

### FEATURES

#### Computes

- True rms value
- Average rectified value
- Absolute value

#### Provides

- 200 mV full-scale input range (larger inputs with input scaling)
- Direct interfacing with 3½ digit CMOS ADCs
- High input impedance:  $10^{12} \Omega$
- Low input bias current: 25 pA maximum
- High accuracy:  $\pm 0.2 \text{ mV} \pm 0.3\%$  of reading
- RMS conversion with signal crest factors up to 5
- Wide power supply range:  $\pm 2.5 \text{ V}$  to  $\pm 16.5 \text{ V}$
- Low power: 25  $\mu\text{A}$  (typical) standby current
- No external trims needed for specified accuracy

The **AD737** output is negative-going; the **AD736** is a positive output-going version of the same basic device

### GENERAL DESCRIPTION

The **AD737** is a low power, precision, monolithic, true rms-to-dc converter. It is laser trimmed to provide a maximum error of  $\pm 0.2 \text{ mV} \pm 0.3\%$  of reading with sine wave inputs. Furthermore, it maintains high accuracy while measuring a wide range of input waveforms, including variable duty cycle pulses and triac (phase) controlled sine waves. The low cost and small physical size of this converter make it suitable for upgrading the performance of non-rms precision rectifiers in many applications. Compared to these circuits, the **AD737** offers higher accuracy at equal or lower cost.

The **AD737** can compute the rms value of both ac and dc input voltages. It can also be operated ac-coupled by adding one external capacitor. In this mode, the **AD737** can resolve input signal levels of 100  $\mu\text{V}$  rms or less, despite variations in temperature or supply voltage. High accuracy is also maintained for input waveforms with crest factors of 1 to 3. In addition, crest factors as high as 5 can be measured (while introducing only 2.5% additional error) at the 200 mV full-scale input level.

The **AD737** has no output buffer amplifier, thereby significantly reducing dc offset errors occurring at the output, which makes the device highly compatible with high input impedance ADCs.

Requiring only 160  $\mu\text{A}$  of power supply current, the **AD737** is optimized for use in portable multimeters and other battery-powered applications. In power-down mode, the standby supply current in is typically 25  $\mu\text{A}$ .

### FUNCTIONAL BLOCK DIAGRAM

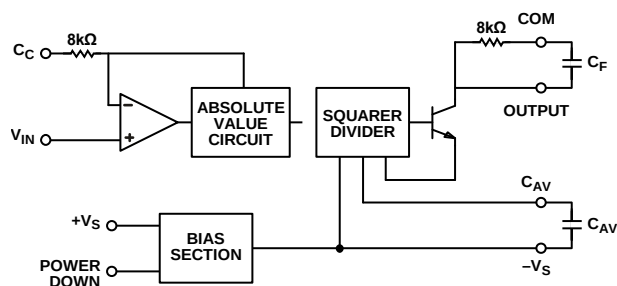


Figure 1.

The **AD737** has both high ( $10^{12} \Omega$ ) and low impedance input options. The high-Z FET input connects high source impedance input attenuators, and a low impedance (8 k $\Omega$ ) input accepts rms voltages to 0.9 V while operating from the minimum power supply voltage of  $\pm 2.5 \text{ V}$ . The two inputs can be used either single ended or differentially.

The **AD737** achieves 1% of reading error bandwidth, exceeding 10 kHz for input amplitudes from 20 mV rms to 200 mV rms, while consuming only 0.72 mW.

The **AD737** is available in two performance grades. The **AD737J** and **AD737K** grades operate over the commercial temperature range of  $0^\circ\text{C}$  to  $70^\circ\text{C}$ . The **AD737JR-5** is tested with supply voltages of  $\pm 2.5 \text{ V}$  dc. The **AD737A** grade operates over the industrial temperature range of  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ . The **AD737** is available in two low cost, 8-lead packages: PDIP and SOIC\_N.

### PRODUCT HIGHLIGHTS

1. Computes average rectified, absolute, or true rms value of a signal regardless of waveform.
2. Only one external component, an averaging capacitor, is required for the **AD737** to perform true rms measurement.
3. The standby power consumption of 125  $\mu\text{W}$  makes the **AD737** suitable for battery-powered applications.

#### Rev. I

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## REVISION HISTORY

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### 12/99—Rev. B to Rev. C

## SPECIFICATIONS

$T_A = 25^\circ\text{C}$ ,  $\pm V_S = \pm 5\text{ V}$  except as noted,  $C_{AV} = 33\text{ }\mu\text{F}$ ,  $C_C = 10\text{ }\mu\text{F}$ ,  $f = 1\text{ kHz}$ , sine wave input applied to Pin 2, unless otherwise specified. Specifications shown in **boldface** are tested on all production units at final electrical test. Results from these tests are used to calculate outgoing quality levels.

Table 1.

| Parameter                                        | Test Conditions/<br>Comments                                                                                                                                            | AD737A, AD737J |         |                        | AD737K   |         |                        | AD737J-5 |         |                | Unit                                                                                               |
|--------------------------------------------------|-------------------------------------------------------------------------------------------------------------------------------------------------------------------------|----------------|---------|------------------------|----------|---------|------------------------|----------|---------|----------------|----------------------------------------------------------------------------------------------------|
|                                                  |                                                                                                                                                                         | Min            | Typ     | Max                    | Min      | Typ     | Max                    | Min      | Typ     | Max            |                                                                                                    |
| ACCURACY                                         |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
| Total Error                                      | $E_{IN} = 0$ to $200\text{ mV rms}$<br>$\pm V_S = \pm 2.5\text{ V}$<br>$\pm V_S = \pm 2.5\text{ V}$ ,<br>input to Pin 1<br>$E_{IN} = 200\text{ mV}$ to $1\text{ V rms}$ |                | 0.2/0.3 | <b>0.4/0.5</b>         |          | 0.2/0.2 | <b>0.2/0.3</b>         |          | 0.2/0.3 | <b>0.4/0.5</b> | $\pm\text{mV}/\pm\text{POR}^1$<br>$\pm\text{mV}/\pm\text{POR}^1$<br>$\pm\text{mV}/\pm\text{POR}^1$ |
| Over Temperature                                 |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
| JN, JR, KR                                       | $E_{IN} = 200\text{ mV rms}$ ,<br>$\pm V_S = \pm 2.5\text{ V}$                                                                                                          |                | 0.007   |                        |          | 0.007   |                        |          | 0.02    |                | $\pm\text{POR}/^\circ\text{C}$                                                                     |
| AN and AR                                        | $E_{IN} = 200\text{ mV rms}$ ,<br>$\pm V_S = \pm 2.5\text{ V}$                                                                                                          |                | 0.014   |                        |          | 0.014   |                        |          |         |                | $\pm\text{POR}/^\circ\text{C}$                                                                     |
| vs. Supply Voltage                               |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
|                                                  | $E_{IN} = 200\text{ mV rms}$ ,<br>$\pm V_S = \pm 2.5\text{ V}$ to $\pm 5\text{ V}$                                                                                      | <b>0</b>       | -0.18   | <b>-0.3</b>            | <b>0</b> | -0.18   | <b>-0.3</b>            | <b>0</b> | -0.18   | <b>-0.3</b>    | %/V                                                                                                |
|                                                  | $E_{IN} = 200\text{ mV rms}$ ,<br>$\pm V_S = \pm 5\text{ V}$ to $\pm 16.5\text{ V}$                                                                                     | <b>0</b>       | 0.06    | <b>0.1</b>             | <b>0</b> | 0.06    | <b>0.1</b>             | <b>0</b> | 0.06    | <b>0.1</b>     | %/V                                                                                                |
| DC Reversal Error                                | DC-coupled,<br>$V_{IN} = 600\text{ mV dc}$<br>$\pm V_S = \pm 2.5\text{ V}$<br>$V_{IN} = 200\text{ mV dc}$                                                               |                | 1.3     | 2.5                    |          | 1.3     | 2.5                    |          | 1.7     | 2.5            | POR<br>POR                                                                                         |
| Nonlinearity <sup>2</sup>                        | $E_{IN} = 0$ mV to<br>$200\text{ mV rms}$ ,<br>@ $100\text{ mV rms}$                                                                                                    | <b>0</b>       | 0.25    | <b>0.35</b>            | <b>0</b> | 0.25    | <b>0.35</b>            |          |         |                | POR                                                                                                |
| Input to Pin 1 <sup>3</sup>                      | AC coupled,<br>$E_{IN} = 100\text{ mV rms}$ , after<br>correction, $\pm V_S = \pm 2.5\text{ V}$                                                                         |                |         |                        |          |         |                        |          | 0.02    | 0.1            | POR                                                                                                |
| Total Error,<br>External Trim                    | $E_{IN} = 0$ mV to<br>$200\text{ mV rms}$                                                                                                                               |                | 0.1/0.2 |                        |          | 0.1/0.2 |                        |          | 0.1/0.2 |                | $\pm\text{mV}/\pm\text{POR}$                                                                       |
| ADDITIONAL<br>CREST FACTOR<br>ERROR <sup>4</sup> |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
| For Crest Factors<br>from 1 to 3                 | $C_{AV} = C_F = 100\text{ }\mu\text{F}$                                                                                                                                 |                | 0.7     |                        |          | 0.7     |                        |          |         |                | %                                                                                                  |
|                                                  | $C_{AV} = 22\text{ }\mu\text{F}$ , $C_F = 100\text{ }\mu\text{F}$ ,<br>$\pm V_S = \pm 2.5\text{ V}$ , input to<br>Pin 1                                                 |                |         |                        |          |         |                        |          | 1.7     |                | %                                                                                                  |
| For Crest Factors<br>from 3 to 5                 | $C_{AV} = C_F = 100\text{ }\mu\text{F}$                                                                                                                                 |                | 2.5     |                        |          | 2.5     |                        |          |         |                | %                                                                                                  |
| INPUT<br>CHARACTERISTICS                         |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
| High-Z Input (Pin 2)<br>Signal Range             |                                                                                                                                                                         |                |         |                        |          |         |                        |          |         |                |                                                                                                    |
| Continuous<br>RMS Level                          | $\pm V_S = +2.5\text{ V}$<br><br>$\pm V_S = +2.8\text{ V}/-3.2\text{ V}$<br>$\pm V_S = \pm 5\text{ V}$ to $\pm 16.5\text{ V}$                                           |                |         | <b>200</b><br><b>1</b> |          |         | <b>200</b><br><b>1</b> |          |         | <b>200</b>     | mV rms<br>mV rms<br>V rms                                                                          |

| Parameter                                  | Test Conditions/<br>Comments                    | AD737A, AD737J |           |          | AD737K    |           |           | AD737J-5  |      |          | Unit           |
|--------------------------------------------|-------------------------------------------------|----------------|-----------|----------|-----------|-----------|-----------|-----------|------|----------|----------------|
|                                            |                                                 | Min            | Typ       | Max      | Min       | Typ       | Max       | Min       | Typ  | Max      |                |
| Peak Transient Input                       | $\pm V_S = +2.5$ V input to Pin 1               |                |           |          |           |           |           | $\pm 0.6$ |      |          | V              |
|                                            | $\pm V_S = +2.8$ V/–3.2 V                       | $\pm 0.9$      |           |          | $\pm 0.9$ |           |           |           |      |          | V              |
|                                            | $\pm V_S = \pm 5$ V                             |                | $\pm 2.7$ |          |           |           | $\pm 2.7$ |           |      |          | V              |
|                                            | $\pm V_S = \pm 16.5$ V                          | $\pm 4.0$      |           |          | $\pm 4.0$ |           |           |           |      |          | V              |
| Input Resistance                           |                                                 |                | 1012      |          |           | 1012      |           |           | 1012 |          | $\Omega$       |
| Input Bias Current                         | $\pm V_S = \pm 5$ V                             |                | 1         | 25       |           | 1         | 25        |           | 1    | 25       | pA             |
| Low-Z Input (Pin 1) Signal Range           |                                                 |                |           |          |           |           |           |           |      |          |                |
| Continuous RMS Level                       | $\pm V_S = +2.5$ V                              |                |           |          |           |           |           |           |      | 300      | mV rms         |
|                                            | $\pm V_S = +2.8$ V/–3.2 V                       |                |           | 300      |           |           | 300       |           |      |          | mV rms         |
|                                            | $\pm V_S = \pm 5$ V to $\pm 16.5$ V             |                |           | 1        |           |           | 1         |           |      |          | V rms          |
| Peak Transient Input                       | $\pm V_S = +2.5$ V                              |                |           |          |           |           |           | $\pm 1.7$ |      |          | V              |
|                                            | $\pm V_S = +2.8$ V/–3.2 V                       |                | $\pm 1.7$ |          |           | $\pm 1.7$ |           |           |      |          | V              |
|                                            | $\pm V_S = \pm 5$ V                             |                | $\pm 3.8$ |          |           | $\pm 3.8$ |           |           |      |          | V              |
|                                            | $\pm V_S = \pm 16.5$ V                          |                | $\pm 11$  |          |           | $\pm 11$  |           |           |      |          | V              |
| Input Resistance                           |                                                 | 6.4            | 8         | 9.6      | 6.4       | 8         | 9.6       | 6.4       | 8    | 9.6      | k $\Omega$     |
| Maximum Continuous Nondestructive Input    | All supply voltages                             |                |           | $\pm 12$ |           |           | $\pm 12$  |           |      | $\pm 12$ | V p-p          |
| Input Offset Voltage <sup>5</sup>          | AC-coupled                                      |                |           | $\pm 3$  |           |           | $\pm 3$   |           |      | $\pm 3$  | mV             |
| Over the Rated Operating Temperature Range |                                                 |                | 8         | 30       |           | 8         | 30        |           | 8    | 30       | $\mu$ V/°C     |
| vs. Supply                                 | $V_S = \pm 2.5$ V to $\pm 5$ V                  |                | 80        |          |           | 80        |           |           | 80   |          | $\mu$ V/V      |
|                                            | $V_S = \pm 5$ V to $\pm 16.5$ V                 |                | 50        | 150      |           | 50        | 150       |           |      |          | $\mu$ V/V      |
| OUTPUT CHARACTERISTICS                     |                                                 |                |           |          |           |           |           |           |      |          |                |
|                                            | No load, output is negative with respect to COM |                |           |          |           |           |           |           |      |          |                |
| Output Voltage Range                       | $\pm V_S = +2.8$ V/–3.2 V                       | –1.6           | –1.7      |          | –1.6      | –1.7      |           |           |      |          | V <sup>6</sup> |
|                                            | $\pm V_S = \pm 5$ V                             | –3.3           | –3.4      |          | –3.3      | –3.4      |           |           |      |          | V <sup>6</sup> |
|                                            | $\pm V_S = \pm 16.5$ V                          | –4             | –5        |          | –4        | –5        |           |           |      |          | V              |
|                                            | $\pm V_S = \pm 2.5$ V, input to Pin 1           |                |           |          |           |           |           | –1.1      | –0.9 |          | V <sup>6</sup> |
| Output Resistance                          | DC                                              | 6.4            | 8         | 9.6      | 6.4       | 8         | 9.6       | 6.4       | 8    | 9.6      | k $\Omega$     |
| FREQUENCY RESPONSE                         |                                                 |                |           |          |           |           |           |           |      |          |                |
| High-Z Input (Pin 2)                       |                                                 |                |           |          |           |           |           |           |      |          |                |
| 1% Additional Error                        | $V_{IN} = 1$ mV rms                             |                | 1         |          |           | 1         |           |           | 1    |          | kHz            |
|                                            | $V_{IN} = 10$ mV rms                            |                | 6         |          |           | 6         |           |           | 6    |          | kHz            |
|                                            | $V_{IN} = 100$ mV rms                           |                | 37        |          |           | 37        |           |           | 37   |          | kHz            |
|                                            | $V_{IN} = 200$ mV rms                           |                | 33        |          |           | 33        |           |           | 33   |          | kHz            |

| Parameter                    | Test Conditions/<br>Comments | AD737A, AD737J |         |            | AD737K        |         |            | AD737J-5  |         |            | Unit    |
|------------------------------|------------------------------|----------------|---------|------------|---------------|---------|------------|-----------|---------|------------|---------|
|                              |                              | Min            | Typ     | Max        | Min           | Typ     | Max        | Min       | Typ     | Max        |         |
| 3 dB Bandwidth               | $V_{IN} = 1$ mV rms          |                | 5       |            |               | 5       |            |           | 5       |            | kHz     |
|                              | $V_{IN} = 10$ mV rms         |                | 55      |            |               | 55      |            |           | 55      |            | kHz     |
|                              | $V_{IN} = 100$ mV rms        |                | 170     |            |               | 170     |            |           | 170     |            | kHz     |
|                              | $V_{IN} = 200$ mV rms        |                | 190     |            |               | 190     |            |           | 190     |            | kHz     |
| Low-Z Input<br>(Pin 1)       |                              |                |         |            |               |         |            |           |         |            |         |
|                              | 1% Additional Error          |                | 1       |            |               | 1       |            |           | 1       |            | kHz     |
|                              | $V_{IN} = 1$ mV rms          |                | 1       |            |               | 1       |            |           | 1       |            | kHz     |
|                              | $V_{IN} = 10$ mV rms         |                | 6       |            |               | 6       |            |           | 6       |            | kHz     |
| 3 dB Bandwidth               | $V_{IN} = 40$ mV rms         |                |         |            |               |         |            |           | 25      |            | kHz     |
|                              | $V_{IN} = 100$ mV rms        |                | 90      |            |               | 90      |            |           | 90      |            | kHz     |
|                              | $V_{IN} = 200$ mV rms        |                | 90      |            |               | 90      |            |           | 90      |            | kHz     |
|                              | $V_{IN} = 1$ mV rms          |                | 5       |            |               | 5       |            |           | 5       |            | kHz     |
|                              | $V_{IN} = 10$ mV rms         |                | 55      |            |               | 55      |            |           | 55      |            | kHz     |
|                              | $V_{IN} = 100$ mV rms        |                | 350     |            |               | 350     |            |           | 350     |            | kHz     |
|                              | $V_{IN} = 200$ mV rms        |                | 460     |            |               | 460     |            |           | 460     |            | kHz     |
|                              |                              |                |         |            |               |         |            |           |         |            |         |
| POWER-DOWN<br>MODE           |                              |                |         |            |               |         |            |           |         |            |         |
| Disable Voltage              |                              |                | 0       |            |               | 0       |            |           |         |            | V       |
| Input Current,<br>PD Enabled | $V_{PD} = V_S$               |                | 11      |            |               | 11      |            |           |         |            | $\mu$ A |
| POWER SUPPLY                 |                              |                |         |            |               |         |            |           |         |            |         |
| Operating<br>Voltage Range   |                              | +2.8/<br>-3.2  | $\pm 5$ | $\pm 16.5$ | +2.8/<br>-3.2 | $\pm 5$ | $\pm 16.5$ | $\pm 2.5$ | $\pm 5$ | $\pm 16.5$ | V       |
| Current                      | No input                     |                | 120     | <b>160</b> |               | 120     | <b>160</b> |           | 120     | <b>160</b> | $\mu$ A |
|                              | Rated input                  |                | 170     | 210        |               | 170     | 210        |           | 170     | 210        | $\mu$ A |
|                              | Powered down                 |                | 25      | 40         |               | 25      | 40         |           | 25      | 40         | $\mu$ A |

<sup>1</sup> POR is % of reading.

<sup>2</sup> Nonlinearity is defined as the maximum deviation (in percent error) from a straight line connecting the readings at 0 V and at 200 mV rms.

<sup>3</sup> After fourth-order error correction using the equation

$$y = -0.31009x^4 - 0.21692x^3 - 0.06939x^2 + 0.99756x + 11.1 \times 10^{-6}$$

where y is the corrected result and x is the device output between 0.01 V and 0.3 V.

<sup>4</sup> Crest factor error is specified as the additional error resulting from the specific crest factor, using a 200 mV rms signal as a reference. The crest factor is defined as  $V_{PEAK}/V_{rms}$ .

<sup>5</sup> DC offset does not limit ac resolution.

<sup>6</sup> Value is measured with respect to COM.

## ABSOLUTE MAXIMUM RATINGS

Table 2.

| Parameter                            | Rating                                      |
|--------------------------------------|---------------------------------------------|
| Supply Voltage                       | $\pm 16.5\text{ V}$                         |
| Internal Power Dissipation           | 200 mW                                      |
| Input Voltage                        |                                             |
| Pin 1                                | $\pm 12\text{ V}$                           |
| Pin 2 to Pin 8                       | $\pm V_s$                                   |
| Output Short-Circuit Duration        | Indefinite                                  |
| Differential Input Voltage           | $+V_s$ and $-V_s$                           |
| Storage Temperature Range            | $-65^\circ\text{C}$ to $+125^\circ\text{C}$ |
| Lead Temperature, Soldering (60 sec) | $300^\circ\text{C}$                         |
| ESD Rating                           | 500 V                                       |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL RESISTANCE

$\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 3. Thermal Resistance

| Package Type        | $\theta_{JA}$ | Unit               |
|---------------------|---------------|--------------------|
| 8-Lead PDIP (N-8)   | 165           | $^\circ\text{C/W}$ |
| 8-Lead SOIC_N (R-8) | 155           | $^\circ\text{C/W}$ |

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

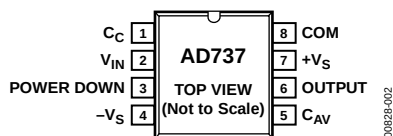


Figure 2. SOIC\_N Pin Configuration (R-8)

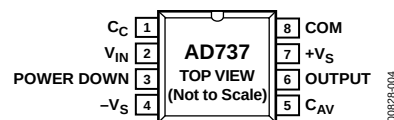


Figure 3. PDIP Pin Configuration (N-8)

Table 4. Pin Function Descriptions

| Pin No. | Mnemonic   | Description                                                                |
|---------|------------|----------------------------------------------------------------------------|
| 1       | $C_C$      | Coupling Capacitor for Indirect DC Coupling.                               |
| 2       | $V_{IN}$   | RMS Input.                                                                 |
| 3       | POWER DOWN | Disables the <a href="#">AD737</a> . Low is enabled; high is powered down. |
| 4       | $-V_S$     | Negative Power Supply.                                                     |
| 5       | $C_{AV}$   | Averaging Capacitor.                                                       |
| 6       | OUTPUT     | Output.                                                                    |
| 7       | $+V_S$     | Positive Power Supply.                                                     |
| 8       | COM        | Common.                                                                    |

## TYPICAL PERFORMANCE CHARACTERISTICS

$T_A = 25^\circ\text{C}$ ,  $\pm V_S = \pm 5\text{ V}$  (except AD737J-5, where  $\pm V_S = \pm 2.5\text{ V}$ ),  $C_{AV} = 33\text{ }\mu\text{F}$ ,  $C_C = 10\text{ }\mu\text{F}$ ,  $f = 1\text{ kHz}$ , sine wave input applied to Pin 2, unless otherwise specified.

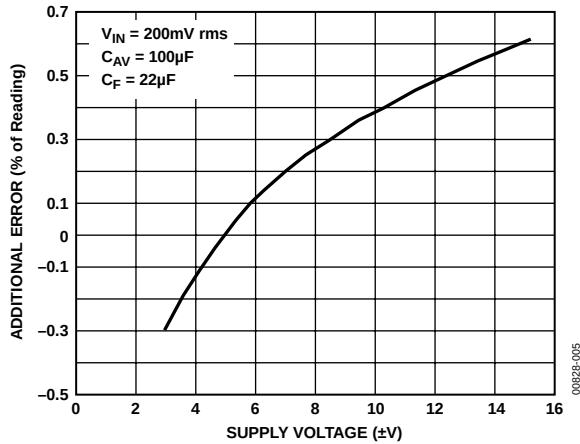


Figure 4. Additional Error vs. Supply Voltage

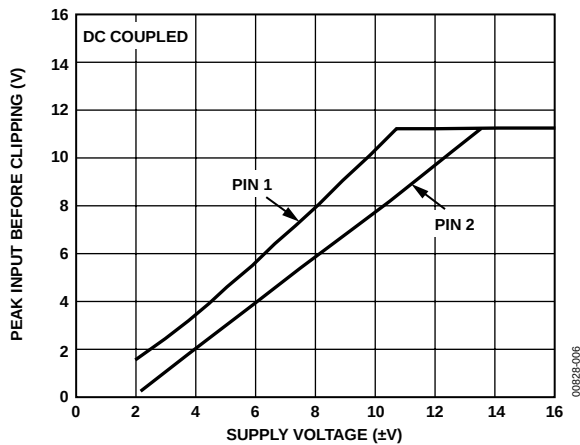


Figure 5. Peak Input Level for 1% Saturation vs. Supply Voltage

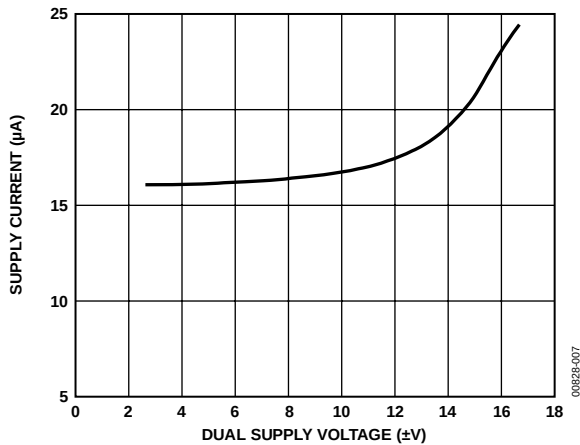


Figure 6. Supply Current (Power-Down Mode) vs. Supply Voltage (Dual)

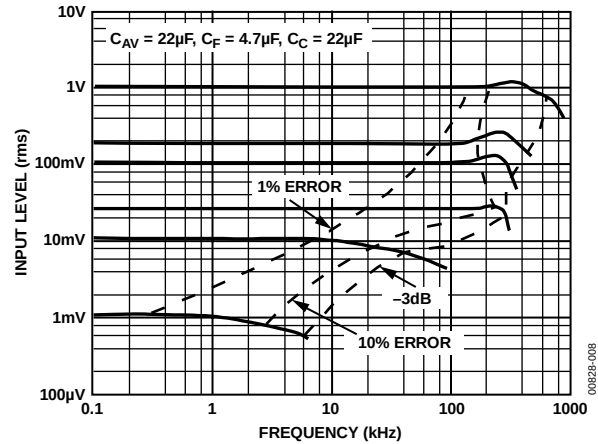


Figure 7. Frequency Response Driving Pin 1; Negative DC Output

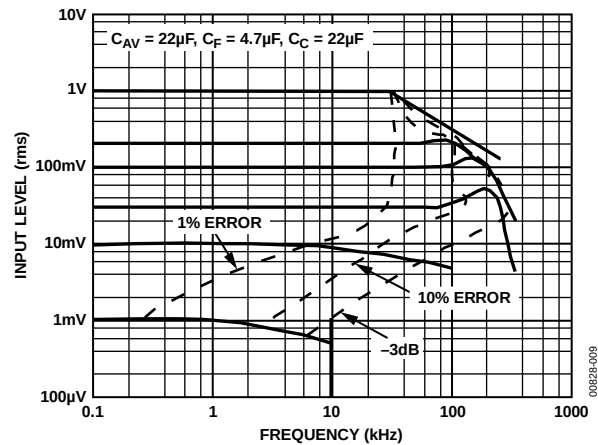


Figure 8. Frequency Response Driving Pin 2; Negative DC Output

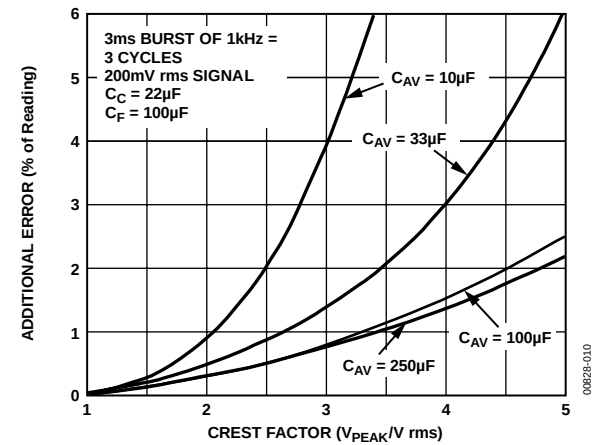


Figure 9. Additional Error vs. Crest Factor



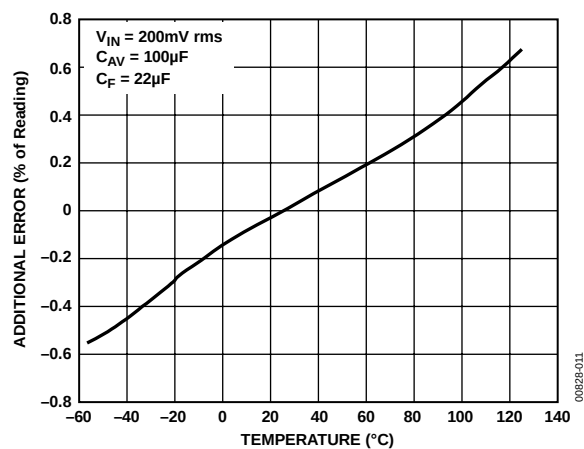


Figure 10. Additional Error vs. Temperature

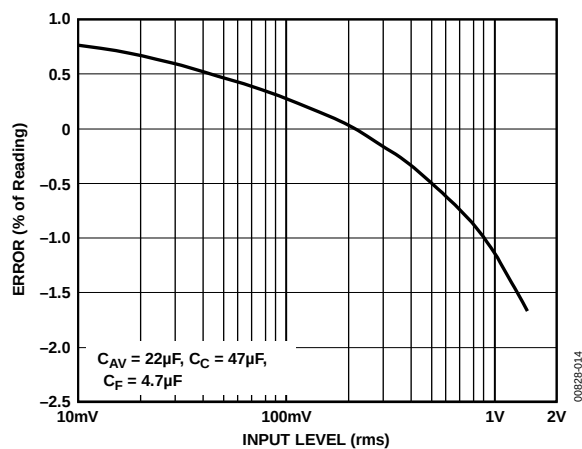


Figure 13. Error vs. RMS Input Level Using Circuit in Figure 29

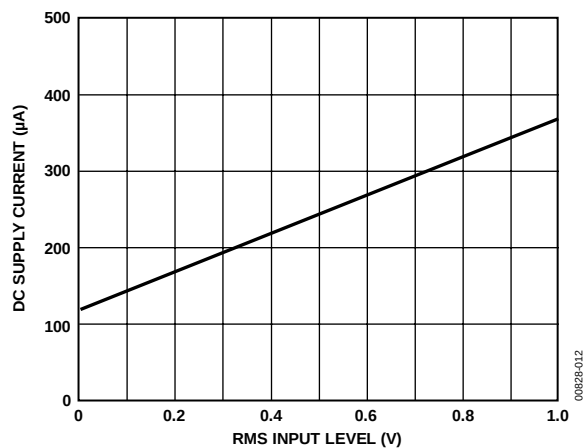


Figure 11. DC Supply Current vs. RMS Input Level

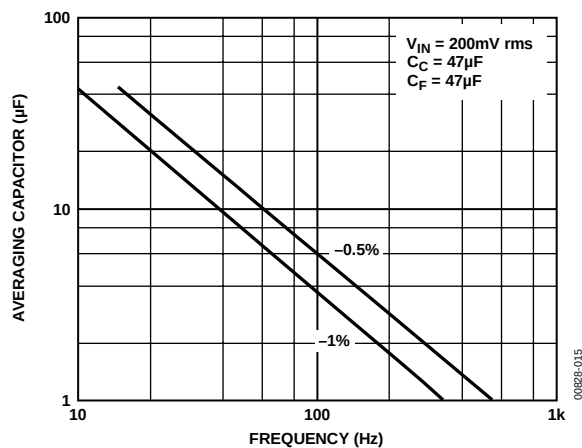


Figure 14. Value of Averaging Capacitor vs. Frequency for Specified Averaging Error

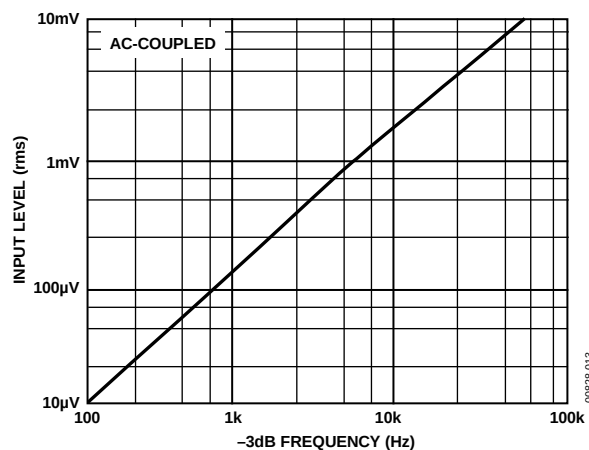


Figure 12. RMS Input Level vs. -3 dB Frequency; Negative DC Output

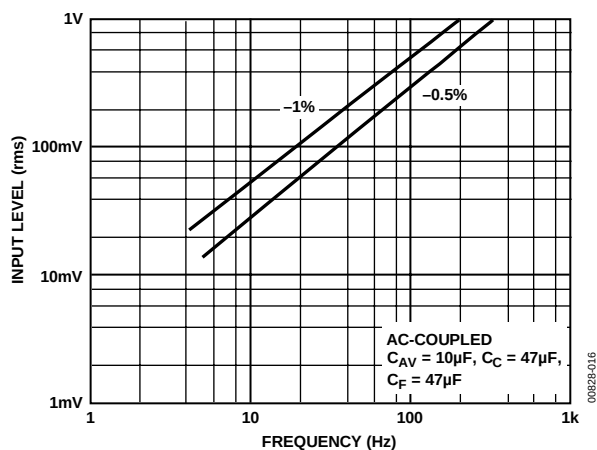


Figure 15. RMS Input Level vs. Frequency for Specified Averaging Error

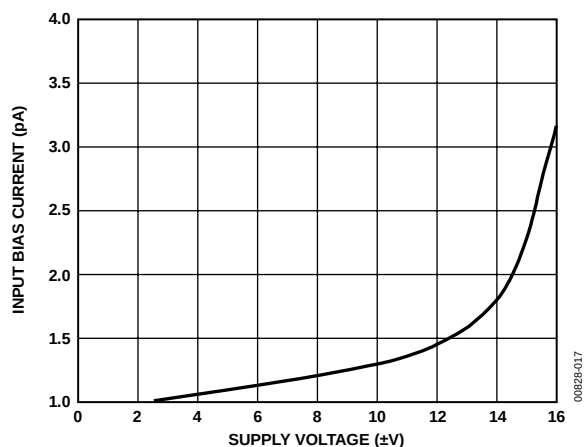


Figure 16. Input Bias Current vs. Supply Voltage

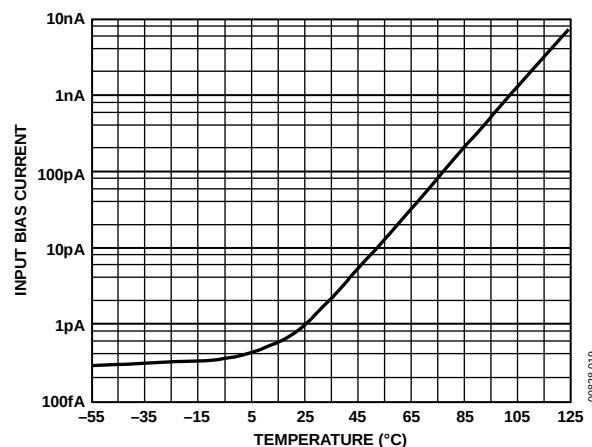


Figure 18. Input Bias Current vs. Temperature

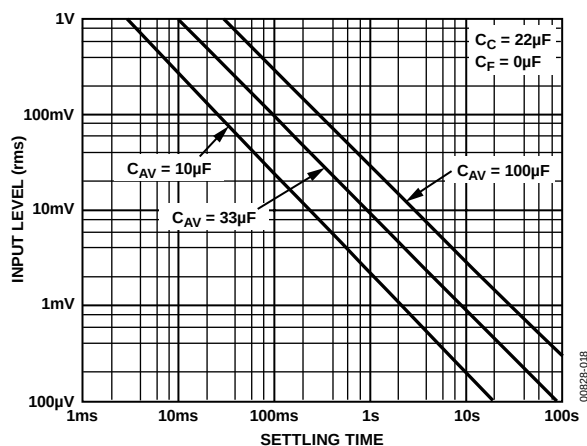
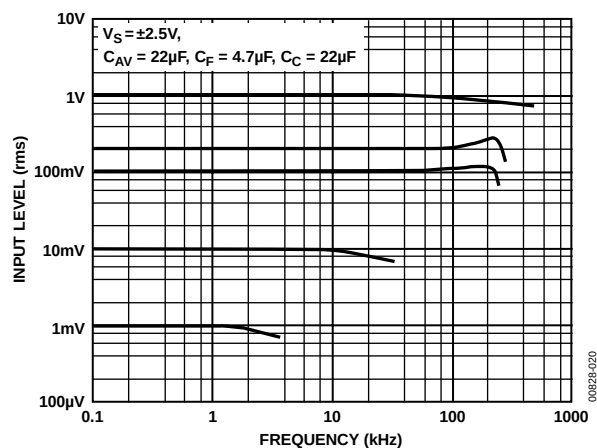
Figure 17. RMS Input Level vs. Settling Time for Three Values of  $C_{AV}$ 

Figure 19. Frequency Response Driving Pin 1; Negative DC Output

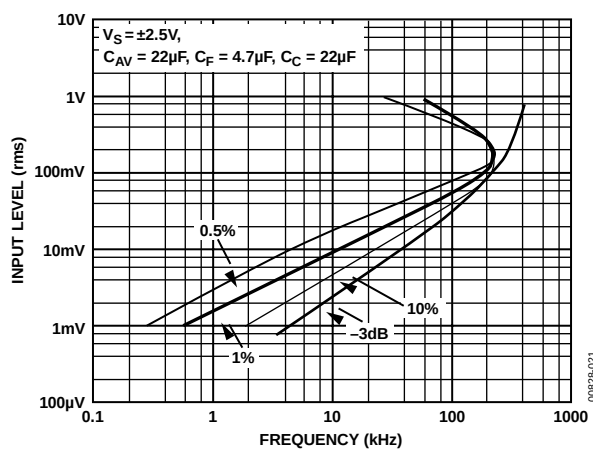


Figure 20. Error Contours Driving Pin 1

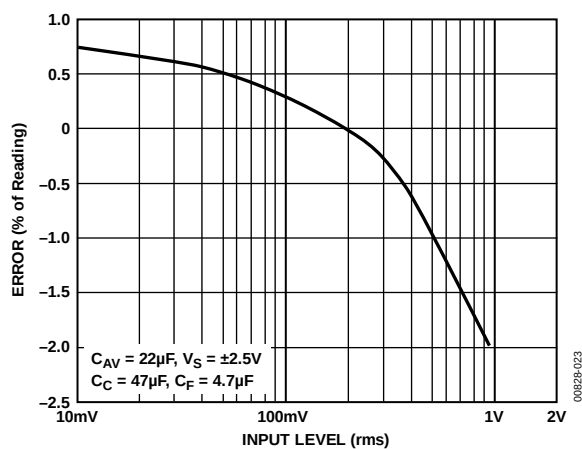
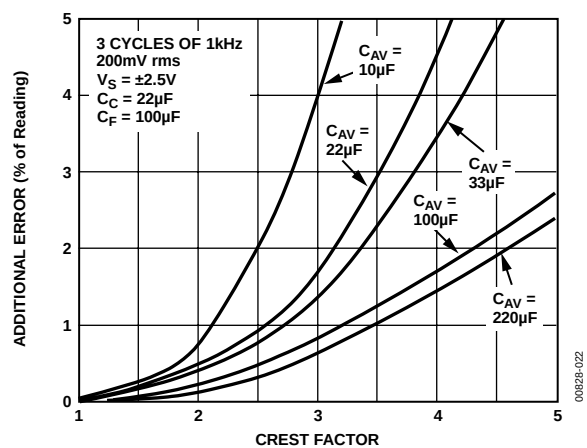


Figure 22. Error vs. RMS Input Level Driving Pin 1

Figure 21. Additional Error vs. Crest Factor for Various Values of  $C_{AV}$

As shown in Figure 23, the [AD737](#) has four functional subsections: an input amplifier, a full-wave rectifier, an rms core, and a bias section. The FET input amplifier allows a high impedance, buffered input at Pin 2 or a low impedance, wide dynamic range input at Pin 1. The high impedance input, with its low input bias current, is ideal for use with high impedance input attenuators. The input signal can be either dc-coupled or ac-coupled to the input amplifier. Unlike other rms converters, the [AD737](#) permits both direct and indirect ac coupling of the inputs. AC coupling is provided by placing a series capacitor between the input signal and Pin 2 (or Pin 1) for direct coupling and between Pin 1 and ground (while driving Pin 2) for indirect coupling.



The output of the input amplifier drives a full-wave precision rectifier, which, in turn, drives the rms core. It is the core that provides the essential rms operations of squaring, averaging, and square rooting, using an external averaging capacitor,  $C_{AV}$ . Without  $C_{AV}$ , the rectified input signal passes through the core unprocessed, as is done with the average responding connection (see Figure 25). In the average responding mode, averaging is carried out by an RC post filter consisting of an 8 k $\Omega$  internal scale factor resistor connected between Pin 6 and Pin 8 and an

$$V_{OUT} = \sqrt{Avg(V_{IN}^2)}$$

## DC ERROR, OUTPUT RIPPLE, AND AVERAGING ERROR

Figure 24 shows the typical output waveform of the AD737 with a sine wave input voltage applied. As with all real-world devices, the ideal output of  $V_{OUT} = V_{IN}$  is never exactly achieved; instead, the output contains both a dc and an ac error component.

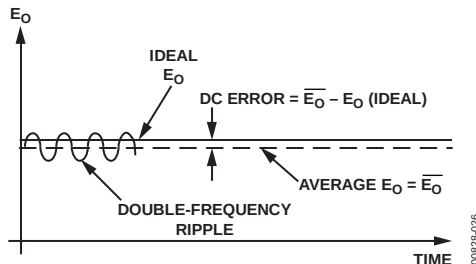


Figure 24. Output Waveform for Sine Wave Input Voltage

As shown, the dc error is the difference between the average of the output signal (when all the ripple in the output has been removed by external filtering) and the ideal dc output. The dc error component is, therefore, set solely by the value of the averaging capacitor used—no amount of post filtering (using a very large postfiltering capacitor,  $C_F$ ) allows the output voltage to equal its ideal value. The ac error component, an output ripple, can be easily removed using a large enough  $C_F$ .

In most cases, the combined magnitudes of the dc and ac error components must be considered when selecting appropriate values for  $C_{AV}$  and  $C_F$  capacitors. This combined error, representing the maximum uncertainty of the measurement, is termed the averaging error and is equal to the peak value of the output ripple plus the dc error. As the input frequency increases, both error components decrease rapidly. If the input frequency doubles, the dc error and ripple reduce to one-quarter and one-half of their original values, respectively, and rapidly become insignificant.

## AC MEASUREMENT ACCURACY AND CREST FACTOR

The crest factor of the input waveform is often overlooked when determining the accuracy of an ac measurement. Crest factor is defined as the ratio of the peak signal amplitude to the rms amplitude (crest factor =  $V_{PEAK}/V_{rms}$ ). Many common waveforms, such as sine and triangle waves, have relatively low crest factors ( $\geq 2$ ). Other waveforms, such as low duty cycle pulse trains and SCR waveforms, have high crest factors. These types of waveforms require a long averaging time constant to average out the long time periods between pulses. Figure 9 shows the additional error vs. the crest factor of the AD737 for various values of  $C_{AV}$ .

## CALCULATING SETTLING TIME

Figure 17 can be used to closely approximate the time required for the AD737 to settle when its input level is reduced in amplitude. The net time required for the rms converter to settle is the difference between two times extracted from the graph: the initial time minus the final settling time. As an example, consider the following conditions: a 33  $\mu F$  averaging capacitor, an initial rms input level of 100 mV, and a final (reduced) input level of 1 mV. From Figure 17, the initial settling time (where the 100 mV line intersects the 33  $\mu F$  line) is approximately 80 ms. The settling time corresponding to the new or final input level of 1 mV is approximately 8 seconds. Therefore, the net time for the circuit to settle to its new value is 8 seconds minus 80 ms, which is 7.92 seconds.

Note that, because of the inherent smoothness of the decay characteristic of a capacitor/diode combination, this is the total settling time to the final value (not the settling time to 1%, 0.1%, and so on, of the final value). Also, this graph provides the worst-case settling time because the AD737 settles very quickly with increasing input levels.

Table 5. Error Introduced by an Average Responding Circuit When Measuring Common Waveforms

| Type of Waveform<br>1 V Peak Amplitude | Crest Factor<br>( $V_{PEAK}/V_{rms}$ ) | True RMS<br>Value (V) | Reading of an Average Responding Circuit<br>Calibrated to an RMS Sine Wave Value (V) | Error (%) |
|----------------------------------------|----------------------------------------|-----------------------|--------------------------------------------------------------------------------------|-----------|
| Undistorted Sine Wave                  | 1.414                                  | 0.707                 | 0.707                                                                                | 0         |
| Symmetrical Square Wave                | 1.00                                   | 1.00                  | 1.11                                                                                 | 11.0      |
| Undistorted Triangle Wave              | 1.73                                   | 0.577                 | 0.555                                                                                | -3.8      |
| Gaussian Noise (98% of Peaks < 1 V)    | 3                                      | 0.333                 | 0.295                                                                                | -11.4     |
| Rectangular                            | 2                                      | 0.5                   | 0.278                                                                                | -44       |
| Pulse Train                            | 10                                     | 0.1                   | 0.011                                                                                | -89       |
| SCR Waveforms                          |                                        |                       |                                                                                      |           |
| 50% Duty Cycle                         | 2                                      | 0.495                 | 0.354                                                                                | -28       |
| 25% Duty Cycle                         | 4.7                                    | 0.212                 | 0.150                                                                                | -30       |

## APPLICATIONS INFORMATION

### RMS MEASUREMENT—CHOOSING AN OPTIMUM VALUE FOR $C_{AV}$

Because the external averaging capacitor,  $C_{AV}$ , holds the rectified input signal during rms computation, its value directly affects the accuracy of the rms measurement, especially at low frequencies. Furthermore, because the averaging capacitor is connected across a diode in the rms core, the averaging time constant ( $\tau_{AV}$ ) increases exponentially as the input signal decreases. It follows that decreasing the input signal decreases errors due to nonideal averaging but increases the settling time approaching the decreased rms-computed dc value. Thus, diminishing input values allow the circuit to perform better (due to increased averaging) while increasing the waiting time between measurements. A trade-off must be made between computational accuracy and settling time when selecting  $C_{AV}$ .

## RAPID SETTLING TIMES VIA THE AVERAGE RESPONDING CONNECTION

Because the average responding connection shown in Figure 25 does not use an averaging capacitor, its settling time does not vary with input signal level; it is determined solely by the RC time constant of  $C_F$  and the internal 8 k $\Omega$  output scaling resistor.

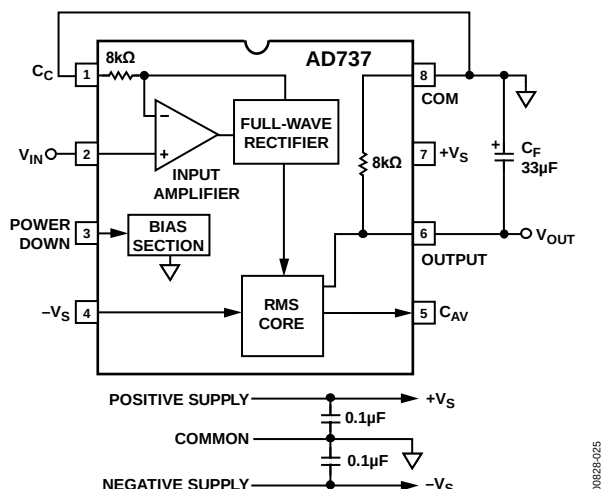


Figure 25. **AD737** Average Responding Circuit

### Selectable Average or RMS Conversion

For some applications, it is desirable to be able to select between rms-value-to-dc conversion and average-value-to-dc conversion. If  $C_{AV}$  is disconnected from the root-mean core, the [AD737](#) full-wave rectifier is a highly accurate absolute value circuit. A CMOS switch whose gate is controlled by a logic level selects between average and rms values.

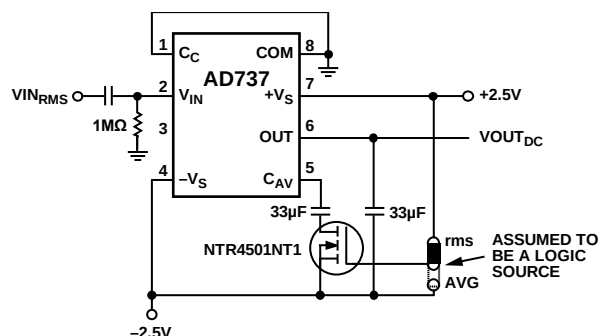


Figure 26. CMOS Switch Is Used to Select RMS or Average Responding Modes

## SELECTING PRACTICAL VALUES FOR CAPACITORS

Table 6 provides practical values of  $C_{AV}$  and  $C_F$  for several common applications.

The input coupling capacitor,  $C_c$ , in conjunction with the 8 k $\Omega$  internal input scaling resistor, determines the  $-3$  dB low frequency roll-off. This frequency,  $F_L$ , is equal to

$$F_L = \frac{1}{2\pi \times 8000 \times C_c \text{ (in Farads)}} \quad (1)$$

Note that, at  $F_L$ , the amplitude error is approximately  $-30\%$  ( $-3$  dB) of reading. To reduce this error to  $0.5\%$  of reading, choose a value of  $C_C$  that sets  $F_L$  at one-tenth of the lowest frequency to be measured.

In addition, if the input voltage has more than 100 mV of dc offset, the ac coupling network at Pin 2 is required in addition to Capacitor Cc.

## SCALING INPUT AND OUTPUT VOLTAGES

The AD737 is an extremely flexible device. With minimal external circuitry, it can be powered with single- or dual-polarity power supplies, and input and output voltages are independently scalable to accommodate nonmatching I/O devices. This section describes a few such applications.

### Extending or Scaling the Input Range

For low supply voltage applications, the maximum peak voltage to the device is extended by simply applying the input voltage to Pin 1 across the internal 8 k $\Omega$  input resistor. The **AD737** input circuit functions quasi-differentially, with a high impedance FET input at Pin 2 (noninverting) and a low impedance input at Pin 1 (inverting, see Figure 25). The internal 8 k $\Omega$  resistor behaves as a voltage-to-current converter connected to the summing node of a feedback loop around the input amplifier. Because the feedback loop acts to servo the summing node voltage to match the voltage at Pin 2, the maximum peak input voltage increases until the internal circuit runs out of headroom, approximately double for a symmetrical dual supply.

### Battery Operation

All the level-shifting for battery operation is provided by the 3½ digit converter, shown in Figure 27. Alternatively, an external op amp adds flexibility by accommodating nonzero common-mode voltages and providing output scaling and offset to zero. When an external operational amplifier is used, the output polarity is positive going.

Figure 28 shows an op amp used in a single-supply application. Note that the combined input resistor value ( $R1 + R2 + 8\text{ k}\Omega$ ) matches that of the  $R5$  feedback resistor. In this instance, the magnitudes of the output dc voltage and the rms of the ac input are equal.  $R3$  and  $R4$  provide current to offset the output to 0 V.

### Scaling the Output Voltage

The output voltage can be scaled to the input rms voltage. For example, assume that the AD737 is retrofitted to an existing application using an averaging responding circuit (full-wave rectifier). The power supply is 12 V, the input voltage is 10 V ac, and the desired output is 6 V dc.

For convenience, use the same combined input resistance as shown in Figure 28. Calculate the rms input current as

$$I_{INMAG} = \frac{10\text{ V}}{69.8\text{ k}\Omega + 2.5\text{ k}\Omega + 8\text{ k}\Omega} = 125\text{ }\mu\text{A} = I_{OUTMAG} \quad (2)$$

Next, using the  $I_{OUTMAG}$  value from Equation 2, calculate the new feedback resistor value ( $R5$ ) required for 6 V output using

$$R5 = \frac{6\text{ V}}{125\text{ }\mu\text{A}} = 48.1\text{ k}\Omega \quad (3)$$

Select the closest-value standard 1% resistor, 47.5 kΩ.

Because the supply is 12 V, the common-mode voltage at the  $R7/R8$  divider is 6 V, and the combined resistor value ( $R3 + R4$ ) is equal to the feedback resistor, or 47.5 kΩ.

$R2$  is used to calibrate the transfer function (gain), and  $R4$  sets the output voltage to zero with no input voltage.

Perform calibration as follows:

1. With no ac input applied, adjust  $R4$  for 0 V.
2. Apply a known input to the input.
3. Adjust the  $R2$  trimmer until the input and output match.

The op amp selected for any single-supply application must be a rail-to-rail type, for example an AD8541, as shown in Figure 28. For higher voltages, a higher voltage part, such as an OP196, can be used. When calibrating to 0 V, the specified voltage above ground for the operational amplifier must be taken into account. Adjust  $R4$  slightly higher as appropriate.

**Table 6. AD737 Capacitor Selection**

| Application                        | RMS Input Level | Low Frequency Cutoff (–3 dB) | Maximum Crest Factor | $C_{AV}$ (μF) | $C_F$ (μF) | Settling Time <sup>1</sup> to 1% |
|------------------------------------|-----------------|------------------------------|----------------------|---------------|------------|----------------------------------|
| General-Purpose RMS Computation    | 0 V to 1 V      | 20 Hz                        | 5                    | 150           | 10         | 360 ms                           |
|                                    |                 | 200 Hz                       | 5                    | 15            | 1          | 36 ms                            |
|                                    | 0 mV to 200 mV  | 20 Hz                        | 5                    | 33            | 10         | 360 ms                           |
|                                    |                 | 200 Hz                       | 5                    | 3.3           | 1          | 36 ms                            |
| General-Purpose Average Responding | 0 V to 1 V      | 20 Hz                        |                      | None          | 33         | 1.2 sec                          |
|                                    |                 | 200 Hz                       |                      | None          | 3.3        | 120 ms                           |
|                                    | 0 mV to 200 mV  | 20 Hz                        |                      | None          | 33         | 1.2 sec                          |
|                                    |                 | 200 Hz                       |                      | None          | 3.3        | 120 ms                           |
| SCR Waveform Measurement           | 0 mV to 200 mV  | 50 Hz                        | 5                    | 100           | 33         | 1.2 sec                          |
|                                    |                 | 60 Hz                        | 5                    | 82            | 27         | 1.0 sec                          |
|                                    | 0 mV to 100 mV  | 50 Hz                        | 5                    | 50            | 33         | 1.2 sec                          |
|                                    |                 | 60 Hz                        | 5                    | 47            | 27         | 1.0 sec                          |
| Audio Applications                 | 0 mV to 200 mV  | 300 Hz                       | 3                    | 1.5           | 0.5        | 18 ms                            |
|                                    | 0 mV to 100 mV  | 20 Hz                        | 10                   | 100           | 68         | 2.4 sec                          |

<sup>1</sup> Settling time is specified over the stated rms input level with the input signal increasing from zero. Settling times are greater for decreasing amplitude input signals.

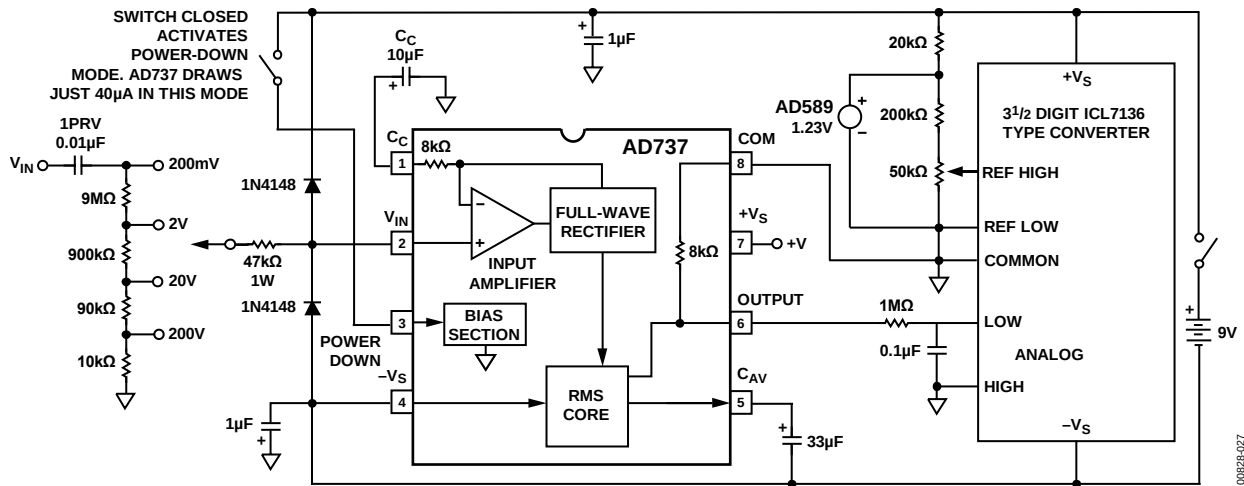


Figure 27. 3 1/2 Digit DVM Circuit

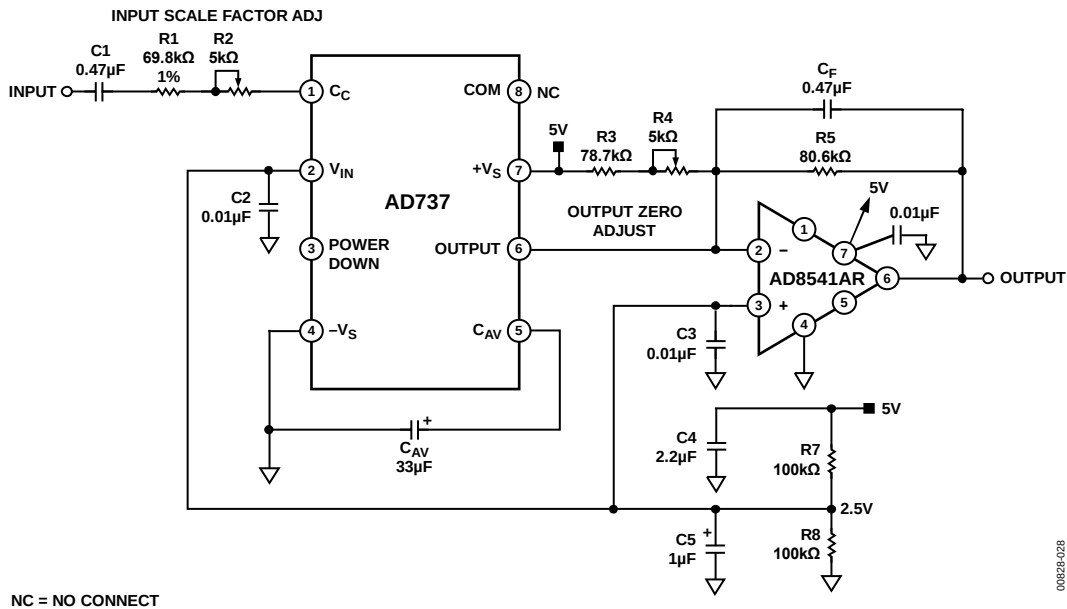


Figure 28. Battery-Powered Operation for 200 mV Maximum RMS Full-Scale Input

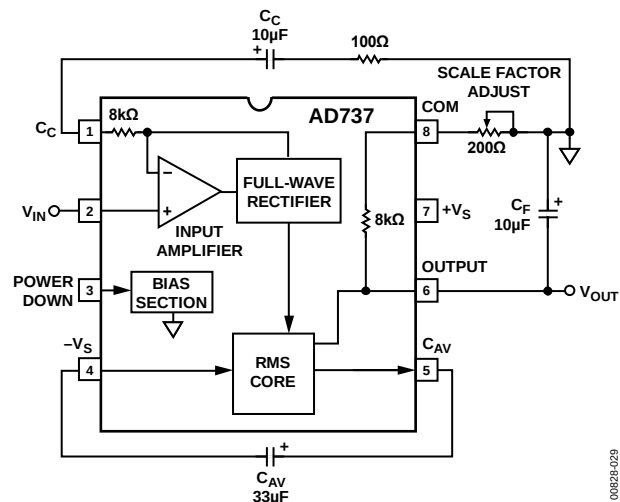


Figure 29. External Scale Factor Trim





Figure 30. dB Output Connection



Figure 31. DC-Coupled Offset Voltage and Scale Factor Trims

## AD737 EVALUATION BOARD

An evaluation board, **AD737-EVALZ**, is available for experiments or for becoming familiar with rms-to-dc converters. Figure 32 is a photograph of the board; Figure 34 to Figure 37 show the signal and power plane copper patterns. The board is designed for multipurpose applications and can be used for the AD736 as well. Although not shipped with the board, an optional socket that accepts the 8-lead surface-mount package is available from Enplas Corp.



Figure 32. AD737 Evaluation Board

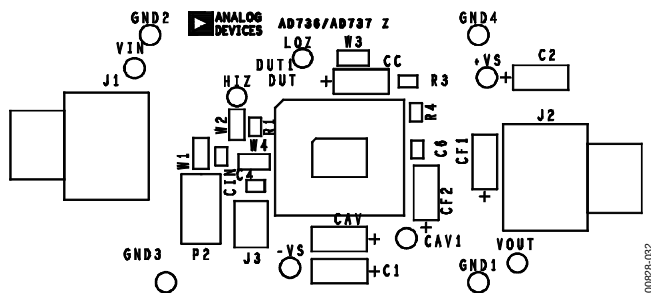


Figure 33. AD737 Evaluation Board—Component-Side Silkscreen

As described in the Applications Information section, the **AD737** can be connected in a variety of ways. As shipped, the board is configured for dual supplies with the high impedance input connected and the power-down feature disabled. Jumpers are provided for connecting the input to the low impedance input (Pin 1) and for dc connections to either input. The schematic with movable jumpers is shown in Figure 38. The jumper positions in black are default connections; the dotted-outline jumpers are optional connections. The board is tested prior to shipment and requires only a power supply connection and a precision meter to perform measurements.

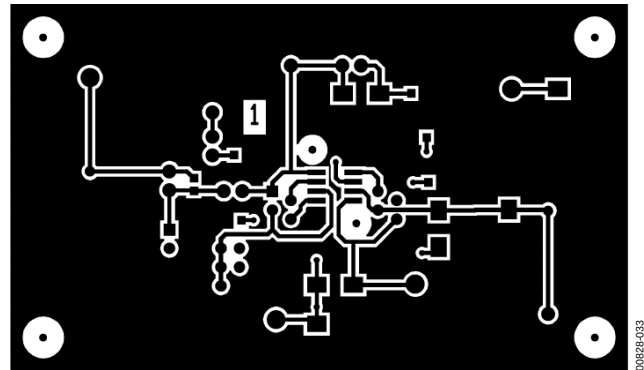


Figure 34. AD737 Evaluation Board—Component-Side Copper

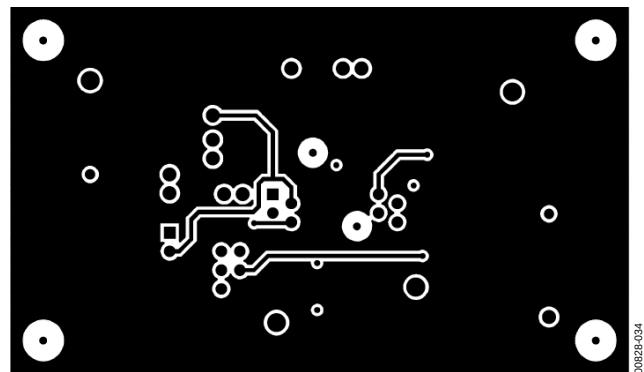


Figure 35. AD737 Evaluation Board—Secondary-Side Copper

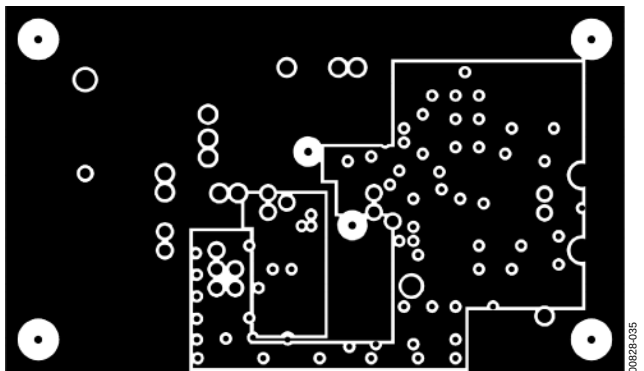


Figure 36. AD737 Evaluation Board—Internal Power Plane

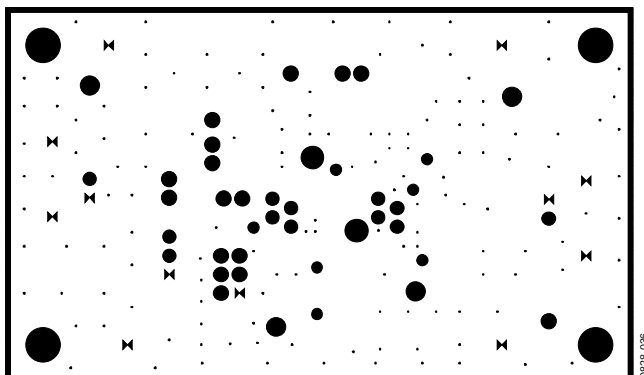


Figure 37. AD737 Evaluation Board—Internal Ground Plane

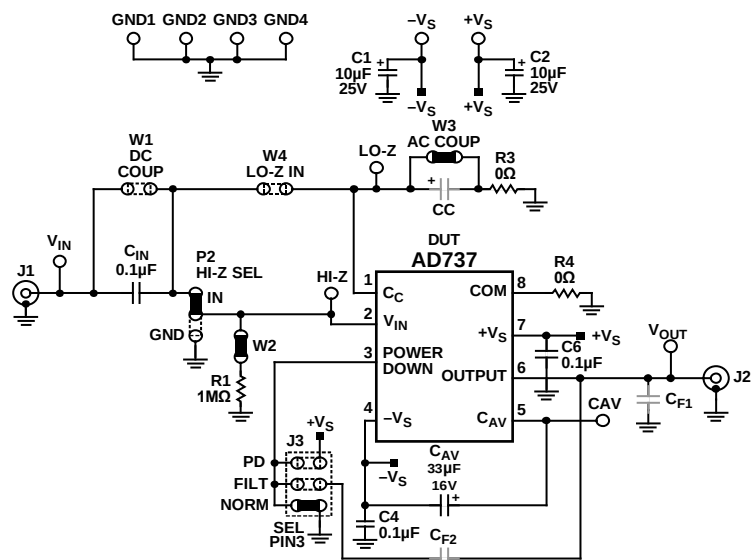
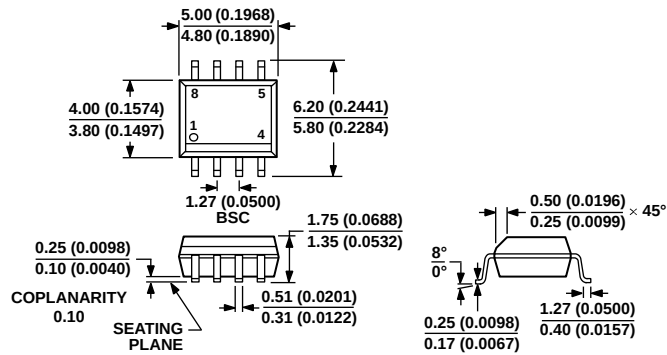


Figure 38. AD737 Evaluation Board Schematic

008228-037

## OUTLINE DIMENSIONS

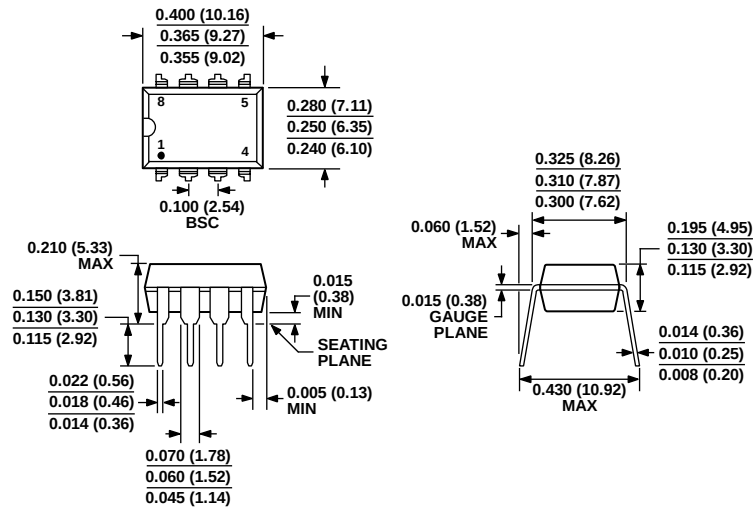


COMPLIANT TO JEDEC STANDARDS MS-012-AA  
CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS  
(IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR  
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

012407-A

Figure 39. 8-Lead Standard Small Outline Package [SOIC\_N]  
Narrow Body  
(R-8)

Dimensions shown in millimeters and (inches)



COMPLIANT TO JEDEC STANDARDS MS-001  
CONTROLLING DIMENSIONS ARE IN INCHES; MILLIMETER DIMENSIONS  
(IN PARENTHESES) ARE ROUNDED-OFF INCH EQUIVALENTS FOR  
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.  
CORNER LEADS MAY BE CONFIGURED AS WHOLE OR HALF LEADS.

070605-A

Figure 40. 8-Lead Plastic Dual-In-Line Package [PDIP]  
(N-8)

Dimensions shown in inches and (millimeters)

## ORDERING GUIDE

| Model <sup>1</sup> | Temperature Range | Package Description                            | Package Option |
|--------------------|-------------------|------------------------------------------------|----------------|
| AD737ANZ           | –40°C to +85°C    | 8-Lead Plastic Dual In-Line Package [PDIP]     | N-8            |
| AD737ARZ           | –40°C to +85°C    | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JNZ           | 0°C to 70°C       | 8-Lead Plastic Dual In-Line Package [PDIP]     | N-8            |
| AD737JRZ           | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JRZ-R7        | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JRZ-RL        | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JRZ-5         | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JRZ-5-R7      | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737JRZ-5-RL      | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737KR-REEL       | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737KR-REEL7      | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737KRZ-RL        | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737KRZ-R7        | 0°C to 70°C       | 8-Lead Standard Small Outline Package [SOIC_N] | R-8            |
| AD737-EVALZ        |                   | Evaluation Board                               |                |

<sup>1</sup> Z = RoHS Compliant Part.

**NOTES**

## NOTES

**NOTES**





Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



#### Как с нами связаться

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