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# 2-A Peak Sink/Source DDR Termination Regulator with VTTREF Buffered Reference For DDR2, DDR3 and DDR3L

Check for Samples: TPS51206

### FEATURES

- Supply Input Voltage: Supports 3.3-V Rail and 5-V Rail
- VLDOIN Input Voltage Range: VTT+0.4 V to 3.5 V
- VTT Termination Regulator
  - Output Voltage Range: 0.5 V to 0.9 V
  - 2-A Peak Sink and Source Current
  - Requires Only 10-µF MLCC Output Capacitor
  - ±20 mV Accuracy
- VTTREF Buffered Reference
  - VDDQ/2 ± 1% Accuracy
  - 10-mA Sink/Source Current
- Supports High-Z in S3 and Soft-Stop in S4/S5 with S3/S5 Inputs
- Over Temperature Protection
- 10-pin 2mm x 2mm SON(DSQ) Package

## APPLICATIONS

- DDR2/DDR3/DDR3L Memory Power Supplies
- SSTL\_18, SSTL\_15, SSTL\_135 and HSTL Termination

#### DESCRIPTION

The TPS51206 is a sink/source double date rate (DDR) termination regulator with VTTREF buffered reference output. It is specifically designed for low input voltage, low cost, low external component count systems where space is a key consideration. The TPS51206 maintains fast transient response and only requires  $1 \times 10$ -µF of ceramic output capacitance. The TPS51206 supports a remote sensing function and all power requirements for DDR2, DDR3 and Low-Power DDR3 (DDR3L) VTT bus. The VTT current capability is ±2A peak. The TPS51206 supports all of the DDR power states, putting VTT to High-Z in S3 state (suspend to RAM) and discharging VTT and VTTREF in S4/S5 state (suspend to disk).

The TPS51206 is available in 10-pin, 2x2, SON (DSQ) PowerPAD<sup>TM</sup> package and specified from -40°C to 85°C.

#### SIMPLIFIED APPLICATION



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet. PowerPAD is a trademark of Texas Instruments.

# TPS51206



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### ORDERING INFORMATION<sup>(1)(2)</sup>

T <sub>A</sub>	PACKAGE	ORDERABLE DEVICE NUMBER	PINS	OUTPUT SUPPLY	QUANTITY
40°C to 95°C	Diantia CON	TPS51206DSQR		Topo and Dool	3000
–40°C to 85°C	Plastic SON	TPS51206DSQT	10	Tape and Reel	250

(1) For the most current package and ordering information, see the *Package Option Addendum* at the end of this document, or visit the TI website at www.ti.com.

(2) Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package

#### **ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>**

			VALU	E	UNIT	
			VALUE   MIN MAX   -0.3 7   -0.3 3.6   -0.3 0.3   -0.3 3.6   2 500   125 150	UNIT		
	VDD, S3, S5		-0.3	7	V	
Input voltage range <sup>(2)</sup>	VLDOIN, VTTSNS, VDDQSNS		-0.3	3.6		
	PGND		-0.3	0.3	V	
Output voltage range <sup>(2)</sup>	Dutput voltage range <sup>(2)</sup> VTT, VTTREF					
Electrostatio discharge	HBM QSS 009-105 (JESD22-A114A)			2	kV	
Electrostatic discharge	CDM QSS 009-147 (JESD22-C101B.01)			500	V	
Junction temperature, $T_J$				125	°C	
Operating free-air temperature, T <sub>A</sub>			-55	150	°C	

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to the network ground terminal unless otherwise noted.

#### **RECOMMENDED OPERATING CONDITIONS**

		MIN	ΤΥΡ ΜΑΧ	UNIT
Supply voltage	VDD	3.1	6.5	V
Input voltage range <sup>(1)</sup>	S3, S5	-0.1	6.5	V
	VLDOIN, VTTSNS, VDDQSNS	-0.1	3.5	
	PGND	-0.1	0.1	
Output voltage range <sup>(1)</sup>	VTT, VTTREF	-0.1	3.5	V
Operating free-air temp	perature, T <sub>A</sub>	-40	85	°C

(1) All voltage values are with respect to the network ground terminal unless otherwise noted.

#### THERMAL INFORMATION

		TPS51206	
	THERMAL METRIC <sup>(1)</sup>	DSQ	UNITS
		10 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	70.3	
θ <sub>JCtop</sub>	Junction-to-case (top) thermal resistance	46.3	
$\theta_{JB}$	Junction-to-board thermal resistance	33.8	80.044
τυΨ	Junction-to-top characterization parameter	2.9	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	33.5	
θ <sub>JCbot</sub>	Junction-to-case (bottom) thermal resistance	16.3	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.



#### **ELECTRICAL CHARACTERISTICS**

over operating free-air temperature range,  $V_{VDD}$ =5 V, VLDOIN is connected to VDDQSNS,  $V_{S3}$ = $V_{S5}$ =5 V (unless otherwise noted)

	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT
SUPPLY CUP	RENT					
I <sub>VDD(S0)</sub>	VDD supply current, in S0	$T_{A}$ = 25°C, No load, $V_{S3}$ = $V_{S5}$ = 5 V, $V_{VDDQSNS}$ = 1.8 V		170		μA
I <sub>VDD(S3)</sub>	VDD supply current, in S3	$T_{A}$ = 25°C, No load, $V_{S3}$ = 0 V, $V_{S5}$ = 5 V, $V_{VDDQSNS}$ = 1.8 V		80		μA
IVDDSDN	VDD shutdown current, in S4/S5	$T_{A}$ = 25°C, No load, $V_{S3}$ = $V_{S5}$ = 0 V, $V_{VDDQSNS}$ = 1.8 V			1	μA
IVLDOIN(S0)	VLDOIN supply current, in S0	$T_A = 25^{\circ}C$ , No load, $V_{S3} = V_{S5} = 5$ V, $V_{LDION} = 1.8$ V			5	μA
I <sub>VLDOIN(s3)</sub>	VLDOIN supply current, in S3	$T_{A}$ = 25°C, No load, $V_{S3}$ = 0 V, $V_{S5}$ = 5 V, $V_{LDION}$ = 1.8 V			5	μA
IVLDOINSDN	VLDOIN shutdown current, in S4/S5	$T_{A}$ = 25°C, No load, $V_{S3}$ = $V_{S5}$ = 0 V, $V_{LDION}$ = 1.8 V			5	μA
VTTREF OUT	PUT				·	
V <sub>VTTREF</sub>	Output voltage		V	VDDQSNS/2		V
	Output voltage tolerance to	$ I_{VTTREF} $ < 10 mA, 1.5 V $\leq$ V <sub>VDDQSNS</sub> $\leq$ 1.8 V	49%	50%	51%	
V <sub>VTTREFTOL</sub>	VVDDQSNS	$ I_{VTTREF} $ < 10 mA, 1.2 V $\leq$ V <sub>VDDQSNS</sub> < 1.5 V	48.75%		51.25%	
IVTTREFSRC	Source current	V <sub>VDDQSNS</sub> = 1.8 V, V <sub>VTTREF</sub> = 0 V	10			mA
IVTTREFSNK	Sink current	V <sub>VDDQSNS</sub> = 0 V, V <sub>VTTREF</sub> = 1.8 V	10			mA
IVTTREFDIS	VTTREF Discharge current	$T_A = 25^{\circ}C, V_{S3} = V_{S5} = 0V, V_{VTTREF} = 0.5V$		1.3		mA
VTT OUTPUT	•	·				
V <sub>VTT</sub>	Output voltage		V	VDDQSNS/2		V
		I <sub>VTT</sub>  ≤ 10 mA, 1.4 V ≤ V <sub>VDDQSNS</sub> ≤ 1.8 V	-20		20	
		$ I_{VTT}  \le 1 \text{ A}, 1.4 \text{ V} \le V_{VDDQSNS} \le 1.8 \text{ V}^{(1)}$	-30		30	
V <sub>VTTTOL</sub>	Output voltage tolerance to	$ I_{VTT} $ < 2 A, 1.4 V ≤ V <sub>VDDQSNS</sub> ≤ 1.8 V <sup>(1)</sup>			40	mV
	V <sub>VDDQSNS</sub> /2	$ I_{VTT}  \le 10 \text{ mA}, 1.2 \text{ V} \le \text{V}_{VDDQSNS} \le 1.4 \text{ V}$ $ I_{VTT}  \le 1 \text{ A}, 1.2 \text{ V} \le \text{V}_{VDDQSNS} \le 1.4 \text{ V}^{(1)}$			20	
					30	
		$ I_{VTT} $ < 1.5 A, 1.2 V ≤ V <sub>VDDQSNS</sub> < 1.4 V <sup>(1)</sup>	-40		40	
IVTTOCLSRC	Source current limit	V <sub>VDDQSNS</sub> = 1.8 V, V <sub>VTT</sub> = V <sub>VTTSNS</sub> = 0.7 V	2			А
IVTTOCLSNK	Sink current limit	V <sub>VDDQSNS</sub> = 1.8 V, V <sub>VTT</sub> = V <sub>VTTSNS</sub> = 1.1 V	2			А
I <sub>VTTLK</sub>	Leakage current	$T_A = 25^{\circ}C$ , $V_{S3} = 0$ V, $V_{S5} = 5$ V, $V_{VTT} = V_{VTTREF}$			5	μA
IVTTSNSBIAS	VTTSNS input bias current	$V_{S3} = 5 V, V_{S5} = 5 V, V_{VTTSNS} = V_{VTTREF}$	-0.1		0.1	μA
IVTTSNSLK	VTTSNS leakage current	$V_{S3} = 0 V$ , $V_{S5} = 5 V$ , $V_{VTTSNS} = V_{VTTREF}$	-0.1		0.1	μA
IVTTDIS	VTT Discharge current	$T_A = 25^{\circ}C, V_{S3} = V_{S5} = V_{VDDQSNS} = 0 V, V_{VTT} = 0.5 V$		7		mA
VDDQ INPUT						
IVDDQSNS	VDDQSNS input current	V <sub>VDDQSNS</sub> = 1.8 V		30		μA
UVLO/LOGIC	THRESHOLD					
		Wake up		2.9		V
V <sub>VDDUV</sub>	VDD UVLO threshold voltage	Hysteresis		0.2		
V <sub>LL</sub>	S3/S5 low-level voltage				0.5	V
V <sub>LH</sub>	S3/S5 high-level voltage		1.8			V
V <sub>LHYST</sub>	S3/S5 hysteresis voltage			0.3		V
I <sub>LHLK</sub>	S3/S5 input leak current		-1		1	μA
	ERATURE PROTECTION	1	1			
		Shutdown temperature <sup>(1)</sup>		150		
T <sub>OTP</sub>	Over temperature protection	Hysteresis <sup>(1)</sup>	-	10		°C

(1) Ensured by design. Not production tested.

TEXAS INSTRUMENTS

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#### **DEVICE INFORMATION**



#### PIN FUNCTIONS

PIN I/O		1/0	DESCRIPTION
NAME	AE NO.		DESCRIPTION
GND	8	-	Signal ground
PGND	4	_	Power GND for VTT LDO
S3	7	I	S3 signal input
S5	9	I	S5 signal input
VDD	10	I	Device power supply input (3.3 V or 5 V)
VDDQSNS	1	I	VDDQ sense input, reference input for VTTREF
VLDOIN	2	I	Power supply input for VTT/ VTTREF
VTT	3	0	Power output for VTT LDO, need to connect 10-µF or greater MLCC for stability
VTTREF	6	0	VTTREF buffered reference output. Need to connect 0.22-µF or greater MLCC for stability
VTTSNS	5	I	VTT LDO voltage sense input
Pad	_	_	Solder to the ground plane for increased thremal performance.



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#### FUNCTIONAL BLOCK DIAGRAM



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Figure 5. VTTREF Load Regulation (0.9 V)





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**TYPICAL CHARACTERISTICS (continued)** 



Figure 7. VTTREF Load Regulation (0.675 V)



Figure 9. VTT Load Regulation (0.9 V)





Figure 8. VTTREF Load Regulation (0.6 V)



Figure 10. VTT Load Regulation (0.75 V)



Figure 12. VTT Load Regulation (0.6 V)





# **TYPICAL CHARACTERISTICS (continued)**

V<sub>VTTREF</sub>(10 mV/div) 0.9 V offset



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80.0ns/bt

I<sub>VTT</sub> (2 A/div)

80.0ns/bt

180 135

90

45

0

-45 -90

-135

Phase (°)

V<sub>VTTREF</sub>(10 mV/div) 0.75 V offset

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Ch1 Ch3

20.0mV 20.0mV

80

60

40

20

-20

-40

-60

Gain (dB) 0

8

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M 2.0s 125S/s A Ch4 \ 2.5Y

8.0ms/pt

Figure 22. VTT Dropout Voltage

Ch1 Ch3

500mV 5.0V

B<sub>W</sub> B<sub>W</sub>

Ch2 Ch4

500mV 5.0V

BW Time (2 s/div) Figure 21. Shutdown Waveforms (S3/ S5: High to Low)

**EXAS** 

**NSTRUMENTS** 



#### DETAILED DESCRIPTION

#### VTT SINK/SOURCE REGULATOR

The TPS51206 is a sink/source tracking termination regulator specifically designed for low input voltage, low cost, and low external component count systems where space is a key application parameter. The TPS51206 integrates a high-performance, low-dropout (LDO) linear regulator (VTT) that has ultimate fast response to track  $\frac{1}{2}$  VDDQSNS within 40 mV at all conditions, and its current capability is 2 A for both sink and source directions. A 10-µF (or greater) ceramic capacitor(s) need to be attached close to the VTT terminal for stable operation; X5R or better grade is recommended. To achieve tight regulation with minimum effect of trace resistance, the remote sensing terminal, VTTSNS, should be connected to the positive terminal of the output capacitor(s) as a separate trace from the high current path from the VTT pin.

The TPS51206 has a dedicated pin, VLDOIN, for VTT power supply to minimize the LDO power dissipation on user application. The minimum VLDOIN voltage is 0.4 V above the ½ VDDQSNS voltage.

#### VTTREF

The VTTREF pin includes 10 mA of sink/source current capability, and tracks  $\frac{1}{2}$  of VDDQSNS with  $\pm 1\%$  accuracy. A 0.22-µF ceramic capacitor needs to be attached close to the VTTREF terminal for stable operation; X5R or better grade is recommended.

#### **POWER STATE CONTROL**

The TPS51206 has two input pins, S3 and S5, to provide simple control of the power state. Table 1 describes S3/S5 terminal logic state and corresponding state of VTTREF/VTT outputs. VTT is turn-off and placed to high impedance (High-Z) state in S3. The VTT output is floated and does not sink or source current in this state. When both S5 and S3 pins are LOW, the power state is set to S4/S5. In S4/S5 state, all the outputs are turn-off and discharged to GND.

STATE	S3	S5	VTTREF	VTT
S0	HI	HI	ON	ON
S3	LO	HI	ON	OFF(High-Z)
S4/S5	LO	LO	OFF(Discharge)	OFF(Discharge)

Table 1. S3 and S5 Control Table

#### VDD UNDERVOLTAGE LOCKOUT PROTECTION

The TPS51206 input voltage (VDD) includes undervoltage lockout protection (UVLO). When the VDD pin voltage is lower than UVLO threshold voltage, VTT and VTTREF are shut off. This is non-latch protection.

#### **OVER-TEMPERATURE PROTECTION**

This device features internal temperature monitoring. If the temperature exceeds the threshold value, VTT and VTTREF are shut off. This is a non-latch protection.



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Figure 23. Typical Timing Diagram



#### **APPLICATION INFORMATION**

#### **VDD CAPACITOR**

Add a ceramic capacitor, with a value 0.1 µF (or greater) and X5R grade (or better), placed close to the VDD terminal, to stabilize the bias supply voltage from any parasitic impedance from the power supply rail.

#### **VLDOIN CAPACITOR**

Depending on the trace impedance between the VLDOIN bulk power supply to the device, a transient increase of source current is supplied mostly by the charge from the VLDOIN input capacitor. Use a  $10-\mu$ F (or greater) and X5R grade (or better) ceramic capacitor to supply this transient charge.

#### VTTREF CAPACITOR

Add a ceramic capacitor, with a value 0.22  $\mu$ F and X5R grade (or better), placed close to the VTTREF terminal for stable operation.

#### VTT CAPACITOR

For stable operation, a  $10-\mu$ F (or greater) and X5R (or better) grade ceramic capacitor(s) need to be attached close to the VTT terminal. This capacitor is recommended to minimize any additional equivalent series resistance (ESR) and/or equivalent series inductance (ESL) of ground trace between the PGND terminal and the VTT capacitor(s).

#### VTTSNS CONNECTION

To achieve tight regulation with minimum effect of trace resistance, a remote sensing terminal, the VTTSNS pin should be connected to the positive terminal of the VTT pin output capacitor(s) as a separate trace from the high-current path from VTT. Consider adding a low-pass R-C filter at the VTTSNS pin in case the ESR of the VTT output capacitor(s) is larger than 2 m $\Omega$ . The R-C filter time constant should be approximately the same or slightly lower than the time constant of the VTT output capacitance and ESR.



Figure 24. R-C Filter for VTTSNS

#### VDDQSNS CONNECTION

VDDQSNS is a reference input of the VTTREF and VTT. Trace should be routed away from noise-generating lines.



#### THERMAL DESIGN

Because the TPS51206 is a linear regulator, the VTT current flows in both source and sink directions, thereby dissipating power from the device. When the device is sourcing current, the voltage difference between VLDOIN and VTT times  $I_{VTT}$  (VTT current) current becomes the power dissipation as shown in Equation 1.

$$P_{DISS(src)} = (V_{VLDOIN} - V_{VTT}) \times I_{VTT(src)}$$

In this case, if the VLDOIN pin is connected to an alternative power supply lower than the VDDQ voltage, overall power loss can be reduced. For the sink phase, VTT voltage is applied across the internal LDO regulator, and the power dissipation can be calculated by Equation 2.

 $P_{DISS(snk)} = V_{VTT} \times I_{VTT(snk)}$ 

Maximum power dissipation allowed by the package is calculated by Equation 3.

$$P_{PKG} = \frac{T_{J(max)} - T_{A(max)}}{\theta_{JA}}$$

where

- T<sub>J(max)</sub> is +125°C
- T<sub>A(max)</sub> is the maximum ambient temperature in the system
- $\theta_{JA}$  is the thermal resistance from junction to ambient

(3)

(1)

(2)

TPS51206

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#### LAYOUT CONSIDERATIONS



Figure 25. PCB Layout Guideline

Consider the following before beginning a TPS51206 layout design.

- The input bypass capacitor for VLDOIN should be placed as close as possible to the terminal with short and wide connections.
- The output capacitor for VTT should be placed close to the terminals (VTT and PGND) with short and wide connection in order to avoid additional ESR and/or ESL trace inductance.
- VTTSNS should be connected to the positive node of VTT output capacitor(s) as a separate trace from the high current VTT power trace. In addition, VTTSNS trace should be routed away from high current trace, on the separate layer is recommended. This configuration is strongly recommended to avoid additional ESR and/or ESL. If sensing the voltage at the point of the load is required, it is recommended to attach the output capacitor(s) at that point. In addition, it is recommended to minimize any additional ESR and/or ESL of ground trace between the GND pin and the VTT capacitor(s).
- The GND pin (and the negative node of the VTTREF output capacitor) and PGND pins (and the negative node of the VTT output capacitor) should be connected to the internal system ground planes (for better result, use at least two internal ground planes) with multiple vias. Use as many vias as possible to reduce the impedance between GND/PGND and the system ground plane.
- In order to effectively remove heat from the package, properly prepare the thermal land. Apply solder directly to the package thermal pad. The wide traces of the component and the side copper connected to the thermal land pad help to dissipate heat. Numerous vias 0.33 mm in diameter connected from the thermal land to the internal/solder side ground plane(s) should also be used to help dissipation. Please consult the *TPS51206-EVM User's Guide* for more detailed layout recommendations.



#### **APPLICATION DIAGRAMS**

Figure 26 shows an application diagram for a configuration where VLDOIN and VDDQ are connected.



Figure 26.	VLDOIN=VDDQ	Configuration
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REFERENCE DESIGNATOR	SPECIFICATION	MANUFACTURER	PART NUMBER
C1, C3	10 µF, 6.3 V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C2	0.1 μF, 6.3 V, X5R, 1005 (0402)	Taiyo Yuden	JWK105BJ104MP
C4	0.22 µF, 6.3 V, X5R, 1005 (0402)	Taiyo Yuden	JMK105BJ224KV

#### Table 2. VLDOIN=VDDQ Configuration Components

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Figure 27 shows an application diagram for a configuration where VLDOIN and VDDQ are separated.



Figure 27. VLDOIN Separated from VDDQ Configuration

REFERNCE DESIGNATOR	SPECIFICATION	MANUFACTURER	PART NUMBER
C1, C3	10 µF, 6.3V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C2	0.1 µF, 6.3V, X5R, 1005 (0402)	Taiyo Yuden	JWK105BJ104MP
C3	10 µF, 6.3V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C4	0.22 µF, 6.3V, X5R, 1005 (0402)	Taiyo Yuden	JMK105BJ224KV

Table 3. VLDOIN Separated from VDDQ Configuration Components



24-Jan-2013

#### PACKAGING INFORMATION

Orderable Device	Status	Package Type	•	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing			(2)		(3)		(4)	
TPS51206DSQR	ACTIVE	SON	DSQ	10	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1206	Samples
TPS51206DSQT	ACTIVE	SON	DSQ	10	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1206	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND**: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> Only one of markings shown within the brackets will appear on the physical device.

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# PACKAGE MATERIALS INFORMATION

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#### TAPE AND REEL INFORMATION





# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS51206DSQR	SON	DSQ	10	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TPS51206DSQT	SON	DSQ	10	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2

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# PACKAGE MATERIALS INFORMATION

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS51206DSQR	SON	DSQ	10	3000	210.0	185.0	35.0
TPS51206DSQT	SON	DSQ	10	250	210.0	185.0	35.0

# **MECHANICAL DATA**



- C. Small Outline No-Lead (SON) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.





NOTES: A. All linear dimensions are in millimeters





NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for solder mask tolerances.



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