

FEATURES

- **Low Noise Voltage:** 1.9nV/ $\sqrt{\text{Hz}}$
- **Low Supply Current:** 1.2mA/Amp Max
- **Low Offset Voltage:** 350 μV Max
- **Gain-Bandwidth Product:**
 - LT6233: 60MHz; $A_V \geq 1$
 - LT6233-10: 375MHz; $A_V \geq 10$
- **Wide Supply Range:** 3V to 12.6V
- **Output Swings Rail-to-Rail**
- **Common Mode Rejection Ratio:** 115dB Typ
- **Output Current:** 30mA
- **Operating Temperature Range:** -40°C to 85°C
- **LT6233 Shutdown to 10 μA Maximum**
- **LT6233/LT6233-10 in a Low Profile (1mm) ThinSOT™ Package**
- **Dual LT6234 in 8-Pin SO and Tiny DFN Packages**
- **LT6235 in a 16-Pin SSOP Package**

APPLICATIONS

- Ultrasound Amplifiers
- Low Noise, Low Power Signal Processing
- Active Filters
- Driving A/D Converters
- Rail-to-Rail Buffer Amplifiers

DESCRIPTION

The LT®6233/LT6234/LT6235 are single/dual/quad low noise, rail-to-rail output unity-gain stable op amps that feature 1.9nV/ $\sqrt{\text{Hz}}$ noise voltage and draw only 1.2mA of supply current per amplifier. These amplifiers combine very low noise and supply current with a 60MHz gain-bandwidth product, a 17V/ μs slew rate and are optimized for low supply voltage signal conditioning systems. The LT6233-10 is a single amplifier optimized for higher gain applications resulting in higher gain bandwidth and slew rate. The LT6233 and LT6233-10 include an enable pin that can be used to reduce the supply current to less than 10 μA .

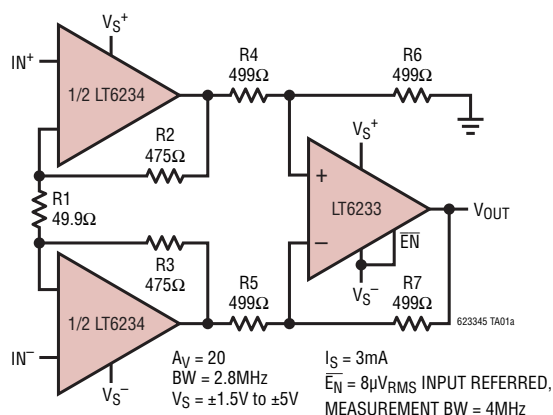
The amplifier family has an output that swings within 50mV of either supply rail to maximize the signal dynamic range in low supply applications and is specified on 3.3V, 5V and $\pm 5\text{V}$ supplies. The $e_n \cdot \sqrt{I_{\text{SUPPLY}}}$ product of 2.1 per amplifier is among the most noise efficient of any op amp.

The LT6233/LT6233-10 are available in the 6-lead SOT-23 package and the LT6234 dual is available in the 8-pin SO package with standard pinouts. For compact layouts, the dual is also available in a tiny dual fine pitch leadless package (DFN). The LT6235 is available in the 16-pin SSOP package.

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TYPICAL APPLICATION

Low Noise Low Power Instrumentation Amplifier



Noise Voltage and Unbalanced Noise Current vs Frequency



623345 TA01b

LT6233/LT6233-10

LT6234/LT6235

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-)	12.6V	Junction Temperature (DD Package).....	125°C
Input Current (Note 2).....	$\pm 40\text{mA}$	Storage Temperature Range	-65°C to 150°C
Output Short-Circuit Duration (Note 3)	Indefinite	Storage Temperature Range	
Operating Temperature Range (Note 4)....	-40°C to 85°C	(DD Package)	-65°C to 125°C
Specified Temperature Range (Note 5)	-40°C to 85°C	Lead Temperature (Soldering, 10 sec).....	300°C
Junction Temperature	150°C		

PIN CONFIGURATION

TOP VIEW S6 PACKAGE 6-LEAD PLASTIC TSOT-23 $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 250^\circ\text{C/W}$	TOP VIEW DD PACKAGE 8-LEAD (3mm $\times$ 3mm) PLASTIC DFN $T_{JMAX} = 125^\circ\text{C}$, $\theta_{JA} = 160^\circ\text{C/W}$ UNDERSIDE METAL CONNECTED TO V^- (PCB CONNECTION OPTIONAL)
TOP VIEW S8 PACKAGE 8-LEAD PLASTIC SO $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 190^\circ\text{C/W}$	TOP VIEW GN PACKAGE 16-LEAD NARROW PLASTIC SSOP $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 135^\circ\text{C/W}$

ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE
LT6233CS6#PBF	LT6233CS6#TRPBF	LTAFL	6-Lead Plastic TSOT-23	0°C to 70°C
LT6233IS6#PBF	LT6233IS6#TRPBF	LTAFL	6-Lead Plastic TSOT-23	-40°C to 85°C
LT6233CS6-10#PBF	LT6233CS6-10#TRPBF	LTAFL	6-Lead Plastic TSOT-23	0°C to 70°C
LT6233IS6-10#PBF	LT6233IS6-10#TRPBF	LTAFL	6-Lead Plastic TSOT-23	-40°C to 85°C
LT6234CS8#PBF	LT6234CS8#TRPBF	6234	8-Lead Plastic SO	0°C to 70°C
LT6234IS8#PBF	LT6234IS8#TRPBF	6234I	8-Lead Plastic SO	-40°C to 85°C
LT6234CDD#PBF	LT6234CDD#TRPBF	LAET	8-Lead (3mm × 3mm) Plastic DFN	0°C to 70°C
LT6234IDD#PBF	LT6234IDD#TRPBF	LAET	8-Lead (3mm × 3mm) Plastic DFN	-40°C to 85°C
LT6235CGN#PBF	LT6235CGN#TRPBF	6235	16-Lead Narrow Plastic SSOP	0°C to 70°C
LT6235IGN#PBF	LT6235IGN#TRPBF	6235I	16-Lead Narrow Plastic SSOP	-40°C to 85°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container. Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>

ELECTRICAL CHARACTERISTICS $T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, 0V ; $V_S = 3.3\text{V}$, 0V ; $V_{CM} = V_{OUT} = \text{half supply}$, ENABLE = 0V, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6233S6, LT6233S6-10		100	500	μV
		LT6234S8, LT6235GN		50	350	μV
		LT6234DD		75	450	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)			80	600	μV
I_B	Input Bias Current			1.5	3	μA
		I_B Match (Channel-to-Channel) (Note 6)		0.04	0.3	μA
I_{OS}	Input Offset Current			0.04	0.3	μA
	Input Noise Voltage	0.1Hz to 10Hz		220		nV _{P-P}
e_n	Input Noise Voltage Density	$f = 10\text{kHz}$, $V_S = 5\text{V}$		1.9	3	nV/ $\sqrt{\text{Hz}}$
i_n	Input Noise Current Density, Balanced Source Input Noise Current Density, Unbalanced Source	$f = 10\text{kHz}$, $V_S = 5\text{V}$, $R_S = 10\text{k}$		0.43		pA/ $\sqrt{\text{Hz}}$
		$f = 10\text{kHz}$, $V_S = 5\text{V}$, $R_S = 10\text{k}$		0.78		pA/ $\sqrt{\text{Hz}}$
	Input Resistance	Common Mode		22		M Ω
		Differential Mode		25		k Ω
C_{IN}	Input Capacitance	Common Mode		2.5		pF
		Differential Mode		4.2		pF
A_{VOL}	Large-Signal Gain	$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 10\text{k}$ to $V_S/2$	73	140		V/mV
		$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 1\text{k}$ to $V_S/2$	18	35		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 10\text{k}$ to $V_S/2$	53	100		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 1\text{k}$ to $V_S/2$	11	20		V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR, $V_S = 5\text{V}$, 0V	1.5		4	V
		Guaranteed by CMRR, $V_S = 3.3\text{V}$, 0V	1.15		2.65	V
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V	90	115		dB
		$V_S = 3.3\text{V}$, $V_{CM} = 1.15\text{V}$ to 2.65V	85	110		dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V	84	115		dB

LT6233/LT6233-10

LT6234/LT6235

ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = 5\text{V}$, 0V ; $V_S = 3.3\text{V}$, 0V ; $V_{CM} = V_{OUT} = \text{half supply}$,
 $\overline{\text{ENABLE}} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
PSRR	Power Supply Rejection Ratio	$V_S = 3\text{V}$ to 10V	90	115		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 3\text{V}$ to 10V	84	115		dB
	Minimum Supply Voltage (Note 7)		3			V
V_{OL}	Output Voltage Swing Low (Note 8)	No Load		4	40	mV
		$I_{SINK} = 5\text{mA}$		75	180	mV
		$V_S = 5\text{V}$, $I_{SINK} = 15\text{mA}$		165	320	mV
		$V_S = 3.3\text{V}$, $I_{SINK} = 10\text{mA}$		125	240	mV
V_{OH}	Output Voltage Swing High (Note 8)	No Load		5	50	mV
		$I_{SOURCE} = 5\text{mA}$		85	195	mV
		$V_S = 5\text{V}$, $I_{SOURCE} = 15\text{mA}$		220	410	mV
		$V_S = 3.3\text{V}$, $I_{SOURCE} = 10\text{mA}$		165	310	mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$	± 40	± 55		mA
		$V_S = 3.3\text{V}$	± 35	± 50		mA
I_S	Supply Current per Amplifier Disabled Supply Current per Amplifier			1.05	1.2	mA
		$\overline{\text{ENABLE}} = V^+ - 0.35\text{V}$		0.2	10	μA
$I_{\overline{\text{ENABLE}}}$	$\overline{\text{ENABLE}}$ Pin Current	$\overline{\text{ENABLE}} = 0.3\text{V}$		-25	-75	μA
V_L	$\overline{\text{ENABLE}}$ Pin Input Voltage Low				0.3	V
V_H	$\overline{\text{ENABLE}}$ Pin Input Voltage High		$V^+ - 0.35$			V
	Output Leakage Current	$\overline{\text{ENABLE}} = V^+ - 0.35\text{V}$, $V_O = 1.5\text{V}$ to 3.5V		0.2	10	μA
t_{ON}	Turn-On Time	$\overline{\text{ENABLE}} = 5\text{V}$ to 0V , $R_L = 1\text{k}$, $V_S = 5\text{V}$		500		ns
t_{OFF}	Turn-Off Time	$\overline{\text{ENABLE}} = 0\text{V}$ to 5V , $R_L = 1\text{k}$, $V_S = 5\text{V}$		76		μs
GBW	Gain-Bandwidth Product	Frequency = 1MHz , $V_S = 5\text{V}$ LT6233-10		55		MHz
				320		MHz
SR	Slew Rate	$V_S = 5\text{V}$, $A_V = -1$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V	10	15		V/ μs
		LT6233-10, $V_S = 5\text{V}$, $A_V = -10$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V		80		V/ μs
FPBW	Full-Power Bandwidth	$V_S = 5\text{V}$, $V_{OUT} = 3V_{P-P}$ (Note 9)	1.06	1.6		MHz
		LT6233-10, $\text{HD2} = \text{HD3} \leq 1\%$		2.2		MHz
t_S	Settling Time (LT6233, LT6234, LT6235)	0.1% , $V_S = 5\text{V}$, $V_{STEP} = 2\text{V}$, $A_V = -1$, $R_L = 1\text{k}$		175		ns

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the $0^{\circ}\text{C} < T_A < 70^{\circ}\text{C}$ temperature range. $V_S = 5\text{V}$, 0V ; $V_S = 3.3\text{V}$, 0V ; $V_{CM} = V_{OUT} = \text{half supply}$, $\overline{\text{ENABLE}} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V_{OS}	Input Offset Voltage	LT6233CS6, LT6233CS6-10	●		600	μV
		LT6234CS8, LT6235CGN	●		450	μV
		LT6234CDD	●		550	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)		●		800	μV
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 10)	$V_{CM} = \text{Half Supply}$	●	0.5	3.0	$\mu\text{V}/^{\circ}\text{C}$
I_B	Input Bias Current		●		3.5	μA
	I_B Match (Channel-to-Channel) (Note 6)		●		0.4	μA
I_{OS}	Input Offset Current		●		0.4	μA
A_{VOL}	Large-Signal Gain	$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 10\text{k}$ to $V_S/2$	●	47		V/mV
		$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 1\text{k}$ to $V_S/2$	●	12		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 10\text{k}$ to $V_S/2$	●	40		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 1\text{k}$ to $V_S/2$	●	7.5		V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR				
		$V_S = 5\text{V}$, 0V $V_S = 3.3\text{V}$, 0V	● ●	1.5 1.15	4 2.65	V V
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V $V_S = 3.3\text{V}$, $V_{CM} = 1.15\text{V}$ to 2.65V	● ●	90 85		dB dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V	●	84		dB
PSRR	Power Supply Rejection Ratio	$V_S = 3\text{V}$ to 10V	●	90		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 3\text{V}$ to 10V	●	84		dB
	Minimum Supply Voltage (Note 7)		●	3		V
V_{OL}	Output Voltage Swing Low (Note 8)	No Load	●		50	mV
		$I_{SINK} = 5\text{mA}$	●		195	mV
		$V_S = 5\text{V}$, $I_{SINK} = 15\text{mA}$	●		360	mV
		$V_S = 3.3\text{V}$, $I_{SINK} = 10\text{mA}$	●		265	mV
V_{OH}	Output Voltage Swing High (Note 8)	No Load	●		60	mV
		$I_{SOURCE} = 5\text{mA}$	●		205	mV
		$V_S = 5\text{V}$, $I_{SOURCE} = 15\text{mA}$	●		435	mV
		$V_S = 3.3\text{V}$, $I_{SOURCE} = 10\text{mA}$	●		330	mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$	●	± 35		mA
		$V_S = 3.3\text{V}$	●	± 30		mA
I_S	Supply Current per Amplifier		●		1.45	mA
	Disabled Supply Current per Amplifier	$\overline{\text{ENABLE}} = V^+ - 0.25\text{V}$	●	1		μA
$I_{\overline{\text{ENABLE}}}$	$\overline{\text{ENABLE}}$ Pin Current	$\overline{\text{ENABLE}} = 0.3\text{V}$	●		-85	μA
V_L	$\overline{\text{ENABLE}}$ Pin Input Voltage Low		●		0.3	V
V_H	$\overline{\text{ENABLE}}$ Pin Input Voltage High		●	$V^+ - 0.25$		V
	Output Leakage Current	$\overline{\text{ENABLE}} = V^+ - 0.25\text{V}$, $V_O = 1.5\text{V}$ to 3.5V	●	1		μA
t_{ON}	Turn-On Time	$\overline{\text{ENABLE}} = 5\text{V}$ to 0V , $R_L = 1\text{k}$, $V_S = 5\text{V}$	●	500		ns
t_{OFF}	Turn-Off Time	$\overline{\text{ENABLE}} = 0\text{V}$ to 5V , $R_L = 1\text{k}$, $V_S = 5\text{V}$	●	120		μs
SR	Slew Rate	$V_S = 5\text{V}$, $A_V = -1$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V	●	9		$\text{V}/\mu\text{s}$
		LT6233-10, $A_V = -10$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V	●	75		$\text{V}/\mu\text{s}$
FPBW	Full-Power Bandwidth (Note 9)	$V_S = 5\text{V}$, $V_{OUT} = 3V_{P-P}$; LT6233C, LT6234C, LT6235C	●	955		kHz

LT6233/LT6233-10

LT6234/LT6235

ELECTRICAL CHARACTERISTICS The ● denotes the specifications which apply over the $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$ temperature range. $V_S = 5\text{V}$, 0V ; $V_S = 3.3\text{V}$, 0V ; $V_{CM} = V_{OUT} = \text{half supply}$, $\overline{\text{ENABLE}} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6233IS6, LT6233IS6-10	●		700	μV
		LT6234IS8, LT6235IGN	●		550	μV
		LT6234IDD	●		650	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)		●		1000	μV
$V_{OS\text{ TC}}$	Input Offset Voltage Drift (Note 10)	$V_{CM} = \text{Half Supply}$	●	0.5	3	$\mu\text{V}/^{\circ}\text{C}$
I_B	Input Bias Current		●		4	μA
	I_B Match (Channel-to-Channel) (Note 6)		●		0.4	μA
I_{OS}	Input Offset Current		●		0.5	μA
A_{VOL}	Large-Signal Gain	$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 10\text{k}$ to $V_S/2$	●	45		V/mV
		$V_S = 5\text{V}$, $V_O = 0.5\text{V}$ to 4.5V , $R_L = 1\text{k}$ to $V_S/2$	●	11		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 10\text{k}$ to $V_S/2$	●	38		V/mV
		$V_S = 3.3\text{V}$, $V_O = 0.65\text{V}$ to 2.65V , $R_L = 1\text{k}$ to $V_S/2$	●	7		V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR				
		$V_S = 5\text{V}$, 0V	●	1.5	4	V
		$V_S = 3.3\text{V}$, 0V	●	1.15	2.65	V
CMRR	Common Mode Rejection Ratio	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V	●	90		dB
		$V_S = 3.3\text{V}$, $V_{CM} = 1.15\text{V}$ to 2.65V	●	85		dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_S = 5\text{V}$, $V_{CM} = 1.5\text{V}$ to 4V	●	84		dB
PSRR	Power Supply Rejection Ratio	$V_S = 3\text{V}$ to 10V	●	90		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = 3\text{V}$ to 10V	●	84		dB
	Minimum Supply Voltage (Note 7)		●	3		V
V_{OL}	Output Voltage Swing Low (Note 8)	No Load	●		50	mV
		$I_{SINK} = 5\text{mA}$	●		195	mV
		$V_S = 5\text{V}$, $I_{SINK} = 15\text{mA}$	●		370	mV
		$V_S = 3.3\text{V}$, $I_{SINK} = 10\text{mA}$	●		275	mV
V_{OH}	Output Voltage Swing High (Note 6)	No Load	●		60	mV
		$I_{SOURCE} = 5\text{mA}$	●		210	mV
		$V_S = 5\text{V}$, $I_{SOURCE} = 15\text{mA}$	●		445	mV
		$V_S = 3.3\text{V}$, $I_{SOURCE} = 10\text{mA}$	●		335	mV
I_{SC}	Short-Circuit Current	$V_S = 5\text{V}$	●	± 30		mA
		$V_S = 3.3\text{V}$	●	± 20		mA
I_S	Supply Current per Amplifier		●		1.5	mA
	Disabled Supply Current per Amplifier	$\overline{\text{ENABLE}} = V^+ - 0.2\text{V}$	●	1		μA
$I_{\overline{\text{ENABLE}}}$	ENABLE Pin Current	$\overline{\text{ENABLE}} = 0.3\text{V}$	●		-100	μA
V_L	ENABLE Pin Input Voltage Low		●		0.3	V
V_H	ENABLE Pin Input Voltage High		●	$V^+ - 0.2$		V
	Output Leakage Current	$\overline{\text{ENABLE}} = V^+ - 0.2\text{V}$, $V_O = 1.5\text{V}$ to 3.5V	●	1		μA
t_{ON}	Turn-On Time	$\overline{\text{ENABLE}} = 5\text{V}$ to 0V , $R_L = 1\text{k}$, $V_S = 5\text{V}$	●	500		ns
t_{OFF}	Turn-Off Time	$\overline{\text{ENABLE}} = 0\text{V}$ to 5V , $R_L = 1\text{k}$, $V_S = 5\text{V}$	●	135		μs
SR	Slew Rate	$V_S = 5\text{V}$, $A_V = -1$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V	●	8		$\text{V}/\mu\text{s}$
		LT6233-10, $A_V = -10$, $R_L = 1\text{k}$, $V_O = 1.5\text{V}$ to 3.5V	●	70		$\text{V}/\mu\text{s}$
FPBW	Full-Power Bandwidth (Note 9)	$V_S = 5\text{V}$, $V_{OUT} = 3V_{P-P}$; LT6233I, LT6234I, LT6235I	●	848		kHz

ELECTRICAL CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_S = \pm 5\text{V}$, $V_{CM} = V_{OUT} = 0\text{V}$, $\overline{\text{ENABLE}} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6233S6, LT6233S6-10		100	500	μV
		LT6234S8, LT6235GN		50	350	μV
		LT6234DD		75	450	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)			100	600	μV
I_B	Input Bias Current			1.5	3	μA
	I_B Match (Channel-to-Channel) (Note 6)			0.04	0.3	μA
I_{OS}	Input Offset Current			0.04	0.3	μA
	Input Noise Voltage	0.1Hz to 10Hz		220		nV_{P-P}
e_n	Input Noise Voltage Density	$f = 10\text{kHz}$		1.9	3.0	$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current Density, Balanced Source Input Noise Current Density, Unbalanced Source	$f = 10\text{kHz}$, $R_S = 10\text{k}$		0.43		$\text{pA}/\sqrt{\text{Hz}}$
		$f = 10\text{kHz}$, $R_S = 10\text{k}$		0.78		$\text{pA}/\sqrt{\text{Hz}}$
	Input Resistance	Common Mode		22		$\text{M}\Omega$
		Differential Mode		25		$\text{k}\Omega$
C_{IN}	Input Capacitance	Common Mode		2.1		pF
		Differential Mode		3.7		pF
A_{VOL}	Large-Signal Gain	$V_O = \pm 4.5\text{V}$, $R_L = 10\text{k}$	97	180		V/mV
		$V_O = \pm 4.5\text{V}$, $R_L = 1\text{k}$	28	55		V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR	-3		4	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -3\text{V}$ to 4V	90	110		dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_{CM} = -3\text{V}$ to 4V	84	120		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	90	115		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	84	115		dB
V_{OL}	Output Voltage Swing Low (Note 8)	No Load		4	40	mV
		$I_{SINK} = 5\text{mA}$		75	180	mV
		$I_{SINK} = 15\text{mA}$		165	320	mV
V_{OH}	Output Voltage Swing High (Note 8)	No Load		5	50	mV
		$I_{SOURCE} = 5\text{mA}$		85	195	mV
		$I_{SOURCE} = 15\text{mA}$		220	410	mV
I_{SC}	Short-Circuit Current		± 40	± 55		mA
I_S	Supply Current per Amplifier Disabled Supply Current per Amplifier			1.15	1.4	mA
		$\overline{\text{ENABLE}} = 4.65\text{V}$		0.2	10	μA
$I_{\overline{\text{ENABLE}}}$	ENABLE Pin Current	$\overline{\text{ENABLE}} = 0.3\text{V}$		-35	-85	μA
V_L	ENABLE Pin Input Voltage Low				0.3	V
V_H	ENABLE Pin Input Voltage High		4.65			V
	Output Leakage Current	$\overline{\text{ENABLE}} = 4.65\text{V}$, $V_O = \pm 1\text{V}$		0.2	10	μA
t_{ON}	Turn-On Time	$\overline{\text{ENABLE}} = 5\text{V}$ to 0V , $R_L = 1\text{k}$		900		ns
t_{OFF}	Turn-Off Time	$\overline{\text{ENABLE}} = 0\text{V}$ to 5V , $R_L = 1\text{k}$		100		μs
GBW	Gain-Bandwidth Product	Frequency = 1MHz LT6233-10	42 260	60 375		MHz MHz
		$A_V = -1$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V LT6233-10, $A_V = -10$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V	12	17 115		$\text{V}/\mu\text{s}$ $\text{V}/\mu\text{s}$
FPBW	Full-Power Bandwidth	$V_{OUT} = 3V_{P-P}$ (Note 9)	1.27	1.8		MHz
		LT6233-10, $\text{HD}_2 = \text{HD}_3 \leq 1\%$		2.2		MHz
t_S	Settling Time (LT6233, LT6234, LT6235)	0.1%, $V_{STEP} = 2\text{V}$, $A_V = -1$, $R_L = 1\text{k}$		170		ns

LT6233/LT6233-10

LT6234/LT6235

ELECTRICAL CHARACTERISTICS The ● denotes the specifications which apply over the $0^{\circ}\text{C} < T_A < 70^{\circ}\text{C}$ temperature range. $V_S = \pm 5\text{V}$, $V_{CM} = V_{OUT} = 0\text{V}$, $\text{ENABLE} = 0\text{V}$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6233CS6, LT6233CS6-10	●			600	μV
		LT6234CS8, LT6235CGN	●			450	μV
		LT6234CDD	●			550	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)		●			800	μV
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 10)		●		0.5	3	$\mu\text{V}/^{\circ}\text{C}$
I_B	Input Bias Current		●			3.5	μA
	I_B Match (Channel-to-Channel) (Note 6)		●			0.4	μA
I_{OS}	Input Offset Current		●			0.4	μA
A_{VOL}	Large-Signal Gain	$V_O = \pm 4.5\text{V}$, $R_L = 10\text{k}$	●	75			V/mV
		$V_O = \pm 4.5\text{V}$, $R_L = 1\text{k}$	●	22			V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR	●	-3		4	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -3\text{V}$ to 4V	●	90			dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_{CM} = -3\text{V}$ to 4V	●	84			dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	90			dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	84			dB
V_{OL}	Output Voltage Swing Low (Note 8)	No Load	●			50	mV
		$I_{SINK} = 5\text{mA}$	●			195	mV
		$I_{SINK} = 15\text{mA}$	●			360	mV
V_{OH}	Output Voltage Swing High (Note 8)	No Load	●			60	mV
		$I_{SOURCE} = 5\text{mA}$	●			205	mV
		$I_{SOURCE} = 15\text{mA}$	●			435	mV
I_{SC}	Short-Circuit Current		●	± 35			mA
I_S	Supply Current per Amplifier		●			1.7	mA
	Disabled Supply Current per Amplifier	$\text{ENABLE} = 4.75\text{V}$	●		1		μA
I_{ENABLE}	ENABLE Pin Current	$\text{ENABLE} = 0.3\text{V}$	●			-95	μA
V_L	ENABLE Pin Input Voltage Low		●			0.3	V
V_H	ENABLE Pin Input Voltage High		●	4.75			V
	Output Leakage Current	$\text{ENABLE} = 4.75\text{V}$, $V_O = \pm 1\text{V}$	●		1		μA
t_{ON}	Turn-On Time	$\text{ENABLE} = 5\text{V}$ to 0V , $R_L = 1\text{k}$	●		900		ns
t_{OFF}	Turn-Off Time	$\text{ENABLE} = 0\text{V}$ to 5V , $R_L = 1\text{k}$	●		150		μs
SR	Slew Rate	$A_V = -1$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V	●	11			$\text{V}/\mu\text{s}$
		LT6233-10, $A_V = -10$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V	●		105		$\text{V}/\mu\text{s}$
FPBW	Full-Power Bandwidth (Note 9)	$V_{OUT} = 3V_{P-P}$; LT6233C, LT6234C, LT6235C	●	1.16			MHz

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the $-40^{\circ}\text{C} < T_A < 85^{\circ}\text{C}$ temperature range. $V_S = \pm 5\text{V}$, $V_{CM} = V_{OUT} = 0\text{V}$, $\text{ENABLE} = 0\text{V}$, unless otherwise noted. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT6233IS6, LT6233IS6-10	●		700	μV
		LT6234IS8, LT6235IGN	●		550	μV
		LT6234IDD	●		650	μV
	Input Offset Voltage Match (Channel-to-Channel) (Note 6)		●		1000	μV
$V_{OS\ TC}$	Input Offset Voltage Drift (Note 10)		●	0.5	3	$\mu\text{V}/^{\circ}\text{C}$
I_B	Input Bias Current		●		4	μA
	I_B Match (Channel-to-Channel) (Note 6)		●		0.4	μA
I_{OS}	Input Offset Current		●		0.5	μA
A_{VOL}	Large-Signal Gain	$V_O = \pm 4.5\text{V}$, $R_L = 10\text{k}$	●	68		V/mV
		$V_O = \pm 4.5\text{V}$, $R_L = 1\text{k}$	●	20		V/mV
V_{CM}	Input Voltage Range	Guaranteed by CMRR	●	-3	4	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -3\text{V}$ to 4V	●	90		dB
	CMRR Match (Channel-to-Channel) (Note 6)	$V_{CM} = -3\text{V}$ to 4V	●	84		dB
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	90		dB
	PSRR Match (Channel-to-Channel) (Note 6)	$V_S = \pm 1.5\text{V}$ to $\pm 5\text{V}$	●	84		dB
V_{OL}	Output Voltage Swing Low (Note 8)	No Load	●		50	mV
		$I_{SINK} = 5\text{mA}$	●		195	mV
		$I_{SINK} = 15\text{mA}$	●		370	mV
V_{OH}	Output Voltage Swing High (Note 8)	No Load	●		70	mV
		$I_{SOURCE} = 5\text{mA}$	●		210	mV
		$I_{SOURCE} = 15\text{mA}$	●		445	mV
I_{SC}	Short-Circuit Current		●	± 30		mA
I_S	Supply Current per Amplifier Disabled Supply Current per Amplifier	$\text{ENABLE} = 4.8\text{V}$	●		1.75	mA
			●	1		μA
I_{ENABLE}	ENABLE Pin Current	$\text{ENABLE} = 0.3\text{V}$	●		-110	μA
V_L	ENABLE Pin Input Voltage Low		●		0.3	V
V_H	ENABLE Pin Input Voltage High		●	4.8		V
	Output Leakage Current	$\text{ENABLE} = 4.8\text{V}$, $V_O = \pm 1\text{V}$	●	1		μA
t_{ON}	Turn-On Time	$\text{ENABLE} = 5\text{V}$ to 0V , $R_L = 1\text{k}$	●	900		ns
t_{OFF}	Turn-Off Time	$\text{ENABLE} = 0\text{V}$ to 5V , $R_L = 1\text{k}$	●	160		μs
SR	Slew Rate	$A_V = -1$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V	●	10		$\text{V}/\mu\text{s}$
		LT6233-10, $A_V = -10$, $R_L = 1\text{k}$, $V_O = -2\text{V}$ to 2V	●	95		$\text{V}/\mu\text{s}$
FPBW	Full-Power Bandwidth (Note 9)	$V_{OUT} = 3V_{P-P}$; LT6233I, LT6234I, LT6235I	●	1.06		MHz

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Inputs are protected by back-to-back diodes. If the differential input voltage exceeds 0.7V , the input current must be limited to less than 40mA .

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum rating when the output is shorted indefinitely.

Note 4: The LT6233C/LT6233I, the LT6234C/LT6234I, and LT6235C/LT6235I are guaranteed functional over the temperature range of -40°C to 85°C .

Note 5: The LT6233C/LT6234C/LT6235C are guaranteed to meet specified performance from 0°C to 70°C . The LT6233C/LT6234C/LT6235C are designed, characterized and expected to meet specified performance from -40°C to 85°C , but are not tested or QA sampled at these temperatures. The LT6233I/LT6234I/LT6235I are guaranteed to meet specified performance from -40°C to 85°C .

Note 6: Matching parameters are the difference between the two amplifiers A and D and between B and C of the LT6235; between the two amplifiers of the LT6234. CMRR and PSRR match are defined as follows: CMRR and PSRR are measured in $\mu\text{V}/\text{V}$ on the matched amplifiers. The difference is calculated between the matching sides in $\mu\text{V}/\text{V}$. The result is converted to dB.

ELECTRICAL CHARACTERISTICS

Note 7: Minimum supply voltage is guaranteed by power supply rejection ratio test.

Note 8: Output voltage swings are measured between the output and power supply rails.

Note 9: Full-power bandwidth is calculated from the slew rate:

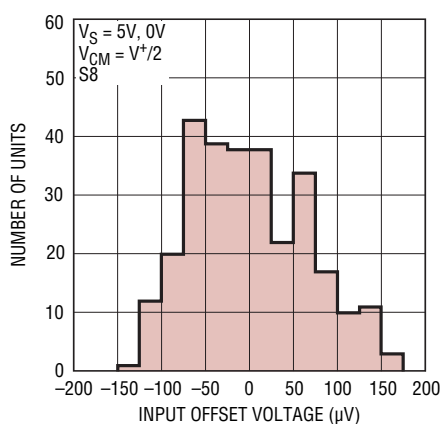
$$FPBW = SR/2\pi V_P$$

Note 10: This parameter is not 100% tested.

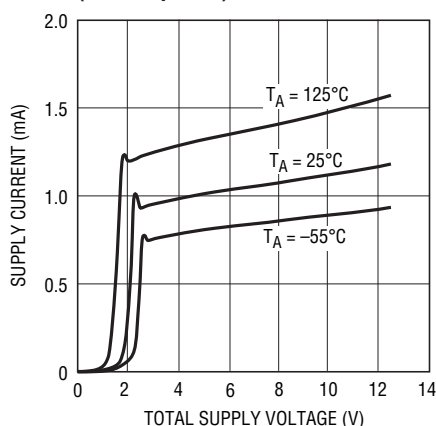
TYPICAL PERFORMANCE CHARACTERISTICS

(LT6233/LT6234/LT6235)

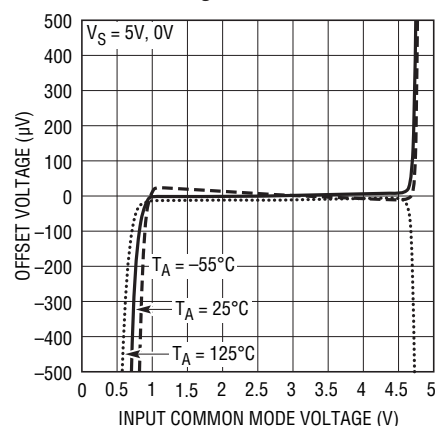
V_{OS} Distribution



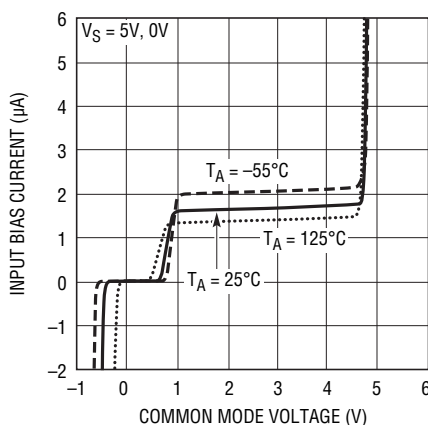
Supply Current vs Supply Voltage (Per Amplifier)



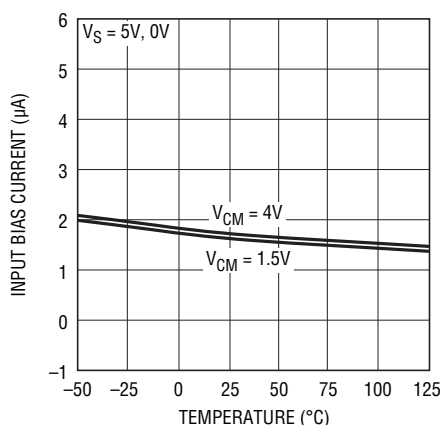
Offset Voltage vs Input Common Mode Voltage



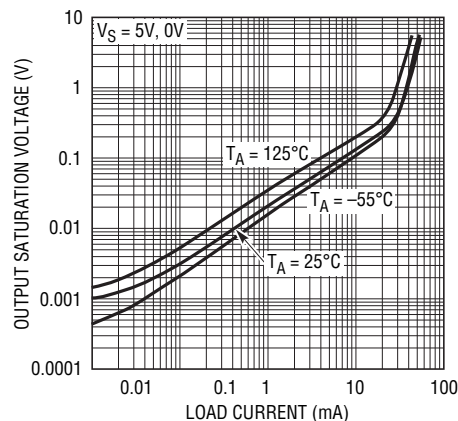
Input Bias Current vs Common Mode Voltage



Input Bias Current vs Temperature



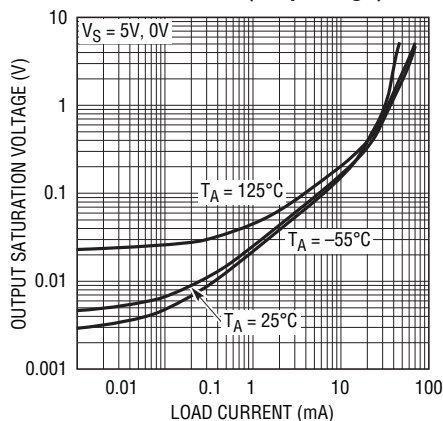
Output Saturation Voltage vs Load Current (Output Low)



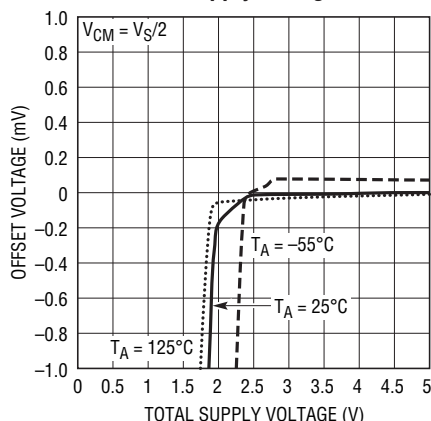
TYPICAL PERFORMANCE CHARACTERISTICS

(LT6233/LT6234/LT6235)

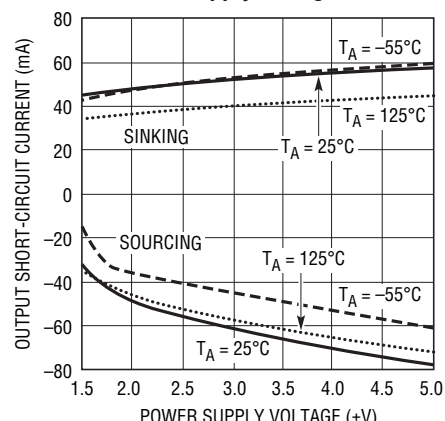
Output Saturation Voltage
vs Load Current (Output High)



Minimum Supply Voltage



Output Short-Circuit Current
vs Power Supply Voltage



Open-Loop Gain



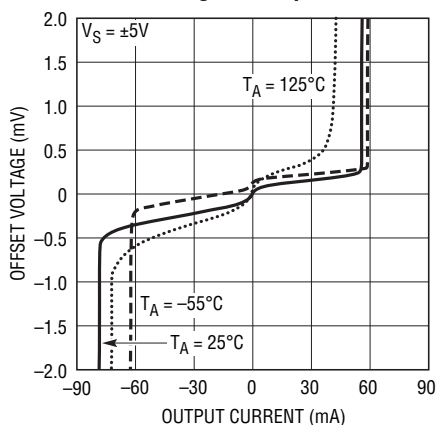
Open-Loop Gain



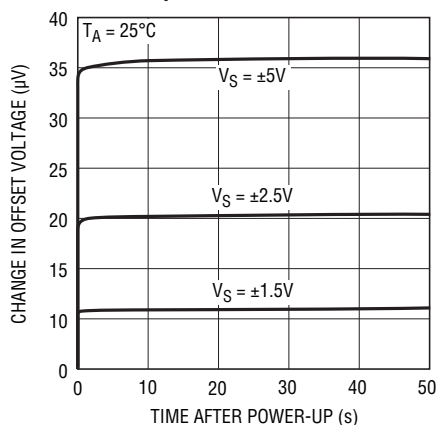
Open-Loop Gain



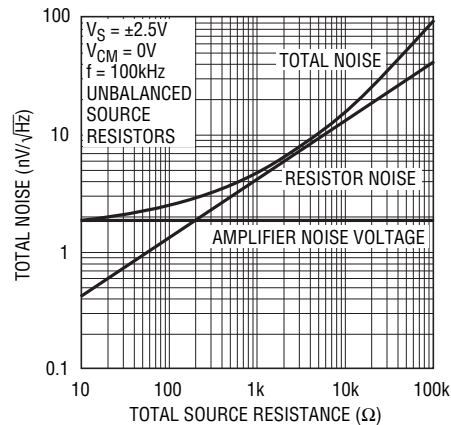
Offset Voltage vs Output Current



Warm-Up Drift vs Time



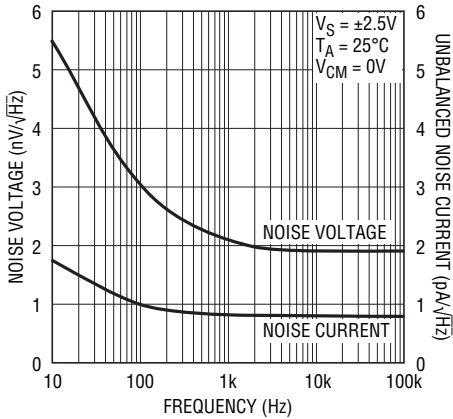
Total Noise vs Total Source
Resistance



TYPICAL PERFORMANCE CHARACTERISTICS

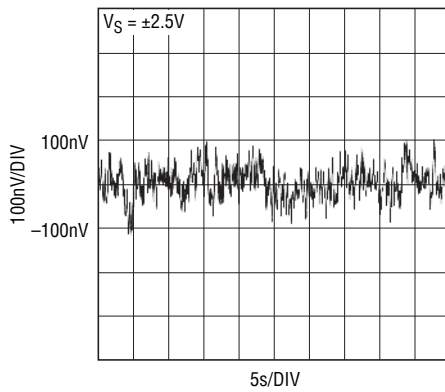
(LT6233/LT6234/LT6235)

Noise Voltage and Unbalanced Noise Current vs Frequency



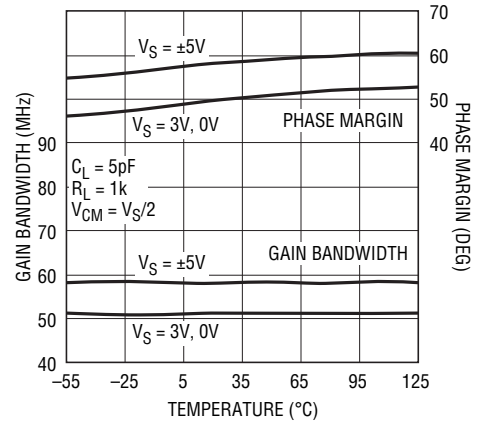
623345 G16

0.1Hz to 10Hz Output Voltage Noise



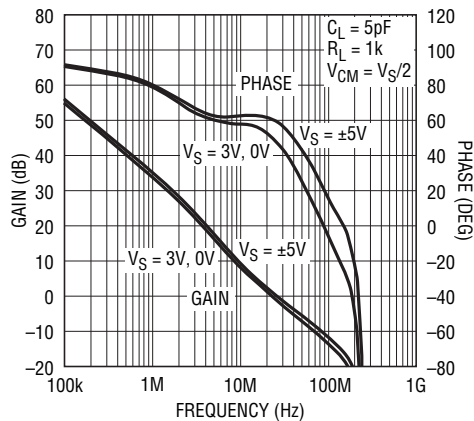
623345 G17

Gain Bandwidth and Phase Margin vs Temperature



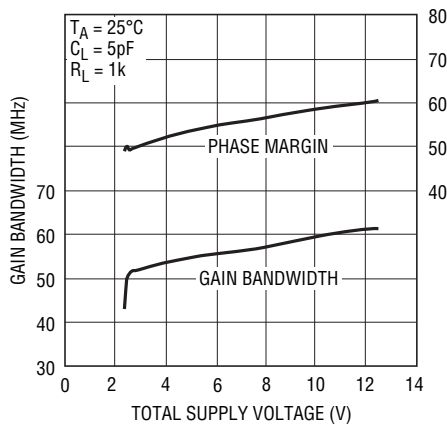
623345 G18

Open-Loop Gain vs Frequency



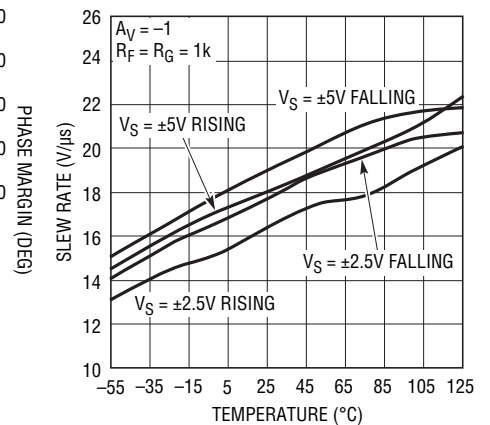
623345 G19

Gain Bandwidth and Phase Margin vs Supply Voltage



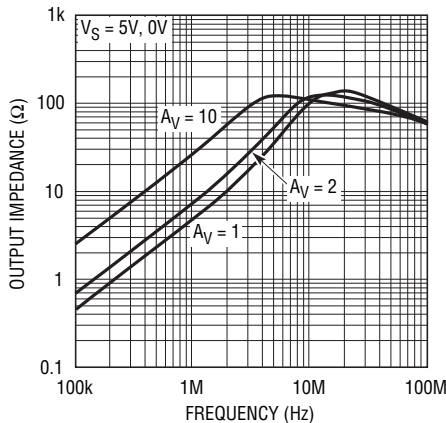
623345 G20

Slew Rate vs Temperature



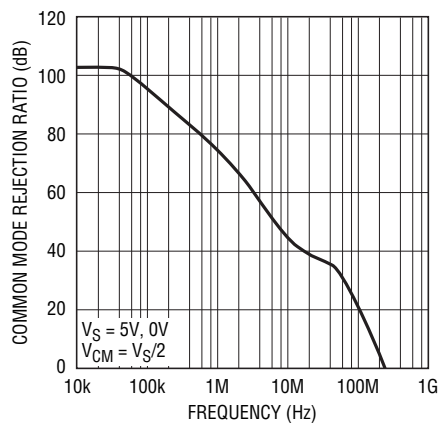
623345 G21

Output Impedance vs Frequency



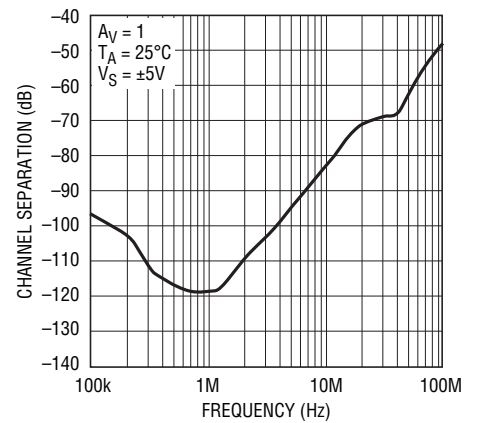
623345 G22

Common Mode Rejection Ratio vs Frequency



623345 G23

Channel Separation vs Frequency



623345 G24

623345fc

TYPICAL PERFORMANCE CHARACTERISTICS

(LT6233/LT6234/LT6235)

Power Supply Rejection Ratio vs Frequency



623345 G25

Series Output Resistance and Overshoot vs Capacitive Load



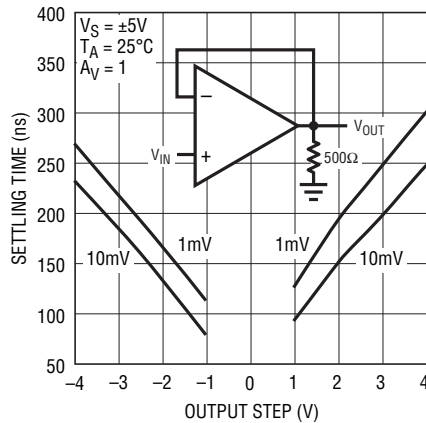
623345 G26

Series Output Resistance and Overshoot vs Capacitive Load



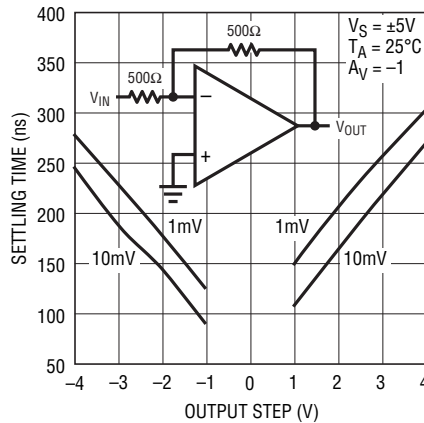
623345 G27

Settling Time vs Output Step (Noninverting)



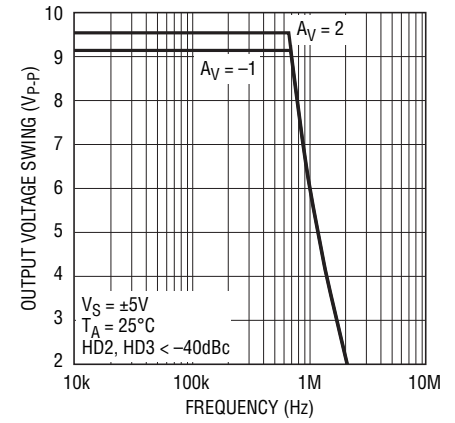
623345 G28

Settling Time vs Output Step (Inverting)



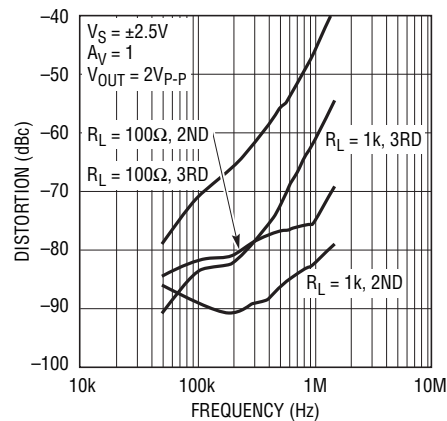
623345 G29

Maximum Undistorted Output Signal vs Frequency



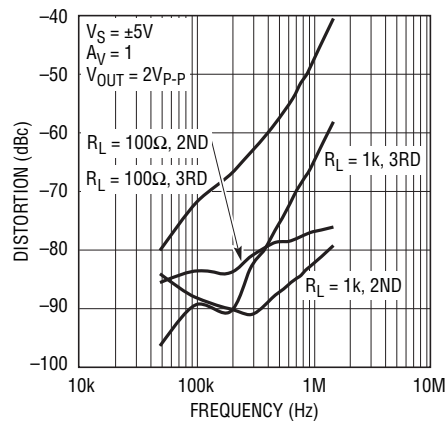
623345 G30

Distortion vs Frequency



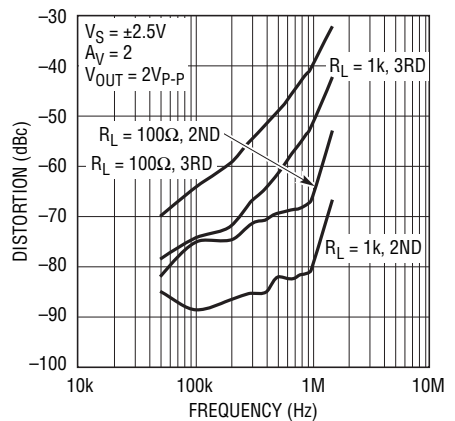
623345 G31

Distortion vs Frequency



623345 G32

Distortion vs Frequency

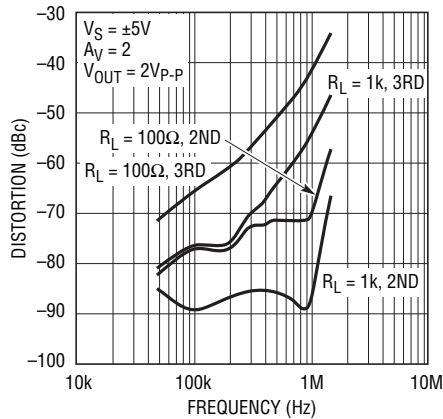


623345 G33

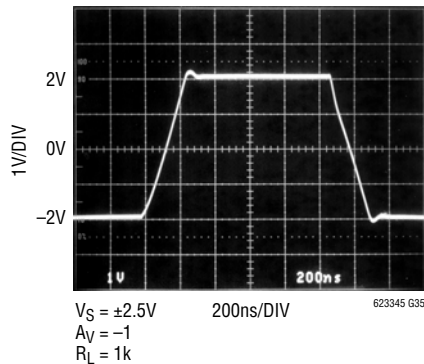
LT6233/LT6233-10 LT6234/LT6235

TYPICAL PERFORMANCE CHARACTERISTICS (LT6233/LT6234/LT6235)

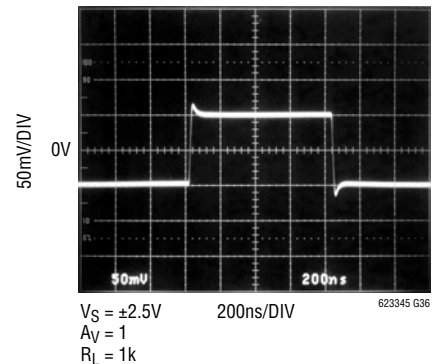
Distortion vs Frequency



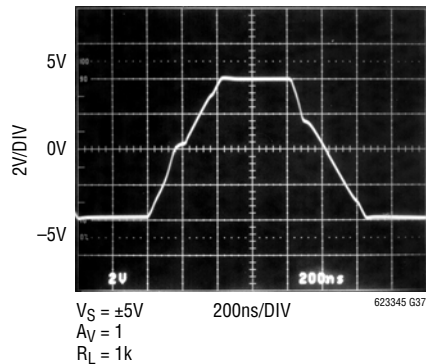
Large-Signal Response



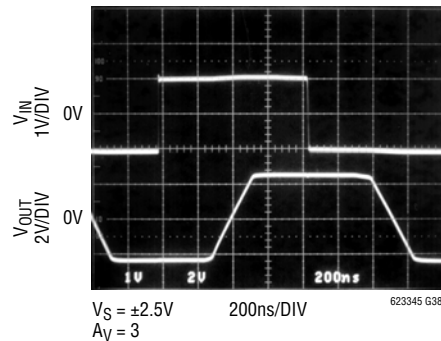
Small-Signal Response



Large-Signal Response

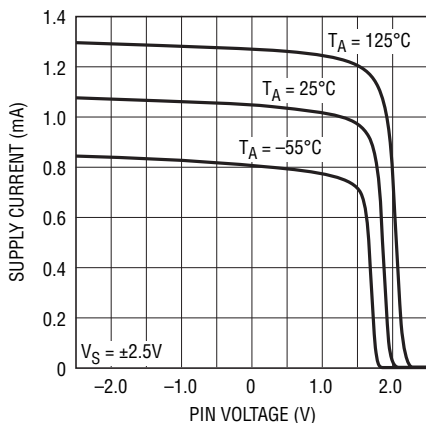


Output Overdrive Recovery



(LT6233) ENABLE Characteristics

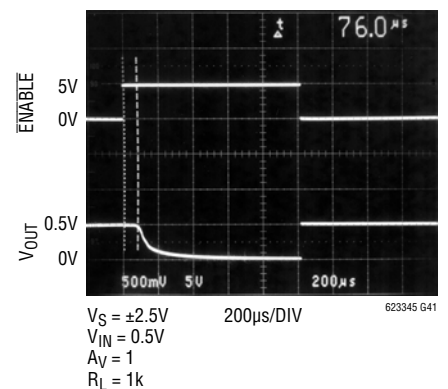
Supply Current vs ENABLE Pin Voltage



ENABLE Pin Current vs ENABLE Pin Voltage



ENABLE Pin Response Time



TYPICAL PERFORMANCE CHARACTERISTICS

(LT6233-10)

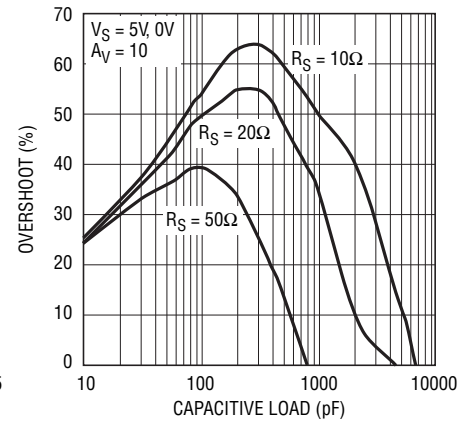
Gain Bandwidth and Phase Margin vs Temperature



Slew Rate vs Temperature



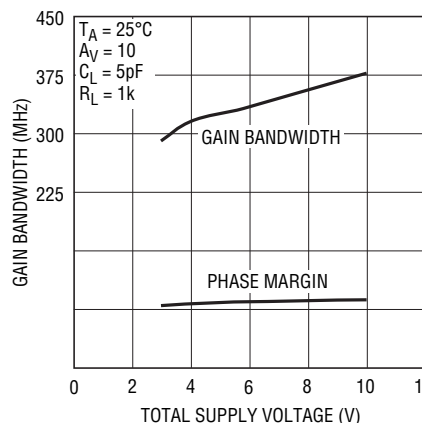
Series Output Resistor and Overshoot vs Capacitive Load



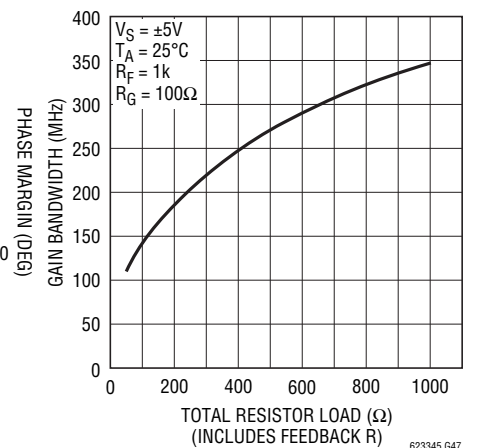
Open-Loop Gain and Phase vs Frequency



Gain Bandwidth and Phase Margin vs Supply Voltage



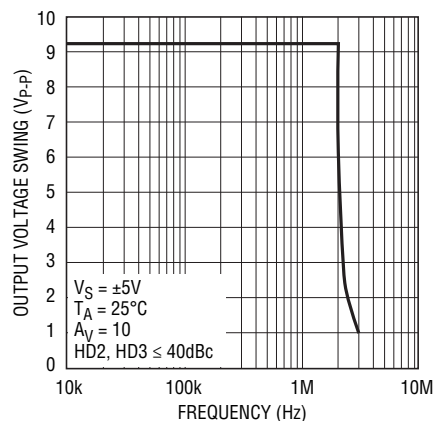
Gain Bandwidth vs Resistor Load



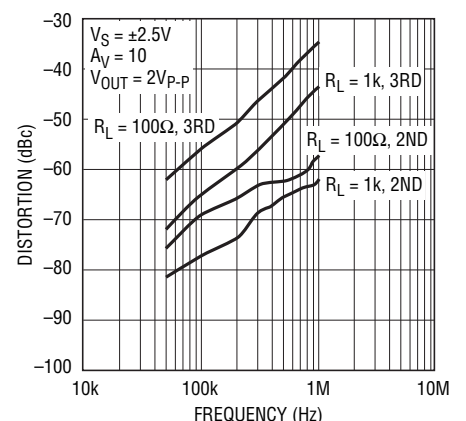
Common Mode Rejection Ratio vs Frequency



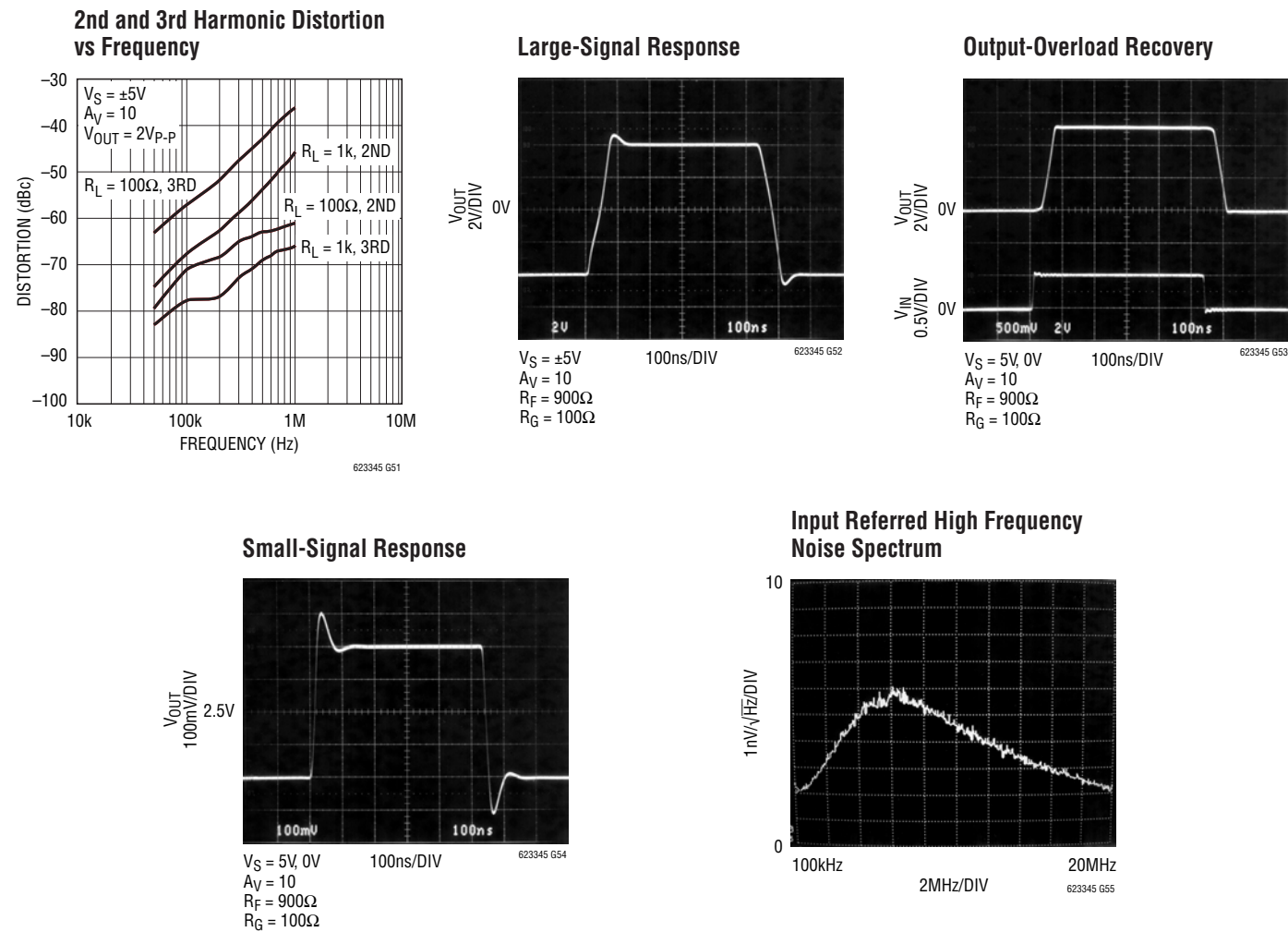
Maximum Undistorted Output vs Frequency



2nd and 3rd Harmonic Distortion vs Frequency



TYPICAL PERFORMANCE CHARACTERISTICS
(LT6233-10)



APPLICATIONS INFORMATION

Amplifier Characteristics

Figure 1 is a simplified schematic of the LT6233/LT6234/LT6235, which has a pair of low noise input transistors Q1 and Q2. A simple current mirror Q3/Q4 converts the differential signal to a single-ended output, and these transistors are degenerated to reduce their contribution to the overall noise.

Capacitor C1 reduces the unity-cross frequency and improves the frequency stability without degrading the gain bandwidth of the amplifier. Capacitor C_M sets the overall amplifier gain bandwidth. The differential drive generator supplies current to transistors Q5 and Q6 that swing the output from rail-to-rail.

Input Protection

There are back-to-back diodes, D1 and D2 across the + and – inputs of these amplifiers to limit the differential input voltage to $\pm 0.7V$. The inputs of the LT6233/LT6234/LT6235 do not have internal resistors in series with the input transistors. This technique is often used to protect the input devices from overvoltage that causes excessive current to flow. The addition of these resistors would significantly degrade the low noise voltage of these amplifiers. For instance, a 100Ω resistor in series with each input would generate $1.8nV/\sqrt{Hz}$ of noise, and the total amplifier noise voltage would rise from $1.9nV/\sqrt{Hz}$ to $2.6nV/\sqrt{Hz}$. Once the input differential voltage exceeds $\pm 0.7V$, steady-state current conducted through the protection diodes should

be limited to $\pm 40mA$. This implies 25Ω of protection resistance is necessary per volt of overdrive beyond $\pm 0.7V$. These input diodes are rugged enough to handle transient currents due to amplifier slew rate overdrive and clipping without protection resistors.

The photo of Figure 2 shows the output response to an input overdrive with the amplifier connected as a voltage follower. With the input signal low, current source I₁ saturates and the differential drive generator drives Q6 into saturation so the output voltage swings all the way to V⁻. The input can swing positive until transistor Q2 saturates into current mirror Q3/Q4. When saturation occurs, the output tries to phase invert, but diode D2 conducts current from the signal source to the output through the feedback connection. The output is clamped a diode drop below the input. In this photo, the input signal generator is limiting at about 20mA.

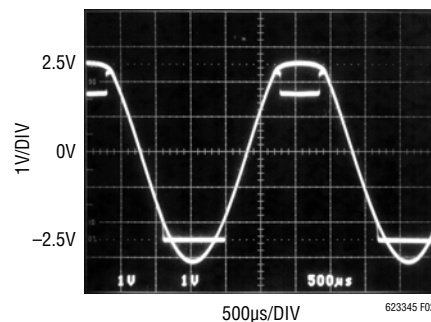


Figure 2. V_S = $\pm 2.5V$, A_V = 1 with Large Overdrive

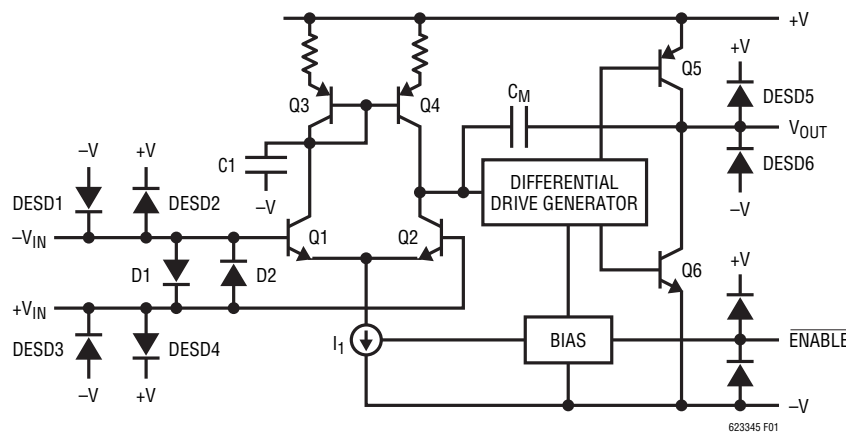


Figure 1. Simplified Schematic

APPLICATIONS INFORMATION

With the amplifier connected in a gain of $A_V \geq 2$, the output can invert with very heavy overdrive. To avoid this inversion, limit the input overdrive to 0.5V beyond the power supply rails.

ESD

The LT6233/LT6234/LT6235 have reverse-biased ESD protection diodes on all inputs and outputs as shown in Figure 1. If these pins are forced beyond either supply, unlimited current will flow through these diodes. If the current is transient and limited to one hundred milliamps or less, no damage to the device will occur.

Noise

The noise voltage of the LT6233/LT6234/LT6235 is equivalent to that of a 225Ω resistor, and for the lowest possible noise it is desirable to keep the source and feedback resistance at or below this value, i.e., $R_S + R_G || R_{FB} \leq 225\Omega$. With $R_S + R_G || R_{FB} = 225\Omega$ the total noise of the amplifier is:

$$e_N = \sqrt{(1.9\text{nV})^2 + (1.9\text{nV})^2} = 2.69\text{nV}/\sqrt{\text{Hz}}$$

Below this resistance value, the amplifier dominates the noise, but in the region between 225Ω and about 30k, the noise is dominated by the resistor thermal noise. As the total resistance is further increased beyond 30k, the amplifier noise current multiplied by the total resistance eventually dominates the noise.

The product of $e_N \cdot \sqrt{I_{\text{SUPPLY}}}$ is an interesting way to gauge low noise amplifiers. Most low noise amplifiers with low e_N have high I_{SUPPLY} current. In applications that require low noise voltage with the lowest possible supply current, this product can prove to be enlightening. The LT6233/LT6234/LT6235 have an $e_N \cdot \sqrt{I_{\text{SUPPLY}}}$ product of only 2.1 per amplifier, yet it is common to see amplifiers with similar noise specifications to have $e_N \cdot \sqrt{I_{\text{SUPPLY}}}$ as high as 13.5.

For a complete discussion of amplifier noise, see the LT1028 data sheet.

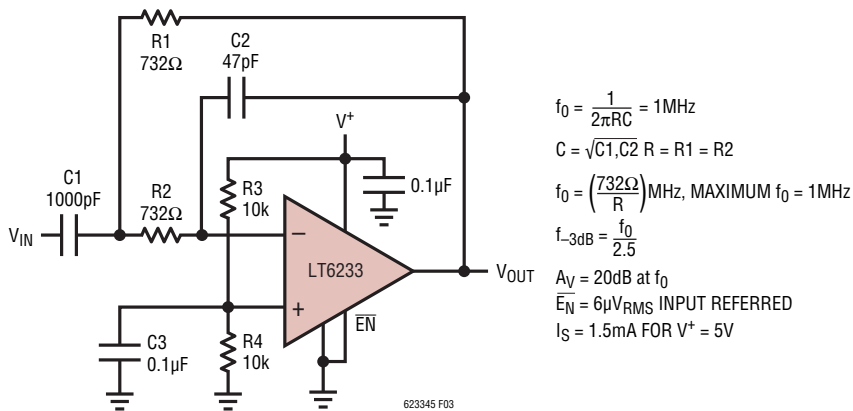
Enable Pin

The LT6233 and LT6233-10 include an $\overline{\text{ENABLE}}$ pin that shuts down the amplifier to 10μA maximum supply current. The $\overline{\text{ENABLE}}$ pin must be driven low to operate the amplifier with normal supply current. The $\overline{\text{ENABLE}}$ pin must be driven high to within 0.35V of V^+ to shut down the supply current. This can be accomplished with simple gate logic; however care must be taken if the logic and the LT6233 operate from different supplies. If this is the case, then open-drain logic can be used with a pull-up resistor to ensure that the amplifier remains off. See Typical Performance Characteristics.

The output leakage current when disabled is very low; however, current can flow into the input protection diodes D1 and D2 if the output voltage exceeds the input voltage by a diode drop.

TYPICAL APPLICATIONS

Single Supply, Low Noise, Low Power, Bandpass Filter with Gain = 10



$$f_0 = \frac{1}{2\pi RC} = 1\text{MHz}$$

$$C = \sqrt{C1 \cdot C2} \quad R = R1 = R2$$

$$f_0 = \left(\frac{732\Omega}{R} \right) \text{MHz, MAXIMUM } f_0 = 1\text{MHz}$$

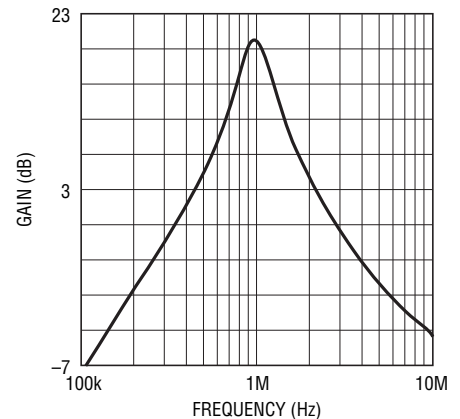
$$f_{-3dB} = \frac{f_0}{2.5}$$

$$A_V = 20\text{dB at } f_0$$

$$E_N = 6\mu\text{VRMS INPUT REFERRED}$$

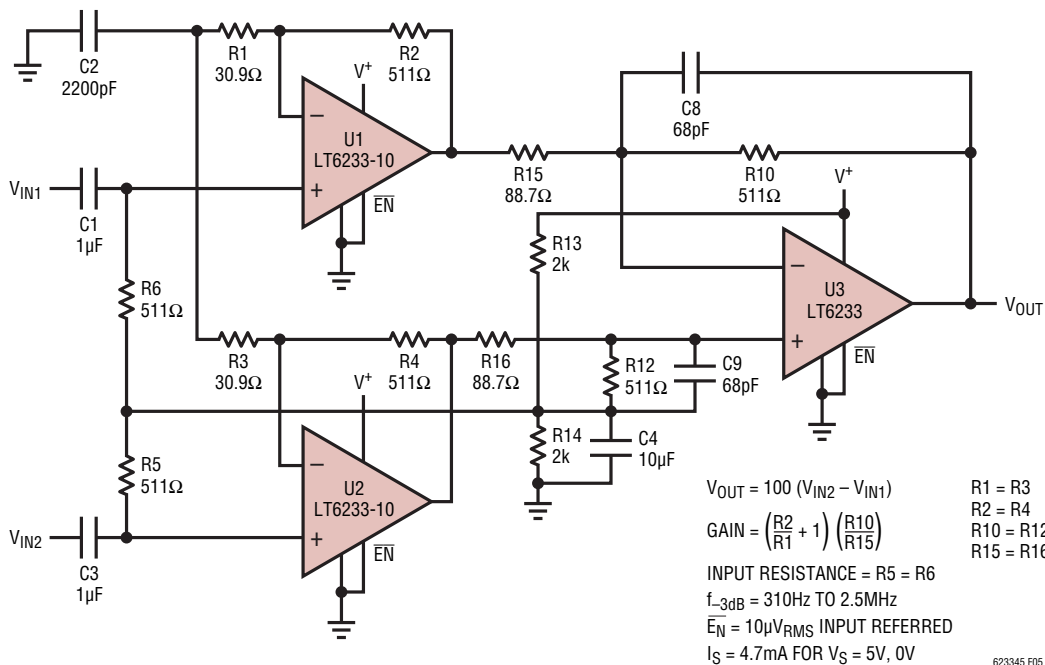
$$I_S = 1.5\text{mA FOR } V^+ = 5\text{V}$$

Frequency Response Plot of Bandpass Filter



623345 F04

Low Power, Low Noise, Single Supply, Instrumentation Amplifier with Gain = 100



$$V_{OUT} = 100 (V_{IN2} - V_{IN1})$$

$$\text{GAIN} = \left(\frac{R2}{R1} + 1 \right) \left(\frac{R10}{R15} \right)$$

$$\text{INPUT RESISTANCE} = R5 = R6$$

$$f_{-3dB} = 310\text{Hz TO } 2.5\text{MHz}$$

$$E_N = 10\mu\text{VRMS INPUT REFERRED}$$

$$I_S = 4.7\text{mA FOR } V_S = 5\text{V, } 0\text{V}$$

$$R1 = R3$$

$$R2 = R4$$

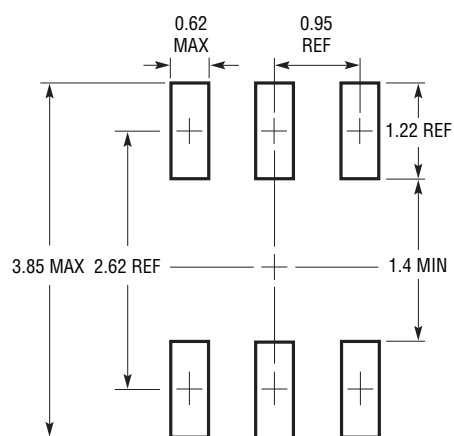
$$R10 = R12$$

$$R15 = R16$$

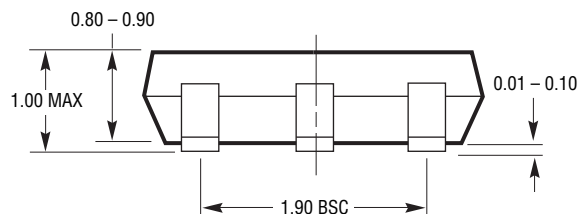
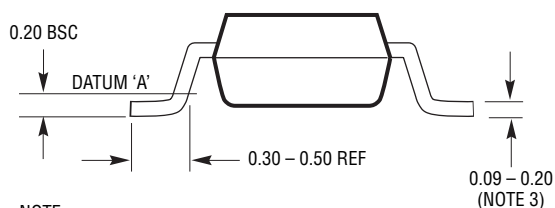
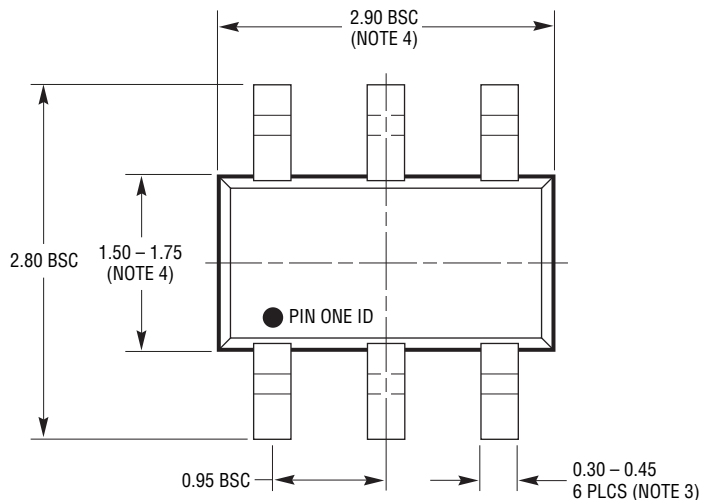
623345 F05

PACKAGE DESCRIPTION

S6 Package
6-Lead Plastic TSOT-23
(Reference LTC DWG # 05-08-1636)



RECOMMENDED SOLDER PAD LAYOUT
PER IPC CALCULATOR

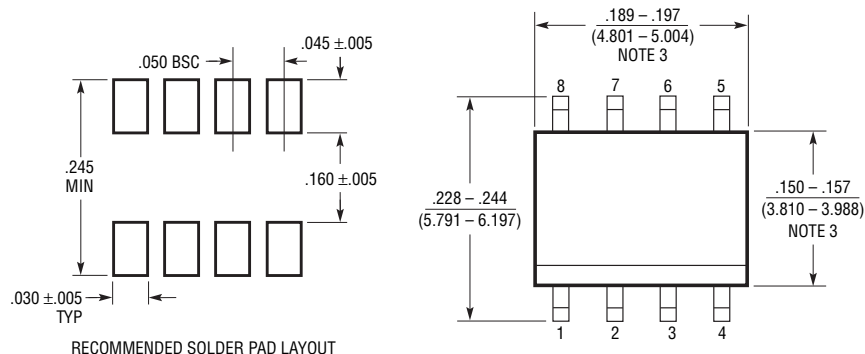


S6 TSOT-23 0302

- NOTE:
1. DIMENSIONS ARE IN MILLIMETERS
 2. DRAWING NOT TO SCALE
 3. DIMENSIONS ARE INCLUSIVE OF PLATING
 4. DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR
 5. MOLD FLASH SHALL NOT EXCEED 0.254mm
 6. JEDEC PACKAGE REFERENCE IS MO-193

PACKAGE DESCRIPTION

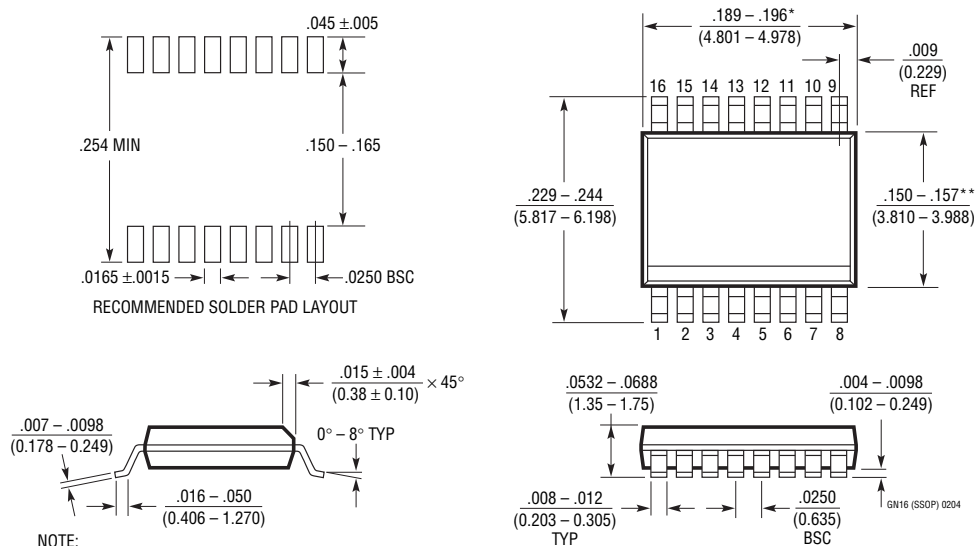
S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)



NOTE:
1. DIMENSIONS IN INCHES
(MILLIMETERS)
2. DRAWING NOT TO SCALE
3. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

S08 0303

GN Package 16-Lead Plastic SSOP (Narrow .150 Inch) (Reference LTC DWG # 05-08-1641)



NOTE:
1. CONTROLLING DIMENSION: INCHES
2. DIMENSIONS ARE IN INCHES
(MILLIMETERS)
3. DRAWING NOT TO SCALE
*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH
SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE
**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD
FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

GN16 (SSOP) 0204

REVISION HISTORY (Revision history begins at Rev C)

REV	DATE	DESCRIPTION	PAGE NUMBER
C	1/11	Revised y-axis label on curve G40 in Typical Performance Characteristics	14
		Updated ENABLE Pin section in Applications Information	18

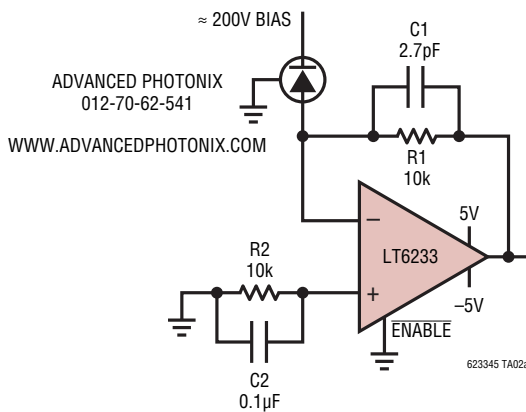
TYPICAL APPLICATIONS

The LT6233 is applied as a transimpedance amplifier with an I-to-V conversion gain of $10\text{k}\Omega$ set by R1. The LT6233 is ideally suited to this application because of its low input offset voltage and current, and its low noise. This is because the $10\text{k}\Omega$ resistor has an inherent thermal noise of $13\text{nV}/\sqrt{\text{Hz}}$ or $1.3\text{pA}/\sqrt{\text{Hz}}$ at room temperature, while the LT6233 contributes only 2nV and $0.8\text{pA}/\sqrt{\text{Hz}}$. So, with respect to both voltage and current noises, the LT6233 is actually quieter than the gain resistor.

The circuit uses an avalanche photodiode with the cathode biased to approximately 200V . When light is incident on

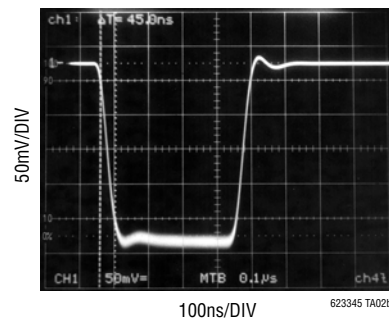
the photodiode, it induces a current I_{PD} which flows into the amplifier circuit. The amplifier output falls negative to maintain balance at its inputs. The transfer function is therefore $V_{OUT} = -I_{PD} \cdot 10\text{k}$. C1 ensures stability and good settling characteristics. Output offset was measured at better than $500\mu\text{V}$, so low in part because R2 serves to cancel the DC effects of bias current. Output noise was measured at below 1mV_{P-P} on a 20MHz measurement bandwidth, with C2 shunting R2's thermal noise. As shown in the scope photo, the rise time is 45ns , indicating a signal bandwidth of 7.8MHz .

Low Power Avalanche Photodiode Transimpedance Amplifier
 $I_S = 1.2\text{mA}$



OUTPUT OFFSET = $500\mu\text{V}$ TYPICAL
BANDWIDTH = 7.8MHz
OUTPUT NOISE = 1mV_{P-P} (20MHz MEASUREMENT BW)

Photodiode Amplifier Time Domain Response



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1028	Single, Ultralow Noise 50MHz Op Amp	$0.85\text{nV}/\sqrt{\text{Hz}}$
LT1677	Single, Low Noise Rail-to-Rail Amplifier	3V Operation, 2.5mA , $4.5\text{nV}/\sqrt{\text{Hz}}$, $60\mu\text{V}$ Max V_{OS}
LT1806/LT1807	Single/Dual, Low Noise 325MHz Rail-to-Rail Amplifier	2.5V Operation, $550\mu\text{V}$ Max V_{OS} , $3.5\text{nV}/\sqrt{\text{Hz}}$
LT6200/LT6201	Single/Dual, Low Noise 165MHz	$0.95\text{nV}/\sqrt{\text{Hz}}$, Rail-to-Rail Input and Output
LT6202/LT6203/LT6204	Single/Dual/Quad, Low Noise, Rail-to-Rail Amplifier	$1.9\text{nV}/\sqrt{\text{Hz}}$, 3mA Max, 100MHz Gain Bandwidth



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- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

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- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



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