



MCP1827/MCP1827S

1.5A, Low-Voltage, Low Quiescent Current LDO Regulator

Features:

- 1.5A Output Current Capability
- Input Operating Voltage Range: 2.3V to 6.0V
- Adjustable Output Voltage Range: 0.8V to 5.0V (MCP1827 only)
- Standard Fixed Output Voltages:
 - 0.8V, 1.2V, 1.8V, 2.5V, 3.0V, 3.3V, 5.0V
- Other Fixed Output Voltage Options Available Upon Request
- Low Dropout Voltage: 330 mV Typical at 1.5A
- Typical Output Voltage Tolerance: 0.5%
- Stable with 1.0 μ F Ceramic Output Capacitor
- Fast response to Load Transients
- Low Supply Current: 120 μ A (typ)
- Low Shutdown Supply Current: 0.1 μ A (typ) (MCP1827 only)
- Fixed Delay on Power Good Output (MCP1827 only)
- Short Circuit Current Limiting and Overtemperature Protection
- 5-Lead Plastic DDPAK, 5-Lead TO-220 Package Options (MCP1827)
- 3-Lead Plastic DDPAK, 3-Lead TO-220 Package Options (MCP1827S)

Applications:

- High-Speed Driver Chipset Power
- Networking Backplane Cards
- Notebook Computers
- Network Interface Cards
- Palmtop Computers
- 2.5V to 1.XV Regulators

Description:

The MCP1827/MCP1827S is a 1.5A Low Dropout (LDO) linear regulator that provides high current and low output voltages. The MCP1827 comes in a fixed or adjustable output voltage version, with an output voltage range of 0.8V to 5.0V. The 1.5A output current capability, combined with the low output voltage capability, make the MCP1827 a good choice for new sub-1.8V output voltage LDO applications that have high current demands. The MCP1827S is a 3-pin fixed voltage version. The MCP1827/MCP1827S is based upon the MCP1727 LDO device.

The MCP1827/MCP1827S is stable using ceramic output capacitors that inherently provide lower output noise and reduce the size and cost of the entire regulator solution. Only 1 μ F of output capacitance is needed to stabilize the LDO.

Using CMOS construction, the quiescent current consumed by the MCP1827/MCP1827S is typically less than 120 μ A over the entire input voltage range, making it attractive for portable computing applications that demand high output current. The MCP1827 versions have a Shutdown ($\overline{\text{SHDN}}$) pin. When shut down, the quiescent current is reduced to less than 0.1 μ A.

On the MCP1827 fixed output versions the scaled-down output voltage is internally monitored and a power good (PWRGD) output is provided when the output is within 92% of regulation (typical). The PWRGD delay is internally fixed at 200 μ s (typical).

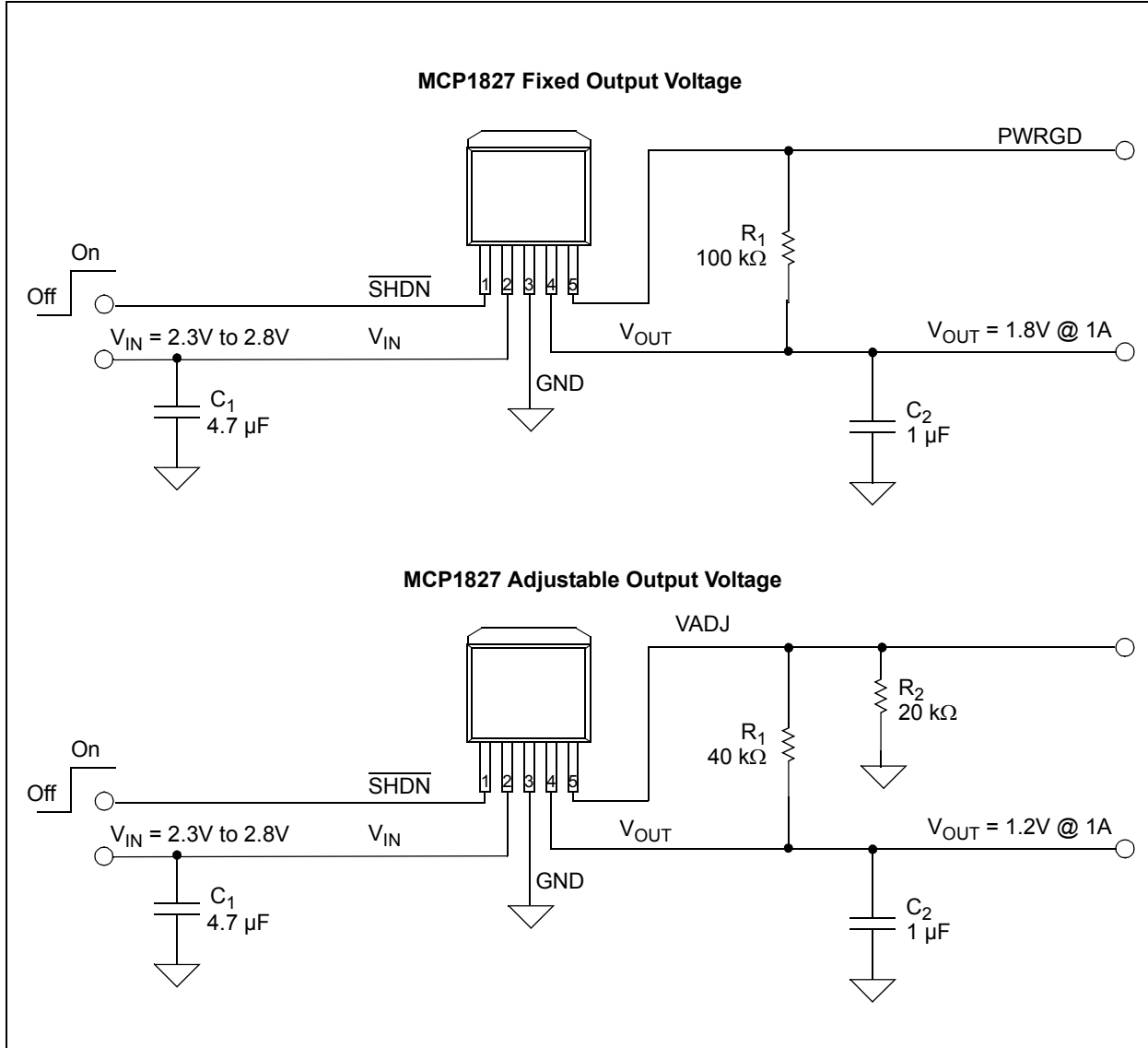
The overtemperature and short circuit current-limiting provide additional protection for the LDO during system Fault conditions.

Package Types



MCP1827/MCP1827S

Typical Application



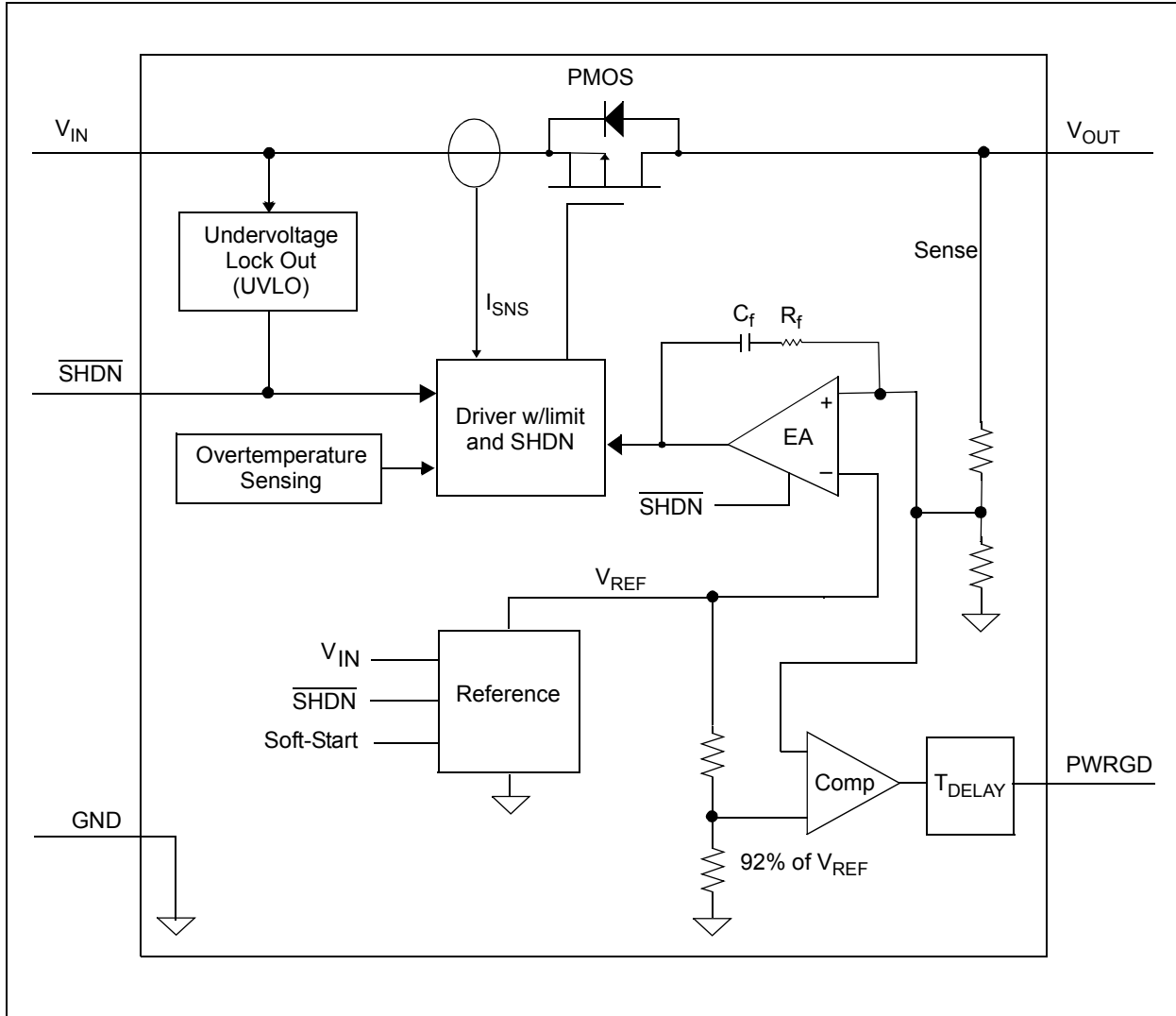
MCP1827/MCP1827S

Functional Block Diagram – Adjustable Output



MCP1827/MCP1827S

Functional Block Diagram – Fixed Output (5-pin)



MCP1827/MCP1827S

Functional Block Diagram – Fixed Output (3-Pin)



MCP1827/MCP1827S

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

V_{IN}	6.5V
Maximum Voltage on Any Pin .. (GND – 0.3V) to (V_{DD} + 0.3)V	
Maximum Power Dissipation	Internally-Limited (Note 6)
Output Short Circuit Duration	Continuous
Storage temperature	-65°C to +150°C
Maximum Junction Temperature, T_J	+150°C
ESD protection on all pins (HBM/MM)	≥ 2 kV; ≥ 200 V

† **Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

AC/DC CHARACTERISTICS

Electrical Specifications: Unless otherwise noted, $V_{IN} = V_{OUT(MAX)} + V_{DROPOUT(MAX)}$ **Note 1**, $V_R = 1.8$ V for Adjustable Output, $I_{OUT} = 1$ mA, $C_{IN} = C_{OUT} = 4.7$ μ F (X7R Ceramic), $T_A = +25^\circ$ C. **Boldface** type applies for junction temperatures, T_J (**Note 7**) of **-40°C to +125°C**

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Input Operating Voltage	V_{IN}	2.3		6.0	V	
Input Quiescent Current	I_q	—	120	220	μ A	$I_L = 0$ mA, $V_{OUT} = 0.8$ V to 5.0V
Input Quiescent Current for SHDN Mode	I_{SHDN}	—	0.1	3	μ A	$\overline{SHDN} = GND$
Maximum Output Current	I_{OUT}	1.5	—	—	A	$V_{IN} = 2.3$ V to 6.0V $V_R = 0.8$ V to 5.0V
Line Regulation	$\frac{\Delta V_{OUT}}{(V_{OUT} \times \Delta V_{IN})}$	—	0.05	0.16	%/V	(Note 1) $\leq V_{IN} \leq 6$ V
Load Regulation	$\Delta V_{OUT}/V_{OUT}$	-1.0	± 0.5	1.0	%	$I_{OUT} = 1$ mA to 1.5A (Note 4)
Output Short Circuit Current	I_{OUT_SC}	—	2.2	—	A	$R_{LOAD} < 0.1\Omega$, Peak Current
Adjust Pin Characteristics (Adjustable Output Only)						
Adjust Pin Reference Voltage	V_{ADJ}	0.402	0.410	0.418	V	$V_{IN} = 2.3$ V to $V_{IN} = 6.0$ V, $I_{OUT} = 1$ mA
Adjust Pin Leakage Current	I_{ADJ}	-10	± 0.01	+10	nA	$V_{IN} = 6.0$ V, $V_{ADJ} = 0$ V to 6V
Adjust Temperature Coefficient	TCV_{OUT}	—	40	—	ppm/°C	Note 3
Fixed-Output Characteristics (Fixed Output Only)						
Voltage Regulation	V_{OUT}	$V_R - 2.5\%$	V_R $\pm 0.5\%$	$V_R + 2.5\%$	V	Note 2

- Note 1:** The minimum V_{IN} must meet two conditions: $V_{IN} \geq 2.3$ V and $V_{IN} \geq V_{OUT(MAX)} + V_{DROPOUT(MAX)}$.
- Note 2:** V_R is the nominal regulator output voltage for the fixed cases. $V_R = 1.2$ V, 1.8V, etc. V_R is the desired set point output voltage for the adjustable cases. $V_R = V_{ADJ} \cdot ((R_1/R_2)+1)$. **Figure 4-1**.
- Note 3:** $TCV_{OUT} = (V_{OUT-HIGH} - V_{OUT-LOW}) \cdot 10^6 / (V_R \cdot \Delta Temperature)$. $V_{OUT-HIGH}$ is the highest voltage measured over the temperature range. $V_{OUT-LOW}$ is the lowest voltage measured over the temperature range.
- Note 4:** Load regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
- Note 5:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of $V_{IN} = V_{OUTMAX} + V_{DROPOUT(MAX)}$.
- Note 6:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e., T_A , T_J , θ_{JA}). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +150°C rating. Sustained junction temperatures above 150°C can impact device reliability.
- Note 7:** The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in the junction temperature over the ambient temperature is not significant.

MCP1827/MCP1827S

AC/DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise noted, $V_{IN} = V_{OUT(MAX)} + V_{DROPOUT(MAX)}$ **Note 1**, $V_R = 1.8V$ for Adjustable Output, $I_{OUT} = 1\text{ mA}$, $C_{IN} = C_{OUT} = 4.7\text{ }\mu\text{F}$ (X7R Ceramic), $T_A = +25^\circ\text{C}$. **Boldface** type applies for junction temperatures, T_J (**Note 7**) of -40°C to $+125^\circ\text{C}$

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Dropout Characteristics						
Dropout Voltage	$V_{IN} - V_{OUT}$	—	330	600	mV	Note 5 , $I_{OUT} = 1.5A$, $V_{IN(MIN)} = 2.3V$
Power Good Characteristics						
PWRGD Input Voltage Operating Range	V_{PWRGD_VIN}	1.0 1.2	—	6.0 6.0	V	$T_A = +25^\circ\text{C}$ $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ For $V_{IN} < 2.3V$, $I_{SINK} = 100\text{ }\mu\text{A}$
PWRGD Threshold Voltage (Referenced to V_{OUT})	V_{PWRGD_TH}	89 90	92 92	95 94	% V_{OUT}	Falling Edge $V_{OUT} < 2.5V$ Fixed, $V_{OUT} = \text{Adj.}$ $V_{OUT} \geq 2.5V$ Fixed
PWRGD Threshold Hysteresis	V_{PWRGD_HYS}	1.0	2.0	3.0	% V_{OUT}	
PWRGD Output Voltage Low	V_{PWRGD_L}	—	0.2	0.4	V	$I_{PWRGD\ SINK} = 1.2\text{ mA}$, $ADJ = 0V$
PWRGD Leakage	P_{PWRGD_LK}	—	1	—	nA	$V_{PWRGD} = V_{IN} = 6.0V$
PWRGD Time Delay	T_{PG}	—	200	—	μs	Rising Edge $R_{PULLUP} = 10\text{ k}\Omega$
Detect Threshold to PWRGD Active Time Delay	$T_{VDET-PWRGD}$	—	200	—	μs	V_{ADJ} or $V_{OUT} = V_{PWRGD_TH} + 20\text{ mV}$ to $V_{PWRGD_TH} - 20\text{ mV}$
Shutdown Input						
Logic High Input	$V_{SHDN-HIGH}$	45			% V_{IN}	$V_{IN} = 2.3V$ to $6.0V$
Logic Low Input	$V_{SHDN-LOW}$			15	% V_{IN}	$V_{IN} = 2.3V$ to $6.0V$
SHDN Input Leakage Current	\overline{SHDN}_{ILK}	-0.1	± 0.001	+0.1	μA	$V_{IN} = 6V$, $\overline{SHDN} = V_{IN}$, $SHDN = GND$
AC Performance						
Output Delay From \overline{SHDN}	T_{OR}		100		μs	$\overline{SHDN} = GND$ to V_{IN} $V_{OUT} = GND$ to $95\% V_R$
Output Noise	e_N	—	2.0	—	$\mu\text{V}/\sqrt{\text{Hz}}$	$I_{OUT} = 200\text{ mA}$, $f = 1\text{ kHz}$, $C_{OUT} = 10\text{ }\mu\text{F}$ (X7R Ceramic), $V_{OUT} = 2.5V$
Power Supply Ripple Rejection Ratio	PSRR	—	60	—	dB	$f = 100\text{ Hz}$, $C_{OUT} = 10\text{ }\mu\text{F}$, $I_{OUT} = 10\text{ mA}$, $V_{INAC} = 30\text{ mV pk-pk}$, $C_{IN} = 0\text{ }\mu\text{F}$

- Note 1:** The minimum V_{IN} must meet two conditions: $V_{IN} \geq 2.3V$ and $V_{IN} \geq V_{OUT(MAX)} + V_{DROPOUT(MAX)}$.
- Note 2:** V_R is the nominal regulator output voltage for the fixed cases. $V_R = 1.2V, 1.8V$, etc. V_R is the desired set point output voltage for the adjustable cases. $V_R = V_{ADJ} * ((R_1/R_2)+1)$. **Figure 4-1**.
- Note 3:** $TCV_{OUT} = (V_{OUT-HIGH} - V_{OUT-LOW}) * 10^6 / (V_R * \Delta\text{Temperature})$. $V_{OUT-HIGH}$ is the highest voltage measured over the temperature range. $V_{OUT-LOW}$ is the lowest voltage measured over the temperature range.
- Note 4:** Load regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
- Note 5:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of $V_{IN} = V_{OUTMAX} + V_{DROPOUT(MAX)}$.
- Note 6:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e., T_A, T_J, θ_{JA}). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum $+150^\circ\text{C}$ rating. Sustained junction temperatures above 150°C can impact device reliability.
- Note 7:** The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in the junction temperature over the ambient temperature is not significant.

MCP1827/MCP1827S

AC/DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise noted, $V_{IN} = V_{OUT(MAX)} + V_{DROPOUT(MAX)}$ **Note 1**, $V_R = 1.8V$ for Adjustable Output, $I_{OUT} = 1\text{ mA}$, $C_{IN} = C_{OUT} = 4.7\ \mu\text{F}$ (X7R Ceramic), $T_A = +25^\circ\text{C}$. **Boldface** type applies for junction temperatures, T_J (**Note 7**) of **-40°C to +125°C**

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Thermal Shutdown Temperature	T_{SD}	—	150	—	°C	$I_{OUT} = 100\ \mu\text{A}$, $V_{OUT} = 1.8V$, $V_{IN} = 2.8V$
Thermal Shutdown Hysteresis	ΔT_{SD}	—	10	—	°C	$I_{OUT} = 100\ \mu\text{A}$, $V_{OUT} = 1.8V$, $V_{IN} = 2.8V$

- Note 1:** The minimum V_{IN} must meet two conditions: $V_{IN} \geq 2.3V$ and $V_{IN} \geq V_{OUT(MAX)} + V_{DROPOUT(MAX)}$.
- 2:** V_R is the nominal regulator output voltage for the fixed cases. $V_R = 1.2V, 1.8V, \text{etc.}$ V_R is the desired set point output voltage for the adjustable cases. $V_R = V_{ADJ} * ((R_1/R_2)+1)$. **Figure 4-1**.
- 3:** $TCV_{OUT} = (V_{OUT-HIGH} - V_{OUT-LOW}) * 10^6 / (V_R * \Delta\text{Temperature})$. $V_{OUT-HIGH}$ is the highest voltage measured over the temperature range. $V_{OUT-LOW}$ is the lowest voltage measured over the temperature range.
- 4:** Load regulation is measured at a constant junction temperature using low duty-cycle pulse testing. Load regulation is tested over a load range from 1 mA to the maximum specified output current.
- 5:** Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below its nominal value that was measured with an input voltage of $V_{IN} = V_{OUTMAX} + V_{DROPOUT(MAX)}$.
- 6:** The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air. (i.e., T_A, T_J, θ_{JA}). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +150°C rating. Sustained junction temperatures above 150°C can impact device reliability.
- 7:** The junction temperature is approximated by soaking the device under test at an ambient temperature equal to the desired junction temperature. The test time is small enough such that the rise in the junction temperature over the ambient temperature is not significant.

TEMPERATURE SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, all limits apply for $V_{IN} = 2.3V$ to 6.0V.

Parameters	Sym.	Min.	Typ.	Max.	Units	Conditions
Temperature Ranges						
Operating Junction Temperature Range	T_J	-40	—	+125	°C	Steady State
Maximum Junction Temperature	T_J	—	—	+150	°C	Transient
Storage Temperature Range	T_A	-65	—	+150	°C	
Thermal Package Resistances						
Thermal Resistance, 5LD DDPK	θ_{JA}	—	31.2	—	°C/W	4-Layer JC51 Standard Board
Thermal Resistance, 3LD DDPK	θ_{JA}	—	31.4	—	°C/W	4-Layer JC51 Standard Board
Thermal Resistance, 5LD TO-220	θ_{JA}	—	29.3	—	°C/W	4-Layer JC51 Standard Board
Thermal Resistance, 3LD TO-220	θ_{JA}	—	29.4	—	°C/W	4-Layer JC51 Standard Board

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 mA$, Temperature = $+25^{\circ}C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 k\Omega$ To V_{IN} .

Note: Junction Temperature (T_J) is approximated by soaking the device under test to an ambient temperature equal to the desired Junction temperature. The test time is small enough such that the rise in Junction temperature over the Ambient temperature is not significant.

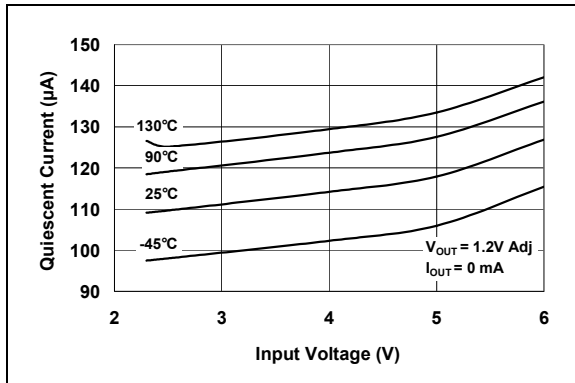


FIGURE 2-1: Quiescent Current vs. Input Voltage (1.2V Adjustable).



FIGURE 2-4: Line Regulation vs. Temperature (1.2V Adjustable).

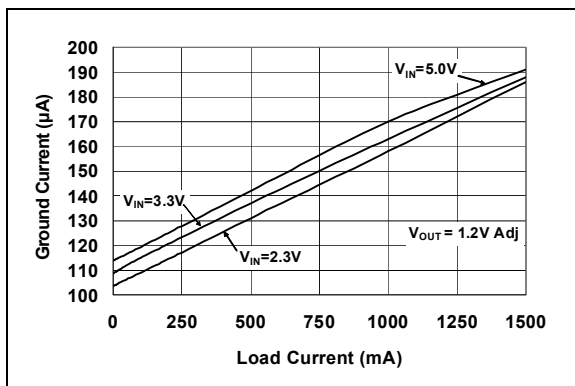


FIGURE 2-2: Ground Current vs. Load Current (1.2V Adjustable).



FIGURE 2-5: Load Regulation vs. Temperature (Adjustable Version).

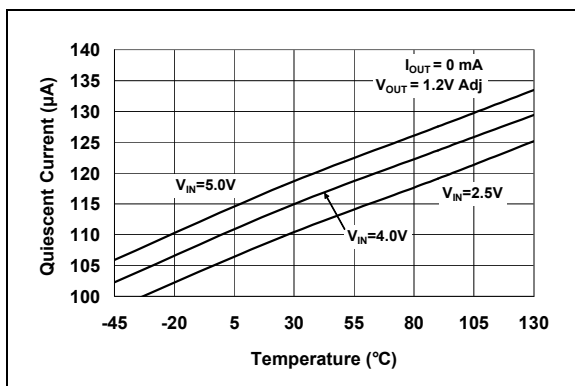


FIGURE 2-3: Quiescent Current vs. Junction Temperature (1.2V Adjustable).

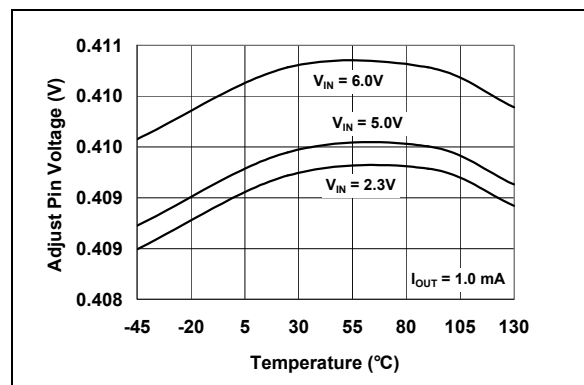


FIGURE 2-6: Adjust Pin Voltage vs. Temperature.

MCP1827/MCP1827S

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 mA$, Temperature = $+25^\circ C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 k\Omega$ To V_{IN} .



FIGURE 2-7: Dropout Voltage vs. Load Current (Adjustable Version).



FIGURE 2-10: Quiescent Current vs. Input Voltage (0.8V Fixed).



FIGURE 2-8: Dropout Voltage vs. Temperature (Adjustable Version).



FIGURE 2-11: Quiescent Current vs. Input Voltage (2.5V Fixed).



FIGURE 2-9: Power Good (PWRGD) Time Delay vs. Temperature (Adjustable Version).



FIGURE 2-12: Ground Current vs. Load Current.

MCP1827/MCP1827S

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 \text{ mA}$, Temperature = $+25^\circ C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 \text{ k}\Omega$ To V_{IN} .



FIGURE 2-13: Quiescent Current vs. Temperature.



FIGURE 2-16: Line Regulation vs. Temperature (2.5V Fixed).



FIGURE 2-14: I_{SHDN} vs. Temperature.



FIGURE 2-17: Load Regulation vs. Temperature ($V_{OUT} < 2.5V$ Fixed).



FIGURE 2-15: Line Regulation vs. Temperature (0.8V Fixed).

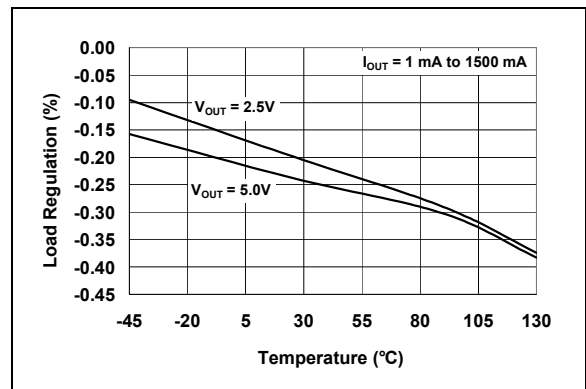


FIGURE 2-18: Load Regulation vs. Temperature ($V_{OUT} \geq 2.5V$ Fixed).

MCP1827/MCP1827S

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 mA$, Temperature = $+25^\circ C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 k\Omega$ To V_{IN} .



FIGURE 2-19: Dropout Voltage vs. Load Current.



FIGURE 2-22: Output Noise Voltage Density vs. Frequency.



FIGURE 2-20: Dropout Voltage vs. Temperature.



FIGURE 2-23: Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 1.2V$ Adj.).



FIGURE 2-21: Short Circuit Current vs. Input Voltage.



FIGURE 2-24: Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 1.2V$ Adj.).

MCP1827/MCP1827S

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 \text{ mA}$, Temperature = $+25^\circ C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 \text{ k}\Omega$ To V_{IN} .

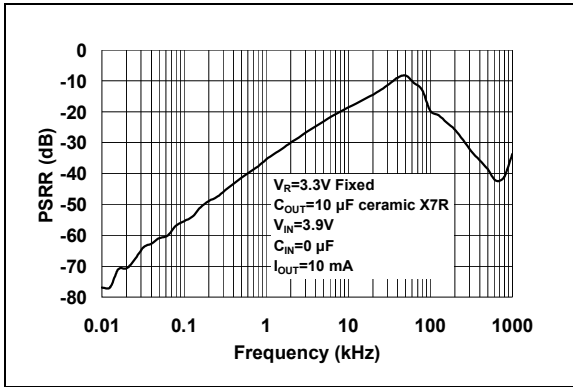


FIGURE 2-25: Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 3.3V$ Fixed).

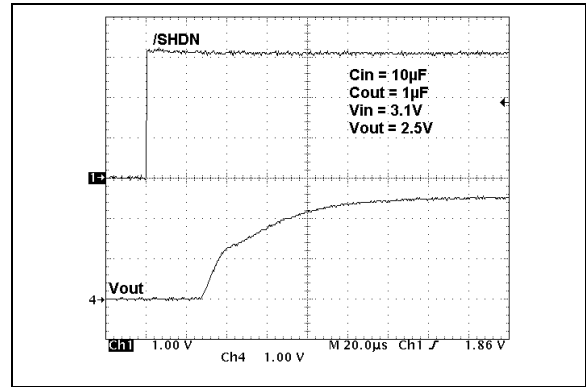


FIGURE 2-28: 2.5V (Adj.) Startup from Shutdown.

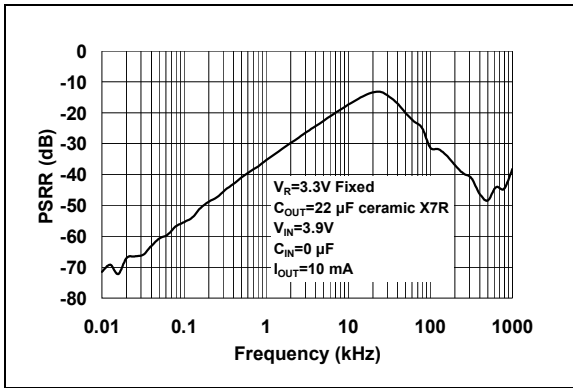


FIGURE 2-26: Power Supply Ripple Rejection (PSRR) vs. Frequency ($V_{OUT} = 3.3V$ Fixed).

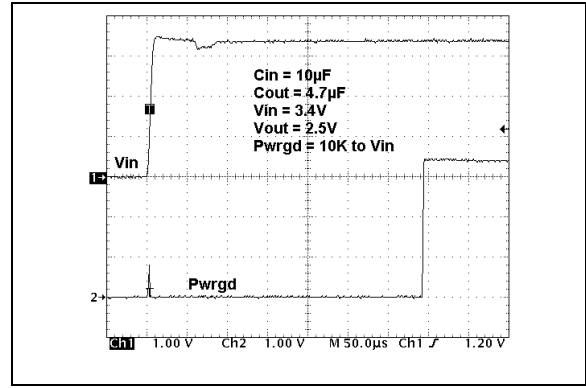


FIGURE 2-29: Power Good (PWRGD) Timing.

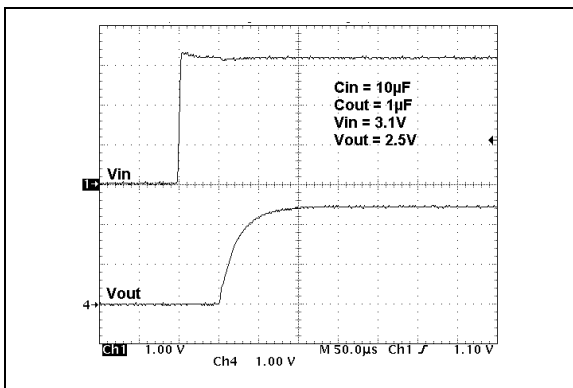


FIGURE 2-27: 2.5V (Adj.) Startup from V_{IN} .

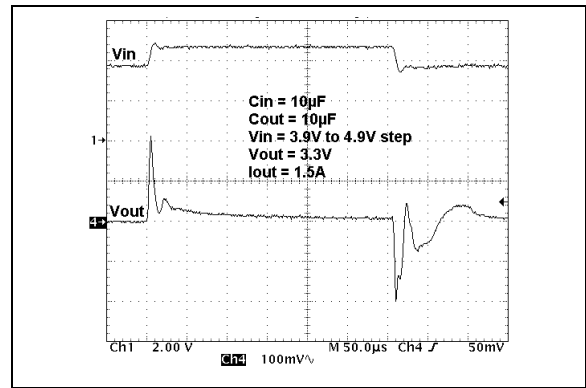


FIGURE 2-30: Dynamic Line Response (3.3V Fixed).

MCP1827/MCP1827S

Note: Unless otherwise indicated, $V_{OUT} = 1.8V$ (Adjustable), $V_{IN} = 2.8V$, $C_{OUT} = 4.7 \mu F$ Ceramic (X7R), $C_{IN} = 4.7 \mu F$ Ceramic (X7R), $I_{OUT} = 1 \text{ mA}$, Temperature = $+25^{\circ}C$, $V_{IN} = V_{OUT} + 0.6V$, $R_{PWRGD} = 10 \text{ k}\Omega$ To V_{IN} .



FIGURE 2-31: Dynamic Load Response (3.3V Fixed, 10 mA to 1500 mA).



FIGURE 2-32: Dynamic Load Response (3.3V Fixed, 100 mA to 1500 mA).

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in [Table 3-1](#).

TABLE 3-1: PIN FUNCTION TABLE

3-Pin Fixed Output	5-Pin Fixed Output	Adjustable Output	Name	Description
—	1	1	$\overline{\text{SHDN}}$	Shutdown Control Input (active-low)
1	2	2	V_{IN}	Input Voltage Supply
2	3	3	GND	Ground
3	4	4	V_{OUT}	Regulated Output Voltage
—	5	—	PWRGD	Power Good Output
—	—	5	ADJ	Voltage Adjust/Sense Input
Pad	Pad	Pad	EP	Exposed Pad of the Package (ground potential)

3.1 Input Voltage Supply (V_{IN})

Connect the unregulated or regulated input voltage source to V_{IN} . If the input voltage source is located several inches away from the LDO, or the input source is a battery, it is recommended that an input capacitor be used. A typical input capacitance value of 1 μF to 10 μF should be sufficient for most applications.

3.2 Shutdown Control Input ($\overline{\text{SHDN}}$)

The $\overline{\text{SHDN}}$ input is used to turn the LDO output voltage on and off. When the $\overline{\text{SHDN}}$ input is at a logic-high level, the LDO output voltage is enabled. When the $\overline{\text{SHDN}}$ input is pulled to a logic-low level, the LDO output voltage is disabled. When the $\overline{\text{SHDN}}$ input is pulled low, the PWRGD output also goes low and the LDO enters a low quiescent current shutdown state where the typical quiescent current is 0.1 μA .

3.3 Ground (GND)

Connect the GND pin of the LDO to a quiet circuit ground. This will help the LDO power supply rejection ratio and noise performance. The ground pin of the LDO only conducts the quiescent current of the LDO (typically 120 μA), so a heavy trace is not required. For applications have switching or noisy inputs tie the GND pin to the return of the output capacitor. Ground planes help lower inductance and voltage spikes caused by fast transient load currents and are recommended for applications that are subjected to fast load transients.

3.4 Power Good Output (PWRGD)

The PWRGD output is an open-drain output used to indicate when the LDO output voltage is within 92% (typically) of its nominal regulation value. The PWRGD threshold has a typical hysteresis value of 2%. The PWRGD output is delayed by 200 μs (typical) from the time the LDO output is within 92% + 3% (max hysteresis) of the regulated output value on power-up. This delay time is internally fixed.

3.5 Output Voltage Adjust Input (ADJ)

For adjustable applications, the output voltage is connected to the ADJ input through a resistor divider that sets the output voltage regulation value. This provides the user the capability to set the output voltage to any value they desire within the 0.8V to 5.0V range of the device.

3.6 Regulated Output Voltage (V_{OUT})

The V_{OUT} pin is the regulated output voltage of the LDO. A minimum output capacitance of 1.0 μF is required for LDO stability. The MCP1827/MCP1827S is stable with ceramic, tantalum and aluminum-electrolytic capacitors. See [Section 4.3 "Output Capacitor"](#) for output capacitor selection guidance.

3.7 Exposed Pad (EP)

The DDPACK and TO-220 package have an exposed tab on the package. A heat sink may be mount to the tab to aid in the removal of heat from the package during operation. The exposed tab is at the ground potential of the LDO.

MCP1827/MCP1827S

4.0 DEVICE OVERVIEW

The MCP1827/MCP1827S is a high output current, Low Dropout (LDO) voltage regulator. The low dropout voltage of 330 mV typical at 1.5A of current makes it ideal for battery-powered applications. Unlike other high output current LDOs, the MCP1827/MCP1827S only draws a maximum of 220 μ A of quiescent current. The MCP1827 has a shutdown control input and a power good output.

4.1 LDO Output Voltage

The 5-pin MCP1827 LDO is available with either a fixed output voltage or an adjustable output voltage. The output voltage range is 0.8V to 5.0V for both versions. The 3-pin MCP1827S LDO is available as a fixed voltage device.

4.1.1 ADJUST INPUT

The adjustable version of the MCP1827 uses the ADJ pin (pin 5) to get the output voltage feedback for output voltage regulation. This allows the user to set the output voltage of the device with two external resistors. The nominal voltage for ADJ is 0.41V.

Figure 4-1 shows the adjustable version of the MCP1827. Resistors R_1 and R_2 form the resistor divider network necessary to set the output voltage. With this configuration, the equation for setting V_{OUT} is:

EQUATION 4-1:

$$V_{OUT} = V_{ADJ} \left(\frac{R_1 + R_2}{R_2} \right)$$

Where:

V_{OUT} = LDO Output Voltage

V_{ADJ} = ADJ Pin Voltage (typically 0.41V)

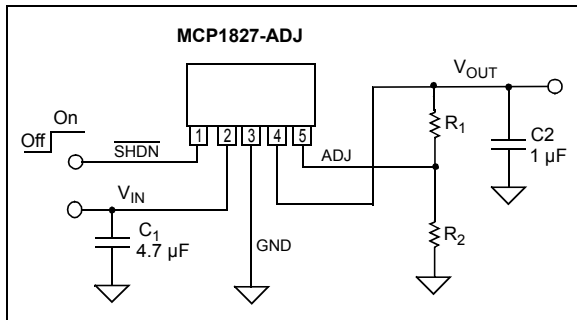


FIGURE 4-1: Typical adjustable output voltage application circuit.

The allowable resistance value range for resistor R_2 is from 10 k Ω to 200 k Ω . Solving the equation for R_1 yields the following equation:

EQUATION 4-2:

$$R_1 = R_2 \left(\frac{V_{OUT} - V_{ADJ}}{V_{ADJ}} \right)$$

Where:

V_{OUT} = LDO Output Voltage

V_{ADJ} = ADJ Pin Voltage (typically 0.41V)

4.2 Output Current and Current Limiting

The MCP1827/MCP1827S LDO is tested and ensured to supply a minimum of 1.5A of output current. The MCP1827/MCP1827S has no minimum output load, so the output load current can go to 0 mA and the LDO will continue to regulate the output voltage to within tolerance.

The MCP1827/MCP1827S also incorporates an output current limit. If the output voltage falls below 0.7V due to an overload condition (usually represents a shorted load condition), the output current is limited to 2.2A (typical). If the overload condition is a soft overload, the MCP1827/MCP1827S will supply higher load currents of up to 3A. The MCP1827/MCP1827S should not be operated in this condition continuously as it may result in failure of the device. However, this does allow for device usage in applications that have higher pulsed load currents having an average output current value of 1.5A or less.

Output overload conditions may also result in an over-temperature shutdown of the device. If the junction temperature rises above 150°C, the LDO will shut down the output voltage. See [Section 4.8 "Overtemperature Protection"](#) for more information on overtemperature shutdown.

4.3 Output Capacitor

The MCP1827/MCP1827S requires a minimum output capacitance of 1 μ F for output voltage stability. Ceramic capacitors are recommended because of their size, cost and environmental robustness qualities.

Aluminum-electrolytic and tantalum capacitors can be used on the LDO output as well. The Equivalent Series Resistance (ESR) of the electrolytic output capacitor must be no greater than 1 ohm. The output capacitor should be located as close to the LDO output as is practical. Ceramic materials X7R and X5R have low temperature coefficients and are well within the acceptable ESR range required. A typical 1 μ F X7R 0805 capacitor has an ESR of 50 milli-ohms.

Larger LDO output capacitors can be used with the MCP1827/MCP1827S to improve dynamic performance and power supply ripple rejection. A maximum of 22 μF is recommended. Aluminum-electrolytic capacitors are not recommended for low-temperature applications of $\leq 25^\circ\text{C}$.

4.4 Input Capacitor

Low input source impedance is necessary for the LDO output to operate properly. When operating from batteries, or in applications with long lead length (> 10 inches) between the input source and the LDO, some input capacitance is recommended. A minimum of 1.0 μF to 4.7 μF is recommended for most applications.

For applications that have output step load requirements, the input capacitance of the LDO is very important. The input capacitance provides the LDO with a good local low-impedance source to pull the transient currents from in order to respond quickly to the output load step. For good step response performance, the input capacitor should be of equivalent (or higher) value than the output capacitor. The capacitor should be placed as close to the input of the LDO as is practical. Larger input capacitors will also help reduce any high-frequency noise on the input and output of the LDO and reduce the effects of any inductance that exists between the input source voltage and the input capacitance of the LDO.

4.5 Power Good Output (PWRGD)

The PWRGD output is used to indicate when the output voltage of the LDO is within 92% (typical value, see [Section 1.0 “Electrical Characteristics”](#) for Minimum and Maximum specifications) of its nominal regulation value.

As the output voltage of the LDO rises, the PWRGD output will be held low until the output voltage has exceeded the power good threshold plus the hysteresis value. Once this threshold has been exceeded, the power good time delay is started (shown as T_{PG} in the Electrical Characteristics table). The power good time delay is fixed at 200 μs (typical). After the time delay period, the PWRGD output will go high, indicating that the output voltage is stable and within regulation limits.

If the output voltage of the LDO falls below the power good threshold, the power good output will transition low. The power good circuitry has a 170 μs delay when detecting a falling output voltage, which helps to increase noise immunity of the power good output and avoid false triggering of the power good output during fast output transients. See [Figure 4-2](#) for power good timing characteristics.

When the LDO is put into Shutdown mode using the $\overline{\text{SHDN}}$ input, the power good output is pulled low immediately, indicating that the output voltage will be out of regulation. The timing diagram for the power good output when using the shutdown input is shown in [Figure 4-3](#).

The power good output is an open-drain output that can be pulled up to any voltage that is equal to or less than the LDO input voltage. This output is capable of sinking 1.2 mA ($V_{PWRGD} < 0.4\text{V}$ maximum).



FIGURE 4-2: Power Good Timing.



FIGURE 4-3: Power Good Timing from Shutdown.

4.6 Shutdown Input ($\overline{\text{SHDN}}$)

The $\overline{\text{SHDN}}$ input is an active-low input signal that turns the LDO on and off. The $\overline{\text{SHDN}}$ threshold is a percentage of the input voltage. The typical value of this shutdown threshold is 30% of V_{IN} , with minimum and maximum limits over the entire operating temperature range of 45% and 15%, respectively.

MCP1827/MCP1827S

The $\overline{\text{SHDN}}$ input will ignore low-going pulses (pulses meant to shut down the LDO) that are up to 400 ns in pulse width. If the shutdown input is pulled low for more than 400 ns, the LDO will enter Shutdown mode. This small bit of filtering helps to reject any system noise spikes on the shutdown input signal.

On the rising edge of the $\overline{\text{SHDN}}$ input, the shutdown circuitry has a 30 μs delay before allowing the LDO output to turn on. This delay helps to reject any false turn-on signals or noise on the $\overline{\text{SHDN}}$ input signal. After the 30 μs delay, the LDO output enters its soft-start period as it rises from 0V to its final regulation value. If the $\overline{\text{SHDN}}$ input signal is pulled low during the 30 μs delay period, the timer will be reset and the delay time will start over again on the next rising edge of the $\overline{\text{SHDN}}$ input. The total time from the $\overline{\text{SHDN}}$ input going high (turn-on) to the LDO output being in regulation is typically 100 μs . See Figure 4-4 for a timing diagram of the $\overline{\text{SHDN}}$ input.



FIGURE 4-4: Shutdown Input Timing Diagram.

4.7 Dropout Voltage and Undervoltage Lockout

Dropout voltage is defined as the input-to-output voltage differential at which the output voltage drops 2% below the nominal value that was measured with a $V_R + 0.6\text{V}$ differential applied. The MCP1827/MCP1827S LDO has a very low dropout voltage specification of 330 mV (typical) at 1.5A of output current. See Section 1.0 “Electrical Characteristics” for maximum dropout voltage specifications.

The MCP1827/MCP1827S LDO operates across an input voltage range of 2.3V to 6.0V and incorporates input Undervoltage Lockout (UVLO) circuitry that keeps the LDO output voltage off until the input voltage reaches a minimum of 2.18V (typical) on the rising edge of the input voltage. As the input voltage falls, the LDO output will remain on until the input voltage level reaches 2.04V (typical).

Since the MCP1827/MCP1827S LDO undervoltage lockout activates at 2.04V as the input voltage is falling, the dropout voltage specification does not apply for output voltages that are less than 1.9V.

For high-current applications, voltage drops across the PCB traces must be taken into account. The trace resistances can cause significant voltage drops between the input voltage source and the LDO. For applications with input voltages near 2.3V, these PCB trace voltage drops can sometimes lower the input voltage enough to trigger a shutdown due to undervoltage lockout.

4.8 Overtemperature Protection

The MCP1827/MCP1827S LDO has temperature-sensing circuitry to prevent the junction temperature from exceeding approximately 150°C. If the LDO junction temperature does reach 150°C, the LDO output will be turned off until the junction temperature cools to approximately 140°C, at which point the LDO output will automatically resume normal operation. If the internal power dissipation continues to be excessive, the device will again shut off. The junction temperature of the die is a function of power dissipation, ambient temperature and package thermal resistance. See Section 5.0 “Application Circuits/Issues” for more information on LDO power dissipation and junction temperature.

5.0 APPLICATION CIRCUITS/ISSUES

5.1 Typical Application

The MCP1827/MCP1827S is used for applications that require high LDO output current and a power good output.



FIGURE 5-1: Typical Application Circuit.

5.1.1 APPLICATION CONDITIONS

Package Type =	TO-220-5
Input Voltage Range =	3.3V ± 5%
V_{IN} maximum =	3.465V
V_{IN} minimum =	3.135V
$V_{DROPOUT (max)}$ =	0.600V
V_{OUT} (typical) =	2.5V
I_{OUT} =	1.5A maximum
P_{DISS} (typical) =	1.2W
Temperature Rise =	35.2°C

5.2 Power Calculations

5.2.1 POWER DISSIPATION

The internal power dissipation within the MCP1827/MCP1827S is a function of input voltage, output voltage, output current and quiescent current. Equation 5-1 can be used to calculate the internal power dissipation for the LDO.

EQUATION 5-1:

$$P_{LDO} = (V_{IN(MAX)} - V_{OUT(MIN)}) \times I_{OUT(MAX)}$$

Where:

P_{LDO} =	LDO Pass device internal power dissipation
$V_{IN(MAX)}$ =	Maximum input voltage
$V_{OUT(MIN)}$ =	LDO minimum output voltage

In addition to the LDO pass element power dissipation, there is power dissipation within the MCP1827/MCP1827S as a result of quiescent or ground current. The power dissipation as a result of the ground current can be calculated using the following equation:

EQUATION 5-2:

$$P_{I(GND)} = V_{IN(MAX)} \times I_{VIN}$$

Where:

$P_{I(GND)}$ =	Power dissipation due to the quiescent current of the LDO
$V_{IN(MAX)}$ =	Maximum input voltage
I_{VIN} =	Current flowing in the V_{IN} pin with no LDO output current (LDO quiescent current)

The total power dissipated within the MCP1827/MCP1827S is the sum of the power dissipated in the LDO pass device and the $P_{I(GND)}$ term. Because of the CMOS construction, the typical I_{GND} for the MCP1827/MCP1827S is 120 µA. Operating at a maximum of 3.465V results in a power dissipation of 0.49 milli-Watts. For most applications, this is small compared to the LDO pass device power dissipation and can be neglected.

The maximum continuous operating junction temperature specified for the MCP1827/MCP1827S is +125°C. To estimate the internal junction temperature of the MCP1827/MCP1827S, the total internal power dissipation is multiplied by the thermal resistance from junction to ambient ($R\theta_{JA}$) of the device. The thermal resistance from junction to ambient for the TO-220-5 package is estimated at 29.3° C/W.

EQUATION 5-3:

$$T_{J(MAX)} = P_{TOTAL} \times R\theta_{JA} + T_{A(MAX)}$$

$T_{J(MAX)}$ =	Maximum continuous junction temperature
P_{TOTAL} =	Total device power dissipation
$R\theta_{JA}$ =	Thermal resistance from junction to ambient
$T_{A(MAX)}$ =	Maximum ambient temperature

MCP1827/MCP1827S

The maximum power dissipation capability for a package can be calculated given the junction-to-ambient thermal resistance and the maximum ambient temperature for the application. Equation 5-4 can be used to determine the package maximum internal power dissipation.

EQUATION 5-4:

$$P_{D(MAX)} = \frac{(T_{J(MAX)} - T_{A(MAX)})}{R\theta_{JA}}$$

$P_{D(MAX)}$ = Maximum device power dissipation

$T_{J(MAX)}$ = maximum continuous junction temperature

$T_{A(MAX)}$ = maximum ambient temperature

$R\theta_{JA}$ = Thermal resistance from junction to ambient

EQUATION 5-5:

$$T_{J(RISE)} = P_{D(MAX)} \times R\theta_{JA}$$

$T_{J(RISE)}$ = Rise in device junction temperature over the ambient temperature

$P_{D(MAX)}$ = Maximum device power dissipation

$R\theta_{JA}$ = Thermal resistance from junction to ambient

EQUATION 5-6:

$$T_J = T_{J(RISE)} + T_A$$

T_J = Junction temperature

$T_{J(RISE)}$ = Rise in device junction temperature over the ambient temperature

T_A = Ambient temperature

5.3 Typical Application

Internal power dissipation, junction temperature rise, junction temperature and maximum power dissipation is calculated in the following example. The power dissipation as a result of ground current is small enough to be neglected.

5.3.1 POWER DISSIPATION EXAMPLE

Package

Package Type = TO-220-5

Input Voltage

$V_{IN} = 3.3V \pm 5\%$

LDO Output Voltage and Current

$V_{OUT} = 2.5V$

$I_{OUT} = 1.5A$

Maximum Ambient Temperature

$T_{A(MAX)} = 60^\circ C$

Internal Power Dissipation

$P_{LDO(MAX)} = (V_{IN(MAX)} - V_{OUT(MIN)}) \times I_{OUT(MAX)}$

$P_{LDO} = ((3.3V \times 1.05) - (2.5V \times 0.975))$
 $\times 1.5A$

$P_{LDO} = 1.54 \text{ Watts}$

5.3.1.1 Device Junction Temperature Rise

The internal junction temperature rise is a function of internal power dissipation and the thermal resistance from junction-to-ambient for the application. The thermal resistance from junction-to-ambient ($R\theta_{JA}$) is derived from EIA/JEDEC standards for measuring thermal resistance. The EIA/JEDEC specification is JESD51. The standard describes the test method and board specifications for measuring the thermal resistance from junction to ambient. The actual thermal resistance for a particular application can vary depending on many factors such as copper area and thickness. Refer to AN792, "A Method to Determine How Much Power a SOT23 Can Dissipate in an Application" (DS00792), for more information regarding this subject.

$T_{J(RISE)} = P_{TOTAL} \times R\theta_{JA}$

$T_{J(RISE)} = 1.54 \text{ W} \times 29.3^\circ \text{ C/W}$

$T_{J(RISE)} = 45.12^\circ C$

5.3.1.2 Junction Temperature Estimate

To estimate the internal junction temperature, the calculated temperature rise is added to the ambient or offset temperature. For this example, the worst-case junction temperature is estimated below:

$$T_J = T_{J(\text{RISE})} + T_{A(\text{MAX})}$$

$$T_J = 45.12^\circ\text{C} + 60.0^\circ\text{C}$$

$$T_J = 105.12^\circ\text{C}$$

As you can see from the result, this application will be operating within the maximum operating junction temperature of 125°C.

5.3.1.3 Maximum Package Power Dissipation at 60°C Ambient Temperature

TO-220-5 (29.3° C/W R_{θJA}):

$$P_{D(\text{MAX})} = (125^\circ\text{C} - 60^\circ\text{C}) / 29.3^\circ\text{C/W}$$

$$P_{D(\text{MAX})} = 2.218\text{W}$$

DDPAK-5 (31.2°C/Watt R_{θJA}):

$$P_{D(\text{MAX})} = (125^\circ\text{C} - 60^\circ\text{C}) / 31.2^\circ\text{C/W}$$

$$P_{D(\text{MAX})} = 2.083\text{W}$$

From this table you can see the difference in maximum allowable power dissipation between the TO-220-5 package and the DDPACK-5 package.

MCP1827/MCP1827S

6.0 PACKAGING INFORMATION

6.1 Package Marking Information

3-Lead DDPAK (MCP1827S)



Example:



3-Lead TO-220 (MCP1827S)



Example:



5-Lead DDPAK (Fixed) (MCP1827)



Example:



5-Lead TO-220 (Adj) (MCP1827)



Example:



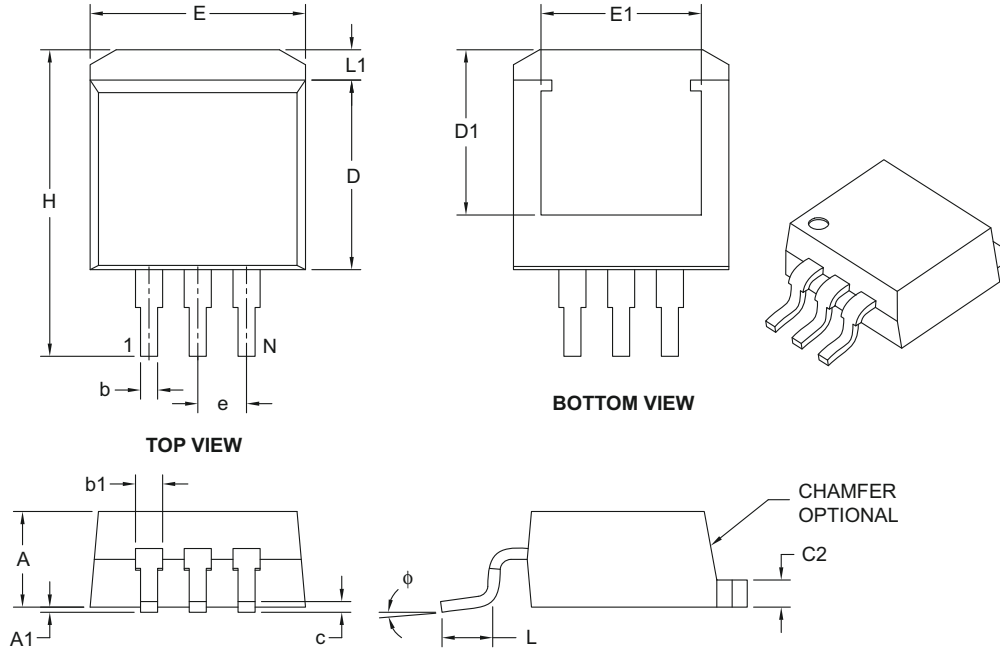
Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	(e3)	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

MCP1827/MCP1827S

3-Lead Plastic (EB) [DDPAK]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Number of Pins	N	3		
Pitch	e	.100 BSC		
Overall Height	A	.160	–	.190
Standoff §	A1	.000	–	.010
Overall Width	E	.380	–	.420
Exposed Pad Width	E1	.245	–	–
Molded Package Length	D	.330	–	.380
Overall Length	H	.549	–	.625
Exposed Pad Length	D1	.270	–	–
Lead Thickness	c	.014	–	.029
Pad Thickness	C2	.045	–	.065
Lower Lead Width	b	.020	–	.039
Upper Lead Width	b1	.045	–	.070
Foot Length	L	.068	–	.110
Pad Length	L1	–	–	.067
Foot Angle	φ	0°	–	8°

Notes:

- § Significant Characteristic.
- Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-011B

MCP1827/MCP1827S

3-Lead Plastic (EB) [DDPAK]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Contact Pitch	E	.100 BSC		
Pad Width	X2			.423
Pad Length	Y2			.327
Contact Pad Spacing	C1		.252	
Contact Pad Width (X3)	X1			.041
Contact Pad Length (X3)	Y1			.157

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2011A

MCP1827/MCP1827S

3-Lead Plastic Transistor Outline (AB) [TO-220]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Number of Pins	N	3		
Pitch	e	.100 BSC		
Overall Pin Pitch	e1	.200 BSC		
Overall Height	A	.140	–	.190
Tab Thickness	A1	.020	–	.055
Base to Lead	A2	.080	–	.115
Overall Width	E	.357	–	.420
Mounting Hole Center	Q	.100	–	.120
Overall Length	D	.560	–	.650
Molded Package Length	D1	.330	–	.355
Tab Length	H1	.230	–	.270
Mounting Hole Diameter	φP	.139	–	.156
Lead Length	L	.500	–	.580
Lead Shoulder	L1	–	–	.250
Lead Thickness	c	.012	–	.024
Lead Width	b	.015	.027	.040
Shoulder Width	b2	.045	.057	.070

Notes:

- Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-034B

MCP1827/MCP1827S

5-Lead Plastic (ET) [DDPAK]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Number of Pins	N	5		
Pitch	e	.067 BSC		
Overall Height	A	.160	–	.190
Standoff §	A1	.000	–	.010
Overall Width	E	.380	–	.420
Exposed Pad Width	E1	.245	–	–
Molded Package Length	D	.330	–	.380
Overall Length	H	.549	–	.625
Exposed Pad Length	D1	.270	–	–
Lead Thickness	c	.014	–	.029
Pad Thickness	C2	.045	–	.065
Lead Width	b	.020	–	.039
Foot Length	L	.068	–	.110
Pad Length	L1	–	–	.067
Foot Angle	ϕ	0°	–	8°

Notes:

- § Significant Characteristic.
- Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

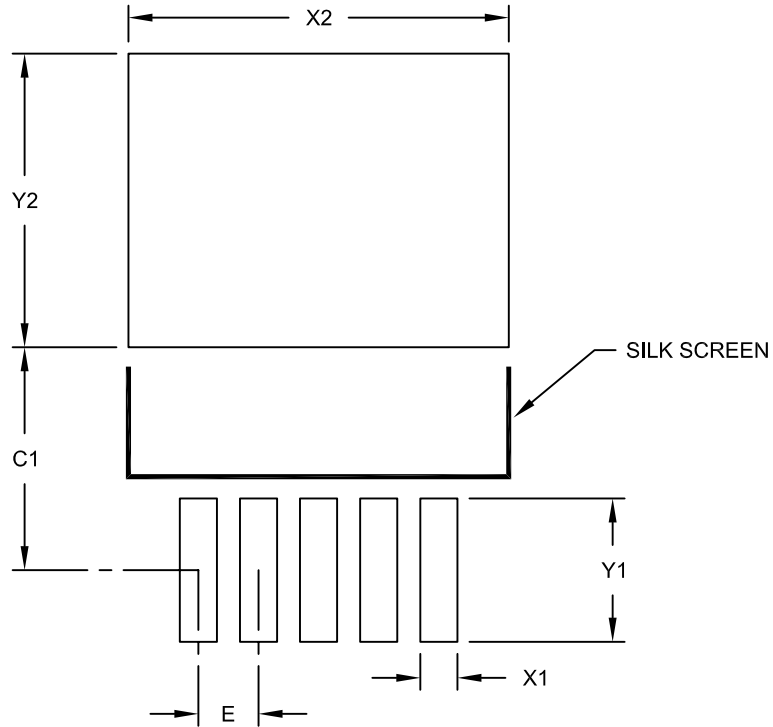
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-012B

MCP1827/MCP1827S

5-Lead Plastic (ET) [DDPAK]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

		Units	INCHES		
		Dimension Limits	MIN	NOM	MAX
Contact Pitch	E		.067 BSC		
Optional Center Pad Width	X2				.423
Optional Center Pad Length	Y2				.327
Contact Pad Spacing	C1			.248	
Contact Pad Width (X5)	X1				.041
Contact Pad Length (X5)	Y1				.159

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2012A

MCP1827/MCP1827S

5-Lead Plastic Transistor Outline (AT) [TO-220]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Number of Pins	N	5		
Pitch	e	.067 BSC		
Overall Pin Pitch	e1	.268 BSC		
Overall Height	A	.140	–	.190
Overall Width	E	.380	–	.420
Overall Length	D	.560	–	.650
Molded Package Length	D1	.330	–	.355
Tab Length	H1	.204	–	.293
Tab Thickness	A1	.020	–	.055
Mounting Hole Center	Q	.100	–	.120
Mounting Hole Diameter	φP	.139	–	.156
Lead Length	L	.482	–	.590
Base to Bottom of Lead	A2	.080	–	.115
Lead Thickness	c	.012	–	.025
Lead Width	b	.015	.027	.040

Notes:

- Dimensions D and E do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-036B

APPENDIX A: REVISION HISTORY

Revision D (March 2013)

The following is the list of modifications:

- Updated the value of $V_{\text{DROPOUT (max)}}$ in [Section 5.1 “Typical Application”](#).
- Updated the 5-lead DDPACK (MCP1827) information in the [Product Identification System](#) section.

Revision C (February 2007)

- [Figure 2-22](#): Revised label on Y-axis.
- [Section 2.0 “Typical Performance Curves”](#): Added note on Junction Temperature.
- Pages 9-14: Revised notes.

Revision B (September 2006)

- Correction to maximum Dropout Voltage in Section 1.0.
- Added additional graphs in Section 2.0.
- Added disclaimer to package outline drawings.

Revision A (July 2006)

- Original Release of this Document.

MCP1827/MCP1827S

NOTES:

MCP1827/MCP1827S

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

<u>PART NO.</u>	<u>XX</u>	<u>X</u>	<u>X</u>	<u>X/</u>	<u>XX</u>
Device	Output Voltage	Feature Code	Tolerance	Temp.	Package
<p>Device:</p> <p>MCP1827: 1.5A Low Dropout Regulator MCP1827T: 1.5A Low Dropout Regulator Tape and Reel MCP1827S: 1.5A Low Dropout Regulator MCP1827ST: 1.5A Low Dropout Regulator Tape and Reel</p> <p>Output Voltage *:</p> <p>08 = 0.8V "Standard" 12 = 1.2V "Standard" 18 = 1.8V "Standard" 25 = 2.5V "Standard" 30 = 3.0V "Standard" 33 = 3.3V "Standard" 50 = 5.0V "Standard"</p> <p>*Contact factory for other output voltage options</p> <p>Extra Feature Code: 0 = Fixed</p> <p>Tolerance: 2 = 2.0% (Standard)</p> <p>Temperature: E = -40°C to +125°C</p> <p>Package Type:</p> <p>AB = Plastic Transistor Outline, TO-220, 3-lead AT = Plastic Transistor Outline, TO-220, 5-lead EB = Plastic, DDPAK, 3-lead ET = Plastic, DDPAK, 5-lead</p>					
<p>Examples:</p> <p>a) MCP1827-0802E/AT: 0.8V LDO Regulator 5LD TO-220</p> <p>b) MCP1827-1002E/ET: 1.0V LDO Regulator 5LD DDPAK</p> <p>c) MCP1827-1202E/AT: 1.2V LDO Regulator 5LD TO-220</p> <p>d) MCP1827-1802E/AT: 1.8V LDO Regulator 5LD TO-220</p> <p>e) MCP1827-2502E/ET: 2.5V LDO Regulator 5LD DDPAK</p> <p>f) MCP1827-3002E/ET: 3.0V LDO Regulator 5LD DDPAK</p> <p>g) MCP1827-3302E/AT: 3.3V LDO Regulator 5LD TO-220</p> <p>h) MCP1827-5002E/ET: 5.0V LDO Regulator 5LD DDPAK</p> <p>i) MCP1827-ADJE/AT: ADJ LDO Regulator 5LD TO-220</p> <p>j) MCP1827-ADJE/ET: ADJ LDO Regulator 5LD DDPAK</p> <p>a) MCP1827S-0802E/EB: 0.8V LDO Regulator 3LD DDPAK</p> <p>b) MCP1827S-0802E/AB: 0.8V LDO Regulator 3LD TO-220</p> <p>c) MCP1827S-1002E/EB: 1.0V LDO Regulator 3LD DDPAK</p> <p>d) MCP1827S-1202E/AB: 1.2V LDO Regulator 3LD TO-220</p> <p>e) MCP1827S-1802E/EB: 1.8V LDO Regulator 3LD DDPAK</p> <p>f) MCP1827S-2502E/EB: 2.5V LDO Regulator 3LD DDPAK</p> <p>g) MCP1827S-2502E/EB: 3.0V LDO Regulator 3LD DDPAK</p> <p>h) MCP1827S-3302E/AB: 3.3V LDO Regulator 3LD TO-220</p> <p>i) MCP1827S-5002E/EB: 5.0V LDO Regulator 3LD DDPAK</p>					

MCP1827/MCP1827S

NOTES:

Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as “unbreakable.”

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE. Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights.

Trademarks

The Microchip name and logo, the Microchip logo, dsPIC, FlashFlex, KEELOQ, KEELOQ logo, MPLAB, PIC, PICmicro, PICSTART, PIC³² logo, rPIC, SST, SST Logo, SuperFlash and UNI/O are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

FilterLab, Hampshire, HI-TECH C, Linear Active Thermistor, MTP, SEEVAL and The Embedded Control Solutions Company are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Silicon Storage Technology is a registered trademark of Microchip Technology Inc. in other countries.

Analog-for-the-Digital Age, Application Maestro, BodyCom, chipKIT, chipKIT logo, CodeGuard, dsPICDEM, dsPICDEM.net, dsPICworks, dsSPEAK, ECAN, ECONOMONITOR, FanSense, HI-TIDE, In-Circuit Serial Programming, ICSP, Mindi, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, mTouch, Omniclient Code Generation, PICC, PICC-18, PICDEM, PICDEM.net, PICkit, PICTail, REAL ICE, rLAB, Select Mode, SQI, Serial Quad I/O, Total Endurance, TSHARC, UniWinDriver, WiperLock, ZENA and Z-Scale are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

GestIC and ULPP are registered trademarks of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2006-2013, Microchip Technology Incorporated, Printed in the U.S.A., All Rights Reserved.

 Printed on recycled paper.

ISBN: 9781620770412

QUALITY MANAGEMENT SYSTEM
CERTIFIED BY DNV
== ISO/TS 16949 ==

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC[®] MCUs and dsPIC[®] DSCs, KEELOQ[®] code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.



MICROCHIP

Worldwide Sales and Service

AMERICAS

Corporate Office
2355 West Chandler Blvd.
Chandler, AZ 85224-6199
Tel: 480-792-7200
Fax: 480-792-7277
Technical Support:
<http://www.microchip.com/support>
Web Address:
www.microchip.com

Atlanta
Duluth, GA
Tel: 678-957-9614
Fax: 678-957-1455

Boston
Westborough, MA
Tel: 774-760-0087
Fax: 774-760-0088

Chicago
Itasca, IL
Tel: 630-285-0071
Fax: 630-285-0075

Cleveland
Independence, OH
Tel: 216-447-0464
Fax: 216-447-0643

Dallas
Addison, TX
Tel: 972-818-7423
Fax: 972-818-2924

Detroit
Farmington Hills, MI
Tel: 248-538-2250
Fax: 248-538-2260

Indianapolis
Noblesville, IN
Tel: 317-773-8323
Fax: 317-773-5453

Los Angeles
Mission Viejo, CA
Tel: 949-462-9523
Fax: 949-462-9608

Santa Clara
Santa Clara, CA
Tel: 408-961-6444
Fax: 408-961-6445

Toronto
Mississauga, Ontario,
Canada
Tel: 905-673-0699
Fax: 905-673-6509

ASIA/PACIFIC

Asia Pacific Office
Suites 3707-14, 37th Floor
Tower 6, The Gateway
Harbour City, Kowloon
Hong Kong
Tel: 852-2401-1200
Fax: 852-2401-3431

Australia - Sydney
Tel: 61-2-9868-6733
Fax: 61-2-9868-6755

China - Beijing
Tel: 86-10-8569-7000
Fax: 86-10-8528-2104

China - Chengdu
Tel: 86-28-8665-5511
Fax: 86-28-8665-7889

China - Chongqing
Tel: 86-23-8980-9588
Fax: 86-23-8980-9500

China - Hangzhou
Tel: 86-571-2819-3187
Fax: 86-571-2819-3189

China - Hong Kong SAR
Tel: 852-2943-5100
Fax: 852-2401-3431

China - Nanjing
Tel: 86-25-8473-2460
Fax: 86-25-8473-2470

China - Qingdao
Tel: 86-532-8502-7355
Fax: 86-532-8502-7205

China - Shanghai
Tel: 86-21-5407-5533
Fax: 86-21-5407-5066

China - Shenyang
Tel: 86-24-2334-2829
Fax: 86-24-2334-2393

China - Shenzhen
Tel: 86-755-8864-2200
Fax: 86-755-8203-1760

China - Wuhan
Tel: 86-27-5980-5300
Fax: 86-27-5980-5118

China - Xian
Tel: 86-29-8833-7252
Fax: 86-29-8833-7256

China - Xiamen
Tel: 86-592-2388138
Fax: 86-592-2388130

China - Zhuhai
Tel: 86-756-3210040
Fax: 86-756-3210049

ASIA/PACIFIC

India - Bangalore
Tel: 91-80-3090-4444
Fax: 91-80-3090-4123

India - New Delhi
Tel: 91-11-4160-8631
Fax: 91-11-4160-8632

India - Pune
Tel: 91-20-2566-1512
Fax: 91-20-2566-1513

Japan - Osaka
Tel: 81-6-6152-7160
Fax: 81-6-6152-9310

Japan - Tokyo
Tel: 81-3-6880-3770
Fax: 81-3-6880-3771

Korea - Daegu
Tel: 82-53-744-4301
Fax: 82-53-744-4302

Korea - Seoul
Tel: 82-2-554-7200
Fax: 82-2-558-5932 or
82-2-558-5934

Malaysia - Kuala Lumpur
Tel: 60-3-6201-9857
Fax: 60-3-6201-9859

Malaysia - Penang
Tel: 60-4-227-8870
Fax: 60-4-227-4068

Philippines - Manila
Tel: 63-2-634-9065
Fax: 63-2-634-9069

Singapore
Tel: 65-6334-8870
Fax: 65-6334-8850

Taiwan - Hsin Chu
Tel: 886-3-5778-366
Fax: 886-3-5770-955

Taiwan - Kaohsiung
Tel: 886-7-213-7828
Fax: 886-7-330-9305

Taiwan - Taipei
Tel: 886-2-2508-8600
Fax: 886-2-2508-0102

Thailand - Bangkok
Tel: 66-2-694-1351
Fax: 66-2-694-1350

EUROPE

Austria - Wels
Tel: 43-7242-2244-39
Fax: 43-7242-2244-393

Denmark - Copenhagen
Tel: 45-4450-2828
Fax: 45-4485-2829

France - Paris
Tel: 33-1-69-53-63-20
Fax: 33-1-69-30-90-79

Germany - Munich
Tel: 49-89-627-144-0
Fax: 49-89-627-144-44

Italy - Milan
Tel: 39-0331-742611
Fax: 39-0331-466781

Netherlands - Drunen
Tel: 31-416-690399
Fax: 31-416-690340

Spain - Madrid
Tel: 34-91-708-08-90
Fax: 34-91-708-08-91

UK - Wokingham
Tel: 44-118-921-5869
Fax: 44-118-921-5820

11/29/12



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)

Факс: 8 (812) 320-02-42

Электронная почта: org@eplast1.ru

Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.