



BGU7004

SiGe:C Low Noise Amplifier MMIC for GPS, GLONASS, Galileo and Compass

Rev. 3 — 18 January 2017

Product data sheet

1. Product profile

1.1 General description

The BGU7004 is, also known as the GPS1103M, an AEC-Q100 qualified Low Noise Amplifier (LNA) for GNSS receiver applications in a plastic leadless 6-pin, extremely small SOT886 package. The BGU7004 requires only one external matching inductor and one external decoupling capacitor.

The BGU7004 adapts itself to the changing environment resulting from co-habitation of different radio systems in modern cellular handsets. It has been designed for low power consumption and optimal performance when jamming signals from co-existing cellular transmitters are present. At low jamming power levels it delivers 16.5 dB gain at a noise figure of 0.85 dB. During high jamming power levels, resulting for example from a cellular transmit burst, it temporarily increases its bias current to improve sensitivity.

1.2 Features and benefits

- AEC-Q100 qualified (see [Section 9.1](#))
- Covers full GNSS L1 band, from 1559 MHz to 1610 MHz
- Noise figure (NF) = 0.85 dB and gain (G_p) = 16.5 dB
- High input 1 dB compression point P_i (1dB) of -11 dBm
- High out of band $IP3_i$ of 9 dBm
- Supply voltage 1.5 V to 2.85 V
- Power-down mode current consumption < 1 μ A
- Optimized performance at low supply current of 4.5 mA
- Integrated matching for the output
- Requires only one input matching inductor and one supply decoupling capacitor
- Input and output DC decoupled
- ESD protection on all pins (HBM > 2 kV)
- Integrated temperature stabilized bias for easy design
- Small 6-pin leadless package 1 mm \times 1.45 mm \times 0.5 mm
- 110 GHz transit frequency - SiGe:C technology

1.3 Applications

- LNA for GPS, GLONASS, Galileo and Compass (BeiDou) in automotive applications like Toll Collection and Emergency Call.



- LNA for GPS, GLONASS, Galileo and Compass (BeiDou) in smart phones, feature phones, tablet PCs, Personal Navigation Devices, Digital Still Cameras, Digital Video Cameras, RF Front End modules, complete GPS chipset modules and theft protection (laptop, ATM).

1.4 Quick reference data

Table 1. Quick reference data

$f = 1559 \text{ MHz to } 1610 \text{ MHz}$; $V_{CC} = 1.8 \text{ V}$; $P_i < -40 \text{ dBm}$; $T_{amb} = 25^\circ\text{C}$; input matched to 50Ω using a 5.6 nH inductor; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{CC}	supply voltage	RF input AC coupled	1.5	-	2.85	V
I_{CC}	supply current	$V_{ENABLE} \geq 0.8 \text{ V}$				
		$P_i < -40 \text{ dBm}$	3.2	4.5	5.7	mA
		$P_i = -20 \text{ dBm}$	8.1	11.6	14.4	mA
G_p	power gain	$P_i < -40 \text{ dBm}$, no jammer	14	16.5	19	dB
		$P_i = -20 \text{ dBm}$, no jammer	15	17.5	20	dB
NF	noise figure	$P_i < -40 \text{ dBm}$, no jammer [1]	-	0.85	1.2	dB
		$P_i < -40 \text{ dBm}$, no jammer [2]	-	0.9	1.3	dB
		$P_i = -20 \text{ dBm}$, no jammer	-	1.2	1.6	dB
$P_{i(1\text{dB})}$	input power at 1 dB gain compression	$f = 1559 \text{ MHz to } 1610 \text{ MHz}$				
		$V_{CC} = 1.5 \text{ V}$	-15	-12	-	dBm
		$V_{CC} = 1.8 \text{ V}$	-14	-11	-	dBm
		$V_{CC} = 2.85 \text{ V}$	-11	-8	-	dBm
IP3 _i	input third-order intercept point	$f = 1.575 \text{ GHz}$				
		$V_{CC} = 1.5 \text{ V}$ [3]	5	8	-	dBm
		$V_{CC} = 1.8 \text{ V}$ [3]	5	9	-	dBm
		$V_{CC} = 2.85 \text{ V}$ [3]	5	12	-	dBm

[1] PCB losses are subtracted.

[2] Including PCB losses.

[3] $f_1 = 1713 \text{ MHz}$; $f_2 = 1851 \text{ MHz}$; $P_1 = P_2 = -30 \text{ dBm}$.

2. Pinning information

Table 2. Pinning

Pin	Description	Simplified outline	Graphic symbol
1	GND	<p>Transparent top view</p>	
2	GND		
3	RF_IN		
4	V_{CC}		
5	ENABLE		
6	RF_OUT		

3. Ordering information

Table 3. Ordering information

Type number	Package		Version
	Name	Description	
BGU7004	XSON6	plastic extremely thin small outline package; no leads; 6 terminals; body 1 × 1.45 × 0.5 mm	SOT886

4. Marking

Table 4. Marking codes

Type number	Marking code
BGU7004	UY

5. Limiting values

Table 5. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	supply voltage	RF input AC coupled	-0.5	3.1	V
V_{ENABLE}	voltage on pin ENABLE	$V_{CC} \geq 2.5$ V	-0.5	3.1	V
		$V_{CC} < 2.5$ V	[2] -0.5	$V_{CC} + 0.6$	V
V_{RF_IN}	voltage on pin RF_IN	DC			
		$V_{CC} \geq 3.0$ V	[3] -0.5	3.6	V
		$V_{CC} < 3.0$ V	[2][3] -0.5	$V_{CC} + 0.6$	V
V_{RF_OUT}	voltage on pin RF_OUT	DC			
		$V_{CC} \geq 1.8$ V	[3] -0.5	3.6	V
		$V_{CC} < 1.8$ V	[2][3] -0.5	$V_{CC} + 1.8$	V
P_i	input power		-	0	dBm
P_{tot}	total power dissipation	$T_{sp} \leq 130$ °C	[1]	55	mW
T_{stg}	storage temperature		-65	150	°C
T_j	junction temperature		-	150	°C

[1] T_{sp} is the temperature at the soldering point of the emitter lead.

[2] Due to internal ESD diode protection, the applied voltage should not exceed the specified maximum in order to avoid excess current.

[3] The RF input and RF output are AC coupled through internal DC blocking capacitors.

6. Thermal characteristics

Table 6. Thermal characteristics

Symbol	Parameter	Conditions	Typ	Unit
$R_{th(j-sp)}$	thermal resistance from junction to solder point		225	K/W

7. Characteristics

Table 7. Characteristics

$f = 1559 \text{ MHz to } 1610 \text{ MHz}$; $V_{CC} = 1.8 \text{ V}$; $V_{ENABLE} \geq 0.8 \text{ V}$; $P_i < -40 \text{ dBm}$; $T_{amb} = 25 \text{ }^\circ\text{C}$; input matched to $50 \text{ } \Omega$ using a 5.6 nH inductor; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{CC}	supply voltage	RF input AC coupled	1.5	-	2.85	V
I_{CC}	supply current	$V_{ENABLE} \geq 0.8 \text{ V}$				
		$P_i < -40 \text{ dBm}$	3.2	4.5	5.7	mA
		$P_i = -20 \text{ dBm}$	8.1	11.6	14.4	mA
		$V_{ENABLE} \leq 0.3 \text{ V}$	-	-	1	μA
T_{amb}	ambient temperature		-40	+25	+125	$^\circ\text{C}$
G_p	power gain	$T_{amb} = 25 \text{ }^\circ\text{C}$				
		$P_i < -40 \text{ dBm}$, no jammer	14	16.5	19	dB
		$P_i = -20 \text{ dBm}$, no jammer	15	17.5	20	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 850 \text{ MHz}$	15	17.5	20	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 1850 \text{ MHz}$	15	17.5	20	dB
		$-40 \text{ }^\circ\text{C} \leq T_{amb} \leq +125 \text{ }^\circ\text{C}$				
		$P_i < -40 \text{ dBm}$, no jammer	13	-	20	dB
		$P_i = -20 \text{ dBm}$, no jammer	14	-	21	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 850 \text{ MHz}$	14	-	21	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 1850 \text{ MHz}$	14	-	21	dB
RL_{in}	input return loss	$P_i < -40 \text{ dBm}$	5	8	-	dB
		$P_i = -20 \text{ dBm}$	6	10	-	dB
RL_{out}	output return loss	$P_i < -40 \text{ dBm}$	10	20	-	dB
		$P_i = -20 \text{ dBm}$	10	14	-	dB
ISL	isolation		20	23	-	dB
NF	noise figure	$T_{amb} = 25 \text{ }^\circ\text{C}$				
		$P_i < -40 \text{ dBm}$, no jammer [1]	-	0.85	1.2	dB
		$P_i < -40 \text{ dBm}$, no jammer [2]	-	0.9	1.3	dB
		$P_i = -20 \text{ dBm}$, no jammer	-	1.2	1.6	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 850 \text{ MHz}$	-	1.1	1.5	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 1850 \text{ MHz}$	-	1.3	1.7	dB
		$-40 \text{ }^\circ\text{C} \leq T_{amb} \leq +125 \text{ }^\circ\text{C}$				
		$P_i < -40 \text{ dBm}$, no jammer	-	-	1.8	dB
		$P_i = -20 \text{ dBm}$, no jammer	-	-	2.0	dB
		$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 850 \text{ MHz}$	-	-	1.9	dB
$P_{jam} = -20 \text{ dBm}$; $f_{jam} = 1850 \text{ MHz}$	-	-	2.1	dB		

Table 7. Characteristics ...continued

$f = 1559 \text{ MHz to } 1610 \text{ MHz}$; $V_{CC} = 1.8 \text{ V}$; $V_{ENABLE} \geq 0.8 \text{ V}$; $P_i < -40 \text{ dBm}$; $T_{amb} = 25 \text{ }^\circ\text{C}$; input matched to $50 \text{ } \Omega$ using a 5.6 nH inductor; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$P_{i(1dB)}$	input power at 1 dB gain compression	$f = 1559 \text{ MHz to } 1610 \text{ MHz}$				
		$V_{CC} = 1.5 \text{ V}$	-15	-12	-	dBm
		$V_{CC} = 1.8 \text{ V}$	-14	-11	-	dBm
		$V_{CC} = 2.85 \text{ V}$	-11	-8	-	dBm
		$f = 806 \text{ MHz to } 928 \text{ MHz}$				
		$V_{CC} = 1.5 \text{ V}$	[3] -15	-12	-	dBm
		$V_{CC} = 1.8 \text{ V}$	[3] -14	-11	-	dBm
		$V_{CC} = 2.85 \text{ V}$	[3] -14	-11	-	dBm
		$f = 1612 \text{ MHz to } 1909 \text{ MHz}$				
		$V_{CC} = 1.5 \text{ V}$	[3] -13	-10	-	dBm
		$V_{CC} = 1.8 \text{ V}$	[3] -12	-9	-	dBm
		$V_{CC} = 2.85 \text{ V}$	[3] -10	-7	-	dBm
$IP3_i$	input third-order intercept point	$f = 1.575 \text{ GHz}$				
		$V_{CC} = 1.5 \text{ V}$	[4] 5	8	-	dBm
		$V_{CC} = 1.8 \text{ V}$	[4] 5	9	-	dBm
		$V_{CC} = 2.85 \text{ V}$	[4] 5	12	-	dBm
t_{on}	turn-on time		[5] -	-	2	μs
t_{off}	turn-off time		[5] -	-	1	μs
K	Rollett stability factor		1	-	-	

- [1] PCB losses are subtracted.
- [2] Including PCB losses.
- [3] Out of band.
- [4] $f_1 = 1713 \text{ MHz}$; $f_2 = 1851 \text{ MHz}$; $P_1 = P_2 = -30 \text{ dBm}$.
- [5] Within 10 % of the final gain.

Table 8. ENABLE (pin 5)

$-40 \text{ }^\circ\text{C} \leq T_{amb} \leq +125 \text{ }^\circ\text{C}$; $1.5 \text{ V} \leq V_{CC} \leq 2.85 \text{ V}$

$V_{ENABLE} \text{ (V)}$	State
≤ 0.3	OFF
≥ 0.8	ON

8. Application information

8.1 GNSS LNA

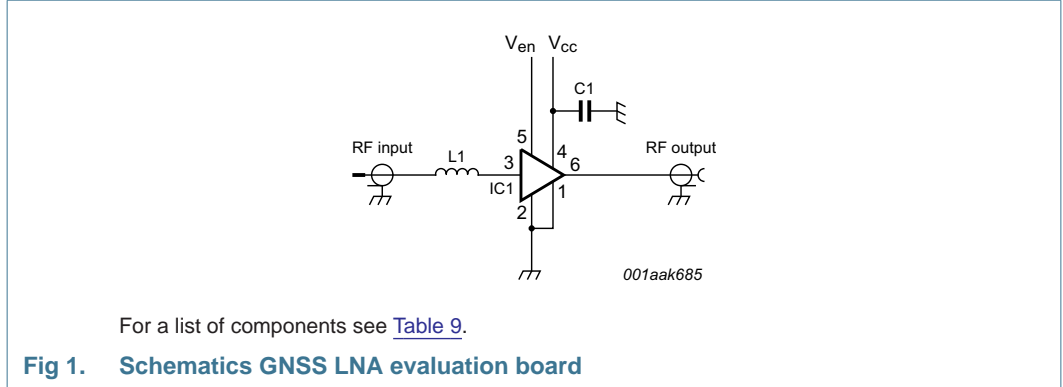
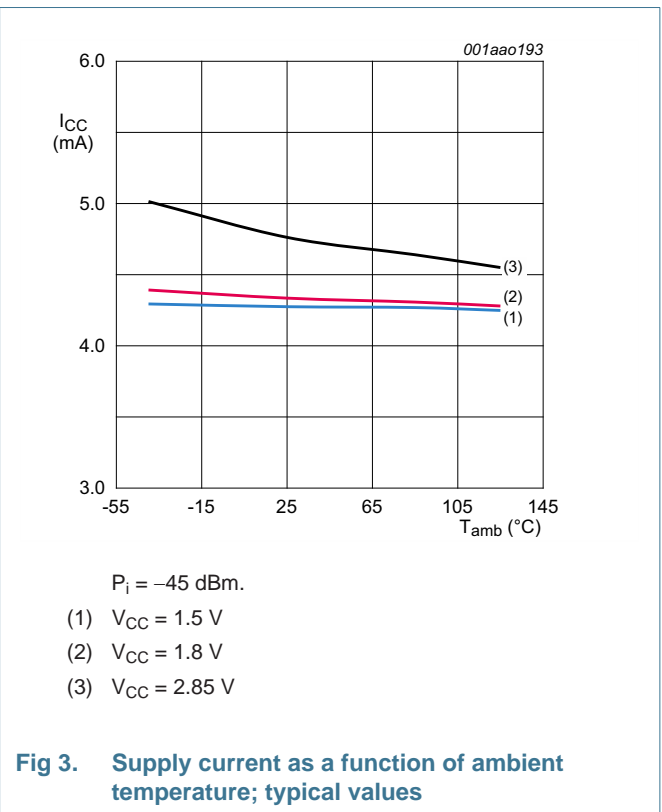
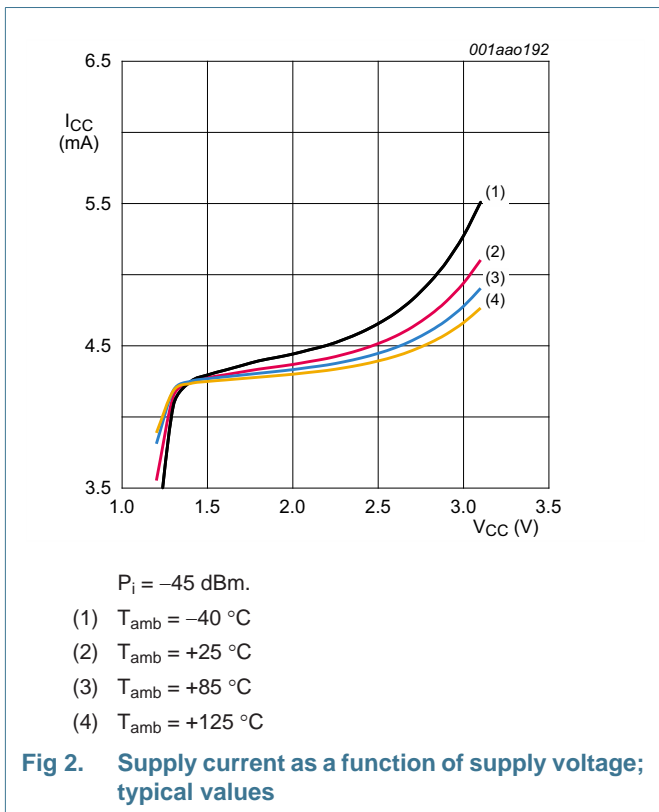
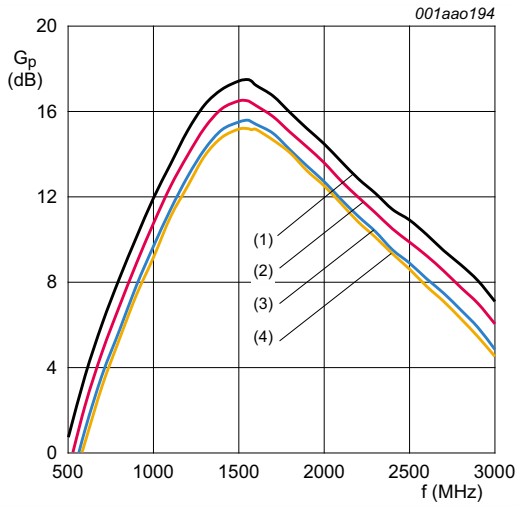


Table 9. List of components

For schematics see [Figure 1](#).

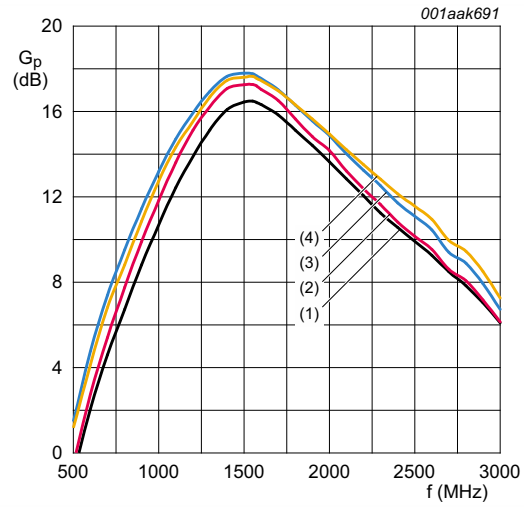
Component	Description	Value	Supplier	Remarks
C1	decoupling capacitor	1 nF	various	
IC1	BGU7004	-	NXP	
L1	high quality matching inductor	5.6 nH	Murata LQW15A	





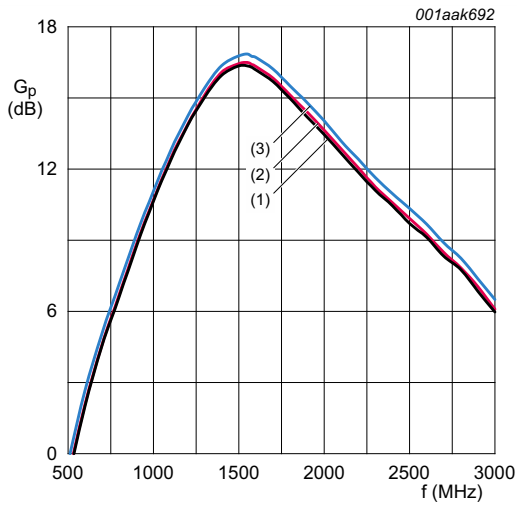
$V_{CC} = 1.8\text{ V}; P_i = -45\text{ dBm}$.
 (1) $T_{amb} = -40\text{ }^\circ\text{C}$
 (2) $T_{amb} = +25\text{ }^\circ\text{C}$
 (3) $T_{amb} = +85\text{ }^\circ\text{C}$
 (4) $T_{amb} = +125\text{ }^\circ\text{C}$

Fig 4. Power gain as a function of frequency; typical values



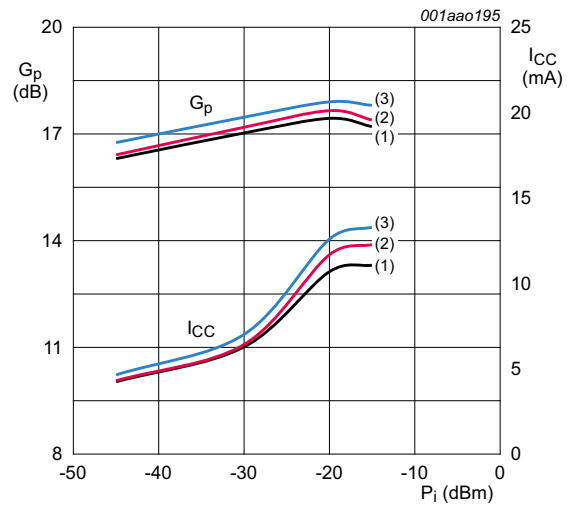
$V_{CC} = 1.8\text{ V}; T_{amb} = 25\text{ }^\circ\text{C}$.
 (1) $P_i = -45\text{ dBm}$
 (2) $P_i = -30\text{ dBm}$
 (3) $P_i = -20\text{ dBm}$
 (4) $P_i = -15\text{ dBm}$

Fig 5. Power gain as a function of frequency; typical values



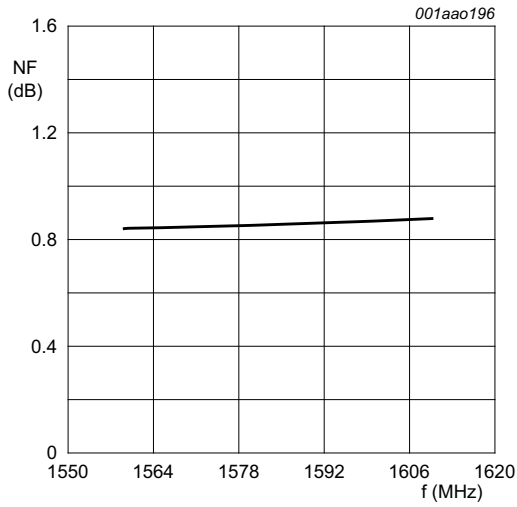
$P_i = -45\text{ dBm}; T_{amb} = 25\text{ }^\circ\text{C}$.
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 6. Power gain as a function of frequency; typical values



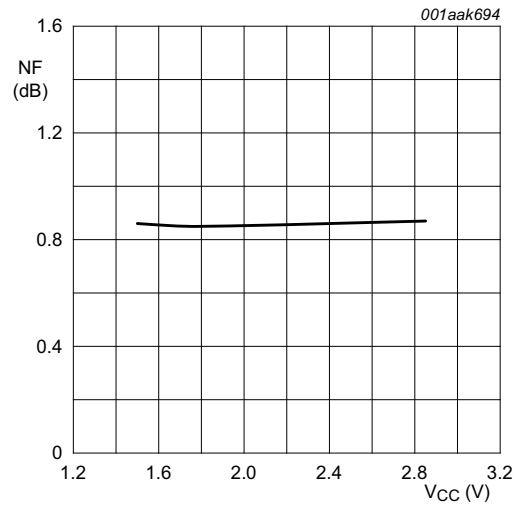
$T_{amb} = 25\text{ }^\circ\text{C}; f = 1575\text{ MHz}$.
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 7. Power gain and supply current as a function of input power; typical values



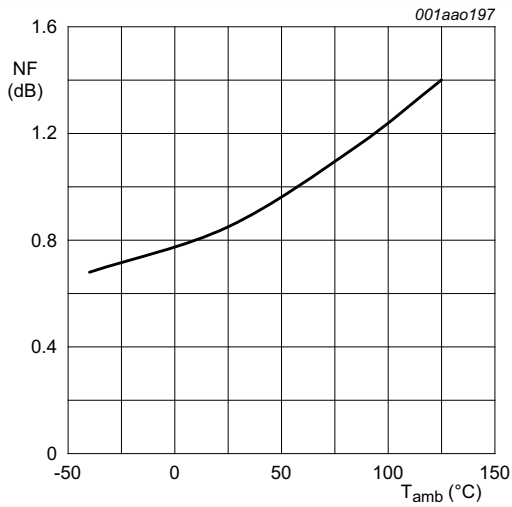
f = 1575 MHz; T_{amb} = 25 °C; no jammer.

Fig 8. Noise figure as a function of supply voltage; typical values



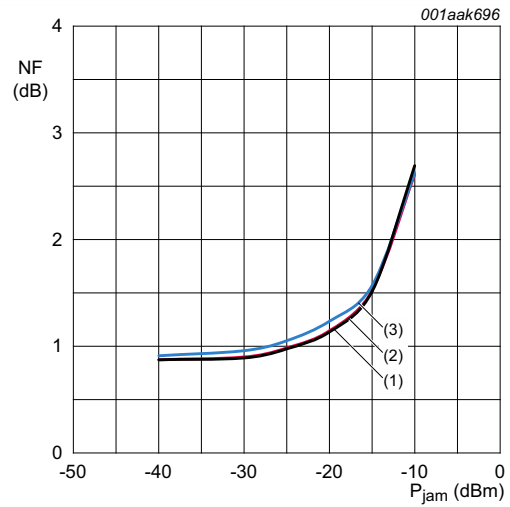
f = 1575 MHz; T_{amb} = 25 °C; no jammer.

Fig 9. Noise figure as a function of supply voltage; typical values



f = 1575 MHz; V_{CC} = 1.8 V; no jammer.

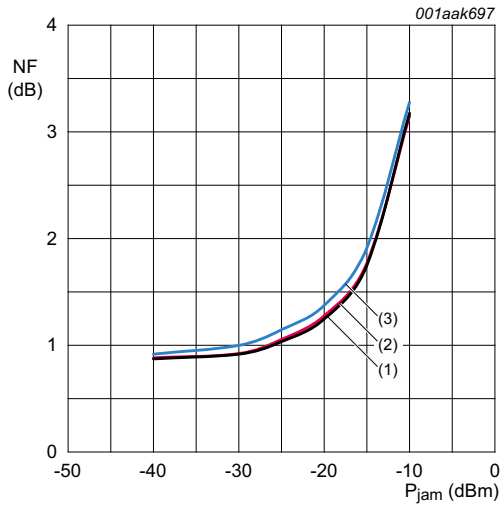
Fig 10. Noise figure as a function of ambient temperature; typical values



f_{jam} = 850 MHz; T_{amb} = 25 °C; f = 1575 MHz.

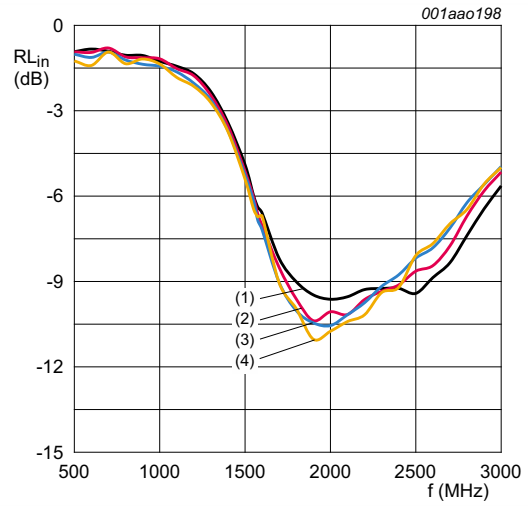
- (1) V_{CC} = 1.5 V
- (2) V_{CC} = 1.8 V
- (3) V_{CC} = 2.85 V

Fig 11. Noise figure as a function of jamming power; typical values



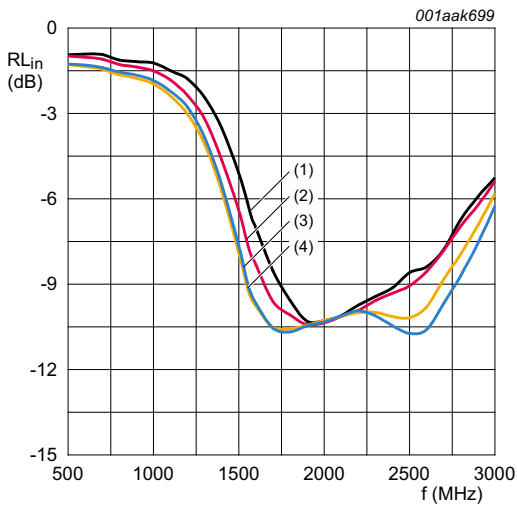
$f_{jam} = 1850 \text{ MHz}; T_{amb} = 25 \text{ }^\circ\text{C}; f = 1575 \text{ MHz}.$
 (1) $V_{CC} = 1.5 \text{ V}$
 (2) $V_{CC} = 1.8 \text{ V}$
 (3) $V_{CC} = 2.85 \text{ V}$

Fig 12. Noise figure as a function of jamming power; typical values



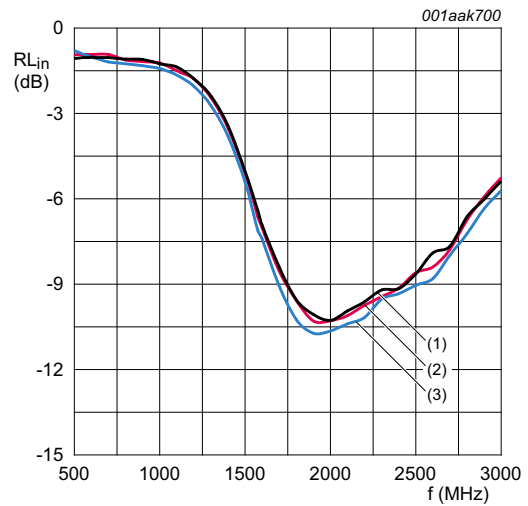
$V_{CC} = 1.8 \text{ V}; P_i = -45 \text{ dBm}.$
 (1) $T_{amb} = -40 \text{ }^\circ\text{C}$
 (2) $T_{amb} = +25 \text{ }^\circ\text{C}$
 (3) $T_{amb} = +85 \text{ }^\circ\text{C}$
 (4) $T_{amb} = +125 \text{ }^\circ\text{C}$

Fig 13. Input return loss as a function of frequency; typical values



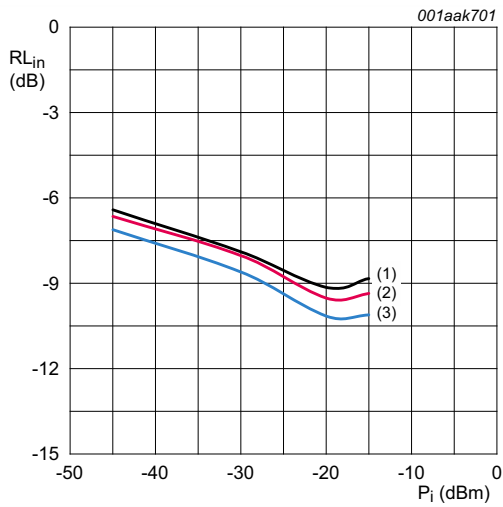
$V_{CC} = 1.8 \text{ V}; T_{amb} = 25 \text{ }^\circ\text{C}.$
 (1) $P_i = -45 \text{ dBm}$
 (2) $P_i = -30 \text{ dBm}$
 (3) $P_i = -20 \text{ dBm}$
 (4) $P_i = -15 \text{ dBm}$

Fig 14. Input return loss as a function of frequency; typical values



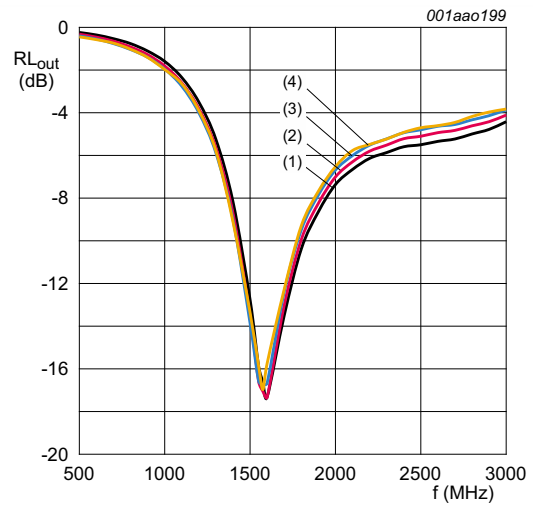
$P_i = -45 \text{ dBm}; T_{amb} = 25 \text{ }^\circ\text{C}.$
 (1) $V_{CC} = 1.5 \text{ V}$
 (2) $V_{CC} = 1.8 \text{ V}$
 (3) $V_{CC} = 2.85 \text{ V}$

Fig 15. Input return loss as a function of frequency; typical values



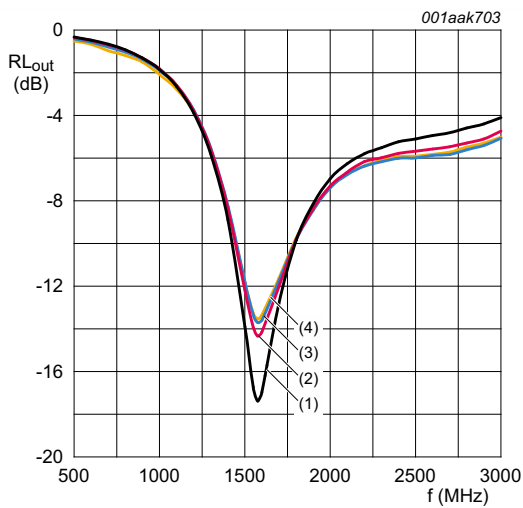
$T_{amb} = 25\text{ }^{\circ}\text{C}; f = 1575\text{ MHz.}$
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 16. Input return loss as a function of input power; typical values



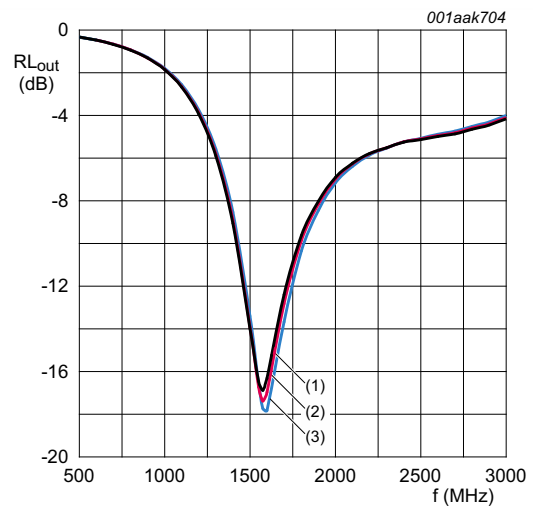
$V_{CC} = 1.8\text{ V}; P_i = -45\text{ dBm.}$
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = +25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = +85\text{ }^{\circ}\text{C}$
 (4) $T_{amb} = +125\text{ }^{\circ}\text{C}$

Fig 17. Output return loss as a function of frequency; typical values



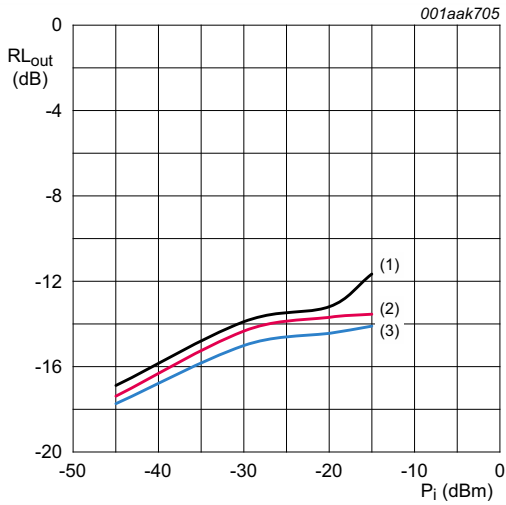
$V_{CC} = 1.8\text{ V}; T_{amb} = 25\text{ }^{\circ}\text{C.}$
 (1) $P_i = -45\text{ dBm}$
 (2) $P_i = -30\text{ dBm}$
 (3) $P_i = -20\text{ dBm}$
 (4) $P_i = -15\text{ dBm}$

Fig 18. Output return loss as a function of frequency; typical values



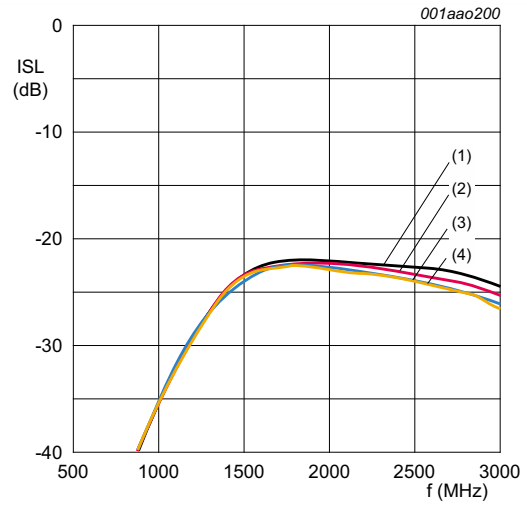
$P_i = -45\text{ dBm}; T_{amb} = 25\text{ }^{\circ}\text{C.}$
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 19. Output return loss as a function of frequency; typical values



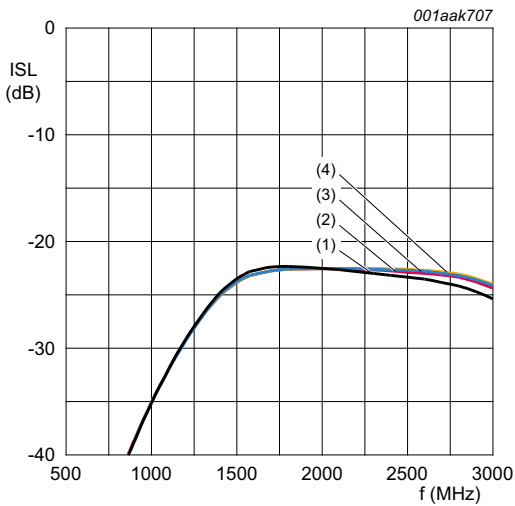
$T_{amb} = 25\text{ }^{\circ}\text{C}$; $f = 1575\text{ MHz}$.
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 20. Output return loss as a function of input power; typical values



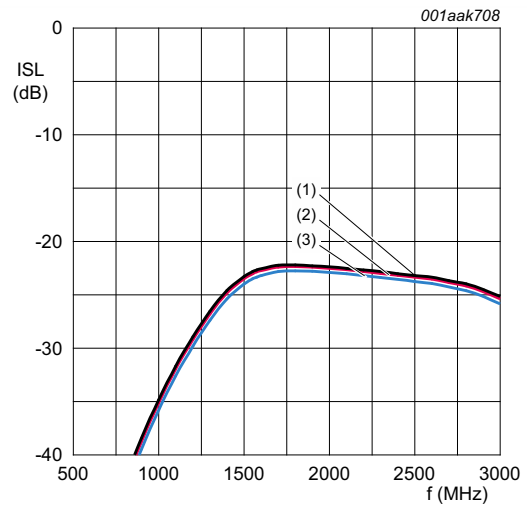
$V_{CC} = 1.8\text{ V}$; $P_i = -45\text{ dBm}$.
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = +25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = +85\text{ }^{\circ}\text{C}$
 (4) $T_{amb} = +125\text{ }^{\circ}\text{C}$

Fig 21. Isolation as a function of frequency; typical values



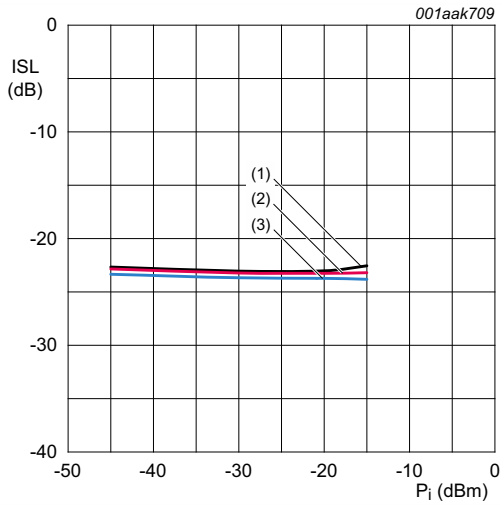
$V_{CC} = 1.8\text{ V}$; $T_{amb} = 25\text{ }^{\circ}\text{C}$.
 (1) $P_i = -45\text{ dBm}$
 (2) $P_i = -30\text{ dBm}$
 (3) $P_i = -20\text{ dBm}$
 (4) $P_i = -15\text{ dBm}$

Fig 22. Isolation as a function of frequency; typical values



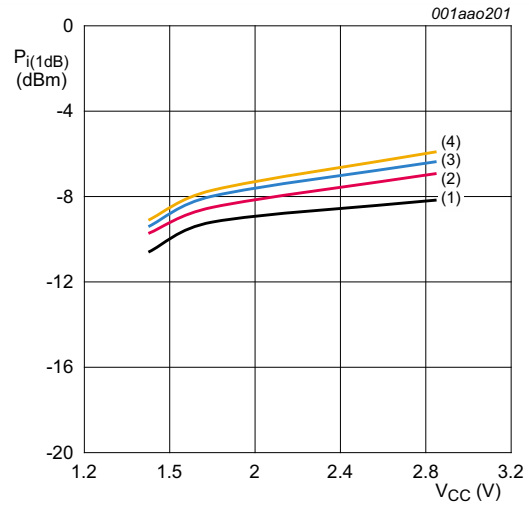
$P_i = -45\text{ dBm}$; $T_{amb} = 25\text{ }^{\circ}\text{C}$.
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 23. Isolation as a function of frequency; typical values



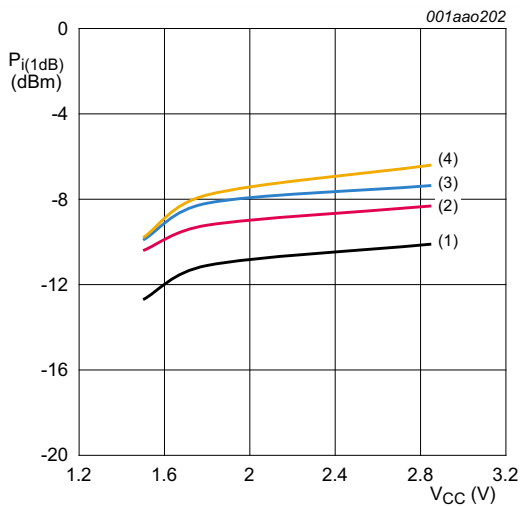
$T_{amb} = 25\text{ }^{\circ}\text{C}; f = 1575\text{ MHz.}$
 (1) $V_{CC} = 1.5\text{ V}$
 (2) $V_{CC} = 1.8\text{ V}$
 (3) $V_{CC} = 2.85\text{ V}$

Fig 24. Isolation as a function of input power; typical values



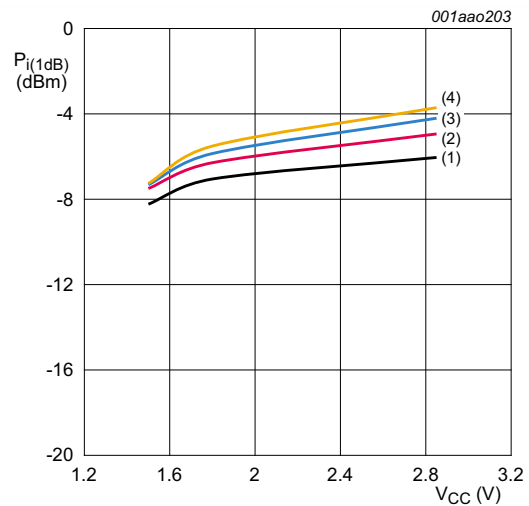
$f = 850\text{ MHz.}$
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = +25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = +85\text{ }^{\circ}\text{C}$
 (4) $T_{amb} = +125\text{ }^{\circ}\text{C}$

Fig 25. Input power at 1 dB gain compression as a function of supply voltage; typical values



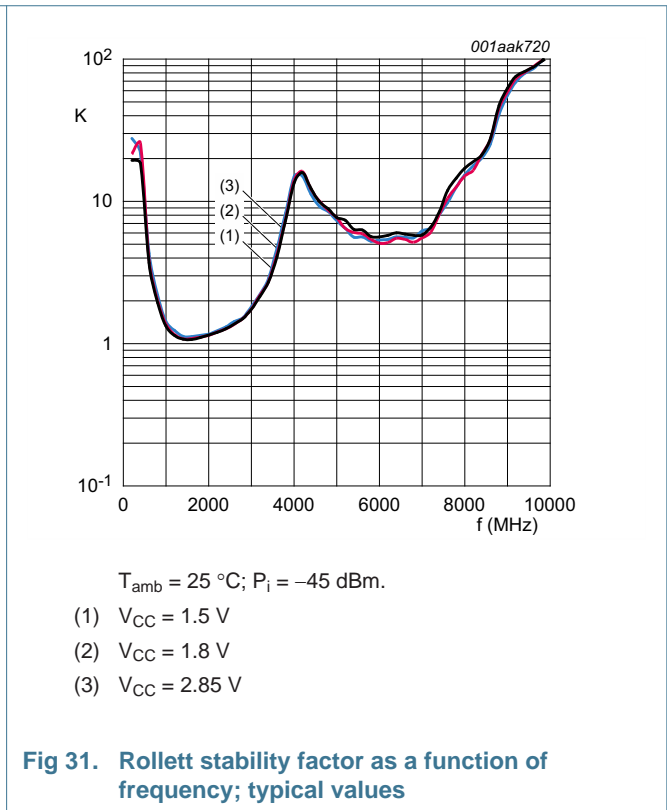
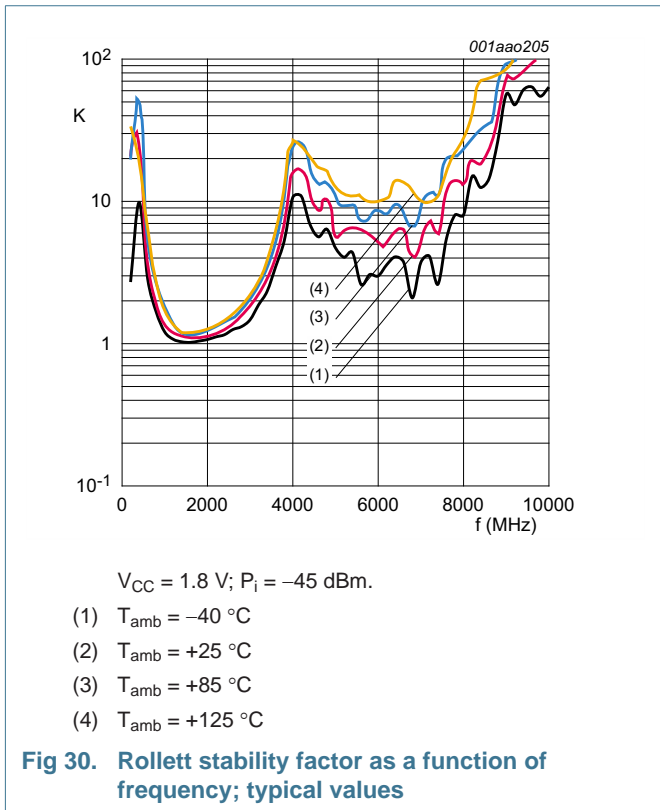
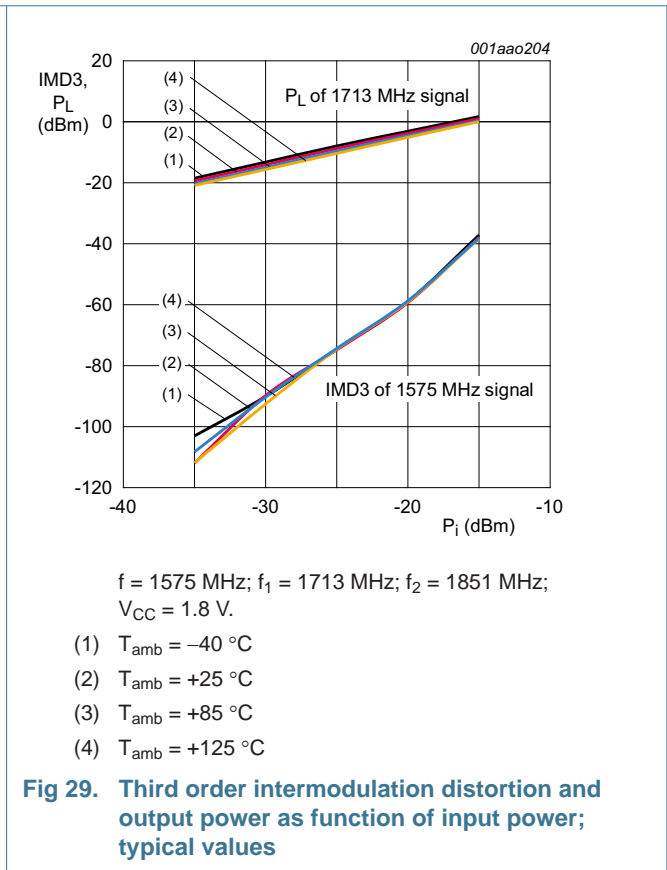
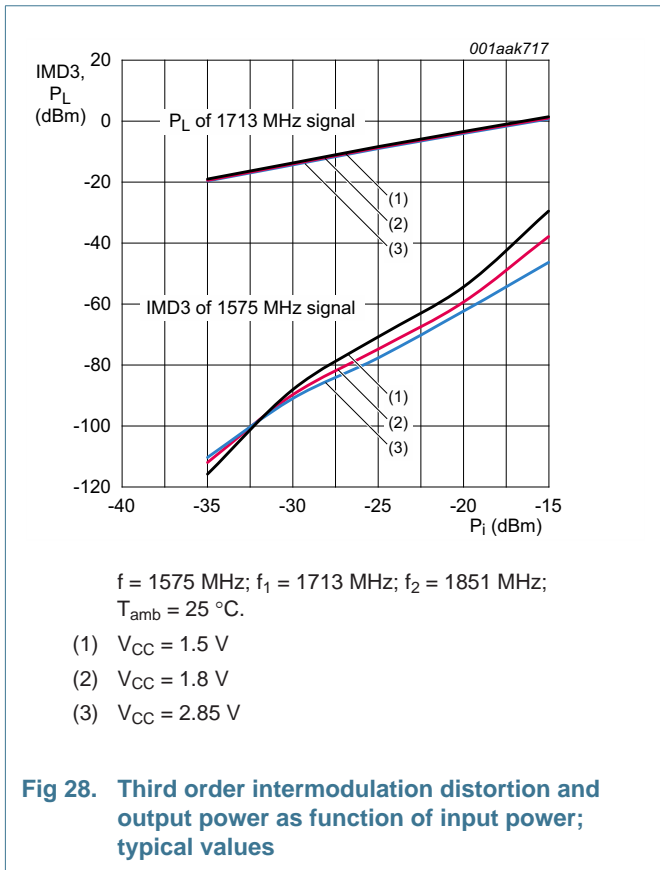
$f = 1850\text{ MHz.}$
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = +25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = +85\text{ }^{\circ}\text{C}$
 (4) $T_{amb} = +125\text{ }^{\circ}\text{C}$

Fig 26. Input power at 1 dB gain compression as a function of supply voltage; typical values



$f = 1575\text{ MHz.}$
 (1) $T_{amb} = -40\text{ }^{\circ}\text{C}$
 (2) $T_{amb} = +25\text{ }^{\circ}\text{C}$
 (3) $T_{amb} = +85\text{ }^{\circ}\text{C}$
 (4) $T_{amb} = +125\text{ }^{\circ}\text{C}$

Fig 27. Input power at 1 dB gain compression as a function of supply voltage; typical values



8.2 GPS front-end

The GPS LNA is typically used in a GPS front-end. A GPS front-end application circuit and its characteristics is provided here.

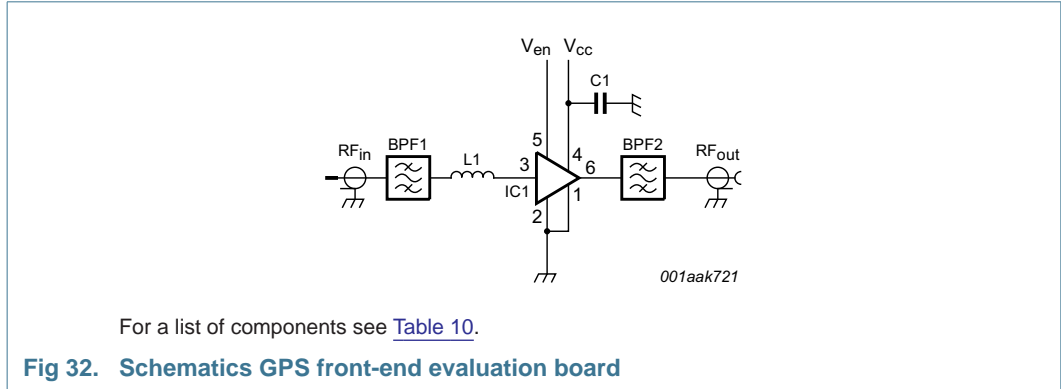


Table 10. List of components
For schematics see [Figure 32](#).

Component	Description	Value	Supplier	Remarks
BPF1, BPF2	GPS SAW filter	-	Murata SAFEA1G57KE0F00	Alternatives from Epcos: <ul style="list-style-type: none"> • B9444 Alternatives from Murata: <ul style="list-style-type: none"> • SAFEA1G57KH0F00 • SAFEA1G57KB0F00 Alternatives from Fujitsu: <ul style="list-style-type: none"> • FAR-F6KA-1G5754-L4AA • FAR-F6KA-1G5754-L4AJ
C1	decoupling capacitor	1 nF	Various	
IC1	BGU7004	-	NXP	
L1	high quality matching inductor	5.6 nH	Murata LQW15A	

8.3 Characteristics GPS front-end

Table 11. Characteristics GPS front-end

$f = 1575 \text{ MHz}$; $V_{CC} = 1.8 \text{ V}$; $V_{ENABLE} \geq 0.8 \text{ V}$; power at LNA input $P_i < -40 \text{ dBm}$; $T_{amb} = 25 \text{ }^\circ\text{C}$; input and output matched to $50 \text{ }\Omega$; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
V_{CC}	supply voltage	RF input AC coupled	1.5	-	2.85	V	
I_{CC}	supply current		-	4.5	-	mA	
G_p	power gain	power at LNA input $P_i < -40 \text{ dBm}$	[1]	-	14.5	-	dB
		power at LNA input $P_i = -20 \text{ dBm}$	[1]	-	15.5	-	dB
RL_{in}	input return loss	power at LNA input $P_i < -40 \text{ dBm}$	[1]	-	8.5	-	dB
		power at LNA input $P_i = -20 \text{ dBm}$	[1]	-	10.5	-	dB
RL_{out}	output return loss	power at LNA input $P_i < -40 \text{ dBm}$	[1]	-	14.5	-	dB
		power at LNA input $P_i = -20 \text{ dBm}$	[1]	-	12.5	-	dB
NF	noise figure	power at LNA input $P_i < -40 \text{ dBm}$	[1]	-	1.8	-	dB
		power at LNA input $P_i = -20 \text{ dBm}$	[1]	-	1.9	-	dB
$P_{i(1dB)}$	input power at 1 dB gain compression	$f = 1575 \text{ MHz}$		-8.2		dBm	
		$f = 806 \text{ MHz to } 928 \text{ MHz}$	[2]	31		dBm	
		$f = 1612 \text{ MHz to } 1909 \text{ MHz}$	[2]	40		dBm	
$IP3_i$	input third-order intercept point		[3]	64		dBm	
α	attenuation	$f = 850 \text{ MHz}$	[4]	95	-	-	dBc
		$f = 1850 \text{ MHz}$	[4]	90	-	-	dBc
t_{on}	turn-on time		[5]	-	2	μs	
t_{off}	turn-off time		[5]	-	1	μs	

[1] Power at GPS front-end input = power at LNA input + attenuation BPF1.

[2] Out of band.

[3] $f_1 = 1713 \text{ MHz}$; $f_2 = 1851 \text{ MHz}$; $P_1 = P_2 = +10 \text{ dBm}$.

[4] Relative to $f = 1575 \text{ MHz}$.

[5] Within 10 % of the final gain.

9. Test information

9.1 Quality information

All qualification tests are performed according AEC-Q100 except for read point testing (final test of qualification sample). Which is done only at room temperature.

As part of the zero defect program, the following is part of the industrial test flow:

- Part Average Testing
- Maverick Lot Handling at assembly factory

10. Package outline

XSON6: plastic extremely thin small outline package; no leads; 6 terminals; body 1 x 1.45 x 0.5 mm

SOT886

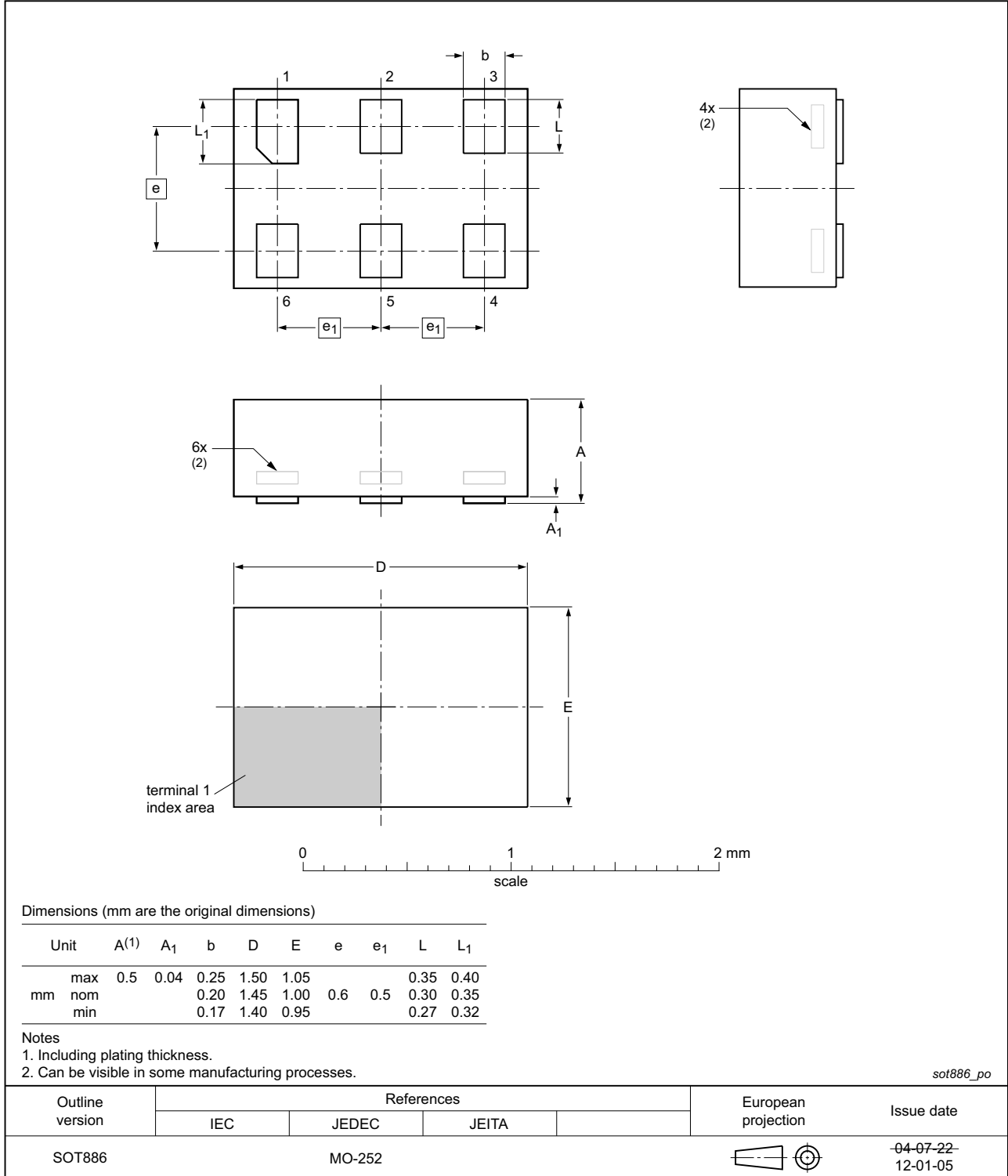


Fig 33. Package outline SOT886 (XSON6)

11. Handling information

CAUTION



This device is sensitive to ElectroStatic Discharge (ESD). Observe precautions for handling electrostatic sensitive devices.

Such precautions are described in the *ANSI/ESD S20.20*, *IEC/ST 61340-5*, *JESD625-A* or equivalent standards.

12. Abbreviations

Table 12. Abbreviations

Acronym	Description
AEC	Automotive Electronics Council
ATM	Automated Teller Machine (cash dispenser)
BPF	Band-Pass Filter
ESD	ElectroStatic Discharge
GLONASS	GLObal NAvigation Satellite System
GNSS	Global Navigation Satellite System
GPS	Global Positioning System
HBM	Human Body Model
MMIC	Monolithic Microwave Integrated Circuit
PCB	Printed Circuit Board
SAW	Surface Acoustic Wave
SiGe:C	Silicon Germanium Carbon

13. Revision history

Table 13. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
BGU7004 v.3	20170118	Product data sheet	-	BGU7004 v.2
Modifications:	<ul style="list-style-type: none"> Section 1: added GPS1103M according to our new naming convention 			
BGU7004 v.2	20150220	Product data sheet	-	BGU7004 v.1
Modifications:	<ul style="list-style-type: none"> The title of this data sheet has been changed. Section 1.3 on page 1: Added GLONASS, Galileo and Compass (BeiDou) to the possible applications. Section 11 on page 17: ESD information has moved from Section 1.1 to this section. Section 14.3 on page 18: Adjusted the disclaimers with respect to "suitability to use in automotive applications" and "Translations". 			
BGU7004 v.1	20110705	Product data sheet	-	-

14. Legal information

14.1 Data sheet status

Document status ^{[1][2]}	Product status ^[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

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