking</i>	Conditions
00b	no blocking
01b	theta blocking
10b	theta blocking or acceleration slope in any axis > 0.2 g
11b	value of orient is not stable for at least 100 ms or theta blocking or acceleration slope in any axis > 0.4 g

The theta blocking is defined by the following inequality:


$$|\tan \theta| < \frac{\sqrt{\text{blocking\_theta}}}{8}.$$

The parameter *blocking\_theta* of the above given equation stands for the contents of the (0x2D) *orient\_theta* bits. Hereby it is possible to define a blocking angle between 0° and 44.8°. The internal blocking algorithm saturates the acceleration values before further processing. As a consequence, the blocking angles are strictly valid only for a device at rest; they can be different if the device is moved.

Example:

To get a maximum blocking angle of 19° the parameter *blocking\_theta* is determined in the following way:  $(8 * \tan(19^\circ))^2 = 7.588$ , therefore, *blocking\_value* = 8dec = 001000b has to be chosen.



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In order to avoid unwanted generation of the orientation interrupt in a nearly flat position ( $z \sim 0$ , sign change due to small movements or noise), a hysteresis of 0.2 g is implemented for the z-axis, i. e. a after a sign change the interrupt is only generated after  $|z| > 0.2$  g.

#### 4.8.8 Flat detection

The flat detection feature gives information about the orientation of the devices' z-axis relative to the g-vector, i. e. it recognizes whether the device is in a flat position or not.

The condition for the device to be in the flat position is

$$|\tan \theta| < \frac{\sqrt{\text{parameter\_theta}}}{8}.$$

Like *blocking\_theta*, used with orientation recognition, the *parameter\_theta* stands for a user-defined setting. In this case the content of the (0x2E) *flat\_theta* bits. The possible flat angles also range from 0° to 44.8°. To ensure proper operation, *parameter\_theta* has to be less than or equal to *blocking\_theta*.


The flat interrupt is enabled (disabled) by writing '1' ('0') to bit (0x16) *flat\_en*. The flat interrupt is generated if the flat value has changed and the new value is stable for at least the time given by the (0x2F) *flat\_hold\_time* bits. The flat value is stored in the (0x0C) *flat* bit if the interrupt is enabled. This value is '1' if the device is in the flat position, it is '0' otherwise. The content of the (0x0C) *flat* bit is changed only if the interrupt is generated. The interrupt is automatically cleared after one sample period. Its status is stored in the (0x09) *flat\_int* bit.

If temporary or latched interrupt mode is used, after the generation of the interrupt the changed (0x0C) *flat* value is kept fixed as long as the interrupt persists (e. g. until the latch time expires or the interrupt is reset). After clearing the interrupt, the (0x0C) *flat* value is only updated with the next following value change (i.e. with the next occurring interrupt).

The meaning of the (0x2F) *flat\_hold\_time* bits can be seen from table 22.

**Table 22: Meaning of flat\_hold\_time**

(0x2F) <i>flat_hold_time</i>	Time
00b	0
01b	512 ms
10b	1024 ms
11b	2048 ms

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#### 4.8.9 Low-g interrupt

This interrupt is based on the comparison of acceleration data against a low-g threshold, which is most useful for free-fall detection.

The interrupt is enabled (disabled) by writing '1' ('0') to the (0x17) *low\_en* bit. There are two modes available, 'single' mode and 'sum' mode. In 'single' mode, the acceleration of each axis is compared with the threshold; in 'sum' mode, the sum of absolute values of all accelerations  $|acc_x| + |acc_y| + |acc_z|$  is compared with the threshold. The mode is selected by the contents of the (0x24) *low\_mode* bit: '0' means 'single' mode, '1' means 'sum' mode.

The low-g threshold is set through the (0x23) *low\_th* register. 1 LSB of (0x23) *low\_th* always corresponds to an acceleration of 7.81 mg (i.e. increment is independent from g-range setting).

A hysteresis can be selected by setting the (0x24) *low\_hy* bits. 1 LSB of (0x24) *low\_hy* always corresponds to an acceleration difference of 125 mg in any g-range (as well, increment is independent from g-range setting).

The low-g interrupt is generated if the absolute values of the acceleration of all axes ('and' relation, in case of single mode) or their sum (in case of sum mode) are lower than the threshold for at least the time defined by the (0x22) *low\_dur* register. The interrupt is reset if the absolute value of the acceleration of at least one axis ('or' relation, in case of single mode) or the sum of absolute values (in case of sum mode) is higher than the threshold plus the hysteresis for at least one data acquisition. In bit (0x09) *low\_int* the interrupt status is stored.

The relation between the content of (0x22) *low\_dur* and the actual delay of the interrupt generation is:  $\text{delay [ms]} = [(0x22) \text{ low\_dur} + 1] \cdot 2 \text{ ms}$ . Therefore, possible delay times range from 2 ms to 512 ms.


#### 4.8.10 High-g interrupt

This interrupt is based on the comparison of acceleration data against a high-g threshold for the detection of shock or other high-acceleration events.

The high-g interrupt is enabled (disabled) per axis by writing '1' ('0') to bits (0x17) *high\_en\_x*, (0x17) *high\_en\_y*, and (0x17) *high\_en\_z*, respectively. The high-g threshold is set through the (0x26) *high\_th* register. The meaning of an LSB of (0x26) *high\_th* depends on the selected g-range: it corresponds to 7.81 mg in 2g-range, 15.63 mg in 4g-range, 31.25 mg in 8g-range, and 62.5 mg in 16g-range (i.e. increment depends from g-range setting).

A hysteresis can be selected by setting the (0x24) *high\_hy* bits. Analogously to (0x26) *high\_th*, the meaning of an LSB of (0x24) *high\_hy* is g-range dependent: it corresponds to an acceleration difference of 125 mg in 2g-range, 250 mg in 4g-range, 500 mg in 8g-range, and 1000mg in 16g-range (as well, increment depends from g-range setting).

The high-g interrupt is generated if the absolute value of the acceleration of at least one of the enabled axes ('or' relation) is higher than the threshold for at least the time defined by the (0x25) *high\_dur* register. The interrupt is reset if the absolute value of the acceleration of all enabled axes ('and' relation) is lower than the threshold minus the hysteresis for at least the

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time defined by the (0x25) *high\_dur* register. In bit (0x09) *high\_int* the interrupt status is stored. The relation between the content of (0x25) *high\_dur* and the actual delay of the interrupt generation is  $\text{delay [ms]} = [(0x22) \text{ low\_dur} + 1] \cdot 2 \text{ ms}$ . Therefore, possible delay times range from 2 ms to 512 ms.

#### 4.8.10.1 Axis and sign information of high-g interrupt

The axis which triggered the interrupt is indicated by bits (0x0C) *high\_first\_x*, (0x0C) *high\_first\_y*, and (0x0C) *high\_first\_z*. The bit corresponding to the triggering axis contains a '1' while the other bits hold a '0'. These bits are cleared together with clearing the interrupt status. The sign of the triggering acceleration is stored in bit (0x0C) *high\_sign*. If (0x0C) *high\_sign* = '0' ('1'), the sign is positive (negative).

## 5. Register description

### 5.1 General remarks


The entire communication with the device is performed by reading from and writing to registers (exception: dedicated mode, see chapter 4.2.2). Registers have a width of 8 bits; they are mapped to a common space of 64 addresses from (0x00) up to (0x3F). Within the used range there are several registers which are either completely or partially marked as 'reserved'. Any reserved bit is ignored when it is written and no specific value is guaranteed when read. It is recommended not to use registers at all which are completely marked as 'reserved'. Furthermore it is recommended to mask out (logical *and* with zero) reserved bits of registers which are partially marked as reserved.

Registers with addresses from (0x00) up to (0x0E) are read-only. Any attempt to write to these registers is ignored. There are bits within some registers that connected with an action to be done and, therefore, are intended for write-only access, e. g. (0x21) *reset\_int* or the entire (0x14) *softreset* register. Such bits always give '0' when read.

## 5.2 Register map

Register Address	Default Value	bit7	bit6	bit5	bit4	bit3	bit2	bit1	bit0	
0x3F	0x00	reserved								
0x3E	0x00	reserved								
0x3D	0x00	offset unfilt z<7:0>								
0x3C	0x00	offset unfilt y<7:0>								
0x3B	0x00	offset unfilt x<7:0>								
0x3A	0x00	offset filt z<7:0>								
0x39	0x00	offset filt y<7:0>								
0x38	0x00	offset filt x<7:0>								
0x37	0x00	reserved	offset target z<1:0>		offset target y<1:0>		offset target x<1:0>		cut off	
0x36	0x10	offset reset	cal trigger<1:0>		cal rdy	reserved	hp z en	hp y en	hp x en	
0x35	0x00	reserved								
0x34	0x00	reserved						i2c_wdt_en	i2c_wdt_sel	spi3
0x33	0x04	reserved				nvm load	nvm rdy	nvm prog trig	nvm prog mode	
0x32	0x70	reserved				self test sign		self test axis		
0x31	0x01	reserved								
0x30	0x00	reserved								
0x2F	0x10	reserved	flat hold time<1:0>			reserved				
0x2E	0x08	reserved	flat theta<5:0>							
0x2D	0x08	reserved	orient theta<5:0>							
0x2C	0x18	reserved	orient hyst<2:0>			orient blocking<1:0>		orient mode<1:0>		
0x2B	0x0A	tap_samp<1:0>		reserved	tap_th<4:0>			tap_dur<2:0>		
0x2A	0x04	tap quiet	tap shock	reserved						
0x29	0x00	reserved								
0x28	0x14	slope_th<7:0>								
0x27	0x00	reserved						slope_dur<1:0>		
0x26	0xC0	high_th<7:0>								
0x25	0x0F	high_dur<7:0>								
0x24	0x81	high_hy<1:0>			reserved		low mode	low_hy<1:0>		
0x23	0x30	low_th<7:0>								
0x22	0x09	low_dur<7:0>								
0x21	0x00	reset_int	reserved				latch_int<3:0>			
0x20	0x05	reserved				int2_od	int2_lv	int1_od	int1_lv	
0x1F	0x00	reserved				reserved				
0x1E	0x00	reserved		int_src_data	int_src_tap	reserved	int_src_slope	int_src_high	int_src_low	
0x1D	0x00	reserved								
0x1C	0x00	reserved								
0x1B	0x00	int2_flat	int2_orient	int2_s_tap	int2_d_tap	reserved	int2_slope	int2_high	int2_low	
0x1A	0x00	int2_data		reserved				int1_data		
0x19	0x00	int1_flat	int1_orient	int1_s_tap	int1_d_tap	reserved	int1_slope	int1_high	int1_low	
0x18	0x00	reserved								
0x17	0x00	reserved				data_en	low_en	high_en_z	high_en_y	
0x16	0x00	flat_en	orient_en	s_tap_en	d_tap_en	reserved	slope_en_z	slope_en_y	slope_en_x	
0x15	0x00	reserved								
0x14	0x00	softreset								
0x13	0x00	data_high_bw	shadow_dis	reserved						
0x12	0x00	reserved								
0x11	0x00	suspend	lowpower_en	reserved	sleep_dur<3:0>				reserved	
0x10	0x1F	reserved				bw<4:0>				
0x0F	0x03	reserved				range<3:0>				
0x0E	0x00	reserved								
0x0D	0x00	reserved								
0x0C	0x00	flat	orient<2:0>			high_sign	high_first_z	high_first_y	high_first_x	
0x0B	0x00	tap_sign	tap_first_z	tap_first_y	tap_first_x	slope_sign	slope_first_z	slope_first_y	slope_first_x	
0x0A	0x00	data_int	reserved							
0x09	0x00	flat_int	orient_int	s_tap_int	d_tap_int	reserved	slope_int	high_int	low_int	
0x08	0x00	temp<7:0>								
0x07	0x00	acc_z_msb<9:2>								
0x06	0x00	acc_z_lsb<1:0>			0				new_data_z	
0x05	0x00	acc_y_msb<9:2>								
0x04	0x00	acc_y_lsb<1:0>			0				new_data_y	
0x03	0x00	acc_x_msb<9:2>								
0x02	0x00	acc_x_lsb<1:0>			0				new_data_x	
0x01	n/a	reserved								
0x00	0x03	Chip ID								

	w/r
	write only
	read only
	reserved

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### 5.3 Chip ID

**Register (0x00)** *Chip ID* contains the chip identification number.

**Table 23: Chip identification number, register (0x00)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	0	0	0	0	0	1	1

**Register (0x01)** *is reserved*

### 5.4 Acceleration data

**Register (0x02)** contains the LSB part of x-axis acceleration data and the new data flag for the x-axis.

**Table 24: LSB part of x-axis acceleration, register (0x02)**

(0x02) Bit	Name	Description
Bit 7	<i>acc_x_lsb &lt;1&gt;</i>	Bit 1 of x-axis acceleration data
Bit 6	<i>acc_x_lsb &lt;0&gt;</i>	Bit 0 of x-axis acceleration data = <b>x LSB</b>
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	<i>new_data_x</i>	New data flag of x-axis

**Register (0x03)** contains the MSB part of x-axis acceleration data.

**Table 25: MSB part of x-axis acceleration, register (0x03)**

(0x03) Bit	Name	Description
Bit 7	<i>acc_x_msb &lt;9&gt;</i>	Bit 9 of x-axis acceleration data = <b>x MSB</b>
Bit 6	<i>acc_x_msb &lt;8&gt;</i>	Bit 8 of x-axis acceleration data
Bit 5	<i>acc_x_msb &lt;7&gt;</i>	Bit 7 of x-axis acceleration data
Bit 4	<i>acc_x_msb &lt;6&gt;</i>	Bit 6 of x-axis acceleration data
Bit 3	<i>acc_x_msb &lt;5&gt;</i>	Bit 5 of x-axis acceleration data
Bit 2	<i>acc_x_msb &lt;4&gt;</i>	Bit 4 of x-axis acceleration data
Bit 1	<i>acc_x_msb &lt;3&gt;</i>	Bit 3 of x-axis acceleration data
Bit 0	<i>acc_x_msb &lt;2&gt;</i>	Bit 2 of x-axis acceleration data

**Register (0x04)** contains the LSB part of y-axis acceleration data and the new data flag for the y-axis.

**Table 26: LSB part of y-axis acceleration, register (0x04)**

(0x04) Bit	Name	Description
Bit 7	<i>acc_y_lsb &lt;1&gt;</i>	Bit 1 of y-axis acceleration data
Bit 6	<i>acc_y_lsb &lt;0&gt;</i>	Bit 0 of y-axis acceleration data = <b>y LSB</b>
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	<i>new_data_y</i>	New data flag of y-axis

**Register (0x05)** contains the MSB part of acceleration data for the y-axis.


**Table 27: MSB part of y-axis acceleration, register (0x05)**

(0x05) Bit	Name	Description
Bit 7	<i>acc_y_msb &lt;9&gt;</i>	Bit 9 of y-axis acceleration data = <b>y MSB</b>
Bit 6	<i>acc_y_msb &lt;8&gt;</i>	Bit 8 of y-axis acceleration data
Bit 5	<i>acc_y_msb &lt;7&gt;</i>	Bit 7 of y-axis acceleration data
Bit 4	<i>acc_y_msb &lt;6&gt;</i>	Bit 6 of y-axis acceleration data
Bit 3	<i>acc_y_msb &lt;5&gt;</i>	Bit 5 of y-axis acceleration data
Bit 2	<i>acc_y_msb &lt;4&gt;</i>	Bit 4 of y-axis acceleration data
Bit 1	<i>acc_y_msb &lt;3&gt;</i>	Bit 3 of y-axis acceleration data
Bit 0	<i>acc_y_msb &lt;2&gt;</i>	Bit 2 of y-axis acceleration data

**Register (0x06)** contains the LSB part of acceleration data and the new data flag for the z-axis.

**Table 28: LSB part of y-axis acceleration, register (0x06)**

(0x06) Bit	Name	Description
Bit 7	<i>acc_z_lsb &lt;1&gt;</i>	Bit 1 of z-axis acceleration data
Bit 6	<i>acc_z_lsb &lt;0&gt;</i>	Bit 0 of z-axis acceleration data = <b>z LSB</b>
Bit 5	-	(fixed to 0)
Bit 4	-	(fixed to 0)
Bit 3	-	(fixed to 0)
Bit 2	-	(fixed to 0)
Bit 1	-	(fixed to 0)
Bit 0	<i>new_data_z</i>	New data flag of z-axis

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**Register (0x07)** contains the MSB part of acceleration data for the z-axis.

**Table 29: MSB part of z-axis acceleration, register (0x07)**

(0x07) Bit	Name	Description
Bit 7	<i>acc_z_msb</i> <9>	Bit 9 of z-axis acceleration data = <b>z MSB</b>
Bit 6	<i>acc_z_msb</i> <8>	Bit 8 of z-axis acceleration data
Bit 5	<i>acc_z_msb</i> <7>	Bit 7 of z-axis acceleration data
Bit 4	<i>acc_z_msb</i> <6>	Bit 6 of z-axis acceleration data
Bit 3	<i>acc_z_msb</i> <5>	Bit 5 of z-axis acceleration data
Bit 2	<i>acc_z_msb</i> <4>	Bit 4 of z-axis acceleration data
Bit 1	<i>acc_z_msb</i> <3>	Bit 3 of z-axis acceleration data
Bit 0	<i>acc_z_msb</i> <2>	Bit 2 of z-axis acceleration data

## 5.5 Temperature data

**Register (0x08) temp** contains temperature data in two's complement representation. Center temperature = 24 °C → i.e. (0x08) temp = 00000000b

1 LSB increment of temperature sensor is 0.5 °C (0.9 °F).

**Table 30: Temperature data, register (0x08)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>Temp</i> <7>	<i>Temp</i> <6>	<i>Temp</i> <5>	<i>Temp</i> <4>	<i>Temp</i> <3>	<i>Temp</i> <2>	<i>Temp</i> <1>	<i>Temp</i> <0>

## 5.6 Status registers

**Register (0x09)** contains the states of several interrupts.

**Table 31: Interrupt status, register (0x09)**

(0x09) Bit	Name	Description
Bit 7	<i>flat_int</i>	Flat interrupt status
Bit 6	<i>orient_int</i>	Orientation interrupt status
Bit 5	<i>s_tap_int</i>	Single tap interrupt status
Bit 4	<i>d_tap_int</i>	Double tap interrupt status
Bit 3	- reserved -	reserved
Bit 2	<i>slope_int</i>	Slope interrupt status
Bit 1	<i>high_int</i>	High-g interrupt status
Bit 0	<i>low_int</i>	Low-g interrupt status

**Register (0x0A)** contains the status of the new data interrupt.

**Table 32: New data status, register (0x0A)**

(0x0A) Bit	Name	Description
Bit 7	<i>data_int</i>	New data interrupt status
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	- reserved -	reserved
Bit 2	- reserved -	reserved
Bit 1	- reserved -	reserved
Bit 0	- reserved -	reserved

**Register (0x0B)** contains the sign and triggering axis information for the tap and slope interrupts. Here tap interrupt comprises both single and double tap interrupt.

**Table 33: Tap and slope interrupts status, register (0x0B)**

(0x0B) Bit	Name	Description
Bit 7	<i>tap_sign</i>	Sign of 1 <sup>st</sup> tap that triggered the interrupt ('0'=positive, '1'=negative)
Bit 6	<i>tap_first_z</i>	'1' indicates that z-axis is triggering axis of tap interrupt
Bit 5	<i>tap_first_y</i>	'1' indicates that y-axis is triggering axis of tap interrupt
Bit 4	<i>tap_first_x</i>	'1' indicates that x-axis is triggering axis of tap interrupt
Bit 3	<i>slope_sign</i>	Sign of slope that triggered the interrupt ('0'=positive, '1'=negative)
Bit 2	<i>slope_first_z</i>	'1' indicates that z-axis is triggering axis of slope interrupt
Bit 1	<i>slope_first_y</i>	'1' indicates that y-axis is triggering axis of slope interrupt
Bit 0	<i>slope_first_x</i>	'1' indicates that x-axis is triggering axis of slope interrupt


**Register (0x0C)** contains the flat and orientation status, and the sign and triggering axis information for the high-g interrupt. Registers (0x0D) and (0x0E) are *reserved*.

**Table 34: Flat and orientation Status, register (0x0C)**

(0x0C) Bit	Name	Description
Bit 7	<i>flat</i>	flat detection ('1' if flat condition is fulfilled, '0' otherwise)
Bit 6	<i>orient &lt;2&gt;</i>	orientation value of z-axis ('0' if upward looking, '1' if downward looking)
Bit 5	<i>orient &lt;1&gt;</i>	orientation value of x-y plane ('00'=portrait upright, '01'=portrait upside-down, '10'=landscape left, '11'=landscape right)
Bit 4	<i>orient &lt;0&gt;</i>	
Bit 3	<i>high_sign</i>	Sign of slope that triggered the interrupt ('0'=positive, '1'=negative)
Bit 2	<i>high_first_z</i>	'1' indicates that z-axis is triggering axis of high-g interrupt
Bit 1	<i>high_first_y</i>	'1' indicates that y-axis is triggering axis of high-g interrupt
Bit 0	<i>high_first_x</i>	'1' indicates that x-axis is triggering axis of high-g interrupt

**Registers (0x0D) and (0x0E)** are *reserved*.



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## 5.7 g-range selection

**Register (0x0F)** contains the selection of the g-range. Proper settings for (0x0F) range are '0011b' (selects  $\pm 2g$  range), '0101b' (selects  $\pm 4g$  range), '1000b' (selects  $\pm 8g$  range), '1100b' (selects  $\pm 16g$  range). All other settings are irregular; if such a setting is used,  $\pm 2g$  range is selected. Default value of (0x0F) range (after reset) is '0011b'.

**Table 35: g-range, register (0x0F)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	range <3>	range <2>	range <1>	range <0>

## 5.8 Bandwidths

**Register (0x10)** contains the selection of the bandwidth for filtered acceleration data. Settings for (0x10) bw are '00xxb' (bandwidth = 7.81 Hz), '01000b' (bandwidth = 7.81 Hz), '01001b' (bandwidth = 15.63 Hz), '01010b' (bandwidth = 31.25 Hz), '01011b' (bandwidth = 62.5 Hz), '01100b' (bandwidth = 125 Hz), '01101b' (bandwidth = 250 Hz), '01110b' (bandwidth = 500 Hz), '01111b' (bandwidth = 1000 Hz), '1xxxxb' (bandwidth = 1000 Hz). Default value of (0x10) bw (after reset) is '11111b'. It is recommended to actively use the range from '01000b' to '01111b' only in order to be compatible with future products.

**Table 36: Bandwidths, register (0x10)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	bw <4>	bw <3>	bw <2>	bw <1>	bw <0>

## 5.9 Power modes

**Register (0x11)** contains the configuration of the power modes. (0x11) suspend = '1' ('0') sets (resets) suspend mode; default value of (0x11) suspend is '0'. (0x11) lowpower\_en = '1' ('0') sets (resets) low-power mode, default value of (0x11) lowpower\_en is '0'.

The settings for (0x11) sleep\_dur are '0000b' to '0101b' (sleep phase duration = 0.5 ms), '0110b' (sleep phase duration = 1 ms), '0111b' (sleep phase duration = 2 ms), '1000b' (sleep phase duration = 4 ms), '1001b' (sleep phase duration = 6 ms), '1010b' (sleep phase duration = 10 ms), '1011b' (sleep phase duration = 25 ms), '1100b' (sleep phase duration = 50 ms), '1101b' (sleep phase duration = 100 ms), '1110b' (sleep phase duration = 500 ms), '1111b' (sleep phase duration = 1 s). Default value of (0x11) sleep\_dur is '0000b'.

**Table 37: Power modes, register (0x11)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
suspend	lowpower_en	reserved	sleep_dur<3>	sleep_dur<2>	sleep_dur<1>	sleep_dur<0>	reserved

## 5.10 Special control settings

**Register (0x12)** is reserved.

**Register (0x13)** contains settings for the configuration of the acceleration data acquisition and the data output format.

(0x13) *data\_high\_bw* = '0' ('1') selects filtered (unfiltered) acceleration data to be written into the data registers (0x02) to (0x07). Default value of (0x13) *data\_high\_bw* is '0'.

(0x13) *shadow\_dis* = '0' ('1') enables (disables) the shadowing procedure. Shadowing means that the MSB register is updated by reading the corresponding LSB register. Default value of (0x13) *shadow\_dis* is '0'.

**Table 38: Acceleration data acquisition & data output format, register (0x13)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>data_high_bw</i>	<i>shadow_dis</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>

**Register (0x14)** is the *softreset* register. A user-triggered reset (*softreset*) of the sensor is performed after writing '0xB6' to the *softreset* register. After that reset all registers return to their default values. Reading (0x14) *softreset* returns 0x00.


**Register (0x15)** is reserved.

## 5.11 Interrupt settings

**Registers (0x16) and (0x17)** contain the enable bits for the interrupts. Default value of each enable bit is '0'.

**Table 39: Interrupt setting, register (0x16)**

(0x16) Bit	Name	Description
Bit 7	<i>flat_en</i>	'1' ('0') enables (disables) flat interrupt
Bit 6	<i>orient_en</i>	'1' ('0') enables (disables) orientation interrupt
Bit 5	<i>s_tap_en</i>	'1' ('0') enables (disables) single tap interrupt
Bit 4	<i>d_tap_en</i>	'1' ('0') enables (disables) double tap interrupt
Bit 3	- reserved -	reserved
Bit 2	<i>slope_en_z</i>	'1' ('0') enables (disables) slope interrupt for z-axis
Bit 1	<i>slope_en_y</i>	'1' ('0') enables (disables) slope interrupt for y-axis
Bit 0	<i>slope_en_x</i>	'1' ('0') enables (disables) slope interrupt for x-axis

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**Table 40: Interrupt setting, register (0x17)**

(0x17) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	<i>data_en</i>	'1' ('0') enables (disables) new data interrupt
Bit 3	<i>low_en</i>	'1' ('0') enables (disables) low-g interrupt
Bit 2	<i>high_en_z</i>	'1' ('0') enables (disables) high-g interrupt for z-axis
Bit 1	<i>high_en_y</i>	'1' ('0') enables (disables) high-g interrupt for y-axis
Bit 0	<i>high_en_x</i>	'1' ('0') enables (disables) high-g interrupt for x-axis

**Register (0x18)** is reserved.

**Registers (0x19) to (0x1B)** contain the mapping of interrupts onto the interrupt pins. Default value of each mapping bit is '0'.

**Table 41: Interrupt mapping, register (0x19)**

(0x19) Bit	Name	Description
Bit 7	<i>int1_flat</i>	'1' ('0') maps (unmaps) flat interrupt to INT1 pin
Bit 6	<i>int1_orient</i>	'1' ('0') maps (unmaps) orientation interrupt to INT1 pin
Bit 5	<i>int1_s_tap</i>	'1' ('0') maps (unmaps) single tap interrupt to INT1 pin
Bit 4	<i>int1_d_tap</i>	'1' ('0') maps (unmaps) double tap interrupt to INT1 pin
Bit 3	- reserved -	reserved
Bit 2	<i>int1_slope</i>	'1' ('0') maps (unmaps) slope interrupt to INT1 pin
Bit 1	<i>int1_high</i>	'1' ('0') maps (unmaps) high-g interrupt to INT1 pin
Bit 0	<i>int1_low</i>	'1' ('0') maps (unmaps) low-g interrupt to INT1 pin

**Table 42: Interrupt mapping, register (0x1A)**

(0x1A) Bit	Name	Description
Bit 7	<i>int2_data</i>	'1' ('0') maps (unmaps) new data interrupt to INT2 pin
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	- reserved -	reserved
Bit 2	- reserved -	reserved
Bit 1	- reserved -	reserved
Bit 0	<i>int1_data</i>	'1' ('0') maps (unmaps) new data interrupt to INT1 pin

**Table 43: Interrupt mapping, register (0x1B)**

(0x1B) Bit	Name	Description
Bit 7	<i>int2_flat</i>	'1' ('0') maps (unmaps) flat interrupt to INT2 pin
Bit 6	<i>int2_orient</i>	'1' ('0') maps (unmaps) orientation interrupt to INT2 pin
Bit 5	<i>int2_s_tap</i>	'1' ('0') maps (unmaps) single tap interrupt to INT2 pin
Bit 4	<i>int2_d_tap</i>	'1' ('0') maps (unmaps) double tap interrupt to INT2 pin
Bit 3	- reserved -	reserved
Bit 2	<i>int2_slope</i>	'1' ('0') maps (unmaps) slope interrupt to INT2 pin
Bit 1	<i>int2_high</i>	'1' ('0') maps (unmaps) high-g interrupt to INT2 pin
Bit 0	<i>int2_low</i>	'1' ('0') maps (unmaps) low-g interrupt to INT2 pin

**Registers (0x1C) and (0x1D) are reserved.**

**Register (0x1E)** contains the data source definition for those interrupts with selectable data source. Default value of each data source selection bit is '0'.

**Table 44: Interrupt data source definition, register (0x1E)**

(0x1E) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	<i>int_src_data</i>	'1' ('0') selects unfiltered (filtered) data for the new data interrupt
Bit 4	<i>int_src_tap</i>	'1' ('0') selects unfiltered (filtered) data for the single tap and double tap interrupts
Bit 3	- reserved -	reserved
Bit 2	<i>int_src_slope</i>	'1' ('0') selects unfiltered (filtered) data for the slope interrupt
Bit 1	<i>int_src_high</i>	'1' ('0') selects unfiltered (filtered) data for the high-g interrupt
Bit 0	<i>int_src_low</i>	'1' ('0') selects unfiltered (filtered) data for the low-g interrupt

**Register (0x1F) is reserved.**

**Register (0x20)** contains the behavioural configuration (electrical behaviour) of the interrupt pins. Default value of (0x20) *int1\_od* and (0x20) *int2\_od* is '0'. Default value of (0x20) *int1\_lvl* and (0x20) *int2\_lvl* is '1'.

**Table 45: Electrical behaviour of interrupt pin, register (0x20)**

(0x20) Bit	Name	Description
Bit 7	- reserved -	reserved
Bit 6	- reserved -	reserved
Bit 5	- reserved -	reserved
Bit 4	- reserved -	reserved
Bit 3	<i>int2_od</i>	'0' selects push-pull, '1' selects open drive for INT2 pin
Bit 2	<i>int2_lvl</i>	'0' ('1') selects active level '0' ('1') for INT2 pin
Bit 1	<i>int1_od</i>	'0' selects push-pull, '1' selects open drive for INT1
Bit 0	<i>int1_lvl</i>	'0' ('1') selects active level '0' ('1') for INT1 pin

**Register (0x21)** contains the interrupt reset bit and the interrupt mode selection. Writing '1' to (0x21) *reset\_int* resets any latched interrupt.

The settings for (0x21) *latch\_int* are '0000b' (non-latched), '0001b' (temporary, 250 ms), '0010b' (temporary, 500 ms), '0011b' (temporary, 1 s), '0100b' (temporary, 2 s), '0101b' (temporary, 4 s), '0110b' (temporary, 8 s), '0111b' (latched), '1000b' (non-latched), '1001b' (temporary, 500 μs), '1010b' (temporary, 500 μs), '1011b' (temporary, 1 ms), '1100b' (temporary, 12.5 ms), '1101b' (temporary, 25 ms), '1110b' (temporary, 50 ms), '1111b' (latched).

Default value of (0x21) *latch\_int* is '0000b'.

**Table 46: Interrupt reset bit and interrupt mode selection, register (0x21)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reset_int</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>latch_int</i> <3>	<i>latch_int</i> <2>	<i>latch_int</i> <1>	<i>latch_int</i> <0>

**Register (0x22)** contains the delay time definition for the low-g interrupt. The physical delay time can be computed from the content of (0x22) *low\_dur* according to:

delay [ms] = [(0x22) *low\_dur* + 1] • 2 ms.

Possible delay times range from 2 ms to 512 ms. Default value of (0x22) *low\_dur* is 0x09, corresponding to a delay of 20 ms.

**Table 47: Delay time definition for the low-g interrupt, register (0x22)**


Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>low_dur</i> <7>	<i>low_dur</i> <6>	<i>low_dur</i> <5>	<i>low_dur</i> <4>	<i>low_dur</i> <3>	<i>low_dur</i> <2>	<i>low_dur</i> <1>	<i>low_dur</i> <0>

**Register (0x23)** contains the threshold definition for the low-g interrupt. An LSB of (0x23) *low\_th* corresponds to an actual acceleration of 7.81 mg. Therefore, the threshold ranges from 0 g to 1.992 g. Default value of (0x23) *low\_th* is 0x30, corresponding to an acceleration of 375 mg.

**Table 48: Threshold definition for the low-g interrupt, register (0x23)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>low_th</i> <7>	<i>low_th</i> <6>	<i>low_th</i> <5>	<i>low_th</i> <4>	<i>low_th</i> <3>	<i>low_th</i> <2>	<i>low_th</i> <1>	<i>low_th</i> <0>

**Register (0x24)** contains the low-g interrupt mode selection, the low-g interrupt hysteresis setting, and the high-g interrupt hysteresis setting. Setting (0x24) *low\_mode* to '0' ('1') selects 'single' mode ('sum' mode). Default value is '0' ('single' mode).

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(0x24) *low\_hy* sets the hysteresis of the low-g interrupt. An LSB of (0x24) *low\_hy* corresponds to an acceleration difference of 125 mg. Default value of (0x24) *low\_hy* is '01b'.

(0x24) *high\_hy* sets the hysteresis of the high-g interrupt. The meaning of an LSB of (0x24) *high\_hy* depends on the selected g-range. It corresponds to an acceleration difference of 125 mg in 2g-range, 250 mg in 4g-range, 500 mg in 8g-range, and 1000mg in 16g-range. Default value of (0x24) *high\_hy* is '10b'.

**Table 49: Threshold definition for the low-g interrupt, register (0x24)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>high_hy</i> <1>	<i>high_hy</i> <0>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>low_mode</i>	<i>low_hy</i> <1>	<i>low_hy</i> <0>

**Register (0x25)** contains the delay time definition for the high-g interrupt. The physical delay time can be computed from the content of (0x25) *high\_dur* according to  $\text{delay [ms]} = [(0x25) \text{ high\_dur} + 1] \cdot 2 \text{ ms}$ . Possible delay times range from 2 ms to 512 ms. Default value of (0x25) *high\_dur* is 0x0F, corresponding to a delay of 32 ms.

**Table 50: Delay time definition for the high-g interrupt, register (0x25)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>high_dur</i> <7>	<i>high_dur</i> <6>	<i>high_dur</i> <5>	<i>high_dur</i> <4>	<i>high_dur</i> <3>	<i>high_dur</i> <2>	<i>high_dur</i> <1>	<i>high_dur</i> <0>

**Register (0x26)** contains the threshold definition for the high-g interrupt. The meaning of an LSB of (0x26) *high\_th* depends on the selected g-range. It corresponds to 7.81 mg in 2g-range, 15.63 mg in 4g-range, 31.25 mg in 8g-range, and 62.5 mg in 16g-range. Default value of (0x26) *high\_th* is 0xC0.


**Table 51: Threshold definition for the high-g interrupt, register (0x26)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>high_th</i> <7>	<i>high_th</i> <6>	<i>high_th</i> <5>	<i>high_th</i> <4>	<i>high_th</i> <3>	<i>high_th</i> <2>	<i>high_th</i> <1>	<i>high_th</i> <0>

**Register (0x27)** contains the definition of the number of samples to be evaluated for the slope interrupt (any-motion detection). The number of samples is  $N = (0x27) \text{ slope\_dur} + 1$ . Default value of (0x27) *slope\_dur* is '00b'.

**Table 52: Samples number definition for the slope interrupt, register (0x27)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>slope_dur</i> <1>	<i>slope_dur</i> <0>

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**Register (0x28)** contains the threshold definition for the slope interrupt. An LSB of (0x28) *slope\_th* corresponds to an LSB of acceleration data. Its meaning therefore depends on the selected g-range. Default value of (0x28) *slope\_th* is 0x14.

**Table 53: Slope threshold for the slope interrupt, register (0x28)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>slope_th&lt;7&gt;</i>	<i>slope_th&lt;6&gt;</i>	<i>slope_th&lt;5&gt;</i>	<i>slope_th&lt;4&gt;</i>	<i>slope_th&lt;3&gt;</i>	<i>slope_th&lt;2&gt;</i>	<i>slope_th&lt;1&gt;</i>	<i>slope_th&lt;0&gt;</i>

**Register (0x29)** is reserved.

**Register (0x2A)** contains the timing definitions for the single tap and double tap interrupts.

(0x2A) *tap\_quiet* = '0' ('1') selects a quiet duration of 30 ms (20 ms). The default value of (0x2A) *tap\_quiet* is '0'.

(0x2A) *tap\_shock* = '0' ('1') selects a shock duration of 50 ms (75 ms). The default value of (0x2A) *tap\_shock* is '0'.

(0x2A) *tap\_dur* selects the length of the time window for the second shock event (for double tap detection). The settings for (0x2A) *tap\_dur* are '000b' (50 ms), '001b' (100 ms), '010b' (150 ms), '011b' (200 ms), '100b' (250 ms), '101b' (375 ms), '110b' (500 ms), '111b' (700 ms). The default value of (0x2A) *tap\_dur* is '100b'.

**Table 54: Tap Quiet duration and tap shock duration, register (0x2A)**


Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>tap_quiet</i>	<i>tap_shock</i>	reserved	reserved	reserved	<i>tap_dur&lt;2&gt;</i>	<i>tap_dur&lt;1&gt;</i>	<i>tap_dur&lt;0&gt;</i>

**Register (0x2B)** contains the definition of the number of samples to be processed after wake-up in low-power mode and the threshold definition for the single and double tap interrupts. (0x2B) *tap\_samp* selects the number of samples that are processed after wake-up in the low-power mode. The settings for (0x2B) *tap\_samp* are '00b' (2 samples), '01b' (4 samples), '10b' (8 samples), and '11b' (16 samples). Default value of (0x2B) *tap\_samp* is '00b'.

The meaning of an LSB of (0x2B) *tap\_th* depends on the selected g-range. It corresponds to an acceleration difference of 62.5mg in 2g-range, 125mg in 4g-range, 250mg in 8g-range, and 500mg in 16g-range. Default value of (0x2B) *tap\_th* is 0x0A.

**Table 55: Samples number after wake-up and threshold tap interrupt, register (0x2B)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>tap_samp&lt;1&gt;</i>	<i>tap_samp&lt;0&gt;</i>	reserved	<i>tap_th&lt;4&gt;</i>	<i>tap_th&lt;3&gt;</i>	<i>tap_th&lt;2&gt;</i>	<i>tap_th&lt;1&gt;</i>	<i>tap_th&lt;0&gt;</i>

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**Register (0x2C)** contains the definition of hysteresis, blocking, and mode for the orientation interrupt. (0x2C) *orient\_hyst* sets the hysteresis of the orientation interrupt; 1 LSB always corresponds to 62.5 mg, in any g-range (i.e. increment is independent from g-range setting). Default value of (0x2C) *orient\_hyst* is '001b'.

(0x2C) *orient\_blocking* selects the kind of blocking that is used for the generation of the orientation interrupt. The settings for (0x2C) *orient\_blocking* are '00b' (no blocking), '01b' (theta blocking), '10b' (theta blocking or slope in any axis > 0.2 g), and '11b' (*orient* value not stable for at least 100 ms or theta blocking or slope in any axis > 0.4 g). Default value of (0x2C) *orient\_blocking* is '10b'.

(0x2C) *orient\_mode* sets the thresholds for switching between the different orientations. The settings for (0x2C) *orient\_mode* are '00b' (symmetrical), '01b' (high-asymmetrical), '10b' (low-asymmetrical), '11b' (symmetrical). Default value of (0x2C) *orient\_mode* is '00b'.

**Table 56: Hysteresis, Blocking for Orientation Interrupt, Register (0x2C)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>orient_hyst</i> <2>	<i>orient_hyst</i> <1>	<i>orient_hyst</i> <0>	<i>orient_blocking</i> <1>	<i>orient_blocking</i> <0>	<i>orient_mode</i> <1>	<i>orient_mode</i> <0>

**Register (0x2D)** contains the definition of the theta blocking angle for the orientation interrupt. (0x2D) *orient\_theta* defines a blocking angle between 0° and 44.8° as described in section "4.8.1.7 Orientation blocking". Default value of (0x2D) *orient\_theta* is 0x08.

**Table 57: Theta blocking angle, register (0x2D)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>reserved</i>	<i>orient_theta</i> <5>	<i>orient_theta</i> <4>	<i>orient_theta</i> <3>	<i>orient_theta</i> <2>	<i>orient_theta</i> <1>	<i>orient_theta</i> <0>

**Register (0x2E)** contains the definition of the flat threshold angle for the flat interrupt. (0x2E) *flat\_theta* defines a blocking angle between 0° and 44.8° as described in section "4.8.8 Flat detection". Default value of (0x2E) *flat\_theta* is 0x08.

**Table 58: Flat threshold angle, register (0x2E)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>reserved</i>	<i>flat_theta</i> <5>	<i>flat_theta</i> <4>	<i>flat_theta</i> <3>	<i>flat_theta</i> <2>	<i>flat_theta</i> <1>	<i>flat_theta</i> <0>

**Register (0x2F)** contains the definition of the flat hold time. (0x2F) *flat\_hold\_time* defines the time a new flat value has to be at least stable for before the interrupt is generated. The settings for (0x2F) *flat\_hold\_time* are '00b' (0), '01b' (512 ms), '10b' (1024 ms), '11b' (2048 ms). Default value of (0x2F) *flat\_hold\_time* is '01b'.



**Table 59: Flat threshold angle, register (0x2F)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	flat_hold_time<1>	flat_hold_time<0>	reserved	reserved	reserved	reserved

**Register (0x30) and (0x31)** are reserved.

## 5.12 Self-test

**Register (0x32)** contains the settings for the activation of the sensor self-test.

(0x32) *self\_test\_sign* sets the sign of the electrostatic excitation. The settings for (0x32) *self\_test\_sign* are '0' (positive sign) and '1' (negative sign). Default value of (0x32) *self\_test\_sign* is '0'.

(0x32) *self\_test\_axis* defines the axis which shall be excited. Only one axis can be excited at the same time. The settings for (0x32) *self\_test\_axis* are '00b' (no self-test), '01' (x-axis), '10' (y-axis), and '11' (z-axis). Default value of (0x32) *self\_test\_axis* is '00b'.

**Table 60: Sensor self-test, register (0x32)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
reserved	reserved	reserved	reserved	reserved	<i>self_test_sign</i>	<i>self_test_axis&lt;1&gt;</i>	<i>self_test_axis&lt;0&gt;</i>

## 5.13 Non-volatile memory control (EEPROM control)

**Register (0x33)** contains the control settings for the non-volatile memory (EEPROM). (0x33) *nvm\_load* is used to perform a user-defined image update. Writing '1' (0x33) *nvm\_load* starts the update procedure. The value '1' is kept as long as the update procedure runs, afterwards it is reset to '0'.

(0x33) *nvm\_rdy* contains the status of writing the EEPROM. (0x33) *nvm\_rdy* is '0' as long as writing the EEPROM endures, it is '1' if currently no write access is performed and, therefore, a new write access can be initiated.

Writing '1' to (0x33) *nvm\_prog\_trig* triggers writing the EEPROM. The EEPROM can only be written if it was unlocked before.

Writing '1' to (0x33) *nvm\_prog\_mode* unlocks the EEPROM.

**Table 61: EEPROM control settings, register (0x33)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>nvm_load</i>	<i>nvm_rdy</i>	<i>nvm_prog_trig</i>	<i>nvm_prog_mode</i>

## 5.14 Interface configuration

**Register (0x34)** contains the settings for the digital interfaces. Writing '1' to (0x34) *i2c\_wdt\_en* enables the watchdog at the SDI pin (= SDA for I<sup>2</sup>C) if I<sup>2</sup>C is selected. Default value of (0x34) *i2c\_wdt\_en* is '0'.

(0x34) *i2c\_wdt\_sel* selects the I<sup>2</sup>C data pad watchdog timer period. The settings for (0x34) *i2c\_wdt\_sel* are '0' (1 ms) and '1' (50 ms). Default value of (0x34) *i2c\_wdt\_sel* is '0'.

(0x34) *spi3* selects the SPI mode. The settings for (0x34) *spi3* are '0' (4-wire SPI) and '1' (3-wire SPI). Default value of (0x34) *spi3* is '0'.

**Table 62: EEPROM control settings, register (0x34)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>reserved</i>	<i>i2c_wdt_en</i>	<i>i2c_wdt_sel</i>	<i>spi3</i>

**Register (0x35)** is *reserved*.

## 5.15 Offset compensation


**Register (0x36)** contains settings for the offset compensation in general, for fast offset compensation, and for slow offset compensation. Writing '1' to (0x36) *offset\_reset* sets all offset compensation registers (0x38 to 0x3D) to zero.

Default value of (0x36) *offset\_reset* is '0'.

(0x36) *cal\_trigger* starts the fast compensation process for the specified axis. The settings for (0x36) *cal\_trigger* are '00b' (no axis selected), '01b' (x-axis), '10b' (y-axis), '11b' (z-axis). A non-zero value is kept until the fast compensation procedure is finished. Default value of (0x36) *cal\_trigger* is '00b'.

(0x36) *cal\_rdy* indicates the state of the fast compensation. (0x36) *cal\_rdy* is '0' when (0x36) *cal\_trigger* has a nonzero value, otherwise (0x36) *cal\_rdy* is '1'.

Writing '1' ('0') to (0x36) *hp\_z\_en* enables (disables) slow offset compensation for the z-axis. Writing '1' ('0') to (0x36) *hp\_y\_en* enables (disables) slow offset compensation for the y-axis.

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Writing '1' ('0') to (0x36) *hp\_x\_en* enables (disables) slow offset compensation for the x-axis. Default value for each of (0x36) *hp\_x\_en*, (0x36) *hp\_y\_en*, and (0x36) *hp\_z\_en* is '0', respectively.

**Table 63: Offset compensation, fast offset compensation, register (0x36)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_reset</i>	<i>cal_trigger&lt;1&gt;</i>	<i>cal_trigger&lt;0&gt;</i>	<i>cal_rdy</i>	<i>reserved</i>	<i>hp_z_en</i>	<i>hp_y_en</i>	<i>hp_x_en</i>

**Register (0x37)** contains settings for the offset compensation in general, and for slow offset compensation. (0x37) *offset\_target\_z* sets the target value for the offset compensation of the z-axis.

(0x37) *offset\_target\_y* sets the target value for the offset compensation of the y-axis.

(0x37) *offset\_target\_x* sets the target value for the offset compensation of the x-axis.

The settings for (0x37) *offset\_target\_x*, (0x37) *offset\_target\_y*, and (0x37) *offset\_target\_z* are '00b' (0 g), '01b' (+1 g), '10b' (-1 g), and '11b' (0 g). Default value of each of (0x37) *offset\_target\_x*, (0x37) *offset\_target\_y*, and (0x37) *offset\_target\_z* is '00b', respectively.

(0x37) *cut\_off* defines the number of samples for comparison by the slow offset compensation. The settings for (0x37) *cut\_off* are '0' (8 samples) and '1' (16 samples). The default value of (0x37) *cut\_off* is '0'.

**Table 64: Offset compensation, slow offset compensation, register (0x37)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>reserved</i>	<i>offset_target_z&lt;1&gt;</i>	<i>offset_target_z&lt;0&gt;</i>	<i>offset_target_y&lt;1&gt;</i>	<i>offset_target_y&lt;0&gt;</i>	<i>offset_target_x&lt;1&gt;</i>	<i>offset_target_x&lt;0&gt;</i>	<i>cut_off</i>

**Register (0x38)** contains the compensation value for filtered data for the x-axis. The contents of each of the registers (0x38) to (0x3D) is added to the corresponding acceleration data; it can be set either automatically by one of the implemented compensation algorithms or manually. These registers are image registers of registers in the EEPROM; the content of the EEPROM is copied to them after every reset.

**Table 65: Filtered data compensation for the x-axis, register (0x38)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_filt_x&lt;7&gt;</i>	<i>offset_filt_x&lt;6&gt;</i>	<i>offset_filt_x&lt;5&gt;</i>	<i>offset_filt_x&lt;4&gt;</i>	<i>offset_filt_x&lt;3&gt;</i>	<i>offset_filt_x&lt;2&gt;</i>	<i>offset_filt_x&lt;1&gt;</i>	<i>offset_filt_x&lt;0&gt;</i>

**Register (0x39)** contains the compensation value for filtered data for the y-axis.

**Table 66: Filtered data compensation for the y-axis, register (0x39)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_ filt_y&lt;7&gt;</i>	<i>offset_ filt_y&lt;6&gt;</i>	<i>offset_ filt_y&lt;5&gt;</i>	<i>offset_ filt_y&lt;4&gt;</i>	<i>offset_ filt_y&lt;3&gt;</i>	<i>offset_ filt_y&lt;2&gt;</i>	<i>offset_ filt_y&lt;1&gt;</i>	<i>offset_ filt_y&lt;0&gt;</i>

**Register (0x3A)** contains the compensation value for filtered data for the z-axis.

**Table 67: Filtered data compensation for the z-axis, register (0x3A)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_ filt_z&lt;7&gt;</i>	<i>offset_ filt_z&lt;6&gt;</i>	<i>offset_ filt_z&lt;5&gt;</i>	<i>offset_ filt_z&lt;4&gt;</i>	<i>offset_ filt_z&lt;3&gt;</i>	<i>offset_ filt_z&lt;2&gt;</i>	<i>offset_ filt_z&lt;1&gt;</i>	<i>offset_ filt_z&lt;0&gt;</i>

**Register (0x3B)** contains the compensation value for unfiltered data for the x-axis.

**Table 68: Unfiltered data compensation for the x-axis, register (0x3B)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_ unfilt_x &lt;7&gt;</i>	<i>offset_ unfilt_x &lt;6&gt;</i>	<i>offset_ unfilt_x &lt;5&gt;</i>	<i>offset_ unfilt_x &lt;4&gt;</i>	<i>offset_ unfilt_x &lt;3&gt;</i>	<i>offset_ unfilt_x &lt;2&gt;</i>	<i>offset_ unfilt_x &lt;1&gt;</i>	<i>offset_ unfilt_x &lt;0&gt;</i>

**Register (0x3C)** contains the compensation value for unfiltered data for the y-axis.


**Table 69: Unfiltered data compensation for the x-axis, register (0x3C)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_ unfilt_y &lt;7&gt;</i>	<i>offset_ unfilt_y &lt;6&gt;</i>	<i>offset_ unfilt_y &lt;5&gt;</i>	<i>offset_ unfilt_y &lt;4&gt;</i>	<i>offset_ unfilt_y &lt;3&gt;</i>	<i>offset_ unfilt_y &lt;2&gt;</i>	<i>offset_ unfilt_y &lt;1&gt;</i>	<i>offset_ unfilt_y &lt;0&gt;</i>


**Register (0x3D)** contains the compensation value for unfiltered data for the z-axis.

**Table 70: Unfiltered data compensation for the y-axis, register (0x3D)**

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
<i>offset_ unfilt_z &lt;7&gt;</i>	<i>offset_ unfilt_z &lt;6&gt;</i>	<i>offset_ unfilt_z &lt;5&gt;</i>	<i>offset_ unfilt_z &lt;4&gt;</i>	<i>offset_ unfilt_z &lt;3&gt;</i>	<i>offset_ unfilt_z &lt;2&gt;</i>	<i>offset_ unfilt_z &lt;1&gt;</i>	<i>offset_ unfilt_z &lt;0&gt;</i>

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**Registers (0x3E) and (0x3F)** are image registers of registers in the EEPROM. They are not linked to any sensor-specific functionality.

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## 6. Digital interfaces

The BMA250 supports two serial digital interface protocols for communication as a slave with a host device (when operating in general mode): SPI and I<sup>2</sup>C. The active interface is selected by the state of the Pin#11 (PS) 'protocol select' pin: '0' ('1') selects SPI (I<sup>2</sup>C). For details see section 4.2 Operational modes.

By default, SPI operates in the standard 4-wire configuration. It can be re-configured by software to work in 3-wire mode instead of standard 4-wire mode.

Both interfaces share the same pins. The mapping for each interface is given in the following table:

**Table 71: Mapping of the interface pins**

Pin#	Name	use w/ SPI	use w/ I <sup>2</sup> C	Description
1	SDO	SDO	address	SPI: Data Output (4-wire mode) I <sup>2</sup> C: Used to set LSB of I <sup>2</sup> C address
2	SDx	SDI	SDA	SPI: Data Input (4-wire mode) Data Input / Output (3-wire mode) I <sup>2</sup> C: Serial Data
10	CSB	CSB	unused	Chip Select (enable)
12	SCx	SCK	SCL	SPI: Serial Clock I <sup>2</sup> C: Serial Clock

The following table shows the electrical specifications of the interface pins:

**Table 72: Electrical specification of the interface pins**

Parameter	Symbol	Condition	Min	Typ	Max	Units
PS Impedance for Tri-state Detection	R <sub>TS</sub>		1			MΩ
	C <sub>TS</sub>				10	pF
PS Impedance for Non-Tri-state	R <sub>NTS</sub>				5	kΩ
Pull-up Resistance	R <sub>up</sub>	Internal Pull-up Resistance to VDDIO	70	120	190	kΩ
Pull-down Resistance	R <sub>down</sub>	Internal Pull-down Resistance to GND	12	20	32	kΩ
Input Capacitance	C <sub>in</sub>			5	10	pF
I <sup>2</sup> C Bus Load Capacitance (max. drive capability)	C <sub>I2C_Load</sub>				400	pF

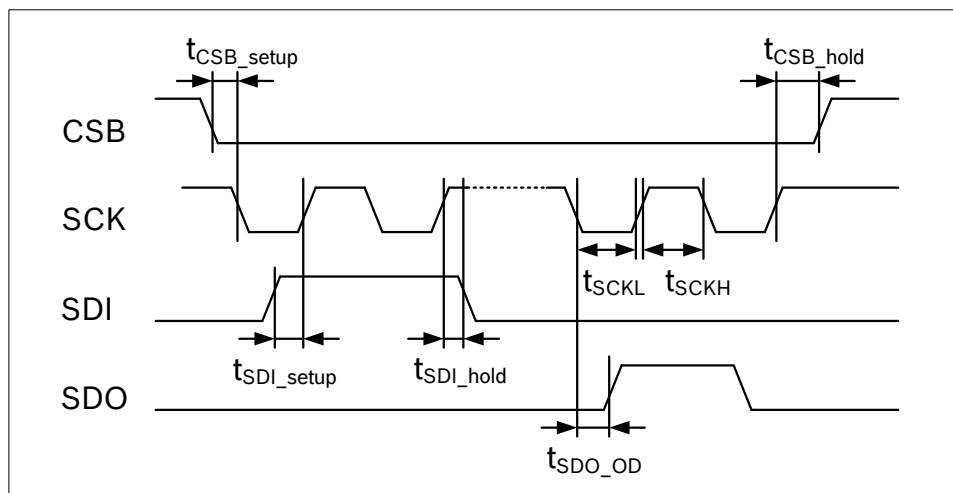
## 6.1 Serial peripheral interface (SPI)

The timing specification for SPI of the BMA250 is given in the following table:

**Table 73: SPI timing**

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	$f_{\text{SPI}}$	Max. Load on SDI or SDO = 25pF		10	MHz
SCK Low Pulse	$t_{\text{SCKL}}$		20		ns
SCK High Pulse	$t_{\text{SCKH}}$		20		ns
SDI Setup Time	$t_{\text{SDI\_setup}}$		20		ns
SDI Hold Time	$t_{\text{SDI\_hold}}$		20		ns
SDO Output Delay	$t_{\text{SDO\_OD}}$	Load = 25pF		30	ns
		Load = 250pF, $V_{\text{DDIO}} = 2.4\text{V}$		40	ns
CSB Setup Time	$t_{\text{CSB\_setup}}$		20		ns
CSB Hold Time	$t_{\text{CSB\_hold}}$		40		ns

The following figure shows the definition of the SPI timings given in table 73:



**Figure 10: SPI timing diagram**

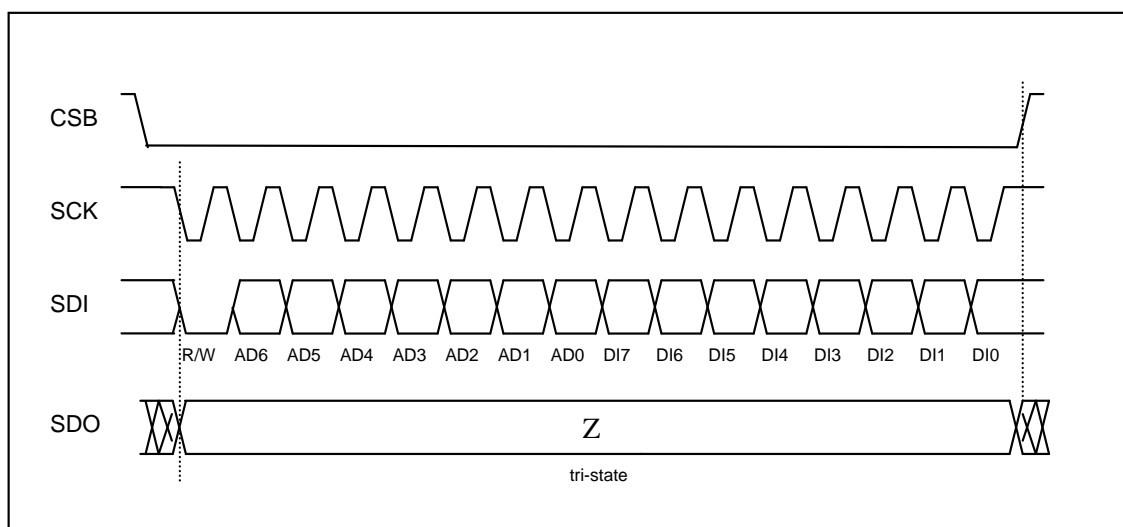
The SPI interface of the BMA250 is compatible with two modes, '00' and '11'. The automatic selection between [CPOL = '0' and CPHA = '0'] and [CPOL = '1' and CPHA = '1'] is done based on the value of SCK after a falling edge of CSB.

Two configurations of the SPI interface are supported by the BMA250: 4-wire and 3-wire. The same protocol is used by both configurations. The device operates in 4-wire configuration by default. It can be switched to 3-wire configuration by writing '1' to (0x34) *spi3*. Pin SDI is used as the common data pin in 3-wire configuration.

For single byte read as well as write operations, 16-bit protocols are used. The BMA250 also supports multiple-byte read operations.

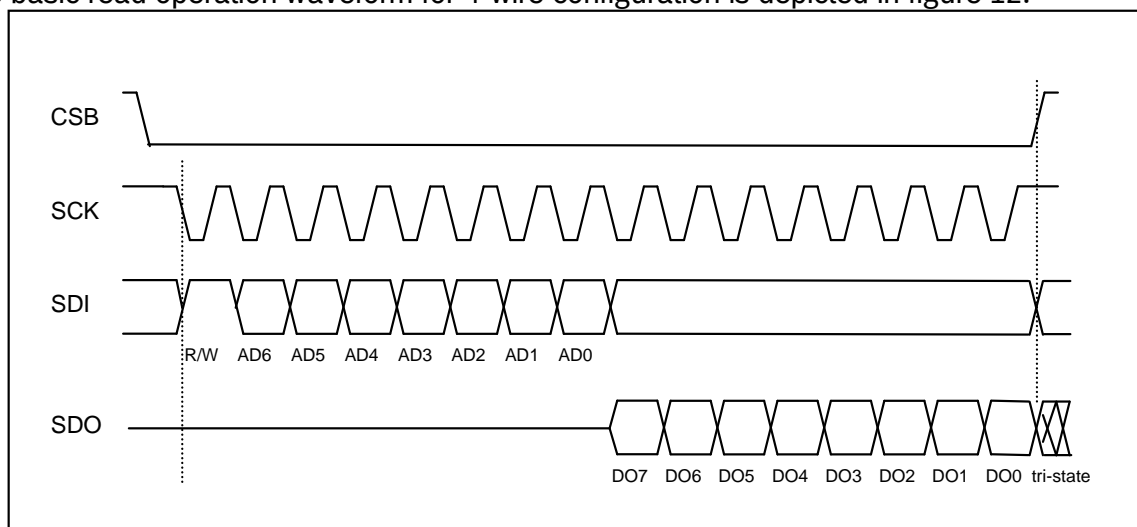
**In SPI 4-wire configuration** CSB (chip select low active), SCK (serial clock), SDI (serial data input), and SDO (serial data output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI and SDO are driven at the falling edge of SCK and should be captured at the rising edge of SCK.

The basic write operation waveform for 4-wire configuration is depicted in figure 11. During the entire write cycle SDO remains in high-impedance state.



**Figure 11: 4-wire basic SPI write sequence (mode '11')**

The basic read operation waveform for 4-wire configuration is depicted in figure 12:



**Figure 12: 4-wire basic SPI read sequence (mode '11')**



The data bits are used as follows:

**Bit0:** Read/Write bit. When 0, the data SDI is written into the chip. When 1, the data SDO from the chip is read.

**Bit1-7:** Address AD(6:0).

**Bit8-15:** when in write mode, these are the data SDI, which will be written into the address. When in read mode, these are the data SDO, which are read from the address.

Multiple read operations are possible by keeping CSB low and continuing the data transfer. Only the first register address has to be written. Addresses are automatically incremented after each read access as long as CSB stays active low.

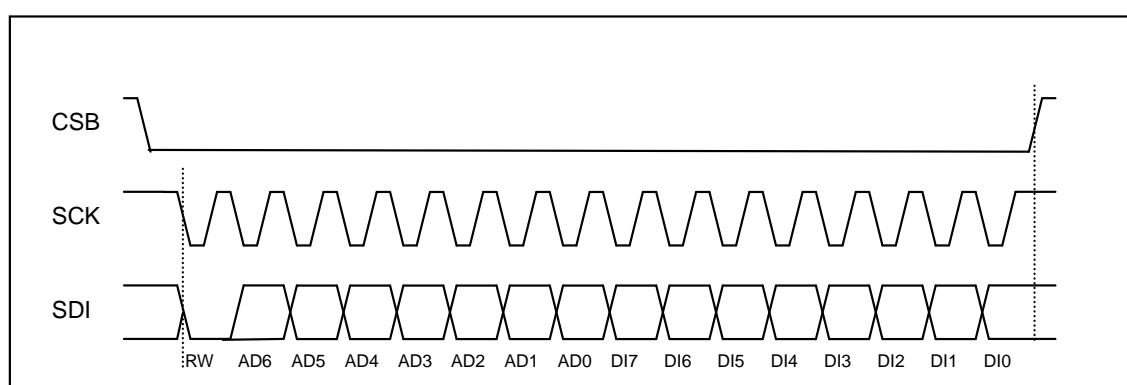
The principle of multiple read is shown in figure 13:

		Control byte								Data byte								Data byte								Data byte																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																									
Start	RW	Register address (02h)								Data register - adress 02h								Data register - adress 03h								Data register - adress 04h								Stop																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																																	
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**Figure 13: SPI multiple read**

**In SPI 3-wire configuration** CSB (chip select low active), SCK (serial clock), and SDI (serial data input and output) pins are used. The communication starts when the CSB is pulled low by the SPI master and stops when CSB is pulled high. SCK is also controlled by SPI master. SDI is driven (when used as input of the device) at the falling edge of SCK and should be captured (when used as the output of the device) at the rising edge of SCK.

The protocol as such is the same in 3-wire configuration as it is in 4-wire configuration. The basic operation waveform (read or write access) for 3-wire configuration is depicted in figure 14:



**Figure 14: 3-wire basic SPI read or write sequence (mode '11')**

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## 6.2 Inter-Integrated Circuit (I<sup>2</sup>C)

The I<sup>2</sup>C bus uses SCL (= SCx pin, serial clock) and SDA (= SDx pin, serial data input and output) signal lines. Both lines are connected to V<sub>DDIO</sub> externally via pull-up resistors so that they are pulled high when the bus is free.

The I<sup>2</sup>C interface of the BMA250 is compatible with the I<sup>2</sup>C Specification UM10204 Rev. 03 (19 June 2007), available at <http://www.nxp.com>. The BMA250 supports I<sup>2</sup>C standard mode and fast mode, only 7-bit address mode is supported. For V<sub>DDIO</sub> = 1.2V to 1.8V the guaranteed voltage output levels are slightly relaxed as described in the Parameter Specification (table 1).

The default I<sup>2</sup>C address of the device is 0011000b (0x18). It is used if the SDO pin is pulled to 'GND'. The alternative address 0011001b (0x19) is selected by pulling the SDO pin to 'V<sub>DDIO</sub>'.

The timing specification for I<sup>2</sup>C of the BMA250 is given in table 74:

**Table 74: I<sup>2</sup>C timings**

Parameter	Symbol	Condition	Min	Max	Units
Clock Frequency	f <sub>SCL</sub>			400	kHz
SCL Low Period	t <sub>LOW</sub>		1.3		μs
SCL High Period	t <sub>HIGH</sub>		0.6		
SDA Setup Time	t <sub>SUDAT</sub>		0.1		
SDA Hold Time	t <sub>HDDAT</sub>		0.0		
Setup Time for a repeated Start Condition	t <sub>SUSTA</sub>		0.6		
Hold Time for a Start Condition	t <sub>HDSTA</sub>		0.6		
Setup Time for a Stop Condition	t <sub>SUSTO</sub>		0.6		
Time before a new Transmission can start	t <sub>BUF</sub>		1.3		

The diagram illustrates the timing parameters for the I2C protocol across three signal lines: SDA (top), SCL (middle), and a second SDA line (bottom). The signals are shown as digital waveforms with various transitions and levels. Vertical dashed lines mark the boundaries of different timing intervals. The parameters shown are:

- $t_{BUF}$ : Buffer time for the first SDA signal transition.
- $t_{HDSTA}$ : Hold time for the first SDA signal after the SCL signal goes low.
- $t_{LOW}$ : Low pulse width for the SCL signal.
- $t_r$ : Rise time for the SCL signal.
- $t_{HDDAT}$ : Hold time for the SCL signal after the SDA signal goes high.
- $t_{HIGH}$ : High pulse width for the SCL signal.
- $t_{SUDAT}$ : Setup time for the SCL signal before the SDA signal goes high.
- $t_{SUSTA}$ : Setup time for the second SDA signal before the SCL signal goes low.
- $t_{SUSTO}$ : Setup time for the second SDA signal after the SCL signal goes high.

The I<sup>2</sup>C protocol works as follows:

**START:** Data transmission on the bus begins with a high to low transition on the SDA line while SCL is held high (start condition (S) indicated by I<sup>2</sup>C bus master). Once the START signal is transferred by the master, the bus is considered busy.

**STOP:** Each data transfer should be terminated by a Stop signal (P) generated by master. The STOP condition is a low to HIGH transition on SDA line while SCL is held high.

**ACK:** Each byte of data transferred must be acknowledged. It is indicated by an acknowledge bit sent by the receiver. The transmitter must release the SDA line (no pull down) during the acknowledge pulse while the receiver must then pull the SDA line low so that it remains stable low during the high period of the acknowledge clock cycle.

In the following diagrams these abbreviations are used:

S	Start
P	Stop
ACKS	Acknowledge by slave
ACKM	Acknowledge by master
NACKM	Not acknowledge by master
RW	Read / Write

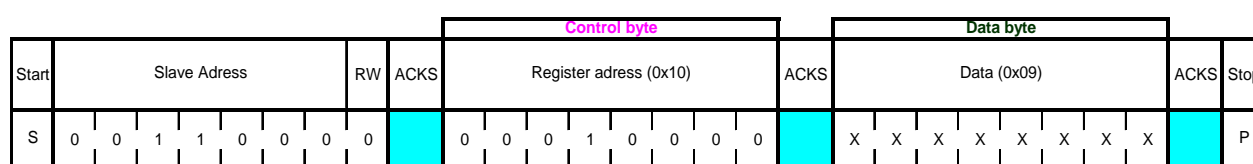
A START immediately followed by a STOP (without SCK toggling from logic “1” to logic “0”) is not supported. If such a combination occurs, the STOP is not recognized by the device.

### I<sup>2</sup>C write access:

I<sup>2</sup>C write access can be used to write a data byte in one sequence.

The sequence begins with start condition generated by the master, followed by 7 bits slave address and a write bit (RW = 0). The slave sends an acknowledge bit (ACK = 0) and releases the bus. Then the master sends the one byte register address. The slave again acknowledges the transmission and waits for the 8 bits of data which shall be written to the specified register address. After the slave acknowledges the data byte, the master generates a stop signal and terminates the writing protocol.

Example of an I<sup>2</sup>C write access:



**Figure 16: I<sup>2</sup>C write**

### I<sup>2</sup>C read access:

I<sup>2</sup>C read access also can be used to read one or multiple data bytes in one sequence.

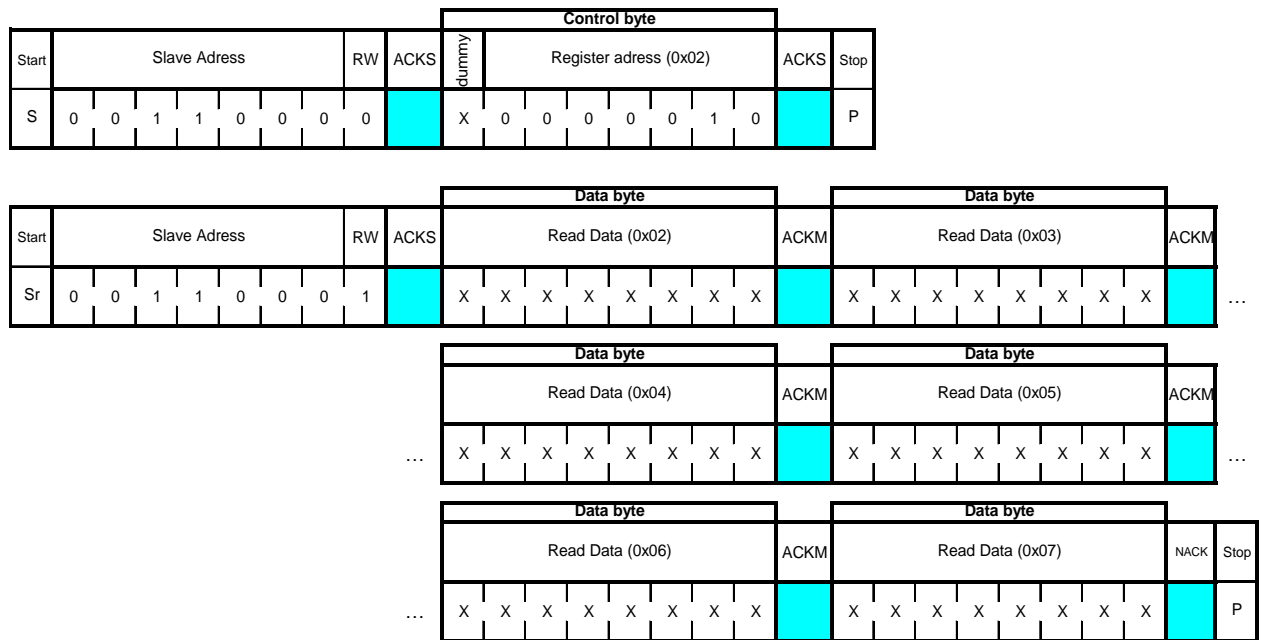
A read sequence consists of a one-byte I<sup>2</sup>C write phase followed by the I<sup>2</sup>C read phase. The two parts of the transmission must be separated by a repeated start condition (Sr). The I<sup>2</sup>C write phase addresses the slave and sends the register address to be read. After slave acknowledges the transmission, the master generates again a start condition and sends the slave address together with a read bit (RW = 1). Then the master releases the bus and waits for the data bytes to be read out from slave. After each data byte the master has to generate an acknowledge bit (ACK = 0) to enable further data transfer. A NACKM (ACK = 1) from the master stops the data being transferred from the slave. The slave releases the bus so that the master can generate a STOP condition and terminate the transmission.

The register address is automatically incremented and, therefore, more than one byte can be sequentially read out. Once a new data read transmission starts, the start address will be set to the register address specified in the latest I<sup>2</sup>C write command. By default the start address is set at 0x00. In this way repetitive multi-bytes reads from the same starting address are possible.


In order to prevent the I<sup>2</sup>C slave of the device to lock-up the I<sup>2</sup>C bus, a watchdog timer (WDT) is implemented. The WDT observes internal I<sup>2</sup>C signals and resets the I<sup>2</sup>C interface if the bus is locked-up by the BMA250. The activity and the timer period of the WDT can be configured through the bits (0x34) *i2c\_wdt\_en* and (0x34) *i2c\_wdt\_sel*.

Writing '1' ('0') to (0x34) *i2c\_wdt\_en* activates (de-activates) the WDT. Writing '0' ('1') to (0x34) *i2c\_wdt\_se* selects a timer period of 1 ms (50 ms).

Example of an I<sup>2</sup>C read access:



**Figure 17: I<sup>2</sup>C multiple read**

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## 7. Pin-out and connection diagram

### 7.1 Pin-out

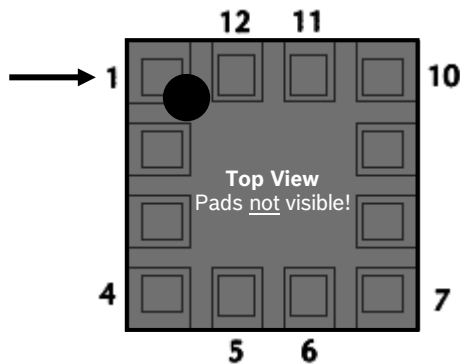


Figure 18: Pin-out top view

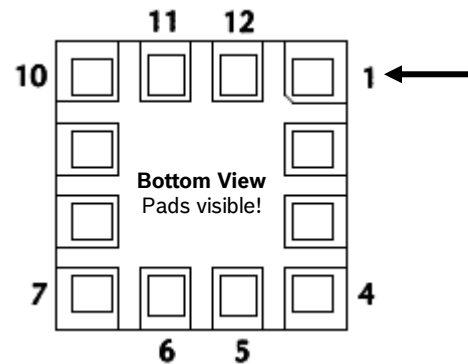



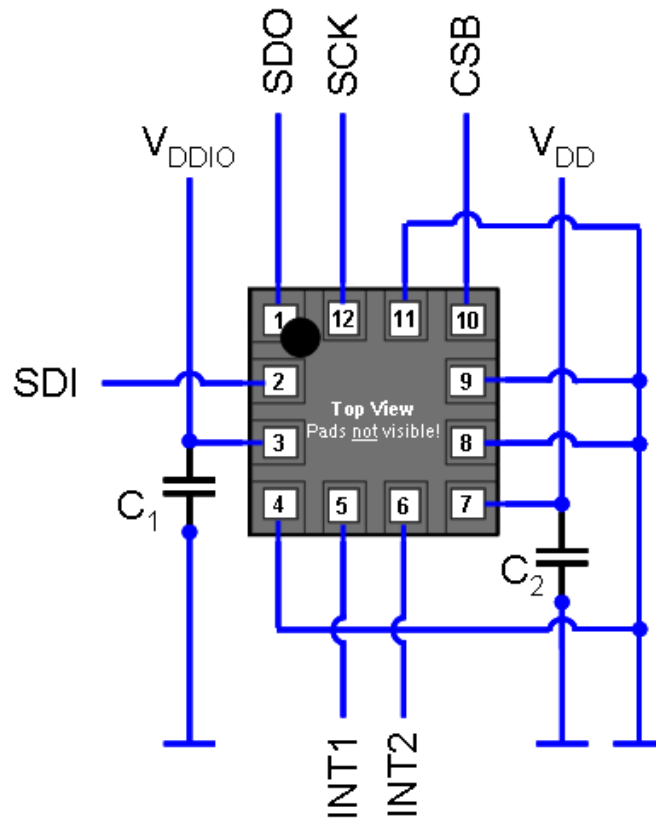
Figure 19 Pin-out bottom view

Table 75: Pin description


Pin#	Name	I/O Type	Description	Connect to		
				in SPI 4W	In SPI 3W	in I <sup>2</sup> C
1	SDO	Digital out	Serial data output in SPI Address select in I <sup>2</sup> C mode see chapter 6.2	SDO	DNC (float)	GND for default addr.
2	SDx	Digital I/O	SDA serial data I/O in I <sup>2</sup> C SDI serial data input in SPI 4W SDA serial data I/O in SPI 3W	SDI	SDA	SDA
3	VDDIO	Supply	Digital I/O supply voltage (1.2V ... 3.6V)	V <sub>DDIO</sub>	V <sub>DDIO</sub>	V <sub>DDIO</sub>
4	NC	--		GND	GND	GND
5	INT1	Digital out	Interrupt output 1	INT1	INT1	INT1
6	INT2	Digital out	Interrupt output 2	INT2	INT2	INT2
7	VDD	Supply	Power supply for analog & digital domain (1.62V ... 3.6V)	V <sub>DD</sub>	V <sub>DD</sub>	V <sub>DD</sub>
8	GNDIO	Ground	Ground for I/O	GND	GND	GND
9	GND	Ground	Ground for digital & analog	GND	GND	GND
10	CSB	Digital in	Chip select for SPI mode	CSB	CSB	DNC (float)
11	PS	Digital in	Protocol select (GND = SPI, V <sub>DDIO</sub> = I <sup>2</sup> C, float = $\mu$ C-less). Pin must not float unless dedicated mode is used, see chapter 4.2.2	GND	GND	V <sub>DDIO</sub>
12	SCx	Digital in	SCK for SPI serial clock SCL for I <sup>2</sup> C serial clock	SCK	SCK	SCL

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## 7.2 Connection diagram 4-wire SPI



**Figure 20: 4-wire SPI connection**

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### 7.3 Connection diagram 3-wire SPI

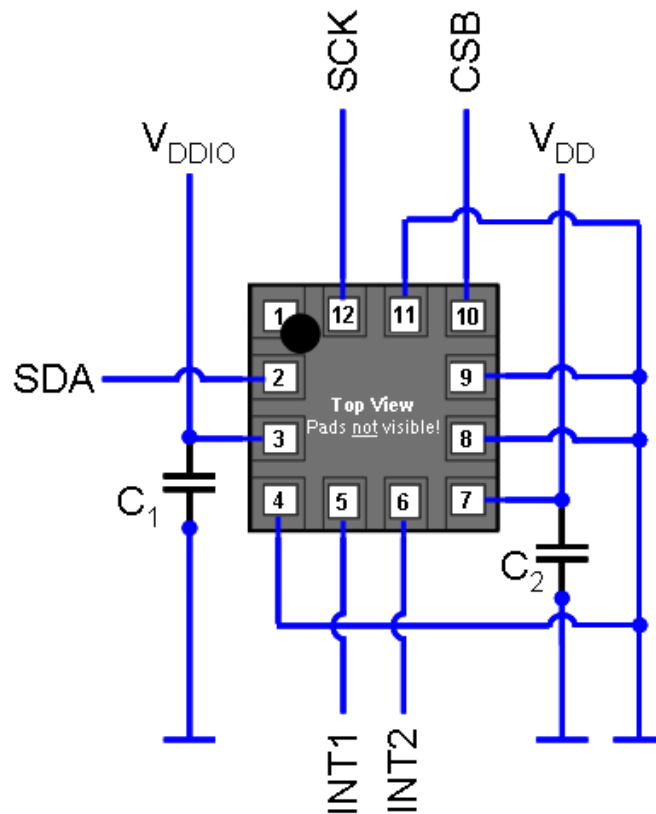

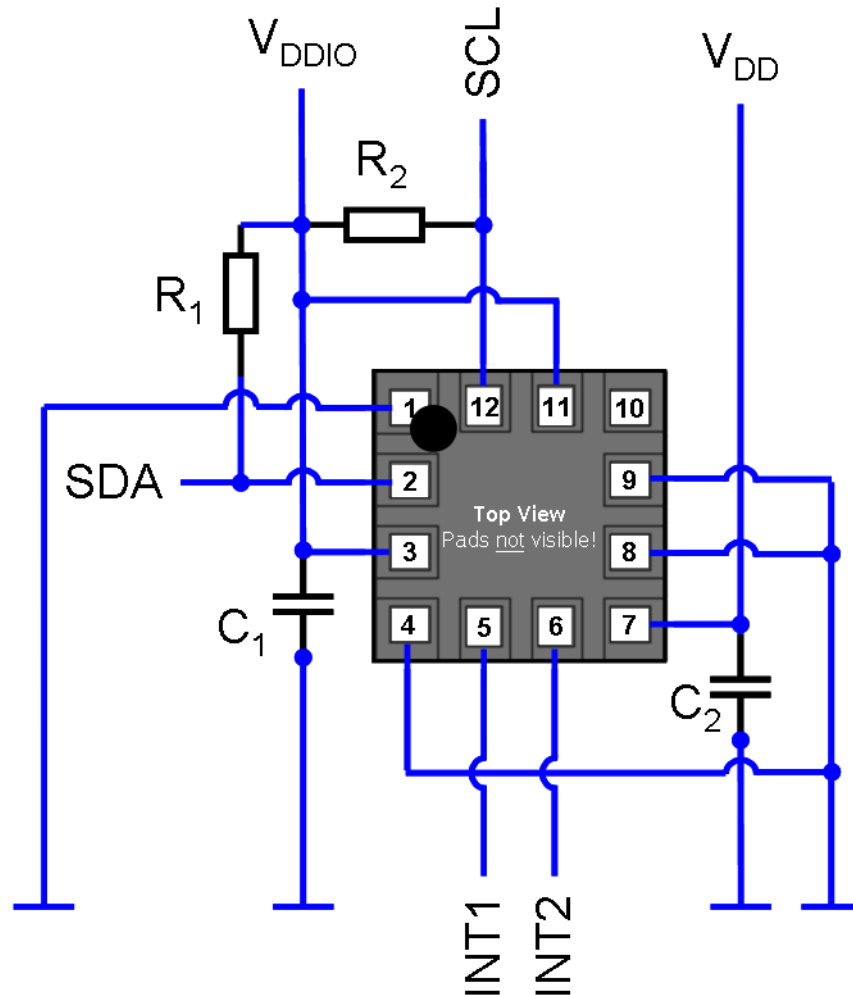


Figure 21: 3-wire SPI connection




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#### 7.4 Connection diagram I<sup>2</sup>C



**Figure 22: I<sup>2</sup>C connection**

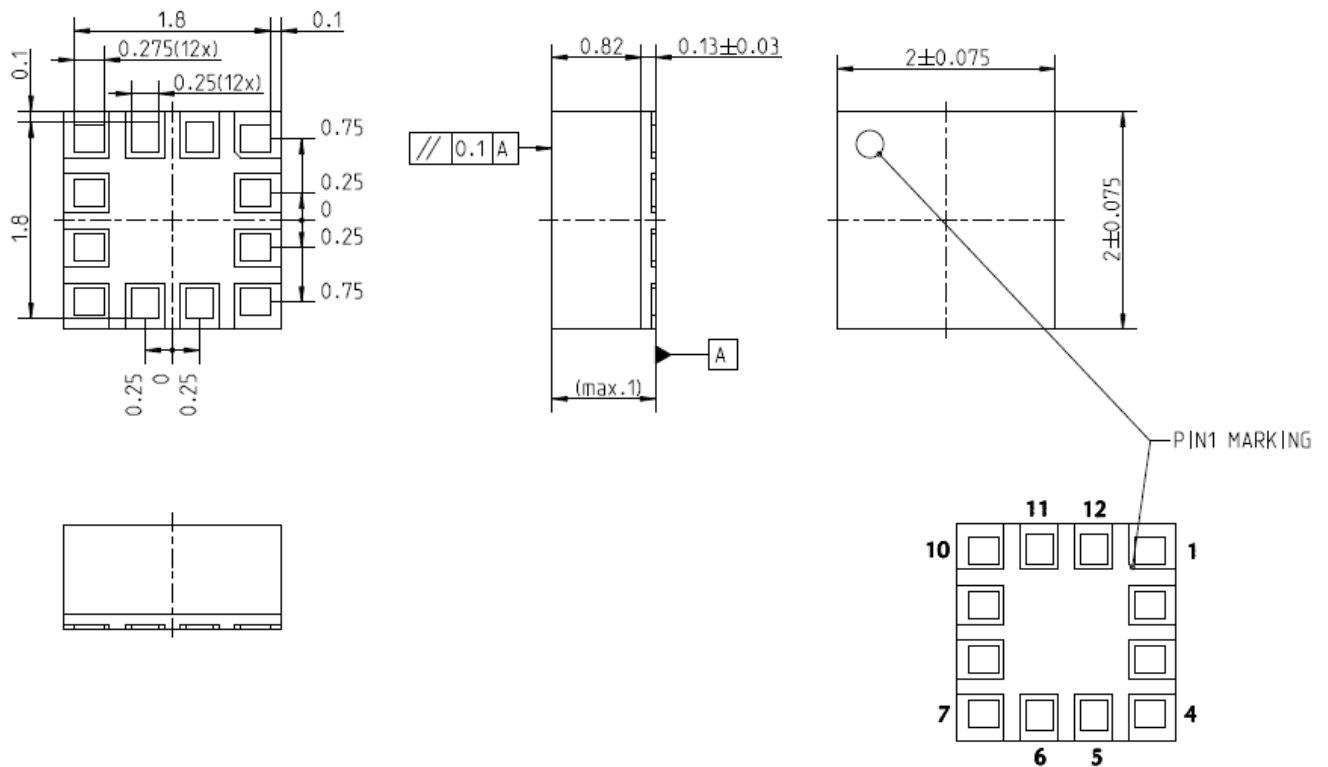
Note: the recommended value for C<sub>1</sub>, C<sub>2</sub> is 100 nF.

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## 8. Package

### 8.1 Outline dimensions

The sensor housing is a standard LGA package. It is compliant with JEDEC Standard MO-229 Type VGGD-3. Its dimensions are the following.



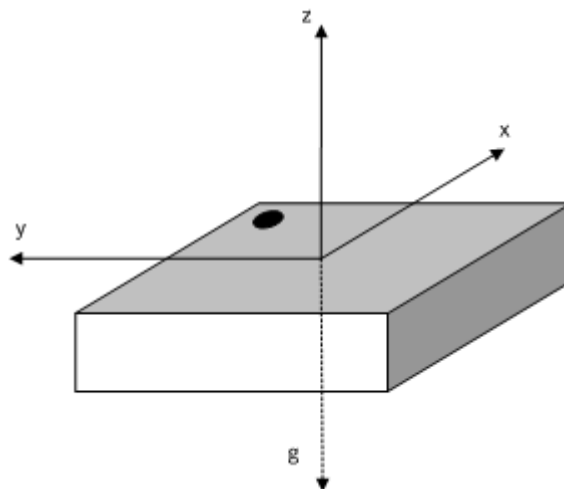
**Figure 23: Package outline dimensions**

## 8.2 Sensing axes orientation

If the sensor is accelerated in the indicated directions, the corresponding channel will deliver a positive acceleration signal (dynamic acceleration). If the sensor is at rest and the force of gravity is acting along the indicated directions, the output of the corresponding channel will be negative (static acceleration).

Example: If the sensor is at rest or at uniform motion in a gravity field according to the figure given below, the output signals are:

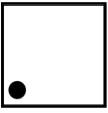
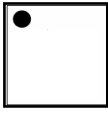
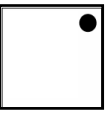
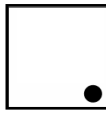
- $\pm 0g$  for the X channel
- $\pm 0g$  for the Y channel
- $+ 1g$  for the Z channel




**Figure 24: Orientation of sensing axis**

The following table lists all corresponding output signals on X, Y, and Z while the sensor is at rest or at uniform motion in a gravity field under assumption of a  $\pm 2g$  range setting and a top down gravity vector as shown above.

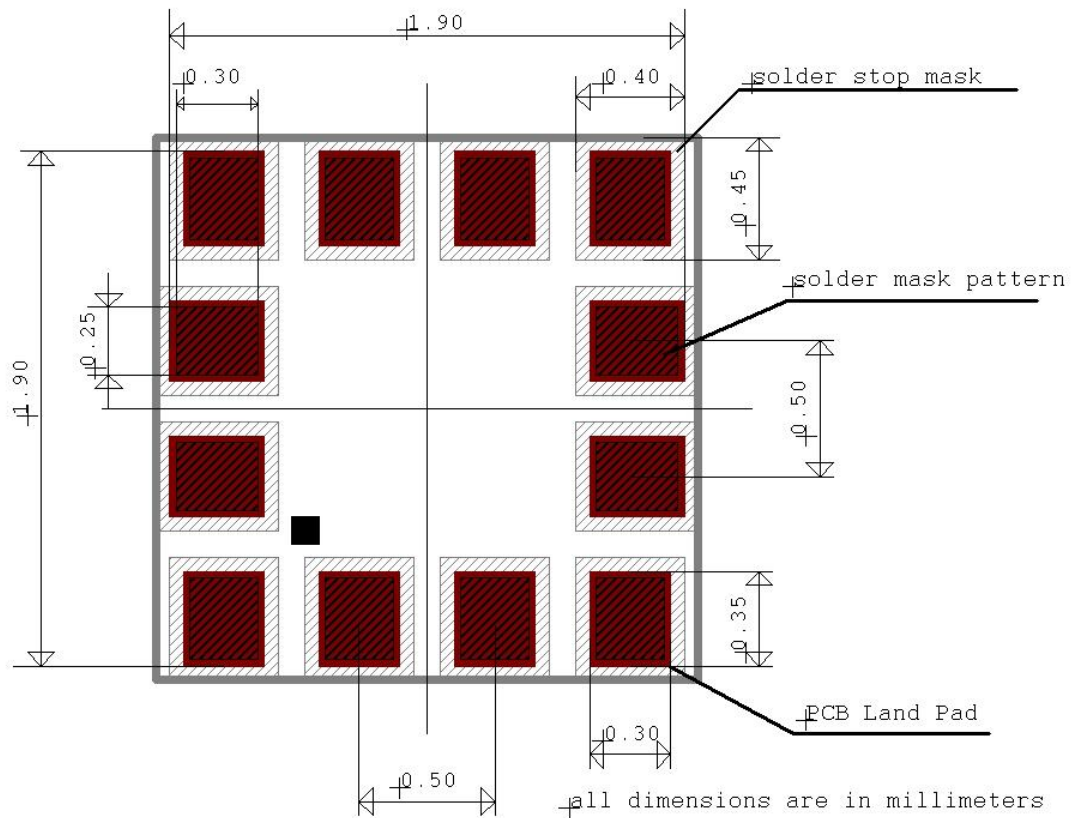
**Table 76: Output signals depending on sensor orientation**

Sensor Orientation (gravity vector ↓)					upright	inverted
Output Signal X	0g / 0LSB	1g / 256LSB	0g / 0LSB	-1g / -256LSB	0g / 0LSB	0g / 0LSB
Output Signal Y	-1g / -256LSB	0g / 0LSB	+1g / 256LSB	0g / 0LSB	0g / 0LSB	0g / 0LSB
Output Signal Z	0g / 0LSB	0g / 0LSB	0g / 0LSB	0g / 0LSB	1g / 256LSB	-1g / -256LSB


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### 8.3 Landing pattern recommendation

For the design of the landing patterns, we recommend the following dimensioning:



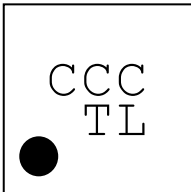
**Figure 25: Landing patterns relative to the device pins, dimensions are in mm**

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## 8.4 Marking

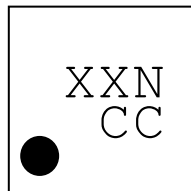
### 8.4.1 Mass production samples

**Table 77: Marking of mass production samples**

Labeling	Name	Symbol	Remark
	Lot counter	CCC	3 alphanumeric digits, variable to generate mass production trace-code
	Product number	T	1 alphanumeric digit, fixed to identify product type, T = "8"
	Sub-con ID	L	1 alphanumeric digit, variable to identify sub-con (L = "A" or L = "U" or L = "P")
	Pin 1 identifier	●	--

### 8.4.2 Engineering samples

**Table 78: Marking of engineering samples**

Labeling	Name	Symbol	Remark
	Eng. sample ID	N	1 alphanumeric digit, fixed to identify engineering sample, N = "e"
	Sample ID	XX	2 alphanumeric digits, variable to generate trace-code
	Counter ID	CC	2 alphanumeric digits, variable to generate trace-code
	Pin 1 identifier	●	--

## 8.5 Soldering guidelines

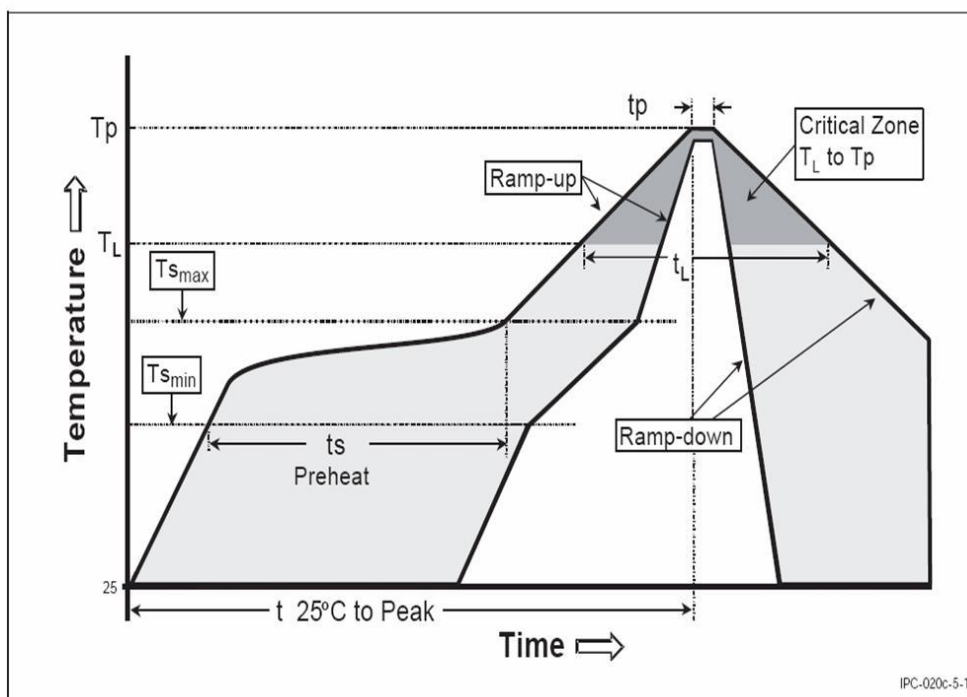
The moisture sensitivity level of the BMA250 sensors corresponds to JEDEC Level 1, see also

- IPC/JEDEC J-STD-020C "Joint Industry Standard: Moisture/Reflow Sensitivity Classification for non-hermetic Solid State Surface Mount Devices"
- IPC/JEDEC J-STD-033A "Joint Industry Standard: Handling, Packing, Shipping and Use of Moisture/Reflow Sensitive Surface Mount Devices".

The sensor fulfils the lead-free soldering requirements of the above-mentioned IPC/JEDEC standard, i.e. reflow soldering with a peak temperature up to 260°C.


Profile Feature		Pb-Free Assembly
Average Ramp-Up Rate ( $T_{s_{max}}$ to $T_p$ )		3° C/second max.
<b>Preheat</b> <ul style="list-style-type: none"> <li>– Temperature Min (<math>T_{s_{min}}</math>)</li> <li>– Temperature Max (<math>T_{s_{max}}</math>)</li> <li>– Time (<math>t_{s_{min}}</math> to <math>t_{s_{max}}</math>)</li> </ul>		150 °C 200 °C 60-180 seconds
Time maintained above: <ul style="list-style-type: none"> <li>– Temperature (<math>T_L</math>)</li> <li>– Time (<math>t_L</math>)</li> </ul>		217 °C 60-150 seconds
Peak/Classification Temperature ( $T_p$ )		260 °C
Time within 5 °C of actual Peak Temperature ( $t_p$ )		20-40 seconds
Ramp-Down Rate		6 °C/second max.
Time 25 °C to Peak Temperature		8 minutes max.

**Note 1:** All temperatures refer to topside of the package, measured on the package body surface.

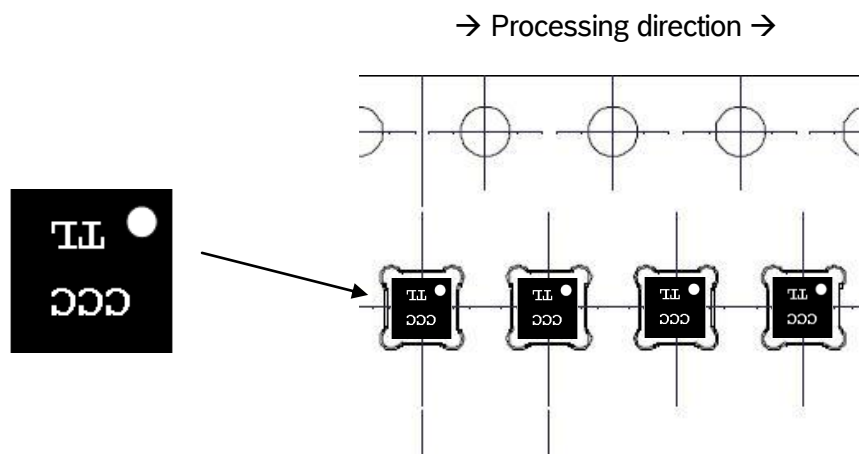


**Figure 26: Soldering profile**



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### 8.7.1 Orientation within the reel



**Figure 28: Orientation of the BMA250 devices relative to the tape**

## 8.8 Environmental safety

The BMA250 sensor meets the requirements of the EC restriction of hazardous substances (RoHS) directive, see also:

*Directive 2002/95/EC of the European Parliament and of the Council of 27 January 2003 on the restriction of the use of certain hazardous substances in electrical and electronic equipment.*

### 8.8.1 Halogen content

Results of chemical analysis indicate that the BMA250 contains less than 900ppm (by weight) of Fluorine, Chlorine, Iodine and Bromine (i.e. < 50ppm per each substance). Therefore the BMA250 can be regarded as halogen-free. For more details on the analysis results please contact your Bosch Sensortec representative.


### 8.8.2 Internal package structure

Within the scope of Bosch Sensortec's ambition to improve its products and secure the mass product supply, Bosch Sensortec qualifies additional sources (e.g. 2<sup>nd</sup> source) for the LGA package of the BMA250.

While Bosch Sensortec took care that all of the technical packages parameters are described above are 100% identical for all sources, there can be differences in the chemical content and the internal structural between the different package sources.

However, as secured by the extensive product qualification process of Bosch Sensortec, this has no impact to the usage or to the quality of the BMA250 product.



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## 9. Legal disclaimer

### 9.1 Engineering samples

Engineering Samples are marked with “e”. Samples may vary from the valid technical specifications of the product series contained in this data sheet. They are therefore not intended or fit for resale to third parties or for use in end products. Their sole purpose is internal client testing. The testing of an engineering sample may in no way replace the testing of a product series. Bosch Sensortec assumes no liability for the use of engineering samples. The Purchaser shall indemnify Bosch Sensortec from all claims arising from the use of engineering samples.

### 9.2 Product use

Bosch Sensortec products are developed for the consumer goods industry. They may only be used within the parameters of this product data sheet. They are not fit for use in life-sustaining or security sensitive systems. Security sensitive systems are those for which a malfunction is expected to lead to bodily harm or significant property damage. In addition, they are not fit for use in products which interact with motor vehicle systems.


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The purchaser must monitor the market for the purchased products, particularly with regard to product safety, and inform Bosch Sensortec without delay of all security relevant incidents.

### 9.3 Application examples and hints

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## 10. Document history and modification

Revision	Chapter	Description of modification/changes	Date
0.8		Document release	17 December 2010
0.9	1	Update table 1	26 January 2011
	4.2.2	Added missing table numbers	
	4.3	Update table 7	
	4.4.1	Update range register 0x0F	
1.0		Document rev. 1.0 update / no changes	03 March 2011
1.05	4.8.3	Typo correction, int1_od, int2_od	17 June 2011
	5.11	Typo correction, register 0x25	
	5.11	Typo correction in table 53 description	
1.10	5.2	Typo correction register map	02 November 2011
1.15	4.8.7	Update orientation interrupt	31 May 2012
	6.2	Update I2C address selection	

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