HSSR-7110, HSSR-7111, HSSR-7112, HSSR-711E 5962-93140



90 V/1.0 Ω, Hermetically Sealed, Power MOSFET Optocoupler

Data Sheet

Description

The HSSR-7110, HSSR-7111, HSSR-7112, HSSR-711E and SMD 5962-93140 are single channel power MOSFET optocouplers, constructed in eight-pin, hermetic, dual-in-line, ceramic packages. The devices operate exactly like a solidstate relay.

The products are capable of operation and storage over the full military temperature range and may be purchased as a standard product (HSSR-7110), with full MIL-PRF-38534 Class H testing (HSSR-7111 and HSSR- 7112), with MIL-PRF-38534 Class E testing (Class K with exceptions) (HSSR-711E) or from the DLA Standard Microcircuit Drawing (SMD) 5962-93140. Details of the Class E program may be found on page 11 of this datasheet.

Functional Diagrams



TRUTH TABLE				
INPUT	OUTPUT			
н	CLOSED			
L	OPEN			

Features

- Dual Marked with Device Part Number and DLA Standard Microcircuit Drawing
- ac/dc Signal & Power Switching
- Compact Solid-State Bidirectional Switch
- Manufactured and Tested on a MIL-PRF-38534Certified
 Line
- QML-38534
- MIL-PRF-38534 Class H
- Modified Space Level Processing Available (Class E)
- Hermetically Sealed 8-Pin Dual In-Line Package
- Small Size and Weight
- Performance Guaranteed over -55°C to 125°C
- Connection A0.8 A, 1.0Ω
- Connection B1.6 A, 0.25 Ω
- 1500 Vdc Withstand Test Voltage
- High Transient Immunity
- 5 Amp Output Surge Current

Applications

- Military and Space
- High Reliability Systems
- Standard 28 Vdc and 48 Vdc Load Driver
- Standard 24 Vac Load Driver
- Aircraft Controls
- ac/dc Electromechanical and Solid State Relay Replacement
- I/O Modules
- Harsh Industrial Environments

CAUTION: It is advised that normal static precautions be taken in handling and assembly of this component to prevent damage and/or degradation which may be induced by ESD.

All devices are manufactured and tested on a MIL-PRF-38534 certified line and are included in the DLA Qualified Manufacturers List, QML-38534 for Hybrid Microcircuits. Each device contains an AlGaAs light emitting diode optically coupled to a photovoltaic diode stack which drives two discrete power MOSFETs. The device operates as a solid-state replacement for single-pole, normally open, (1 Form A) relay used for general purpose switching of signals and loads in high reliability applications.

The devices feature logic level input control and very low output on-resistance, making them suitable for both ac and dc loads. Connection A, as shown in the Functional Diagram, allows the device to switch either ac or dc loads. Connection B, with the polarity and pin configuration as shown, allows the device to switch dc loads only. The advantage of Connection B is that the on-resistance is significantly reduced, and the output current capability increases by a factor of two. The devices are convenient replacements for mechanical and solid state relays where high component reliability with standard footprint lead configuration is desirable. Devices may be purchased with a variety of lead bend and plating options. See Selection Guide table for details. Standard Microcircuit Drawing (SMD) parts are available for each package and lead style.

The HSSR-7110, HSSR-7111, HSSR-7112, HSSR-711E and SMD 5962-93140 are designed to switch loads on 28 Vdc power systems. They meet 80 V surge and \pm 600 V spike requirements.

Avago Technologies's Part Number and	Options		
Commercial	HSSR-7110		
MIL-PRF-38534 Class H	HSSR-7111	HSSR-7112	
MIL-PRF-38534 Class E			HSSR-711E
Standard Lead Finish	Gold Plate	Gold Plate	Gold Plate
Solder Dipped*	Option #200	Option -200	Option -200
Butt Joint/Gold Plate	Option #100	Option -100	
Gull Wing/Soldered*	Option #300	Option -300	
Crew Cut/Gold Plate	Option #600		
SMD Part Number			
Prescript for all below	5962- 5962-		
Gold Plate	9314001HPC	9314002HPC	9314001EPC
Solder Dipped*	9314001HPA	9314002HPA	9314001EPA
Butt Joint/Gold Plate	9314001HYC	9314002HYC	
Butt Joint/Soldered*	9314001HYA	9314002HYA	
Gull Wing/Soldered*	9314001HXA	9314002HXA	
Crew Cut/Gold Plate	9314001HZC		
Crew Cut/Soldered*	9314001HZA		

Selection Guide – Lead Configuration Options

* Solder Contains Lead

CAUTION: Maximum Switching Frequency – Care should be taken during repetitive switching of loads so as not to exceed the maximum output current, maximum output power dissipation, maximum case temperature, and maximum junction temperature.

Outline Drawing

8-pin DIP Through Hole



Device Marking



Thermal Resistance

Maximum Output MOSFET Junction to Case – θJC = 15°C/W

ESD Classification

(MIL-STD-883, Method 3015) (AA), Class 2

Absolute Maximum Ratings

Parameter	Symbol	Min.	Max.	Units	Note
Storage Temperature Range	Ts	-65°	+150°	C	
Operating Ambient Temperature	TA	-55°	+125°	C	
Junction Temperature	Тј		+150°	C	
Operating Case Temperature	Tc		+145°	C	1
Lead Solder Temperature (1.6 mm below seating plane)			260° for 10 s	C	
Average Input Current	IF		20	mA	
Peak Repetitive Input Current (Pulse Width < 100 ms; duty cycle < 50%)	Ігрк		40	mA	
Peak Surge Input Current (Pulse Width < 0.2 ms; duty cycle < 0.1%)	IFPK surge		100	mA	
Reverse Input Voltage	V _R		5	٧	
Average Output Current - Figure 2 Connection A Connection B	I ₀		0.8 1.6	A A	
Single Shot Output Current - Figure 3 Connection A (Pulse width < 10 ms) Connection B (Pulse width < 10 ms)	I _{OPK surge}		5.0 10.0	A A	
Output Voltage Connection A Connection B	V ₀	-90 0	90 90	V V	
Average Output Power Dissipation - Figure 4			800	mW	2

Recommended Operating Conditions

Parameter	Symbol	Min.	Max.	Units	Note
Input Current (on)	I _{F(ON)}	5	20	mA	10
Input Current (on)	I _{F(ON)}	10	20	mA	11
Input Voltage (off)	V _{F(OFF)}	0	0.6	V	
Operating Temperature	T _A	-55	+125	°C	

Hermetic Optocoupler Options

Note: Dimensions in millimeters (inches).



Note: Solder contains lead.

Electrical Specifications

Parameter	Sym.	Group A, Sub-group	Test Conditions	Min.	Typ.*	Max.	Units	Fig.	Notes
Output Withstand Voltage	V _{O(OFF)}	1, 2, 3	$V_F = 0.6 V$, $I_0 = 10 \mu A$	90	110		V	5	
Output On-Resistance R ₍ Connection A	R _(ON)	1, 2, 3	$I_F = 10 \text{ mA}, I_0 = 800 \text{ mA},$ (pulse duration $\leq 30 \text{ ms}$)		0.40	1.0	Ω	6, 7	3, 11
			$I_F = 5 \text{ mA}, I_0 = 800 \text{ mA},$ (pulse duration $\leq 30 \text{ ms}$)			1.0			3, 10
Output On-Resistance Connection B	R _(ON)	1, 2, 3	$I_F = 10 \text{ mA}, I_0 = 1.6 \text{ A},$ (pulse duration $\leq 30 \text{ ms}$)		0.12	0.25	Ω	6, 7	3, 11
			$I_F = 5 \text{ mA}, I_0 = 1.6 \text{ A},$ (pulse duration $\leq 30 \text{ ms}$)			0.25			3, 10
Output Leakage Current	I _{O(OFF)}	1, 2, 3	$V_{\rm F} = 0.6 \rm V, V_0 = 90 \rm V$		10-4	10	μA	8	
Input Forward	nput Forward V _F	1, 2, 3	$I_F = 10 \text{ mA}$	1.0	1.24	1.7	V	9	11
Voltage			$I_F = 5 \text{ mA}$						10
Input Reverse Breakdown Voltage	V _R	1, 2, 3	$I_R = 100 \ \mu A$	5.0			V		
Input-Output Insulation	I _{I-0}	1	$RH \le 65\%, t = 5 s,$ $V_{I-0} = 1500 Vdc, T_A = 25^{\circ}C$			1.0	μA		4, 5
Turn On Time	ırn On Time t _{ON}	t _{ON} 9, 10, 11	$I_F = 10 \text{ mA}, V_{DD} = 28 \text{ V},$ $I_0 = 800 \text{ mA}$		1.25	6.0		1, 10, 11, 12, 13	11
			$I_F = 5 \text{ mA}, V_{DD} = 28 \text{ V},$ $I_0 = 800 \text{ mA}$			6.0			10
Turn Off Time t _{OFF}	t _{OFF}	t _{OFF} 9, 10, 11	$I_F = 10 \text{ mA}, V_{DD} = 28 \text{ V},$ $I_0 = 800 \text{ mA}$		0.02	0.25	ms	1, 10, 14, 15	11
			$I_F = 5 \text{ mA}, V_{DD} = 28 \text{ V},$ $I_0 = 800 \text{ mA}$			0.25			10
Output Transient Rejection	dVo dt	9	$\label{eq:VPEAK} \begin{split} V_{PEAK} &= 50 \text{ V}, \text{C}_{M} = 1000 \text{pF}, \\ \text{C}_{L} &= 15 \text{pF}, \text{R}_{M} \geq 1 \text{M} \Omega \end{split}$	1000			V/µs	17	
Input-Output Transient Rejection	dVio dt	9	$V_{DD} = 5 V, V_{I-0(PEAK)} = 50 V,$ $R_L = 20 k\Omega, C_L = 15 pF$	500			V/µs	18	

Typical Characteristics

All typical values are at $T_A = 25^{\circ}$ C, $I_F(ON) = 10$ mA, $V_F(OFF) = 0.6$ V unless otherwise specified.

Parameter	Symbol	Test Conditions	Тур.	Units	Fig.	Notes
Output Off-Capacitance	C _{O(OFF)}	$V_0 = 28 V, f = 1 MHz$	145	pF	16	
Output Offset Voltage	V _{OS}	$I_F = 10 \text{ mA}, I_0 = 0 \text{ mA}$	2	μV	19	7
Input Diode Temperature Coefficient	$\Delta V_F / \Delta T_A$	$I_F = 10 \text{ mA}$	-1.4	mV/°C		
Input Capacitance	C _{IN}	$V_{\rm F} = 0 V, f = 1 M Hz$	20	pF		8
Input-Output Capacitance	C _{I-0}	$V_{I-0} = 0 V, f = 1 MHz$	1.5	pF		4
Input-Output Resistance	R _{I-0}	$V_{I-0} = 500 \text{ V}, t = 60 \text{ s}$	10 ¹³	Ω		4
Turn On Time With Peaking	t _{ON}	$I_{FPK} = 100 \text{ mA}, I_{FSS} = 10 \text{ mA}$ $V_{DD} = 28 \text{ V}, I_0 = 800 \text{ mA}$	0.22	ms	1	6

Notes:

1. Maximum junction to case thermal resistance for the device is 15°C/W, where case temperature, T_C, is measured at the center of the package bottom.

2. For rating, see Figure 4. The output power P_O rating curve is obtained when the part is handling the maximum average output current I_O as shown in Figure 2.

3. During the pulsed R_{ON} measurement (I_O duration <30 ms), ambient (T_A) and case temperature (T_C) are equal.

4. Device considered a two terminal device: pins 1 through 4 shorted together and pins 5 through 8 shorted together.

5. This is a momentary withstand test, not an operating condition.

6. For a faster turn-on time, the optional peaking circuit shown in Figure 1 may be implemented.

7. V_{OS} is a function of I_F, and is defined between pins 5 and 8, with pin 5 as the reference. V_{OS} must be measured in a stable ambient (free of temperature gradients).

8. Zero-bias capacitance measured between the LED anode and cathode.

9. Standard parts receive 100% testing at 25°C (Subgroups 1 and 9). SMD, Class H and Class E parts receive 100% testing at 25°C, 125°C and -55°C (Subgroups 1 and 9, 2 and 10, 3 and 11 respectively).

10. Applies to HSSR-7112 and 5962-9314002Hxx devices only.

11. Applies to HSSR-7110, HSSR-7111, HSSR-711E, 5962-9314001Hxx and 5962-9314001Exx devices only.



Figure 1. Recommended Input Circuit.





 P_0 - OUTPUT POWER DISSIPATION - W 0.8 0.6 0.4 **CONNECTION - A** $\frac{I_F \ge 10 \text{ mA}}{\Theta_{CA}} = 40^{\circ} \text{ C/W}$ 0.2 = 80° C/W - Ө**с**а 0 _ -55 -25 35 65 95 5 T_A - AMBIENT TEMPERATURE - °C

1.0

Figure 2. Maximum Average Output Current Rating vs. Ambient Temperature.









125

155



Figure 5. Normalized Typical Output Withstand Voltage vs. Temperature.



Figure 8. Typical Output Leakage Current vs. Temperature.

Figure 6. Normalized Typical Output Resistance vs. Temperature.

Figure 7. Typical On State Output I-V Characteristics.



Figure 9. Typical Input Forward Current vs. Input Forward Voltage.



Figure 10. Switching Test Circuit for t_{ON}, t_{OFF}.







CONNECTION A

 $V_{DD} = 28 V$ $I_0 = 800 mA$ $T_A = 25^{\circ}C$

3.0

2.6

2.2



Figure 11. Typical Turn On Time vs. Temperature.



Figure 13. Typical Turn On Time vs. Voltage.









rent.

Figure 16. Typical Output Off Capacitance vs. **Output Voltage.**









Figure 18. Input-Output Transient Rejection Test Circuit.



Figure 19. Voltage Offset Test Setup.



NOTE: IN ORDER TO DETERMINE V_{OUT} CORRECTLY, THE CASE TO AMBIENT THERMAL IMPEDANCE MUST BE MEASURED FOR THE BURN-IN BOARDS TO BE USED. THEN, KNOWING Θ_{ca} , DETERMINE THE CORRECT OUTPUT CURRENT PER FIGURES 2 AND 4 TO INSURE THAT THE DEVICE MEETS THE DEBATING REQUIREMENTS AS SHOWN

Figure 20. Burn-In Circuit.

Applications Information

Thermal Model

The steady state thermal model for the HSSR-7110 is shown in Figure 21. The thermal resistance values given in this model can be used to calculate the temperatures at each node for a given operating condition. The thermal resistances between the LED and other internal nodes are very large in comparison with the other terms and are omitted for simplicity. The components do, however, interact indirectly through θCA , the case-to-ambient thermal resistance. All heat generated flows through θ_{CA} which raises the case temperature TC accordingly. The value of θ CA depends on the conditions of the board design and is, therefore, determined by the designer.

The maximum value for each output MOSFET junctionto-case thermal resistance is specified as 15°C/W. The thermal resistance from FET driver junction-to-case is also 15°C/W/W. The power dissipation in the FET driver, however, is negligible in comparison to the MOSFETs.



- T_{je} = LED JUNCTION TEMPERATURE T_{if1} = FET 1 JUNCTION TEMPERATURE T_{if2} = FET 2 JUNCTION TEMPERATURE T_{id} = FET DRIVER JUNCTION TEMPERATURE = CASE TEMPERATURE (MEASURED AT CENTER Тc **OF PACKAGE BOTTOM)**
- T_A = AMBIENT TEMPERATURE (MEASURED 6" AWAY FROM THE PACKAGE)
- = CASE-TO-AMBIENT THERMAL RESISTANCE

ALL THERMAL RESISTANCE VALUES ARE IN °C/W

Figure 21. Thermal Model.

On-Resistance and Rating Curves

The output on-resistance, RON, specified in this data sheet, is the resistance measured across the output contact when a pulsed current signal ($I_0 = 800$ mA) is applied to the output pins. The use of a pulsed signal (\leq 30 ms) implies that each junction temperature is equal to the ambient and case temperatures. The steadystate resistance, R_{SS}, on the other hand, is the value of the resistance measured across the output contact when a DC current signal is applied to the output pins for a duration sufficient to reach thermal equilibrium. R_{SS} includes the effects of the temperature rise of each element in the thermal model. Rating curves are shown in Figures 2 and 4. Figure 2 specifies the maximum average output current allowable for a given ambient temperature. Figure 4 specifies the output power dissipation allowable for a given ambient temperature. Above 55°C (for $\theta_{CA} = 80^{\circ}$ C/W) and 107°C (for $\theta_{CA} =$ 40°C/W/W), the maximum allowable output current and power dissipation are related by the expression RSS = $P_{O(max)}/(I_{O(max)})^2$ from which R_{SS} can be calculated. Staying within the safe area assures that the steady-state junction temperatures remain less than 150°C. As an example, for $T_A = 95^{\circ}C$ and $\theta_{CA} = 80^{\circ}C/W$, Figure 2 shows that the output current should be limited to less than 610 mA. A check with Figure 4 shows that the output power dissipation at $T_A = 95^{\circ}C$ and $I_O = 610$ mA, will be limited to less than 0.35 W. This yields an RSS of 0.94 Ω .

Design Considerations for Replacement of Electro-Mechanical Relays

The HSSR-7110 family can replace electro-mechanical relays with comparable output voltage and current ratings. The following design issues need to be considered in the replacement circuit.

Input Circuit: The drive circuit of the electro-mechanical relay coil needs to be modified so that the average forward current driving the LED of the HSSR- 7110 does not exceed 20 mA. A nominal forward drive current of 10 mA is recommended. A recommended drive circuit with 5 volt VCC and CMOS logic gates is shown in Figure 1. If higher VCC voltages are used, adjust the current limiting resistor to a nominal LED forward current of 10 mA. One important consideration to note is that when the LED is turned off, no more than 0.6 volt forward bias should be applied across the LED. Even a few microamps of current may be sufficient to turn on the HSSR-7110, although it may take a considerable time. The drive circuit should maintain at least 5 mA of LED current during the ON condition. If the LED forward current is less than the 5 mA level, it will cause the HSSR-7110 to turn on with a longer delay. In addition, the power dissipation in the output power MOSFETs increases, which, in turn, may violate the power dissipation guidelines and affect the reliability of the device.

Output Circuit: Unlike electromechanical relays, the designer should pay careful attention to the output on-resistance of solid state relays. The previous section, "On- Resistance and Rating Curves" describes the issues that need to be considered. In addition, for strictly dc applications the designer has an advantage using Connection B which has twice the output current rating as Connection A. Furthermore, for dc-only applications, with Connection B the on-resistance is considerably less when compared to Connection A.

Output over-voltage protection is yet another important design consideration when replacing electro-mechanical relays with the HSSR-7110. The output power MOSFETs can be protected using Metal oxide varistors (MOVs) or TransZorbs against voltage surges that exceed the 90 volt output withstand voltage rating. Examples of sources of voltage surges are inductive load kickbacks, lightning strikes, and electro-static voltages that exceed the specifications on this data sheet. For more information on output load and protection refer to Application Note 1047.

References:

- 1. Application Note 1047, "Low On-Resistance Solid State Relays for High Reliability Applications."
- 2. Reliability Data for HSSR-7111, HSSR-7112, and HSSR-711E.

MOV is a registered trademark of GE/RCA Solid State.

TransZorb is a registered trademark of General Semiconductor.

MIL-PRF-38534 Class H, Class E and DLA SMD Test Program

Class H:

Avago Technologies' s Hi-Rel Optocouplers are in compliance with MIL-PRF-38534 Class H. Class H devices are also in compliance with DLA drawing 5962-93140.

Testing consists of 100% screening and quality conformance inspection to MIL-PRF-38534.

Class E:

Class E devices are in compliance with DLA drawing 5962-9314001Exx. Avago Technologies has defined the Class E device on this drawing to be based on the Class K requirements of MIL-PRF-38534 with exceptions. The exceptions are as follows:

- 1. Nondestructive Bond Pull, Test method 2023 of MIL-STD-883 in device screening is not required.
- 2. Particle Impact Noise Detection (PIND), Test method 2020 of MIL-STD-883 in device screening and group C testing is not required.
- 3. Die Shear Strength, Test method 2019 of MIL-STD-883 in group B testing is not required.
- 4. Internal Water Vapor Content, Test method 1018 of MIL-STD-883 in group C testing is not required.
- 5. Scanning Electron Microscope (SEM) inspections, Test method 2018 of MIL-STD-883 in element evaluation is not required.

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