

FEATURES

- 3.8nV/ $\sqrt{\text{Hz}}$ Input Noise Voltage
- 3.7mA Supply Current
- 200MHz Gain Bandwidth
- Low Total Harmonic Distortion: -85dBc at 1MHz
- 70V/ μs Slew Rate
- 400 μV Maximum Input Offset Voltage
- 300nA Maximum Input Bias Current
- Unity-Gain Stable
- Capacitive Load Stable Up to 100pF
- 23mA Minimum Output Current
- Specified at $\pm 5\text{V}$ and Single 5V
- Low Profile (1mm) SOT-23 (ThinSot™) Package

APPLICATIONS

- Video and RF Amplification
- ADSL, HDSL II, VDSL Receivers
- Active Filters
- Wideband Amplifiers
- Buffers
- Data Acquisition Systems

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DESCRIPTION

The LT®1722/LT1723/LT1724 are single/dual/quad, low noise, low power, high speed operational amplifiers. These products feature lower input offset voltage, lower input bias current and higher DC gain than devices with comparable bandwidth. The 200MHz gain bandwidth ensures high open-loop gain at video frequencies.

The low input noise voltage is achieved with reduced supply current. The total noise is optimized for a source resistance between 0.8k and 12k. Due to the input bias current cancellation technique used, the resistance seen by each input does not need to be balanced.

The output drives a 150 Ω load to $\pm 3\text{V}$ with $\pm 5\text{V}$ supplies. On a single 5V supply the output swings from 1.5V to 3.5V with a 500 Ω load connected to 2.5V. The amplifier is unity-gain stable ($C_{\text{LOAD}} \leq 100\text{pF}$).

The LT1722/LT1723/LT1724 are manufactured on Linear Technology's advanced low voltage complementary bipolar process. The LT1722 is available in the SO-8 and 5-pin SOT-23 packages. The LT1723 is available in the SO-8 and MS8 packages. The LT1724 is available in the 14-lead SO package.

TYPICAL APPLICATION

Differential Video Line Driver



Line Driver Multiburst Video Signal



LT1722/LT1723/LT1724

ABSOLUTE MAXIMUM RATINGS

(Note 1)

Total Supply Voltage (V^+ to V^-) 12.6V
 Input Voltage..... $\pm V_S$
 Differential Input Voltage (Note 2) $\pm 0.7V$
 Input Current (Note 2)..... $\pm 10mA$
 Output Short-Circuit Duration (Note 3) Indefinite

Operating Temperature Range (Note 4) ... $-40^\circ C$ to $85^\circ C$
 Specified Temperature Range (Note 5) $-40^\circ C$ to $85^\circ C$
 Maximum Junction Temperature $150^\circ C$
 Storage Temperature Range $-65^\circ C$ to $150^\circ C$
 Lead Temperature (Soldering, 10 sec)..... $300^\circ C$

PIN CONFIGURATION



ORDER INFORMATION

| LEAD FREE FINISH | TAPE AND REEL | PART MARKING* | PACKAGE DESCRIPTION | SPECIFIED TEMPERATURE RANGE |
|------------------|------------------|---------------|------------------------|-----------------------------|
| LT1722CS8#PBF | LT1722CS8#TRPBF | 1722 | 8-Lead Plastic SO | 0°C to 70°C |
| LT1722IS8#PBF | LT1722IS8#TRPBF | 1722I | 8-Lead Plastic SO | -40°C to 85°C |
| LT1722CS5#PBF | LT1722CS5#TRPBF | LTZB | 5-Lead Plastic TSOT-23 | 0°C to 70°C |
| LT1722IS5#PBF | LT1722IS5#TRPBF | LTZB | 5-Lead Plastic TSOT-23 | -40°C to 85°C |
| LT1723CS8#PBF | LT1723CS8#TRPBF | 1723 | 8-Lead Plastic SO | 0°C to 70°C |
| LT1723IS8#PBF | LT1723IS8#TRPBF | 1723I | 8-Lead Plastic SO | -40°C to 85°C |
| LT1723CMS8#PBF | LT1723CMS8#TRPBF | LTYC | 8-Lead Plastic MSOP | 0°C to 70°C |
| LT1723IMS8#PBF | LT1723IMS8#TRPBF | LTZA | 8-Lead Plastic MSOP | -40°C to 85°C |
| LT1724CS#PBF | LT1724CS#TRPBF | LT1724CS | 14-Lead Plastic SO | 0°C to 70°C |
| LT1724IS#PBF | LT1724IS#TRPBF | LT1724IS | 14-Lead Plastic SO | -40°C to 85°C |

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container. Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreeel/>

LT1722/LT1723/LT1724

ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$, $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|---------------|--|--|------------------------|------------------------|------------|--|
| V_{OS} | Input Offset Voltage | (Note 6) LT1722 SOT-23 and LT1723 MS8 | | 100 150 | 400 650 | μV μV |
| I_{OS} | Input Offset Current | | | 40 | 300 | nA |
| I_B | Input Bias Current | | | 40 | 300 | nA |
| e_n | Input Noise Voltage | $f = 10\text{kHz}$ | | 3.8 | | $\text{nV}/\sqrt{\text{Hz}}$ |
| i_n | Input Noise Current | $f = 10\text{kHz}$ | | 1.2 | | $\text{pA}/\sqrt{\text{Hz}}$ |
| R_{IN} | Input Resistance | $V_{CM} = \pm 3.5\text{V}$ Differential | 5 | 35 50 | | $\text{M}\Omega$ $\text{k}\Omega$ |
| C_{IN} | Input Capacitance | | | 2 | | pF |
| | Input Voltage Range + Input Voltage Range – | | 3.5 | 4 –4 | –3.5 | V V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 3.5\text{V}$ | 80 | 100 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 2.3\text{V}$ to $\pm 5.5\text{V}$ | 78 | 90 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$ $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | 10 7 | 17 14 | | V/mV V/mV |
| V_{OUT} | Output Swing | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$ | ± 3.2 ± 3.1 | ± 3.8 ± 3.4 | | V V |
| I_{OUT} | Output Current | $V_{OUT} = \pm 3\text{V}$, 10mV Overdrive | 23 | 50 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$ | 35 | 90 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | 45 | 70 | | $\text{V}/\mu\text{s}$ |
| | Full Power Bandwidth | 3V Peak, (Note 8) | | 3.7 | | MHz |
| GBW | Gain Bandwidth | $f = 200\text{kHz}$ | 115 | 200 | | MHz |
| t_S | Settling Time | $A_V = -1$, 2V, 0.1% $A_V = -1$, 2V, 0.01% | | 91 112 | | ns ns |
| t_r , t_f | Rise Time, Fall Time | $A_V = 1$, 10% to 90%, $V_{IN} = 0.2\text{V}_{P-P}$, $R_L = 150\Omega$ | | 6 | | ns |
| | Overshoot | $A_V = 1$, $V_{IN} = 0.2\text{V}_{P-P}$, $R_L = 150\Omega$, $R_F = 0\Omega$ | | 15 | | % |
| | Propagation Delay | 50% V_{IN} to 50% $V_{OUT} = 0.2\text{V}_{P-P}$, $R_L = 150\Omega$ | | 3 | | ns |
| R_O | Output Resistance | $A_V = 1$, $f = 1\text{MHz}$ | | 0.15 | | Ω |
| | Channel Separation | $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | 82 | 90 | | dB |
| I_S | Supply Current | Per Amplifier | | 3.7 | 4.5 | mA |

$T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V, unless otherwise noted.

| | | | | | | |
|----------|--|---|-----|------------|------------|--------------------------------------|
| V_{OS} | Input Offset Voltage | (Note 6) LT1722 SOT-23 and LT1723 MS8 | | 250 350 | 550 800 | μV μV |
| I_{OS} | Input Offset Current | | | 20 | 300 | nA |
| I_B | Input Bias Current | | | 20 | 300 | nA |
| e_n | Input Noise Voltage | $f = 10\text{kHz}$ | | 4 | | $\text{nV}/\sqrt{\text{Hz}}$ |
| i_n | Input Noise Current | $f = 10\text{kHz}$ | | 1.1 | | $\text{pA}/\sqrt{\text{Hz}}$ |
| R_{IN} | Input Resistance | $V_{CM} = 1.5\text{V}$ to 3.5V Differential | 5 | 32 55 | | $\text{M}\Omega$ $\text{k}\Omega$ |
| C_{IN} | Input Capacitance | | | 2 | | pF |
| | Input Voltage Range + Input Voltage Range – | | 3.5 | 4 1 | 1.5 | V V |

ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V , unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|---------------|--------------------------------|--|-----|------------|-----|------------------|
| CMRR | Common Mode Rejection Ratio | $V_{CM} = 1.5\text{V}$ to 3.5V | 80 | 100 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$ | 4 | 10 | | V/mV |
| V_{OUT} | Output Swing+ Output Swing- | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | 3.6 | 3.8 0.9 | 1.4 | V V |
| I_{OUT} | Output Current | $V_{OUT} = 3.5\text{V}$ or 1.5V , 10mV Overdrive | 10 | 20 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$ | 22 | 55 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | 40 | 70 | | V/ μs |
| | Full Power Bandwidth | 1V Peak, (Note 8) | | 8.7 | | MHz |
| GBW | Gain Bandwidth (Note 10) | $f = 200\text{kHz}$ | 115 | 180 | | MHz |
| t_r , t_f | Rise Time, Fall Time | $A_V = 1$, 10% to 90%, $V_{IN} = 0.2\text{V}_{P-P}$, $R_L = 500\Omega$ | | 5 | | ns |
| | Overshoot | $A_V = 1$, $V_{IN} = 0.2\text{V}_{P-P}$, $R_L = 500\Omega$ | | 16 | | % |
| | Propagation Delay | 50% V_{IN} to 50% V_{OUT} , 0.1V , $R_L = 500\Omega$ | | 3 | | ns |
| R_O | Output Resistance | $A_V = 1$, $f = 1\text{MHz}$ | | 0.19 | | Ω |
| | Channel Separation | $V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$ | 82 | 90 | | dB |
| I_S | Supply Current | Per Amplifier | | 3.8 | 5 | mA |

The ● denotes the specifications which apply over the temperature range of $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$. $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted. (Note 5)

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------|--|--|--------|--------------------------|------------|--------------------------------|
| V_{OS} | Input Offset Voltage | (Note 6) LT1722 SOT-23 and LT1723 MS8 | | | 700 850 | μV μV |
| | Input V_{OS} Drift | (Note 9) | | 3 | 7 | $\mu\text{V}/^\circ\text{C}$ |
| I_{OS} | Input Offset Current | | | | 350 | nA |
| I_B | Input Bias Current | | | | 350 | nA |
| | Input Voltage Range + Input Voltage Range - | | 3.5 | | -3.5 | V V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 3.5\text{V}$ | ● | 75 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 2.3\text{V}$ to $\pm 5.5\text{V}$ | ● | 76 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$ $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | ● ● | 9 6 | | V/mV V/mV |
| V_{OUT} | Output Swing | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ $R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● ● | ± 3.15 ± 3.05 | | V V |
| I_{OUT} | Output Current | $V_{OUT} = \pm 3\text{V}$, 10mV Overdrive | ● | 22 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$ | ● | 30 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | ● | 35 | | V/ μs |
| GBW | Gain Bandwidth | $f = 200\text{kHz}$ | ● | 100 | | MHz |
| | Channel Separation | $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | ● | 81 | | dB |
| I_S | Supply Current | Per Amplifier | ● | | 5.45 | mA |

LT1722/LT1723/LT1724

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range of $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5Ω , unless otherwise noted. (Note 5)

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------|--|--|--------|------|------------|--------------------------------|
| V_{OS} | Input Offset Voltage | (Note 6) LT1722 SOT-23 and LT1723 MS8 | ● | | 850 950 | μV μV |
| | Input V_{OS} Drift | (Note 9) | ● | 3 | 7 | $\mu\text{V}/^{\circ}\text{C}$ |
| I_{OS} | Input Offset Current | | ● | | 350 | nA |
| I_B | Input Bias Current | | ● | | 350 | nA |
| | Input Voltage Range + Input Voltage Range - | | ● ● | 3.5 | 1.5 | V V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = 1.5\text{V}$ to 3.5V | ● | 75 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$ | ● | 3 | | V/mV |
| V_{OUT} | Output Swing+ | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | 3.55 | | V |
| | Output Swing- | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | | 1.45 | V |
| I_{OUT} | Output Current | $V_{OUT} = 3.5\text{V}$, or 1.5V , 10mV Overdrive | ● | 9 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$ | ● | 11 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | ● | 30 | | V/ μs |
| GBW | Gain Bandwidth (Note 10) | $f = 200\text{kHz}$ | ● | 100 | | MHz |
| | Channel Separation | $V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$ | ● | 81 | | dB |
| I_S | Supply Current | | ● | | 5.95 | mA |

The ● denotes the specifications which apply over the temperature range of $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$. $V_S = \pm 5\text{V}$, $V_{CM} = 0\text{V}$, unless otherwise noted. (Note 5)

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------|--|---|--------|-----------|-------------|--------------------------------|
| V_{OS} | Input Offset Voltage | (Note 6) LT1722 SOT-23 and LT1723 MS8 | ● | | 900 1100 | μV μV |
| | Input V_{OS} Drift | (Note 9) | ● | 3 | 10 | $\mu\text{V}/^{\circ}\text{C}$ |
| I_{OS} | Input Offset Current | | ● | | 400 | nA |
| I_B | Input Bias Current | | ● | | 400 | nA |
| | Input Voltage Range + Input Voltage Range - | | ● ● | 3.5 | -3.5 | V V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 3.5\text{V}$ | ● | 75 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 2.0\text{V}$ to $\pm 5.5\text{V}$ | ● | 75 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = \pm 3\text{V}$, $R_L = 500\Omega$ | ● | 8 | | V/mV |
| | | $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | ● | 5 | | V/mV |
| V_{OUT} | Output Swing | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | ± 3.1 | | V |
| | | $R_L = 150\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | ± 3.0 | | V |
| I_{OUT} | Output Current | $V_{OUT} = \pm 3\text{V}$, 10mV Overdrive | ● | 20 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 0\text{V}$, $V_{IN} = \pm 1\text{V}$ | ● | 25 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | ● | 25 | | V/ μs |
| GBW | Gain Bandwidth | $f = 200\text{kHz}$ | ● | 90 | | MHz |
| | Channel Separation | $V_{OUT} = \pm 3\text{V}$, $R_L = 150\Omega$ | ● | 80 | | dB |
| I_S | Supply Current | | ● | | 5.95 | mA |

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the temperature range of $-40^{\circ}\text{C} \leq T_A \leq 85^{\circ}\text{C}$. $V_S = 5\text{V}$, $V_{CM} = 2.5\text{V}$, R_L to 2.5V , unless otherwise noted. (Note 5)

| SYMBOL | PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|-----------|--------------------------------|--|--------|-----|------|--------------------------------|
| V_{OS} | Input Offset Voltage | (Note 6) | ● | | 1000 | μV |
| | | LT1722 SOT-23 and LT1723 MS8 | ● | | 1200 | μV |
| | Input V_{OS} Drift | (Note 9) | ● | 3 | 10 | $\mu\text{V}/^{\circ}\text{C}$ |
| I_{OS} | Input Offset Current | | ● | | 400 | nA |
| I_B | Input Bias Current | | ● | | 400 | nA |
| | | Input Voltage Range + Input Voltage Range - | ● ● | 3.5 | 1.5 | V V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = 1.5\text{V}$ to 3.5V | ● | 75 | | dB |
| A_{VOL} | Large-Signal Voltage Gain | $V_{OUT} = 1.5\text{V}$ to 3.5V , $R_L = 500\Omega$ | ● | 2 | | V/mV |
| V_{OUT} | Output Swing+ Output Swing- | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | 3.5 | | V |
| | | $R_L = 500\Omega$, $V_{IN} = \pm 10\text{mV}$ | ● | | 1.5 | V |
| I_{OUT} | Output Current | $V_{OUT} = 3.5\text{V}$ or 1.5V , 30mV Overdrive | ● | 8 | | mA |
| I_{SC} | Short-Circuit Current | $V_{OUT} = 2.5\text{V}$, $V_{IN} = \pm 1\text{V}$ | ● | 10 | | mA |
| SR | Slew Rate | $A_V = -1$, (Note 7) | ● | 20 | | V/ μs |
| GBW | Gain Bandwidth (Note 10) | $f = 200\text{kHz}$ | ● | 90 | | MHz |
| | | Channel Separation | ● | 80 | | dB |
| I_S | Supply Current | | ● | | 6.45 | mA |

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The inputs are protected by back-to-back diodes. If the differential input voltage exceeds 0.7V, the input current should be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

Note 4: The LT1722C/LT1722I, LT1723C/LT1723I, LT1724C/LT1724I are guaranteed functional over the operating temperature range of -40°C to 85°C .

Note 5: The LT1722C/LT1723C/LT1724C are guaranteed to meet specified performance from 0°C to 70°C . The LT1722C/LT1723C/LT1724C are

designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures. The LT1722I/LT1723I/LT1724I are guaranteed to meet specified performance from -40°C to 85°C .

Note 6: Input offset voltage is pulse tested and is exclusive of warm-up drift.

Note 7: Slew rate is measured between $\pm 2\text{V}$ on the output with $\pm 3\text{V}$ input for $\pm 5\text{V}$ supplies and $\pm 1\text{V}$ on the output with $\pm 1.5\text{V}$ input for single 5V supply. (For 5V supply, the voltage levels are 2.5V referred.)

Note 8: Full power bandwidth is calculated from the slew rate:

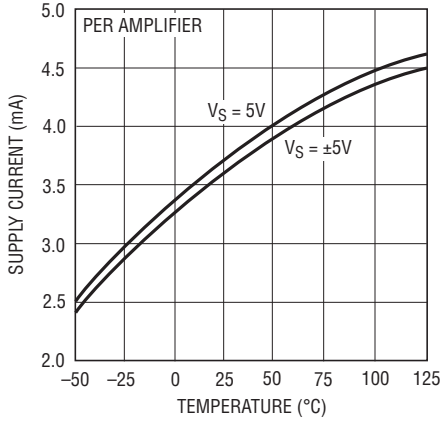
$$\text{FPBW} = \text{SR}/2\pi V_P$$

Note 9: This parameter is not 100% tested.

Note 10: This parameter is guaranteed through correlation with slew rate.

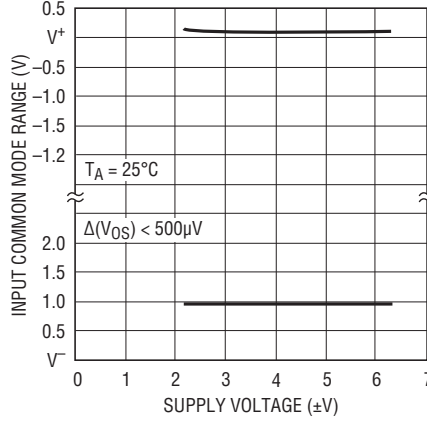
TYPICAL PERFORMANCE CHARACTERISTICS

Supply Current vs Temperature



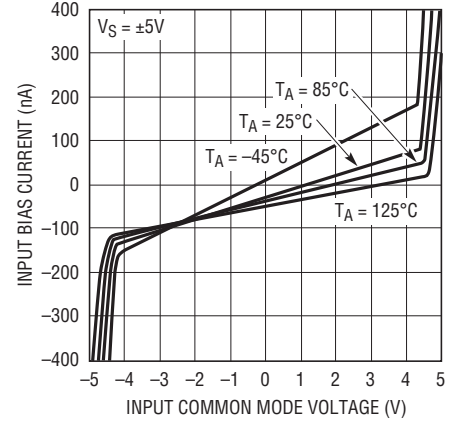
1723 G01

Input Common Mode Range vs Supply Voltage



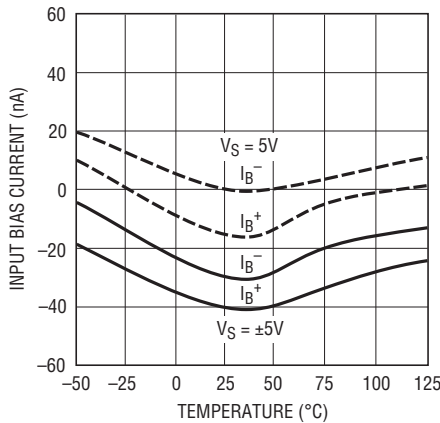
1723 G02

Input Bias Current vs Common Mode Voltage



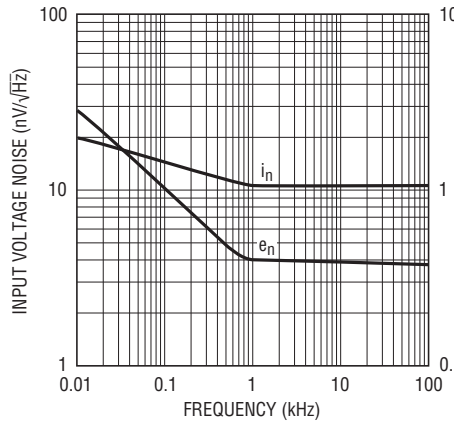
1723 G03

Input Bias Current vs Temperature



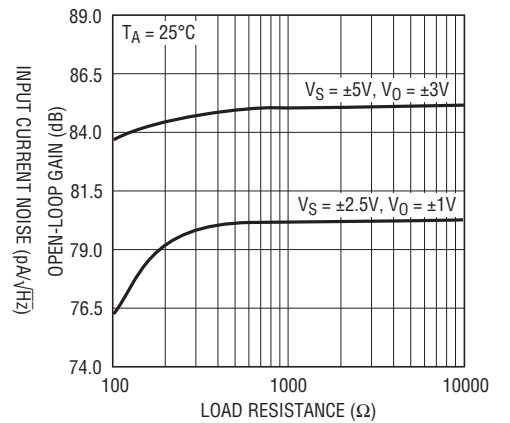
1723 G04

Input Noise Spectral Density



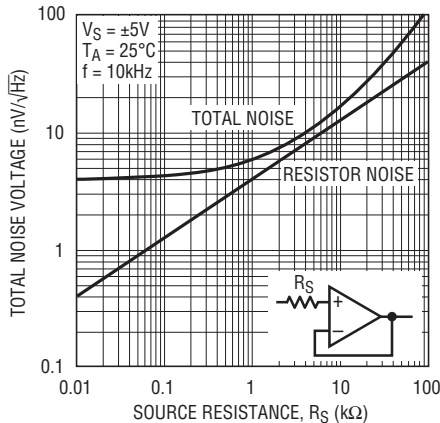
1723 G05

Open-Loop Gain vs Resistive Load



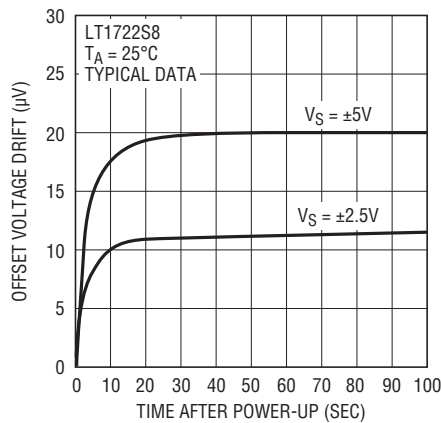
1723 G06

Total Noise vs Unmatched Source Resistance



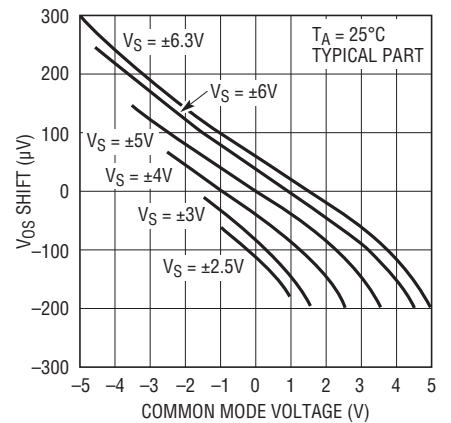
1723 G07

Warm-Up Drift vs Time



1723 G08

VOS Shift vs VCM and VS

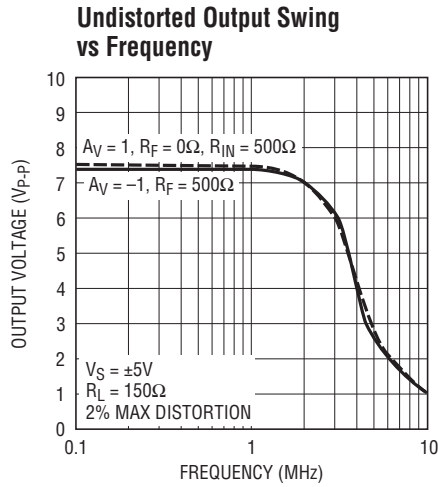


1723 G09

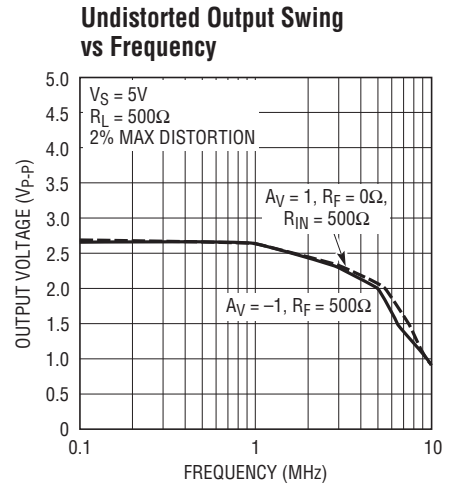
TYPICAL PERFORMANCE CHARACTERISTICS



1723 G10



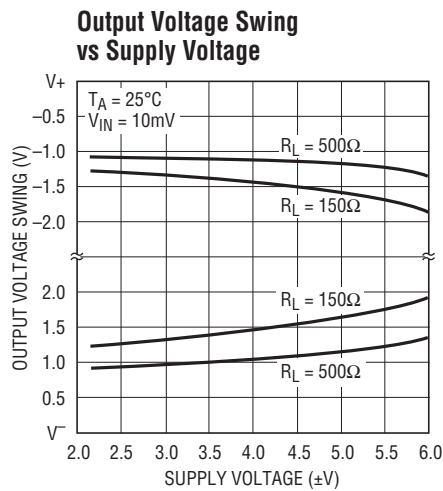
1723 G11



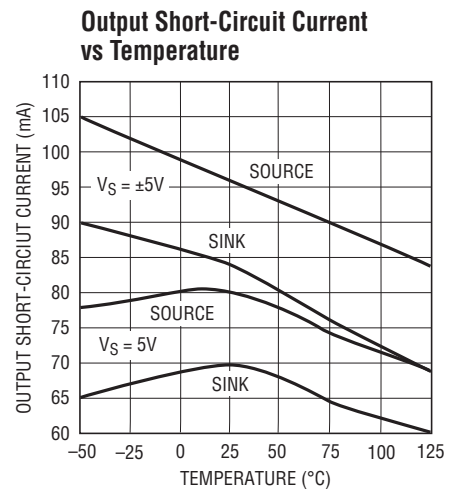
1723 G12



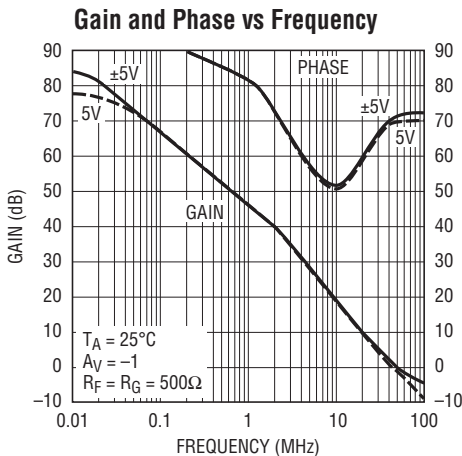
1723 G13



1723 G08



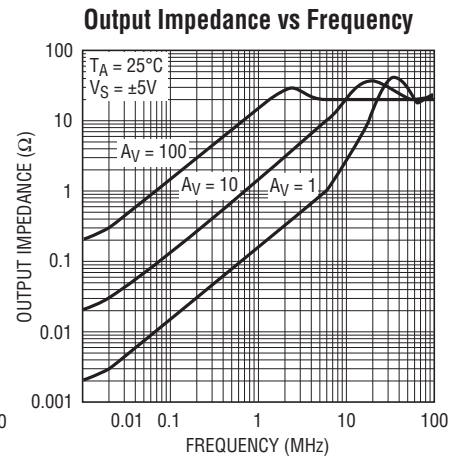
1723 G15



1723 G16

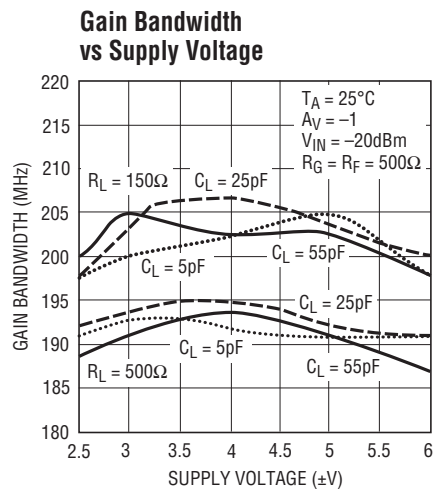
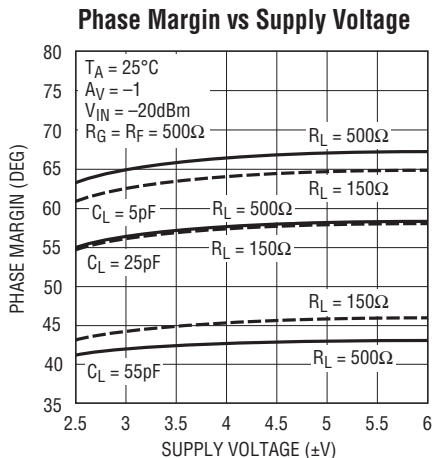
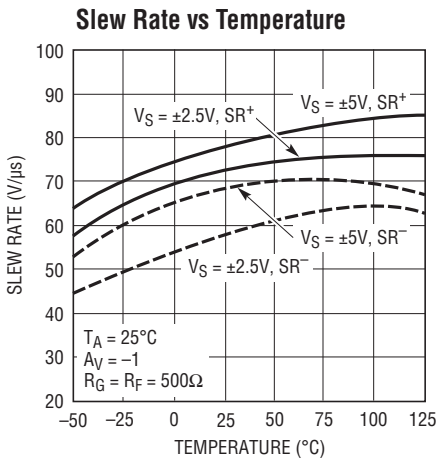
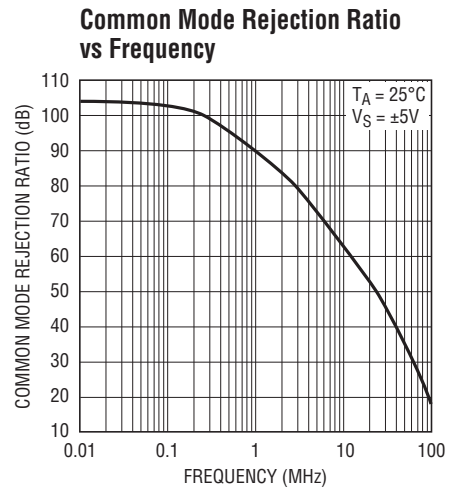
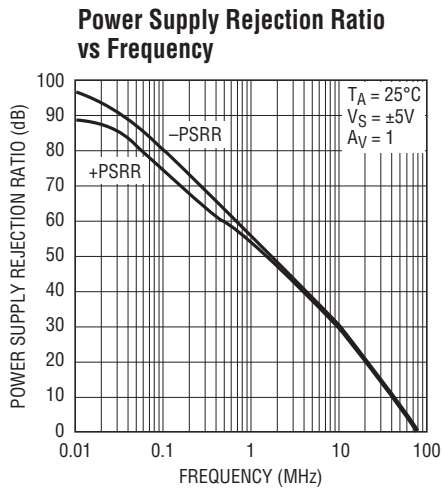
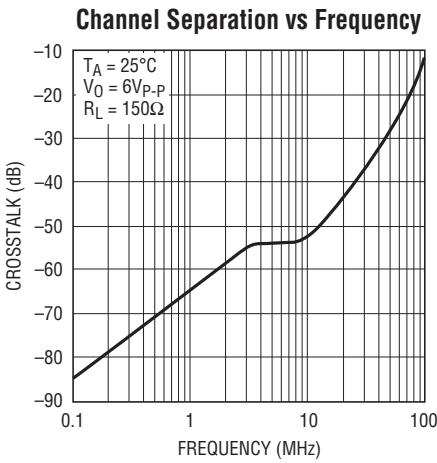
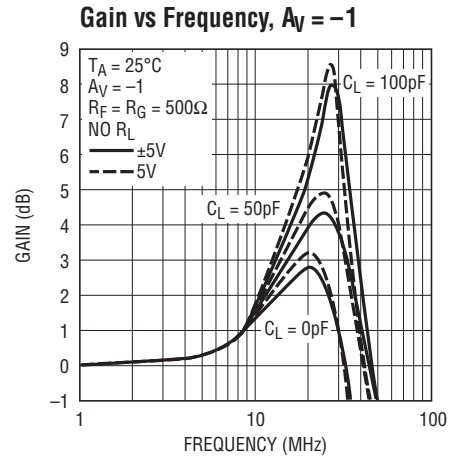
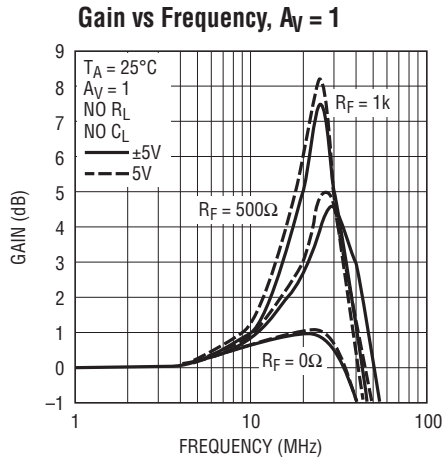
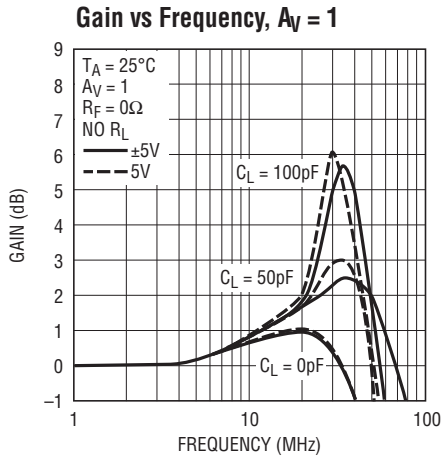


1723 G17



1723 G18

TYPICAL PERFORMANCE CHARACTERISTICS



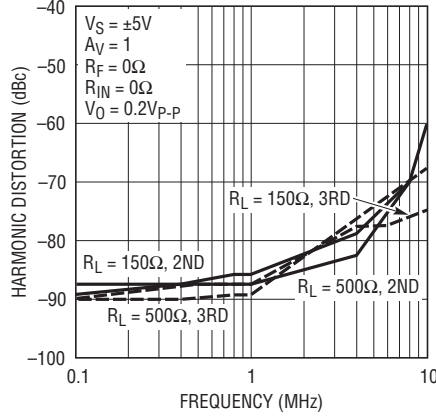
TYPICAL PERFORMANCE CHARACTERISTICS

Slew Rate vs Supply Voltage



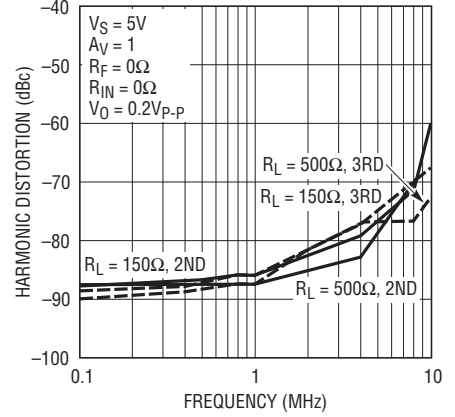
1723 G25

Harmonic Distortion vs Frequency
 $A_V = 1, V_O = 0.2V_{P-P}$



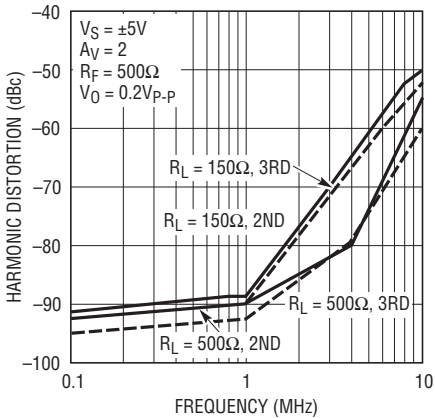
1723 G26

Harmonic Distortion vs Frequency
 $A_V = 1, V_O = 0.2V_{P-P}$



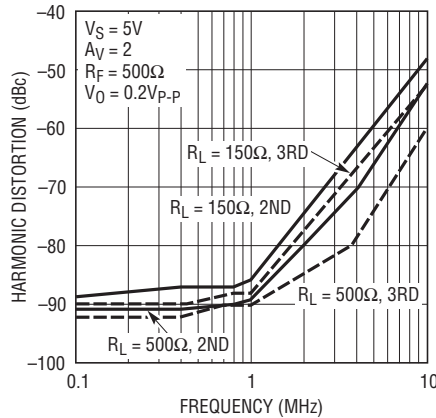
1723 G27

Harmonic Distortion vs Frequency
 $A_V = 2, V_O = 0.2V_{P-P}$



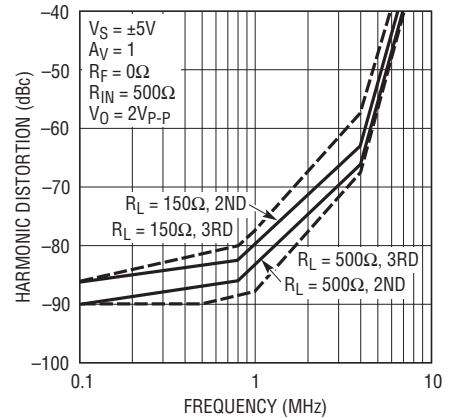
1723 G28

Harmonic Distortion vs Frequency
 $A_V = 2, V_O = 0.2V_{P-P}$



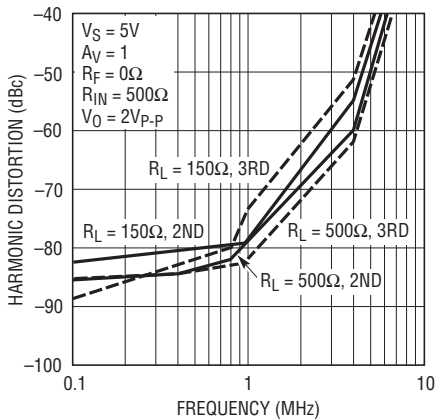
1723 G29

Harmonic Distortion vs Frequency
 $A_V = 1, V_O = 2V_{P-P}$



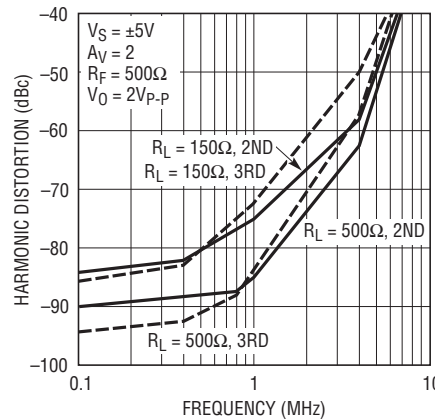
1723 G30

Harmonic Distortion vs Frequency
 $A_V = 1, V_O = 2V_{P-P}$



1723 G31

Harmonic Distortion vs Frequency
 $A_V = 2, V_O = 2V_{P-P}$

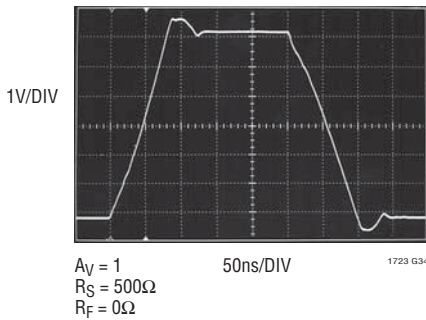


1723 G32

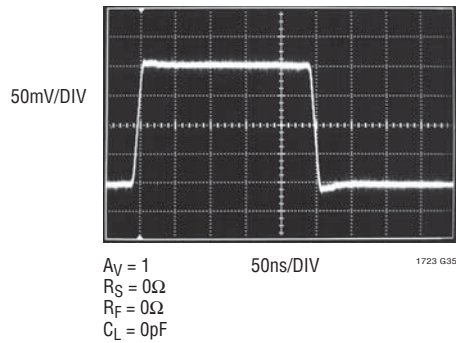
TYPICAL PERFORMANCE CHARACTERISTICS



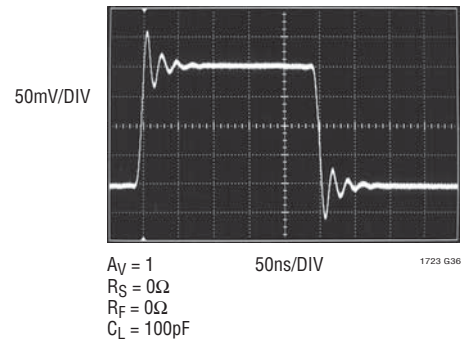
Large-Signal Transient, $A_V = 1$



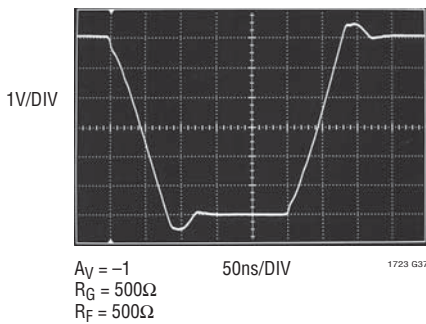
Small-Signal Transient, $A_V = 1$



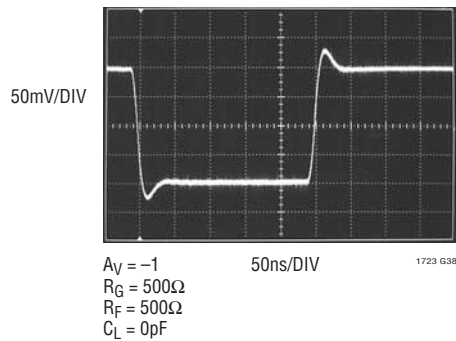
Small-Signal Transient, $A_V = 1$



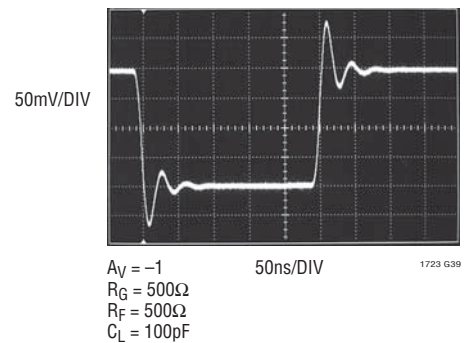
Large-Signal Transient, $A_V = -1$



Small-Signal Transient, $A_V = -1$



Small-Signal Transient, $A_V = -1$



APPLICATIONS INFORMATION

The LT1722/LT1723/LT1724 may be inserted directly into many operational amplifier applications improving both DC and AC performance, as well as noise and distortion.

Layout and Passive Components

The LT1722/LT1723/LT1724 amplifiers are more tolerant of less than ideal layouts than other high speed amplifiers. For maximum performance (for example, fast settling time) use a ground plane, short lead lengths and RF quality bypass capacitors (0.01μF to 0.1μF). For high drive current applications, use low ESR supply bypass capacitors (1μF to 10μF tantalum). The output/input parasitic coupling should be minimized when high frequency performance is required.

The parallel combination of the feedback resistor and gain setting resistor on the inverting input combine with the input capacitance to form a pole that can cause peaking or even oscillations. In parallel with the feedback resistor, a capacitor of value:

$$C_F > R_G \cdot C_{IN}/R_F$$

should be used to cancel the input pole and optimize dynamic performance. For unity-gain applications where a feedback resistor is used, such as an I-to-V converter, C_F should be five times greater than C_{IN} ; an optimum value for C_F is 10pF.

Input Considerations

Each of the LT1722/LT1723/LT1724 inputs is protected with back-to-back diodes across the bases of the NPN input devices. If greater than 0.7V differential input voltages are anticipated, the input current must be limited to less than 10mA with an external series resistor. Each input also has two ESD clamp diodes—one to each supply. If an input is driven beyond the supply, limit the current with an external resistor to less than 10mA. The input stage protection circuit is shown in Figure 1.

The input currents of the LT1722/LT1723/LT1724 are typically in the tens of nA range due to the bias current cancellation technique used at the input. As the input offset current can be greater than either input current,

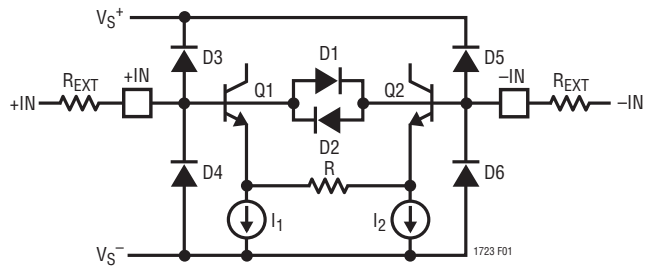


Figure 1. Input Stage Protection

adding resistance to balance source resistance is not recommended. The value of the source resistor should be below 12k as it actually degrades DC accuracy and also increases noise.

Total Input Noise

The total input noise of the LT1722/LT1723/LT1724 is optimized for a source resistance between 0.8k and 12k. Within this range, the total input noise is dominated by the noise of the source resistance itself. When the source resistance is below 0.8k, voltage noise of the amplifier dominates. When the source resistance is above 12k, the input noise current is the dominant contributor.

Capacitive Loading

The LT1722/LT1723/LT1724 drive capacitive loads up to 100pF with unity gain. As the capacitive load increases, both the bandwidth and the phase margin decrease causing peaking in the frequency response and overshoot in the transient response. When there is a need to drive a larger capacitive load, a 25Ω series resistance assures stability with any value of load capacitor. A feedback capacitor also helps to reduce any peaking.

Power Dissipation

The LT1722/LT1723/LT1724 combine high speed and large output drive in a small package. Maximum junction temperature (T_J) is calculated from the ambient temperature (T_A), power dissipation per amplifier (P_D) and number of amplifiers (n) as follows:

$$T_J = T_A + (n \cdot P_D \cdot \theta_{JA})$$

APPLICATIONS INFORMATION

Power dissipation is composed of two parts. The first is due to the quiescent supply current and the second is due to on-chip dissipation caused by the load current.

Worst-case instantaneous power dissipation for a given resistive load in one amplifier occurs at the maximum supply current and when the output voltage is at half of either supply voltage (or the maximum swing if less than half supply voltage).

Therefore $P_{D(MAX)}$ in one amplifier is:

$$P_{D(MAX)} = (V^+ - V^-)(I_{S(MAX)}) + (V^+/2)^2/R_L$$

or

$$P_{D(MAX)} = (V^+ - V^-)(I_{S(MAX)}) + (V^+ - V_{O(MAX)})(V_{O(MAX)}/R_L)$$

Example. Worst-case conditions are: both op amps in the LT1723IS8 are at $T_A = 85^\circ\text{C}$, $V_S = \pm 5\text{V}$, $R_L = 150\Omega$, $V_{OUT} = 2.5\text{V}$.

$$P_{D(MAX)} = 2 \cdot [(10\text{V})(5.95\text{mA}) + (2.5\text{V})^2/150\Omega] = 203\text{mW}$$

$$T_{J(MAX)} = 85^\circ\text{C} + (203\text{mW})(190^\circ\text{C}/\text{W}) = 124^\circ\text{C}$$

which is less than the absolute maximum rating at 150°C .

Circuit Operation

The LT1722/LT1723/LT1724 circuit topology is a voltage feedback amplifier. The operation of the circuit can be understood by referring to the Simplified Schematic. The first stage is a folded cascode formed by the transistors Q1 through Q4. A degeneration resistor, R, is used in the input stage. The current mirror Q5, Q6 is bootstrapped by Q7. The capacitor, C, assures the bandwidth and the slew rate performance. The output stage is formed by complementary emitter followers, Q8 through Q11. The diodes D1 and D2 protect against input reversed biasing. The remaining part of the circuit assures optimum voltage and current biases for all stages.

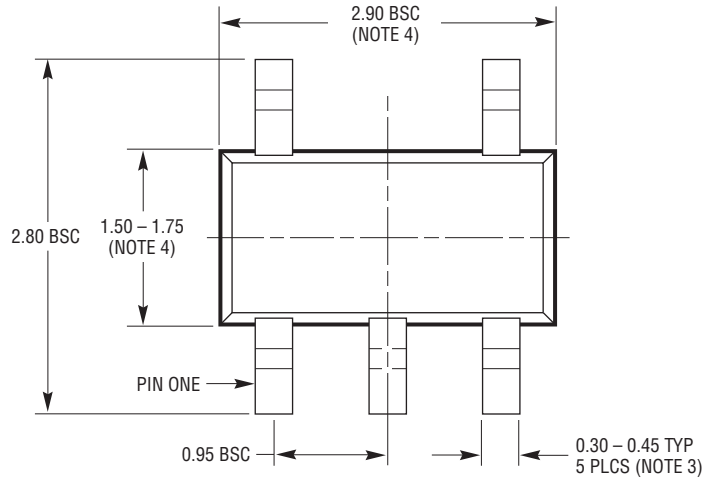
Low noise, reduced current supply, high speed and DC accurate parameters are distinctive features of the LT1722/LT1723/LT1724.

SIMPLIFIED SCHEMATIC



PACKAGE DESCRIPTION

S5 Package
5-Lead Plastic TSOT-23
 (Reference LTC DWG # 05-08-1635)

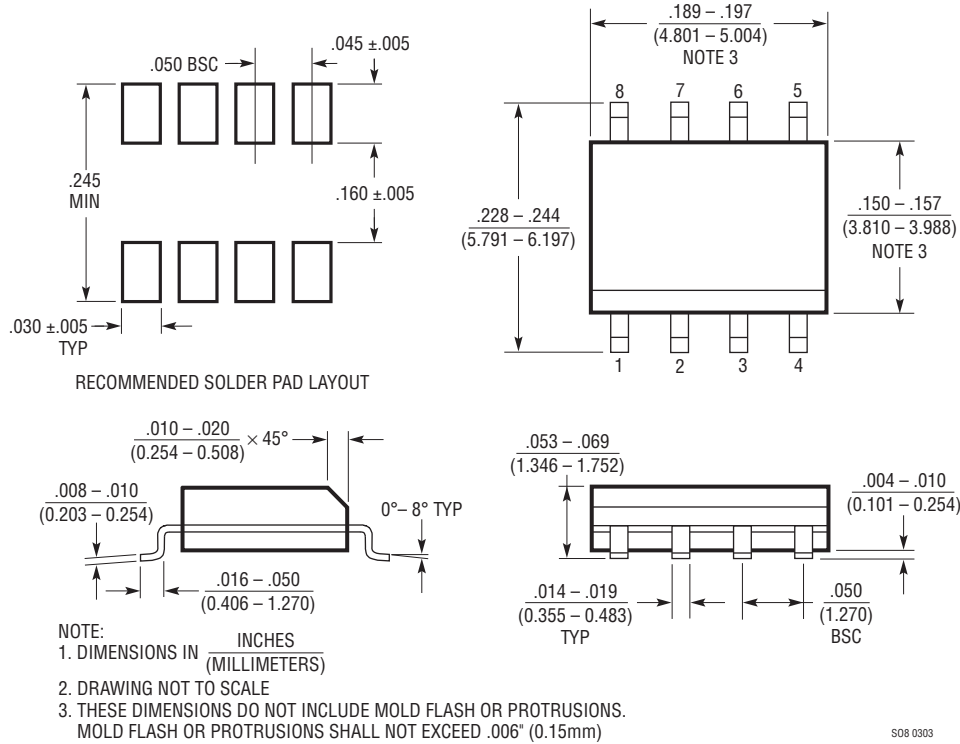


- NOTE:
1. DIMENSIONS ARE IN MILLIMETERS
 2. DRAWING NOT TO SCALE
 3. DIMENSIONS ARE INCLUSIVE OF PLATING
 4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
 5. MOLD FLASH SHALL NOT EXCEED 0.254mm
 6. JEDEC PACKAGE REFERENCE IS MO-193

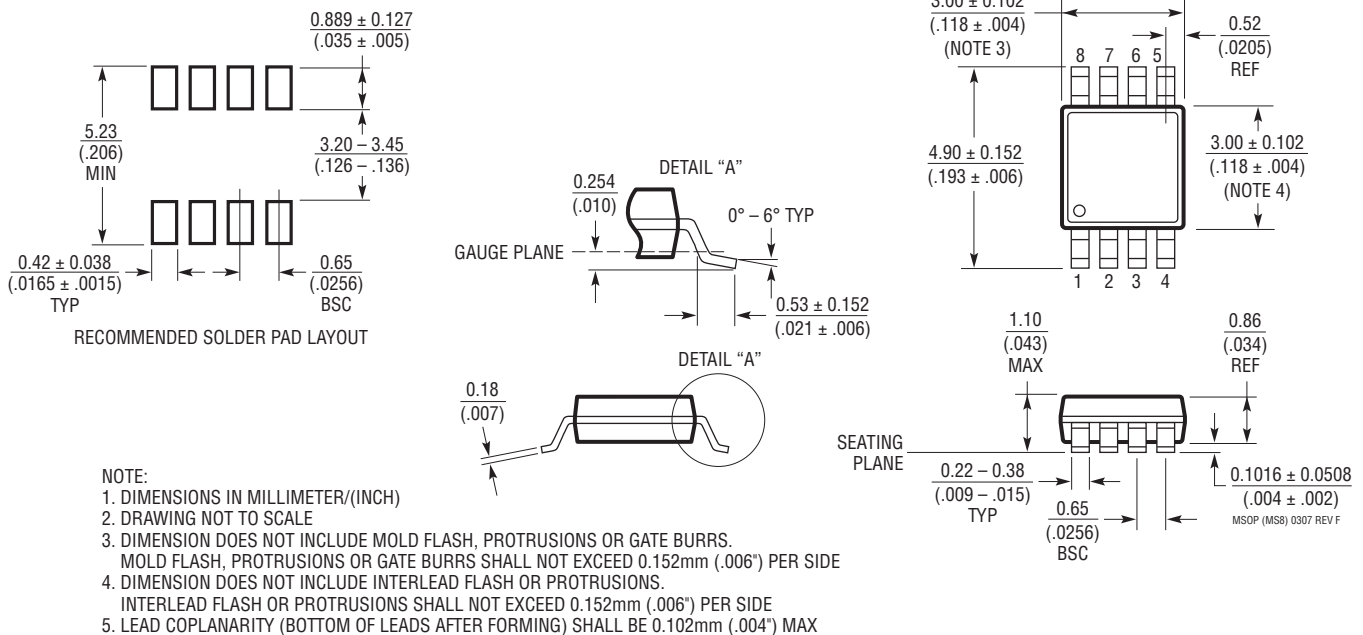
S5 TSOT-23 0302 REV B

PACKAGE DESCRIPTION

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)

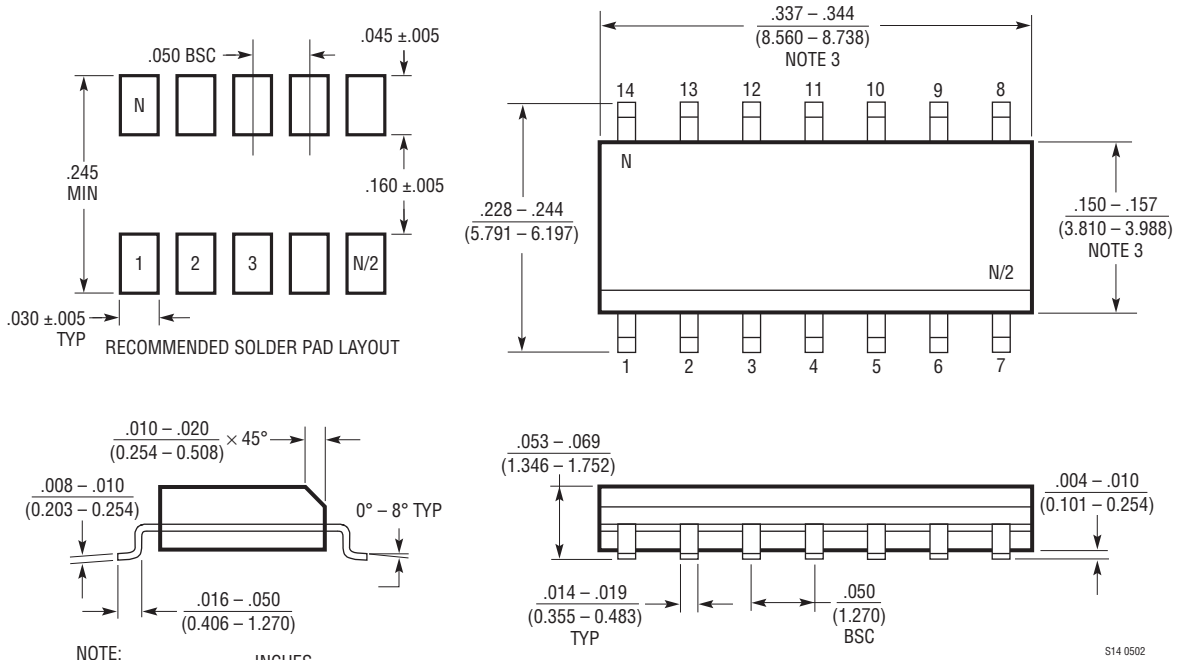


MS8 Package 8-Lead Plastic MSOP (Reference LTC DWG # 05-08-1660 Rev F)



PACKAGE DESCRIPTION

S Package
14-Lead Plastic Small Outline (Narrow .150 Inch)
 (Reference LTC DWG # 05-08-1610)

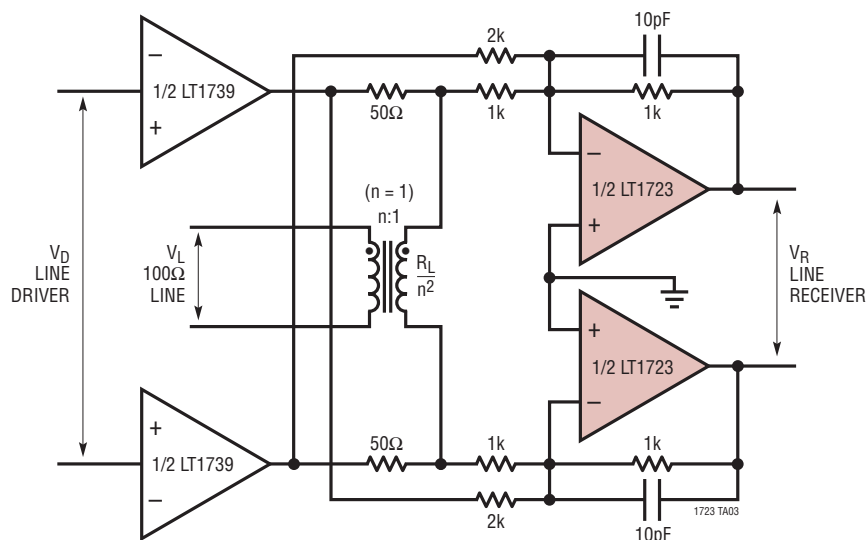


- NOTE:
 1. DIMENSIONS IN INCHES (MILLIMETERS)
 2. DRAWING NOT TO SCALE
 3. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

S14 0502

TYPICAL APPLICATION

4- to 2-Wire Local Echo Cancellation Differential Receiver Amplifier



RELATED PARTS

| PART NUMBER | DESCRIPTION | COMMENTS |
|----------------------|---|--|
| LT1677 | Single, Low Noise Rail-to-Rail Amplifier | 3V Operation, 2.5mA Supply Current, $4.5\text{nV}/\sqrt{\text{Hz}}$ Max e_n , $60\mu\text{V}$ Max V_{OS} |
| LT1800/LT1801/LT1802 | Single/Dual/Quad, Low Power, 80MHz Rail-to-Rail Precision Amplifier | 1.6mA Supply Current, $350\mu\text{V}$ V_{OS} , 2.3V Operation |
| LT1806/LT1807 | Single/Dual, Low Noise 325MHz Rail-to-Rail Amplifiers | 2.5V Operation, $550\mu\text{V}_{MAX}$ V_{OS} , $3.5\text{nV}/\sqrt{\text{Hz}}$ |
| LT1809/LT1810 | Single/Dual, Low Distortion 180MHz Rail-to-Rail Amplifiers | 2.5V Operation, -90dBc at 5MHz Distortion |
| LT1812/LT1813/LT1814 | Single/Dual/Quad, 3mA, 750V/ μs Amplifiers | 5V Operation, 3.6mA Supply Current, 40mA Min Output Current |
| LT6202/LT6203/LT6204 | Single/Dual/Quad, 100MHz, Low Noise Rail-to-Rail Op Amps | $2\text{nV}/\sqrt{\text{Hz}}$, 2.5mA on Single 3V Supply |



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

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Факс: 8 (812) 320-02-42

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