

2 wire interface Real-Time Clock ICs with Battery Backup switch-over Function

NO.EA-104-070626

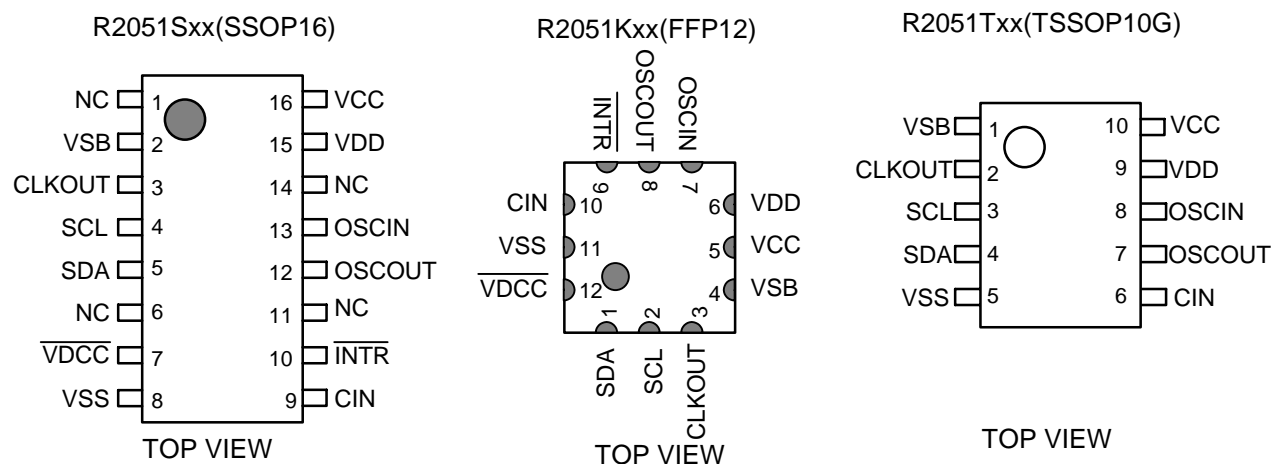
OUTLINE

The R2051 is a CMOS real-time clock IC connected to the CPU by two signal lines, SCL and SDA, and configured to perform serial transmission of time and calendar data to the CPU. Further, battery backup switchover circuit and a voltage detector are incorporated. The periodic interrupt circuit is configured to generate interrupt signals with six selectable interrupts ranging from 0.5 seconds to 1 month. The 2 alarm interrupt circuits generate interrupt signals at preset times. As the oscillation circuit is driven under constant voltage, fluctuation of the oscillator frequency due to supply voltage is small, and the time keeping current is small (TYP. 0.4 μ A at 3V). The oscillation halt sensing circuit can be used to judge the validity of internal data in such events as power-on; The supply voltage monitoring circuit is configured to record a drop in supply voltage below two selectable supply voltage monitoring threshold settings. The 32.768kHz clock output function (CMOS output) is intended to output sub-clock pulses for the external microcomputer. The oscillation adjustment circuit is intended to adjust time counts with high precision by correcting deviations in the oscillation frequency of the quartz crystal unit. Battery backup switchover function is the automatic switchover circuit between a main power supply and a backup battery of primary or secondary battery. Switchover is executed by monitoring the voltage of a main power supply, therefore the voltage of a backup battery voltage is not relevant. Since the package for these ICs is SSOP16 (5.0x6.4x1.25: R2051Sxx), FFP12 (2.0x2.0x1.0: R2051Kxx), or TSSOP10G (4.0x2.9x1.0: R2051Txx), high density mounting of ICs on boards is possible.

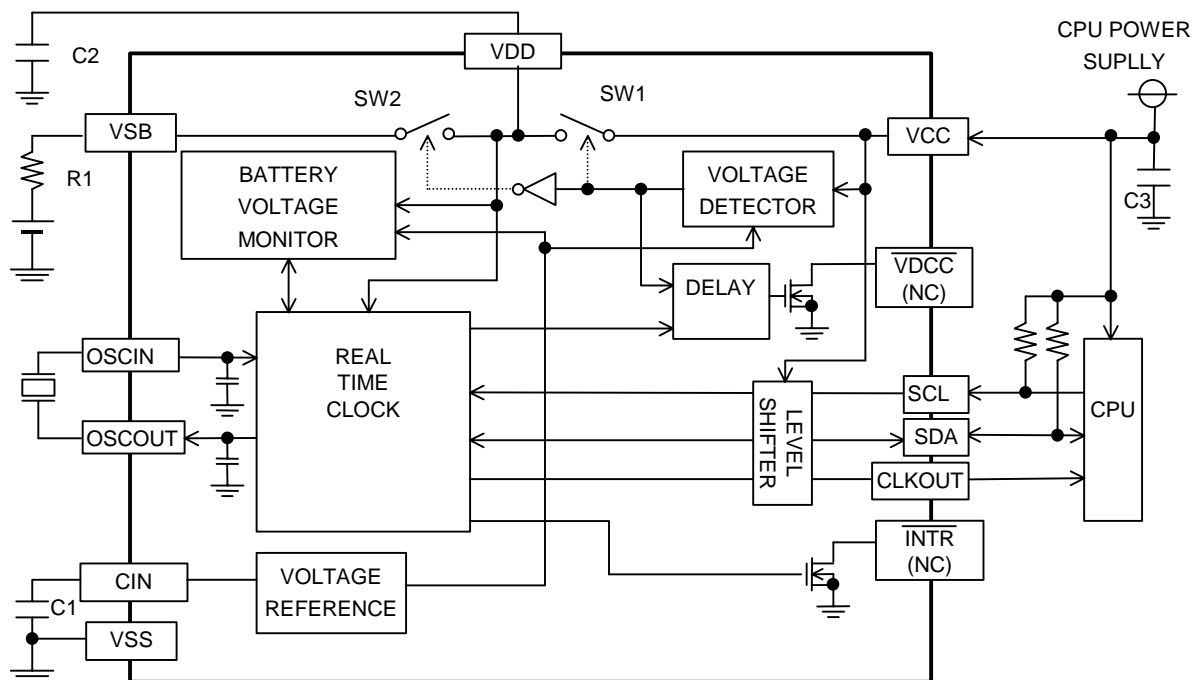
FEATURES

- Minimum Timekeeping supply voltage Typ. 0.75V (Max. 1.00V); VDD pin
- Low power consumption 0.4 μ A TYP (1.0 μ A MAX.) at VDD=3V
- Built-in Backup switchover circuit (can be used for a primary battery, a secondary battery, or an electric double layer capacitor)
- Only two signal lines (SCL and SDA) required for connection to the CPU. (I²C-Bus Interface, 400kHz)
- Time counters (counting hours, minutes, and seconds) and calendar counters (counting years, months, days, and weeks) (in BCD format)
- Interrupt circuit configured to generate interrupt signals (with interrupts ranging from 0.5 seconds to 1 month) to the CPU and provided with an interrupt flag and an interrupt halt (except R2051Txx)
- 2 alarm interrupt circuits (Alarm_W for week, hour, and minute alarm settings and Alarm_D for hour and minute alarm settings) (except R2051Txx)
- Built-in voltage detector with delay
- With Power-on flag to prove that the power supply starts from 0V
- 32-kHz clock output pin (CMOS output. "H" level is always equal to VCC.)
- Supply voltage monitoring circuit with two supply voltage monitoring threshold settings
- Automatic identification of leap years up to the year 2099
- Selectable 12-hour and 24-hour mode settings
- High precision oscillation adjustment circuit
- Built-in oscillation stabilization capacitors (CG and CD)
- CMOS process
- Package SSOP16 (5.0mm x 6.4mm x 1.25mm : R2051Sxx), FFP12 (2.0mm x 2.0mm x 1.0mm : R2051Kxx)
TSSOP10G (4.0x2.9x1.0: R2051Txx)

PIN CONFIGURATION



BLOCK DIAGRAM



() are for the R2051Txx only

SELECTION GUIDE

In the R2051xxx Series, output voltage and options can be designated.

Part Number is designated as follows:

R2051K01-E2 ← Part Number

↑ ↑ ↑

R2051abb-cc

| Code | Description |
|------|---|
| a | Designation of the package. K: FFP12 S: SSOP16 T: TSSOP10G |
| bb | Serial number of Voltage detector setting etc. |
| cc | Designation of the taping type. Only E2 is available. |

| Part Number | Package | -V _{DET1} (switch-over threshold) | DC Electrical Characteristics |
|-------------|----------|--|-------------------------------|
| R2051K01-E2 | FFP12 | 2.40(Typ.) | P.6 |
| R2051K02-E2 | FFP12 | 2.80(Typ.) | P.7 |
| R2051S01-E2 | SSOP16 | 2.40(Typ.) | P.6 |
| R2051S02-E2 | SSOP16 | 2.80(Typ.) | P.7 |
| R2051S03-E2 | SSOP16 | 4.00(Typ.) | P.8 |
| R2051T01-E2 | TSSOP10G | 2.40(Typ.) | P.6 |

PIN DESCRIPTION

| PIN | | | Symbol | Item | Description |
|---------------------|----------------------|------------------------|--------------------------|---|--|
| R2051Kxx (FFP12) | R2051Sxx (SSOP16) | R2051Txx (TSSOP10G) | | | |
| 2 | 4 | 3 | SCL | Serial Clock Line | The SCL pin is used to input clock pulses synchronizing the input and output of data to and from the SDA pin. Allows a maximum input voltage of 5.5 volts regardless of supply voltage. |
| 1 | 5 | 4 | SDA | Serial Data Line | The SDA pin is used to input or output data intended for writing or reading in synchronization with the SCL pin. Up to 5.5v beyond VDD may be input. This pin functions as an Nch open drain output. |
| 9 | 10 | - | $\overline{\text{INTR}}$ | Interrupt Output | The $\overline{\text{INTR}}$ pin is used to output alarm interrupt (Alarm_W) and alarm interrupt (Alarm_D) and output periodic interrupt signals to the CPU. Disabled at power-on from 0V. Nch. open drain output. |
| 3 | 3 | 2 | CLKOUT | 32kHz Clock Output | The CLKOUT pin is used to output 32.768-kHz clock pulses. CMOS output. "H" level is always equal to VCC. |
| 5 | 16 | 10 | VCC | Main Battery input | Supply power to the IC. |
| 4 | 2 | 1 | VSB | Power Supply Input for Backup Battery | Connect a primary battery for backup. Normally, power is supplied from VCC to the IC. If VCC level is equal or less than $-V_{\text{DET1}}$, power is supplied from this pin. |
| 7 | 13 | 8 | OSCIN | Oscillation Circuit Input / Output | The OSCIN and OSCOUT pins are used to connect the 32.768-kHz quartz crystal unit (with all other oscillation circuit components built into the R2051). |
| 8 | 12 | 7 | OSCOUT | | |
| 6 | 15 | 9 | VDD | Positive Power Supply Input | The VDD pin is connected to the power supply. Connect a capacitor as much as 0.1 μ F between VDD and VSS. In the case of using a secondary battery, connecting the secondary battery to this pin is possible. |
| 12 | 7 | - | $\overline{\text{VDCC}}$ | VCC Power Supply Monitoring Result Output | While monitoring VCC Power supply, if the voltage is equal or lower than $-V_{\text{DET1}}$, this output level is "L". When $\overline{\text{VDCC}}$ becomes "L", SW1 turns off and SW2 turns on. As a result, power is supplied from VSB pin to the internal real time clock. When VCC is equal to $+V_{\text{DET1}}$ or more, SW1 turns on and SW2 turns off. After t DELAY passed, $\overline{\text{VDCC}}$ output becomes off, or "H".ch Open-drain output. |
| 10 | 9 | 6 | CIN | Noise Bypass Pin | To stabilize the internal reference, connect a capacitor as much as 0.1 μ F between this pin and VSS. |
| 11 | 8 | 5 | VSS | Negative Power Supply Input | The VSS pin is grounded. |
| - | 1,6, 11,14 | - | | NC | No Connection |

ABSOLUTE MAXIMUM RATINGS

(V_{SS}=0V)

| Symbol | Item | Pin Name | Description | Unit |
|------------------|------------------------|---|------------------------------|------|
| V _{CC} | Supply Voltage 1 | VCC | -0.3 to +6.5 | V |
| V _{DD} | Supply Voltage 2 | VDD | -0.3 to +6.5 | V |
| V _{SB} | Supply Voltage 3 | VSB | -0.3 to +6.5 | V |
| V _I | Input Voltage 1 | SCL, SDA | -0.3 to +6.5 | V |
| | Input Voltage 2 | CIN | -0.3 to V _{DD} +0.3 | V |
| V _O | Output Voltage 1 | $\overline{\text{INTR}}$, $\overline{\text{VDCC}}$ *1) | -0.3 to +6.5 | V |
| | Output Voltage 2 | CLKOUT | -0.3 to V _{CC} +0.3 | V |
| I _{OUT} | Maximum Output Current | VDD | 10 | mA |
| P _D | Power Dissipation | T _{opt} = +25°C | 300 | mW |
| T _{opt} | Operating Temperature | | -40 to +85 | °C |
| T _{stg} | Storage Temperature | | -55 to +125 | °C |

*1) Except R2051Txx

RECOMMENDED OPERATING CONDITIONS

(V_{SS}=0V, T_{opt}=-40 to +85°C)

| Symbol | Item | Pin Name | Min. | Typ. | Max. | Unit |
|---------------------|---|---|---------------------------|--------|------|------|
| V _{access} | Supply Voltage | VCC power supply voltage for interfacing with CPU | -V _{DET1} *1) | | 5.5 | V |
| V _{CLK} | Minimum Timekeeping Voltage CGout, CDout=0pF *2), *3) | | | 0.75 | 1.00 | V |
| f _{XT} | Oscillation Frequency | | | 32.768 | | kHz |
| V _{PUP} | Pull-up Voltage | $\overline{\text{INTR}}$, $\overline{\text{VDCC}}$ *4) | | | 5.5 | V |

*1) -V_{DET1} in V_{access} specification is guaranteed by design.

*2) CGout is connected between OSCIN and VSS, CDout is connected between OSCOUT and VSS.
R2051 series incorporates the capacitors between OSCIN and VSS, between OSCOUT and VSS.
Then normally, CGout and CDout are not necessary.

*3) Quartz crystal unit: CL=6-8pF, R1=30KΩ

*4) Except R2051Txx

DC ELECTRICAL CHARACTERISTICS

• R2051K01, R2051S01, R2051T01

(Unless otherwise specified: $V_{SS}=0V$, $V_{CC}=V_{SB}=3.0V$, 0.1 μF between VDD and VSS, CIN and VSS,
 $T_{opt}=-40$ to $+85^{\circ}C$)

| Symbol | Item | Pin Name | Conditions | Min. | Typ. | Max. | Unit |
|---|---|--|---|---------------------|-------------------|---------------------|------------------|
| V_{IH} | "H" Input Voltage | SCL, SDA | | $0.8 \times V_{CC}$ | | 5.5 | V |
| V_{IL} | "L" Input Voltage | | | -0.3 | | $0.2 \times V_{CC}$ | |
| I_{OH} | "H" Output Current | CLKOUT | $V_{OH}=V_{CC}-0.5V$ | | | -0.5 | mA |
| I_{OL1} | "L" Output Current | CLKOUT | $V_{OL}=0.4V$ | 0.5 | | | mA |
| $I_{OL2}^{*2)}$ | | \overline{INTR} | | 2.0 | | | |
| I_{OL4} | | SDA | | 3.0 | | | |
| $I_{OL3}^{*2)}$ | | \overline{VDCC} | $V_{DD}, V_{SB}, V_{CC}=2.0V$ $V_{OL}=0.4V$ | 0.5 | | | |
| I_{IL} | Input Leakage Current | SCL | $V_i=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{OZ1} | Output Off-state Current 1 | SDA | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| $I_{OZ2}^{*2)}$ | Output Off-state Current 2 | \overline{INTR} , \overline{VDCC} | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{SB} | Time Keeping Current at Backup mode | VS \overline{B} | $V_{CC}=0V$, $V_{SB}=3.0V$, V_{DD} , Output=OPEN Time keeping | | 0.4 | 1.0 | μA |
| I_{SBL} | Leakage Current of Backup pin at VCC_on | VS \overline{B} | $V_{CC}=3.0V$, $V_{SB}=5.5V$ or $0V$, V_{DD} , Output=OPEN | -1.00 | | 1.00 | μA |
| V_{DETH} | Supply Voltage Monitoring Voltage "H" | VDD | $T_{opt}=+25^{\circ}C$ | 1.90 | 2.10 | 2.30 | V |
| V_{DETL} | Supply Voltage Monitoring Voltage "L" | VDD | $T_{opt}=+25^{\circ}C$ | 1.20 | 1.35 | 1.50 | V |
| $-V_{DET1}$ | Detector Threshold Voltage (falling edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 2.34 | 2.40 | 2.46 | V |
| $+V_{DET1}$ | Detector Released Voltage (rising edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 2.44 | 2.52 | 2.60 | V |
| $\frac{\Delta V_{DET}}{\Delta T_{opt}}$ | Detector Threshold and Released Voltage Temperature coefficient | VCC, VS \overline{B} | $T_{opt}=-40$ to $+85^{\circ}C$ *1) | | ± 100 | | ppm/ $^{\circ}C$ |
| V_{DDOUT1} | VDD Output Voltage 1 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=3.0V$, $I_{out}=1.0mA$ | V_{CC} -0.12 | V_{CC} -0.04 | | V |
| V_{DDOUT2} | VDD Output Voltage 2 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=2.0V$, $V_{SB}=3.0V$, $I_{out}=0.1mA$ | V_{SB} -0.08 | V_{SB} -0.02 | | V |
| CG | Internal Oscillation Capacitance 1 | OSCIN | | | 10 | | pF |
| CD | Internal Oscillation Capacitance 2 | OS \overline{COUT} | | | 10 | | |

*1) Guaranteed by design.

*2) Except R2051T01

● R2051K02, R2051S02

(Unless otherwise specified: $V_{SS}=0V$, $V_{CC}=3.3V$, $V_{SB}=3.0V$, $0.1\mu F$ between VDD and VSS, CIN and VSS, $T_{opt}=-40$ to $+85^{\circ}C$)

| Symbol | Item | Pin Name | Conditions | Min. | Typ. | Max. | Unit |
|---|---|--|---|---------------------|-------------------|---------------------|------------------|
| V_{IH} | "H" Input Voltage | SCL, SDA | | $0.8 \times V_{CC}$ | | 5.5 | V |
| V_{IL} | "L" Input Voltage | | | -0.3 | | $0.2 \times V_{CC}$ | |
| I_{OH} | "H" Output Current | CLKOUT | $V_{OH}=V_{CC}-0.5V$ | | | -0.5 | mA |
| I_{OL1} | "L" Output Current | CLKOUT | $V_{OL}=0.4V$ | 0.5 | | | mA |
| I_{OL2} | | \overline{INTR} | | 2.0 | | | |
| I_{OL4} | | SDA | | 3.0 | | | |
| I_{OL3} | | \overline{VDCC} | $V_{DD}, V_{SB}, V_{CC}=2.0V$ $V_{OL}=0.4V$ | 0.5 | | | |
| I_{IL} | Input Leakage Current | SCL | $V_I=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{OZ1} | Output Off-state Current 1 | SDA | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{OZ2} | Output Off-state Current 2 | \overline{INTR} , \overline{VDCC} | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{SB} | Time Keeping Current at Backup mode | VSB | $V_{CC}=0V$, $V_{SB}=3.0V$, V_{DD} , Output=OPEN Time keeping | | 0.4 | 1.0 | μA |
| I_{SBL} | Leakage Current of Backup pin at VCC_on | VSB | $V_{CC}=3.3V$, $V_{SB}=5.5V$ or $0V$, V_{DD} , Output=OPEN | -1.00 | | 1.00 | μA |
| V_{DETH} | Supply Voltage Monitoring Voltage "H" | VDD | $T_{opt}=+25^{\circ}C$ | 1.90 | 2.10 | 2.30 | V |
| V_{DETL} | Supply Voltage Monitoring Voltage "L" | VDD | $T_{opt}=+25^{\circ}C$ | 1.20 | 1.35 | 1.50 | V |
| $-V_{DET1}$ | Detector Threshold Voltage (falling edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 2.73 | 2.80 | 2.87 | V |
| $+V_{DET1}$ | Detector Released Voltage (rising edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 2.85 | 2.94 | 3.03 | V |
| $\frac{\Delta V_{DET}}{\Delta T_{opt}}$ | Detector Threshold and Released Voltage Temperature coefficient | VCC, VSB | $T_{opt}=-40$ to $+85^{\circ}C$ *1) | | ± 100 | | ppm/ $^{\circ}C$ |
| V_{DDOUT1} | VDD Output Voltage 1 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=3.3V$, $I_{out}=1.0mA$ | V_{CC} -0.12 | V_{CC} -0.04 | | V |
| V_{DDOUT2} | VDD Output Voltage 2 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=2.0V$, $V_{SB}=3.3V$, $I_{out}=0.1mA$ | V_{SB} -0.08 | V_{SB} -0.02 | | V |
| CG | Internal Oscillation Capacitance 1 | OSCIN | | | 10 | | pF |
| CD | Internal Oscillation Capacitance 2 | OSCOOUT | | | 10 | | |

*1) Guaranteed by design.

• R2051S03

(Unless otherwise specified: $V_{SS}=0V$, $V_{CC}=5.0V$, $V_{SB}=3.0V$, $0.1\mu F$ between VDD and VSS, CIN and VSS, $T_{opt}=-40$ to $+85^{\circ}C$)

| Symbol | Item | Pin Name | Conditions | Min. | Typ. | Max. | Unit |
|---|---|--|---|---------------------|-------------------|---------------------|------------------|
| V_{IH} | "H" Input Voltage | SCL, SDA | | $0.8 \times V_{CC}$ | | 5.5 | V |
| V_{IL} | "L" Input Voltage | | | -0.3 | | $0.2 \times V_{CC}$ | |
| I_{OH} | "H" Output Current | CLKOUT | $V_{OH}=V_{CC}-0.5V$ | | | -0.5 | mA |
| I_{OL1} | "L" Output Current | CLKOUT | $V_{OL}=0.4V$ | 0.5 | | | mA |
| I_{OL2} | | \overline{INTR} | | 2.0 | | | |
| I_{OL4} | | SDA | | 3.0 | | | |
| I_{OL3} | | \overline{VDCC} | $V_{DD}, V_{SB}, V_{CC}=2.0V$ $V_{OL}=0.4V$ | 0.5 | | | |
| I_{IL} | Input Leakage Current | SCL | $V_I=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{OZ1} | Output Off-state Current 1 | SDA | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{OZ2} | Output Off-state Current 2 | \overline{INTR} , \overline{VDCC} | $V_O=5.5V$ or V_{SS} | -1.0 | | 1.0 | μA |
| I_{SB} | Time Keeping Current at Backup mode | VSB | $V_{CC}=0V$, $V_{SB}=3.0V$, V_{DD} , Output=OPEN Time keeping | | 0.4 | 1.0 | μA |
| I_{SBL} | Leakage Current of Backup pin at VCC_on | VSB | $V_{CC}=5.0V$, $V_{SB}=5.5V$ or $0V$, V_{DD} , Output=OPEN | -1.00 | | 1.00 | μA |
| V_{DETH} | Supply Voltage Monitoring Voltage "H" | VDD | $T_{opt}=+25^{\circ}C$ | 1.90 | 2.10 | 2.30 | V |
| V_{DETL} | Supply Voltage Monitoring Voltage "L" | VDD | $T_{opt}=+25^{\circ}C$ | 1.20 | 1.35 | 1.50 | V |
| $-V_{DET1}$ | Detector Threshold Voltage (falling edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 3.90 | 4.00 | 4.10 | V |
| $+V_{DET1}$ | Detector Released Voltage (rising edge of VCC) | VCC | $T_{opt}=+25^{\circ}C$ | 4.07 | 4.20 | 4.33 | V |
| $\frac{\Delta V_{DET}}{\Delta T_{opt}}$ | Detector Threshold and Released Voltage Temperature coefficient | VCC, VSB | $T_{opt}=-40$ to $+85^{\circ}C$ *1) | | ± 100 | | ppm/ $^{\circ}C$ |
| V_{DDOUT1} | VDD Output Voltage 1 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=5.0V$, $I_{out}=1.0mA$ | V_{CC} -0.12 | V_{CC} -0.04 | | V |
| V_{DDOUT2} | VDD Output Voltage 2 | VDD | $T_{opt}=+25^{\circ}C$, $V_{CC}=2.0V$, $V_{SB}=3.0V$, $I_{out}=0.1mA$ | V_{SB} -0.08 | V_{SB} -0.02 | | V |
| CG | Internal Oscillation Capacitance 1 | OSCIN | | | 10 | | pF |
| CD | Internal Oscillation Capacitance 2 | OSCOUT | | | 10 | | |

*1) Guaranteed by design.

AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified: $V_{SS}=0V$, $T_{opt}=-40$ to $+85^{\circ}C$

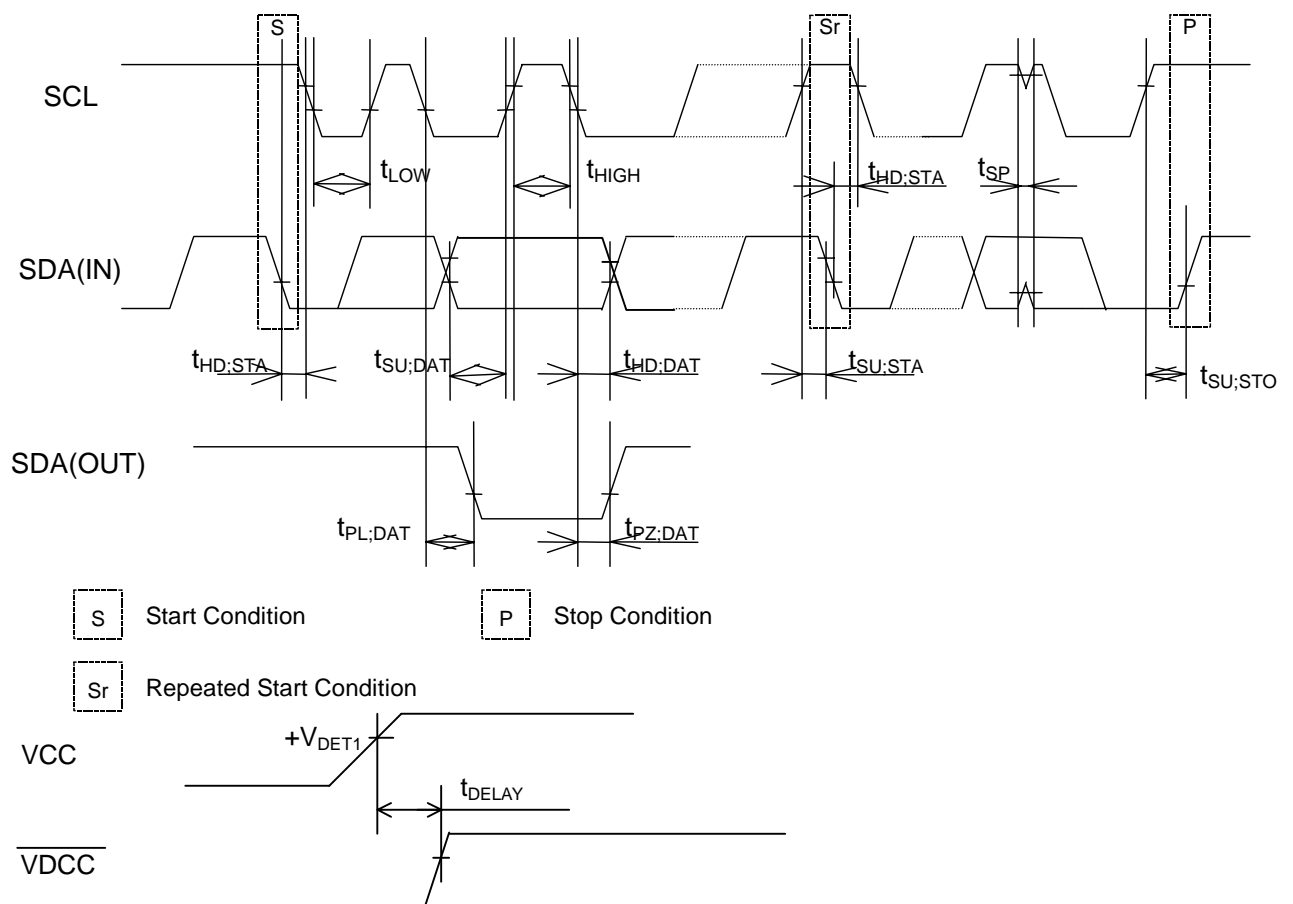
Input and Output Conditions: $V_{IH}=0.8 \times V_{CC}$, $V_{IL}=0.2 \times V_{CC}$, $V_{OH}=0.8 \times V_{CC}$, $V_{OL}=0.2 \times V_{CC}$, $C_L=50pF$

| Sym -bol | Item | Condi- Tions | $V_{CC} \geq 1.7V$ *1) | | | $V_{CC} \geq 2.5V$ *1) | | | Unit |
|-----------------|--|-----------------|------------------------|------|------|------------------------|------|------|---------|
| | | | Min. | Typ. | Max. | Min. | Typ. | Max. | |
| f_{SCL} | SCL Clock Frequency | | | | 100 | | | 400 | kHz |
| t_{LOW} | SCL Clock Low Time | | 4.7 | | | 1.3 | | | μs |
| t_{HIGH} | SCL Clock High Time | | 4.0 | | | 0.6 | | | μs |
| $t_{HD;STA}$ | Start Condition Hold Time | | 4.0 | | | 0.6 | | | μs |
| $t_{SU;STO}$ | Stop Condition Set Up Time | | 4.0 | | | 0.6 | | | μs |
| $t_{SU;STA}$ | Start Condition Set Up Time | | 4.7 | | | 0.6 | | | μs |
| $t_{SU;DAT}$ | Data Set Up Time | | 250 | | | 200 | | | ns |
| $t_{HD;DAT}$ | Data Hold Time | | 0 | | | 0 | | | ns |
| $t_{PL;DAT}$ | SDA "L" Stable Time After Falling of SCL | | | | 2.0 | | | 0.9 | μs |
| $t_{PZ;DAT}$ | SDA off Stable Time After Falling of SCL | | | | 2.0 | | | 0.9 | μs |
| t_R | Rising Time of SCL and SDA (input) | | | | 1000 | | | 300 | ns |
| t_F | Falling Time of SCL and SDA (input) | | | | 300 | | | 300 | ns |
| t_{SP} | Spike Width that can be removed with Input Filter | | | | 50 | | | 50 | ns |
| t_{RCV} | Recovery Time from Stop Condition to Start Condition | | 62 | | | 62 | | | μs |
| t_{DELAY} *2) | Output Delay Time of Voltage Detector | Time Keeping | 100 | 105 | 110 | 100 | 105 | 110 | ms |

*1) VCC voltage interfacing with CPU is defined by Vaccess (P.5 RECOMMENDED OPERATING CONDITIONS)

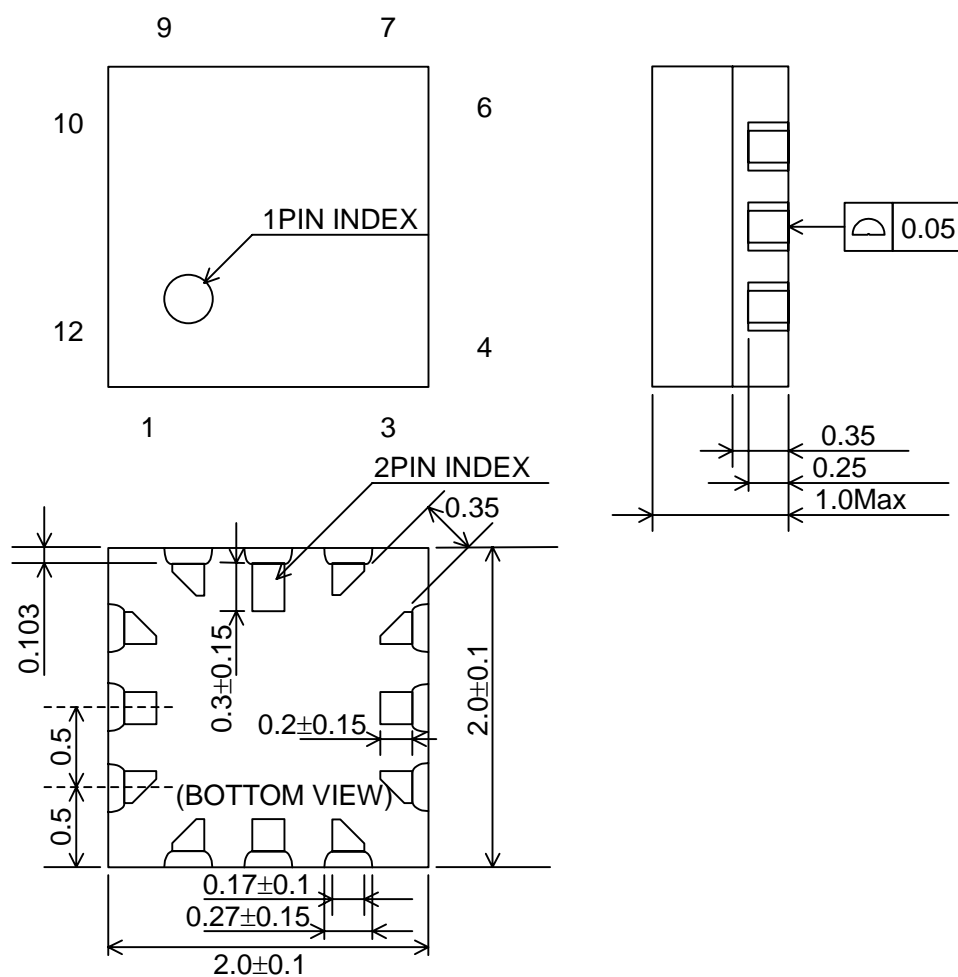
*2) Except R2051Txx

*) For reading/writing timing, see "P.34 Interfacing with the CPU •Data Transmission under Special Condition".

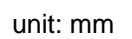


PACKAGE DIMENSIONS

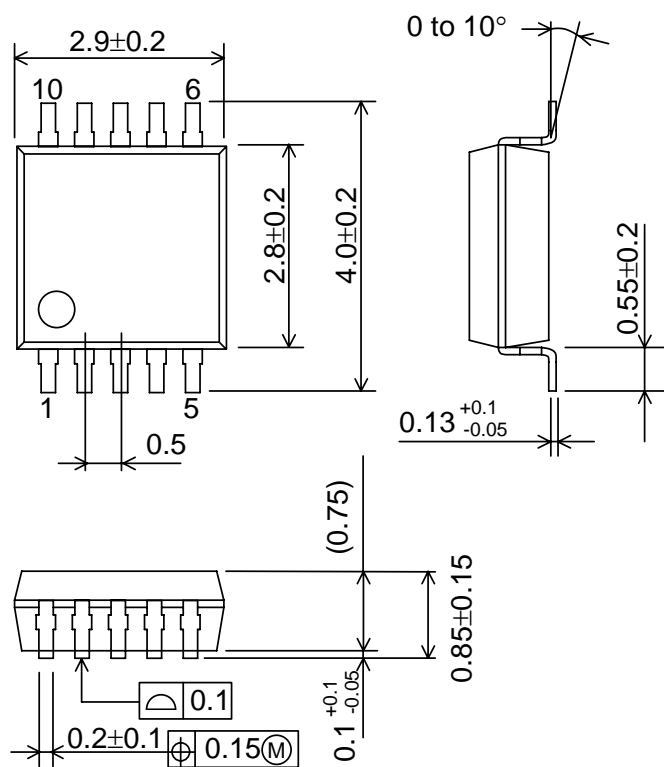
• R2051Kxx



unit: mm



● R2051Txx



unit: mm

GENERAL DESCRIPTION

• Battery Backup Switchover Function

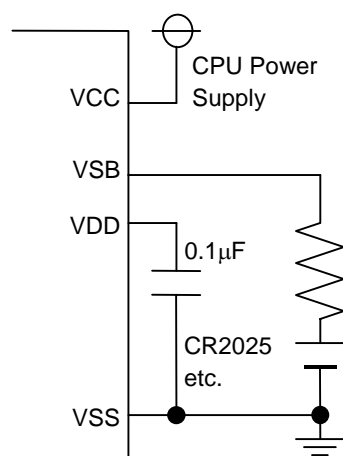
The R2051 has two power supply input, or VCC and VSB. With monitoring input voltage of VCC pin by internal Voltage Detector, it is selected which power supply of VCC or VSB is used for the internal power source.

Refer to the next table to see the state of the backup battery and internal power supply's state of the IC by each condition.

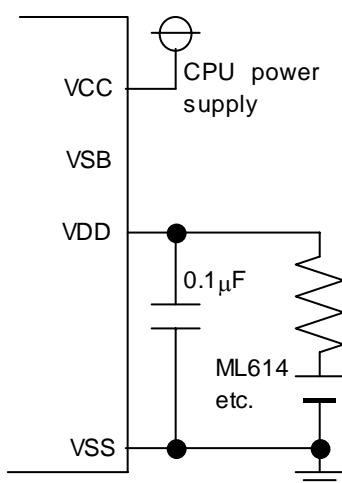
| $V_{CC} \geq V_{DET1}$ | $V_{CC} < V_{DET1}$ |
|--|---|
| VCC→RTC, VDD VDCC =OFF(H) (except R2051Txx) | VSB→RTC, VDD VDCC =L (except R2051Txx) |

As a backup battery, not only a primary battery such as CR2025, LR44, or a secondary battery such as ML614, TC616, but also an electric double layered capacitor or an aluminum capacitor can be used. Switchover point is judged with the voltage of the main power (VCC), therefore, if the backup voltage is higher than main supply voltage, switchover can be realized without extra load to the backup power supply.

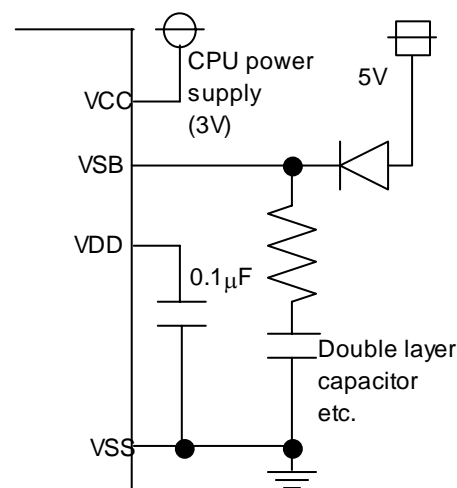
The case of back-up by primary battery



The case of back-up by capacitor or secondary battery (Charging voltage is equal to CPU power supply voltage)



The case of back-up by capacitor or secondary battery (Charging voltage is not equal to CPU power supply voltage)



• Interface with CPU

The R2051 is connected to the CPU by two signal lines SCL and SDA, through which it reads and writes data from and to the CPU. Since the output of the I/O pin of SDA is open drain, data interfacing with a CPU different supply voltage is possible by applying pull-up resistors on the circuit board. The maximum clock frequency of 400kHz (at $V_{DD}=3V$) of SCL enables data transfer in I2C-Bus fast mode. VCC falls down under $-V_{DET1}$, the R2051 stops accessing with CPU.

• Clock and Calendar Function

The R2051 reads and writes time data from and to the CPU in units ranging from seconds to the last two digits of the calendar year. The calendar year will automatically be identified as a leap year when its last two digits are a multiple of 4. Consequently, leap years up to the year 2099 can automatically be identified as such.

*) The year 2000 is a leap year while the year 2100 is not a leap year.

● Alarm Function

The R2051 incorporates the alarm interrupt circuit configured to generate interrupt signals to the CPU at preset times. The alarm interrupt circuit allows two types of alarm settings specified by the Alarm_W registers and the Alarm_D registers. The Alarm_W registers allow week, hour, and minute alarm settings including combinations of multiple day-of-week settings such as "Monday, Wednesday, and Friday" and "Saturday and Sunday". The Alarm_D registers allow hour and minute alarm settings. The Alarm_W outputs from $\overline{\text{INTR}}$ pin, and the Alarm_D outputs also from $\overline{\text{INTR}}$ pin. Each alarm function can be checked from the CPU by using a polling function. R2051Txx has Alarm_D and Alarm_W registers, but does not have $\overline{\text{INTR}}$ output pin.

● High-precision Oscillation Adjustment Function

The R2051 has built-in oscillation stabilization capacitors (CG and CD), that can be connected to an quartz crystal unit to configure an oscillation circuit. Two kinds of accuracy for this function are alternatives. To correct deviations in the oscillator frequency of the crystal, the oscillation adjustment circuit is configured to allow correction of a time count gain or loss (up to $\pm 1.5\text{ppm}$ or $\pm 0.5\text{ppm}$ at 25°C) from the CPU. The maximum range is approximately $\pm 189\text{ppm}$ (or $\pm 63\text{ppm}$) in increments of approximately 3ppm (or 1ppm). Such oscillation frequency adjustment in each system has the following advantages:

- * Allows timekeeping with much higher precision than conventional RTCs while using a quartz crystal unit with a wide range of precision variations.
- * Corrects seasonal frequency deviations through seasonal oscillation adjustment.
- * Allows timekeeping with higher precision particularly with a temperature sensing function out of RTC, through oscillation adjustment in tune with temperature fluctuations.

● Power-on Reset, Oscillation Halt Sensing Function and Supply Voltage Monitoring Function

The R2051 has 3 power supply pins (VCC, VSB, VDD), among them, VCC pin and VDD pin have monitoring function of supply voltage. VCC power supply monitoring circuit makes $\overline{\text{VDCC}}$ pin "L" when VCC power supply pin becomes equal or lower than $-V_{\text{DET1}}$. At the power-on of VCC, this circuit makes $\overline{\text{VDCC}}$ pin turn off, or "H" after the delay time, t_{DELAY} from when the VCC power supply pin becomes equal or more than $+V_{\text{DET1}}$. R2051Txx does not have $\overline{\text{VDCC}}$ output pin.

The R2051 incorporates an oscillation halt sensing circuit equipped with internal registers configured to record any past oscillation halt, the oscillation halt sensing circuit, VDD monitoring flag, and power-on reset flag are useful for judging the validity of time data.

Power on reset function reset the control registers when the system is powered on from 0V. At the same time, the fact is memorized to the register as a flag, thereby identifying whether they are powered on from 0V or battery backed-up.

The R2051 also incorporates a supply voltage monitoring circuit equipped with internal registers configured to record any drop in supply voltage below a certain threshold value. Supply voltage monitoring threshold settings can be selected between 2.1V and 1.35V through internal register settings. The sampling rate is normally 1s. The oscillation halt sensing circuit is configured to confirm the established invalidation of time data in contrast to the supply voltage monitoring circuit intended to confirm the potential invalidation of time data. Further, the supply voltage monitoring circuit can be applied to battery supply voltage monitoring.

● Periodic Interrupt Function

The R2051 incorporates the periodic interrupt circuit configured to generate periodic interrupt signals aside from interrupt signals generated by the periodic interrupt circuit for output from the $\overline{\text{INTR}}$ pin. Periodic interrupt signals have five selectable frequency settings of 2 Hz (once per 0.5 seconds), 1 Hz (once per 1 second), 1/60 Hz (once per 1 minute), 1/3600 Hz (once per 1 hour), and monthly (the first day of every month). Further, periodic interrupt signals also have two selectable waveforms, a normal pulse form (with a frequency of 2 Hz or 1 Hz) and special

form adapted to interruption from the CPU in the level mode (with second, minute, hour, and month interrupts). The condition of periodic interrupt signals can be monitored with using a polling function. R2051Txx has the periodic interrupt registers, but does not have $\overline{\text{INTR}}$ output pin.

- **32kHz Clock Output**

The R2051 incorporates a 32-kHz clock circuit configured to generate clock pulses with the oscillation frequency of a 32.768kHz quartz crystal unit for output from the CLKOUT pin (CMOS push-pull output). The 32-kHz clock output is always enabled and the “H” level of the CLKOUT pin is same as VCC power supply.

Address Mapping

| | Address | Register Name | Data | | | | | | | |
|---|----------|-------------------------------------|------------|------|----------------------------|--------------|--------------|------|------|------|
| | A3A2A1A0 | | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| 0 | 0 0 0 0 | Second Counter | - *2) | S40 | S20 | S10 | S8 | S4 | S2 | S1 |
| 1 | 0 0 0 1 | Minute Counter | - | M40 | M20 | M10 | M8 | M4 | M2 | M1 |
| 2 | 0 0 1 0 | Hour Counter | - | - | H20 P/ \overline{A} | H10 | H8 | H4 | H2 | H1 |
| 3 | 0 0 1 1 | Day-of-week Counter | - | - | - | - | - | W4 | W2 | W1 |
| 4 | 0 1 0 0 | Day-of-month Counter | - | - | D20 | D10 | D8 | D4 | D2 | D1 |
| 5 | 0 1 0 1 | Month Counter and Century Bit | 19 /20 | - | - | MO10 | MO8 | MO4 | MO2 | MO1 |
| 6 | 0 1 1 0 | Year Counter | Y80 | Y40 | Y20 | Y10 | Y8 | Y4 | Y2 | Y1 |
| 7 | 0 1 1 1 | Oscillation Adjustment Register *3) | DEV *4) | F6 | F5 | F4 | F3 | F2 | F1 | F0 |
| 8 | 1 0 0 0 | Alarm_W (Minute Register) | - | WM40 | WM20 | WM10 | WM8 | WM4 | WM2 | WM1 |
| 9 | 1 0 0 1 | Alarm_W (Hour Register) | - | - | WH20 WP/ \overline{A} | WH10 | WH8 | WH4 | WH2 | WH1 |
| A | 1 0 1 0 | Alarm_W (Day-of-week Register) | - | WW6 | WW5 | WW4 | WW3 | WW2 | WW1 | WW0 |
| B | 1 0 1 1 | Alarm_D (Minute Register) | - | DM40 | DM20 | DM10 | DM8 | DM4 | DM2 | DM1 |
| C | 1 1 0 0 | Alarm_D (Hour Register) | - | - | DH20 DP/ \overline{A} | DH10 | DH8 | DH4 | DH2 | DH1 |
| D | 1 1 0 1 | - | - | - | - | - | - | - | - | - |
| E | 1 1 1 0 | Control Register 1 *3) | WALE | DALE | 12 /24 | SCRA TCH2 | TEST | CT2 | CT1 | CT0 |
| F | 1 1 1 1 | Control Register 2 *3) | VDSL | VDET | \overline{XST} | PON *5) | SCRA TCH1 | CTFG | WAFG | DAFG |

Notes:

* 1) All the data listed above accept both reading and writing.

* 2) The data marked with "-" is invalid for writing and reset to 0 for reading.

* 3) When the PON bit is set to 1 in Control Register 2, all the bits are reset to 0 in Oscillation Adjustment Register, Control Register 1 and Control Register 2 excluding the \overline{XST} bit.

* 4) When DEV=0, the oscillation adjustment circuit is configured to allow correction of a time count gain or loss up to ± 1.5 ppm. When DEV=1, the oscillation adjustment circuit is configured to allow correction of a time count gain or loss up to or ± 0.5 ppm.

* 5) PON is a power-on-reset flag.

Register Settings

• Control Register 1 (Address Eh)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|------|------|--------|--------------|------|-----|-----|-----|---------------------|
| WALE | DALE | 12 /24 | SCRA TCH2 | TEST | CT2 | CT1 | CT0 | (For Writing) |
| WALE | DALE | 12 /24 | SCRA TCH2 | TEST | CT2 | CT1 | CT0 | (For Reading) |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

(1) WALE, DALE Alarm_W Enable Bit, Alarm_D Enable Bit

| WALE,DALE | Description | |
|-----------|---|-----------|
| 0 | Disabling the alarm interrupt circuit (under the control of the settings of the Alarm_W registers and the Alarm_D registers). | (Default) |
| 1 | Enabling the alarm interrupt circuit (under the control of the settings of the Alarm_W registers and the Alarm_D registers) | |

(2) 12 /24 12 /24-hour Mode Selection Bit

| 12 /24 | Description | |
|--------|--|-----------|
| 0 | Selecting the 12-hour mode with a.m. and p.m. indications. | (Default) |
| 1 | Selecting the 24-hour mode | |

Setting the 12 /24 bit to 0 and 1 specifies the 12-hour mode and the 24-hour mode, respectively.

| 24-hour mode | 12-hour mode | 24-hour mode | 12-hour mode |
|--------------|--------------|--------------|--------------|
| 00 | 12 (AM12) | 12 | 32 (PM12) |
| 01 | 01 (AM 1) | 13 | 21 (PM 1) |
| 02 | 02 (AM 2) | 14 | 22 (PM 2) |
| 03 | 03 (AM 3) | 15 | 23 (PM 3) |
| 04 | 04 (AM 4) | 16 | 24 (PM 4) |
| 05 | 05 (AM 5) | 17 | 25 (PM 5) |
| 06 | 06 (AM 6) | 18 | 26 (PM 6) |
| 07 | 07 (AM 7) | 19 | 27 (PM 7) |
| 08 | 08 (AM 8) | 20 | 28 (PM 8) |
| 09 | 09 (AM 9) | 21 | 29 (PM 9) |
| 10 | 10 (AM10) | 22 | 30 (PM10) |
| 11 | 11 (AM11) | 23 | 31 (PM11) |

Setting the 12 /24 bit should precede writing time data

(3) SCRATCH2 Scratch Bit 2

| SCRATCH2 | Description | |
|----------|-------------|-----------|
| 0 | | (Default) |
| 1 | | |

The SCRATCH2 bit is intended for scratching and accepts the reading and writing of 0 and 1.
The SCRATCH2 bit will be set to 0 when the PON bit is set to 1 in the Control Register 1.

(4) TEST Test Bit

| TEST | Description | |
|------|------------------------|-----------|
| 0 | Normal operation mode. | (Default) |
| 1 | Test mode. | |

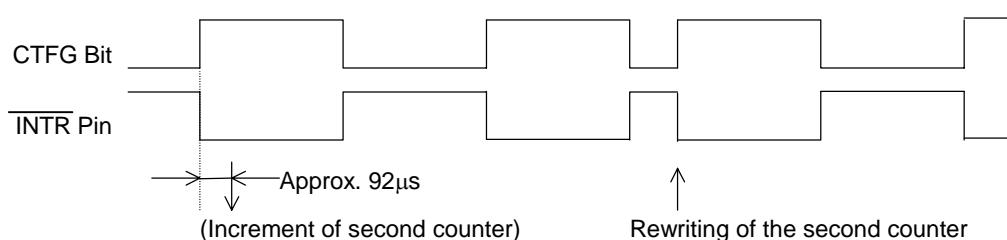
The TEST bit is used only for testing in the factory and should normally be set to 0.

(5) CT2, CT1, and CT0 Periodic Interrupt Selection Bits

| CT2 | CT1 | CT0 | Description | |
|-----|-----|-----|-------------------|--|
| | | | Wave form mode | Interrupt Cycle and Falling Timing |
| 0 | 0 | 0 | - | OFF(H) |
| 0 | 0 | 1 | - | Fixed at "L" |
| 0 | 1 | 0 | Pulse Mode *1) | 2Hz (Duty50%) |
| 0 | 1 | 1 | Pulse Mode *1) | 1Hz (Duty50%) |
| 1 | 0 | 0 | Level Mode *2) | Once per 1 second (Synchronized with second counter increment) |
| 1 | 0 | 1 | Level Mode *2) | Once per 1 minute (at 00 seconds of every minute) |
| 1 | 1 | 0 | Level Mode *2) | Once per hour (at 00 minutes and 00 seconds of every hour) |
| 1 | 1 | 1 | Level Mode *2) | Once per month (at 00 hours, 00 minutes, and 00 seconds of first day of every month) |

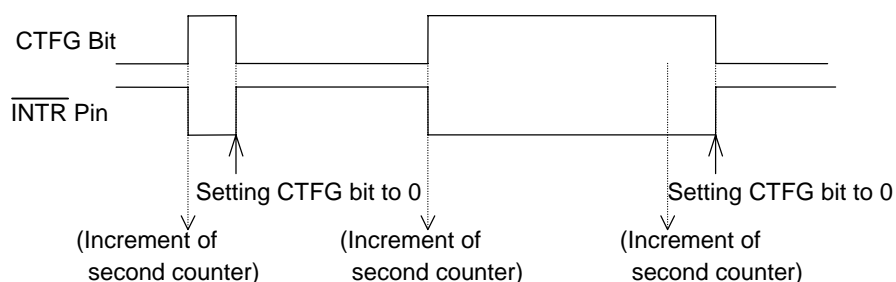
(Default)

* 1) Pulse Mode: 2-Hz and 1-Hz clock pulses are output in synchronization with the increment of the second counter as illustrated in the timing chart below.



In the pulse mode, the increment of the second counter is delayed by approximately 92 μ s from the falling edge of clock pulses. Consequently, time readings immediately after the falling edge of clock pulses may appear to lag behind the time counts of the real-time clocks by approximately 1 second. Rewriting the second counter will reset the other time counters of less than 1 second, driving the $\overline{\text{INTR}}$ pin low.

* 2) Level Mode: Periodic interrupt signals are output with selectable interrupt cycle settings of 1 second, 1 minute, 1 hour, and 1 month. The increment of the second counter is synchronized with the falling edge of periodic interrupt signals. For example, periodic interrupt signals with an interrupt cycle setting of 1 second are output in synchronization with the increment of the second counter as illustrated in the timing chart below.



*1), *2) When the oscillation adjustment circuit is used, the interrupt cycle will fluctuate once per 20sec. or 60sec. as follows:

Pulse Mode: The “L” period of output pulses will increment or decrement by a maximum of ± 3.784 ms. For example, 1-Hz clock pulses will have a duty cycle of $50 \pm 0.3784\%$.

Level Mode: A periodic interrupt cycle of 1 second will increment or decrement by a maximum of ± 3.784 ms.

R2051Tx does not have $\overline{\text{INTR}}$ output pin

• Control Register 2 (Address Fh)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|------|------|-------------------------|-----|--------------|------|------|------|---------------------|
| VDSL | VDET | $\overline{\text{XST}}$ | PON | SCRA TCH1 | CTFG | WAFG | DAFG | (For Writing) |
| VDSL | VDET | $\overline{\text{XST}}$ | PON | SCRA TCH1 | CTFG | WAFG | DAFG | (For Reading) |
| 0 | 0 | Indefinite | 1 | 0 | 0 | 0 | 0 | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to “1” due to VDD power-on from 0 volts.

(1) VDSL VDD Supply Voltage Monitoring Threshold Selection Bit

| VDSL | Description | |
|------|---|-----------|
| 0 | Selecting the VDD supply voltage monitoring threshold setting of 2.1v. | (Default) |
| 1 | Selecting the VDD supply voltage monitoring threshold setting of 1.35v. | |

The VDSL bit is intended to select the VDD supply voltage monitoring threshold settings.

(2) VDET Supply Voltage Monitoring Result Indication Bit

| VDET | Description | |
|------|---|-----------|
| 0 | Indicating supply voltage above the supply voltage monitoring threshold settings. | (Default) |
| 1 | Indicating supply voltage below the supply voltage monitoring threshold settings. | |

Once the VDET bit is set to 1, the supply voltage monitoring circuit will be disabled while the VDET bit will hold the setting of 1. The VDET bit accepts only the writing of 0, which restarts the supply voltage monitoring circuit. Conversely, setting the VDET bit to 1 causes no event.

(3) $\overline{\text{XST}}$ Oscillation Halt Sensing Monitor Bit

| $\overline{\text{XST}}$ | Description |
|-------------------------|---|
| 0 | Sensing a halt of oscillation |
| 1 | Sensing a normal condition of oscillation |

The $\overline{\text{XST}}$ accepts the reading and writing of 0 and 1. The $\overline{\text{XST}}$ bit will be set to 0 when the oscillation halt sensing. The $\overline{\text{XST}}$ bit will hold 0 even after the restart of oscillation.

(4) PON Power-on-reset Flag Bit

| PON | Description | |
|-----|-------------------------------|-----------|
| 0 | Normal condition | (Default) |
| 1 | Detecting VDD power-on -reset | |

The PON bit is for sensing power-on reset condition.

* The PON bit will be set to 1 when VDD power-on from 0 volts. The PON bit will hold the setting of 1 even after power-on.

* When the PON bit is set to 1, all bits will be reset to 0, in the Oscillation Adjustment Register, Control Register 1, and Control Register 2, except $\overline{\text{XST}}$ and PON. As a result, $\overline{\text{INTR}}$ pin stops outputting.

* The PON bit accepts only the writing of 0. Conversely, setting the PON bit to 1 causes no event.

(5) SCRATCH1 Scratch Bit 1

| SCRATCH1 | Description |
|----------|-------------|
| 0 | |
| 1 | |

(Default)

The SCRATCH1 bit is intended for scratching and accepts the reading and writing of 0 and 1. The SCRATCH1 bit will be set to 0 when the PON bit is set to 1 in the Control Register 2.

(6) CTFG Periodic Interrupt Flag Bit

| CTFG | Description |
|------|---------------------------------|
| 0 | Periodic interrupt output = "H" |
| 1 | Periodic interrupt output = "L" |

(Default)

The CTFG bit is set to 1 when the periodic interrupt signals are output from the $\overline{\text{INTR}}$ pin ("L"). The CTFG bit accepts only the writing of 0 in the level mode, which disables ("H") the $\overline{\text{INTR}}$ pin until it is enabled ("L") again in the next interrupt cycle. Conversely, setting the CTFG bit to 1 causes no event. R2051Txx has CTFG bit, but does not have $\overline{\text{INTR}}$ output pin

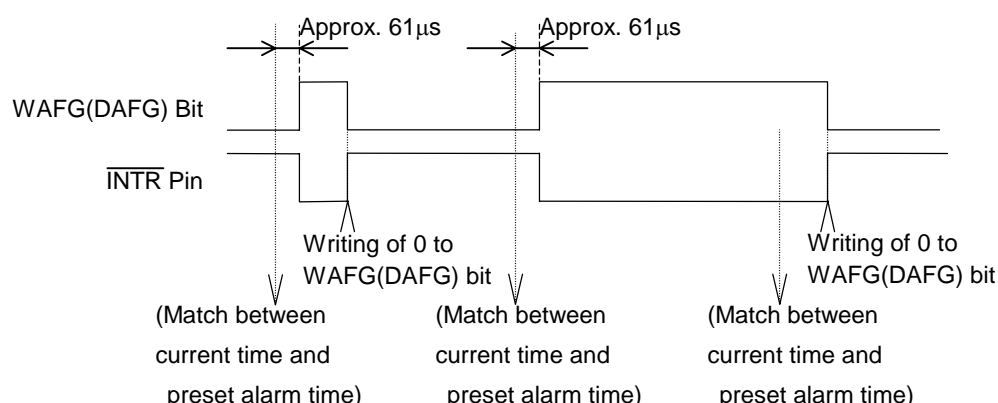
(7) WAFG,DAFG Alarm_W Flag Bit and Alarm_D Flag Bit

| WAFG,DAFG | Description |
|-----------|--|
| 0 | Indicating a mismatch between current time and preset alarm time |
| 1 | Indicating a match between current time and preset alarm time |

(Default)

The WAFG and DAFG bits are valid only when the WALE and DALE have the setting of 1, which is caused approximately 61 μ s after any match between current time and preset alarm time specified by the Alarm_W registers and the Alarm_D registers. The WAFG (DAFG) bit accepts only the writing of 0. $\overline{\text{INTR}}$ pin outputs off ("H") when this bit is set to 0. And $\overline{\text{INTR}}$ pin outputs "L" again at the next preset alarm time. Conversely, setting the WAFG and DAFG bits to 1 causes no event. The WAFG and DAFG bits will have the reading of 0 when the alarm interrupt circuit is disabled with the WALE and DALE bits set to 0. The settings of the WAFG and DAFG bits are synchronized with the output of the $\overline{\text{INTR}}$ pin as shown in the timing chart below.

R2051Txx has WAFG, DAFG bits. But does not have $\overline{\text{INTR}}$ output pin.



• **Time Counter (Address 0-2h)**

Second Counter (Address 0h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|------------|------------|------------|------------|------------|------------|------------|---------------------|
| - | S40 | S20 | S10 | S8 | S4 | S2 | S1 | (For Writing) |
| 0 | S40 | S20 | S10 | S8 | S4 | S2 | S1 | (For Reading) |
| 0 | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Default Settings *) |

Minute Counter (Address 1h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|------------|------------|------------|------------|------------|------------|------------|---------------------|
| - | M40 | M20 | M10 | M8 | M4 | M2 | M1 | (For Writing) |
| 0 | M40 | M20 | M10 | M8 | M4 | M2 | M1 | (For Reading) |
| 0 | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Default Settings *) |

Hour Counter (Address 2h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|----|---------------------------|------------|------------|------------|------------|------------|---------------------|
| - | - | P/ \bar{A} or H20 | H10 | H8 | H4 | H2 | H1 | (For Writing) |
| 0 | 0 | P/ \bar{A} or H20 | H10 | H8 | H4 | H2 | H1 | (For Reading) |
| 0 | 0 | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

* Time digit display (BCD format) as follows:

The second digits range from 00 to 59 and are carried to the minute digit in transition from 59 to 00.

The minute digits range from 00 to 59 and are carried to the hour digits in transition from 59 to 00.

The hour digits range as shown in "P18 • Control Register 1 (ADDRESS Eh) (2) 12 /24: 12 /24-hour Mode Selection Bit" and are carried to the day-of-month and day-of-week digits in transition from PM11 to AM12 or from 23 to 00.

* Any writing to the second counter resets divider units of less than 1 second.

* Any carry from lower digits with the writing of non-existent time may cause the time counters to malfunction. Therefore, such incorrect writing should be replaced with the writing of existent time data.

• **Day-of-week Counter (Address 3h)**

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|----|----|----|----|------------|------------|------------|---------------------|
| - | - | - | - | - | W4 | W2 | W1 | (For Writing) |
| 0 | 0 | 0 | 0 | 0 | W4 | W2 | W1 | (For Reading) |
| 0 | 0 | 0 | 0 | 0 | Indefinite | Indefinite | Indefinite | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

* The day-of-week counter is incremented by 1 when the day-of-week digits are carried to the day-of-month digits.

* Day-of-week display (incremented in septimal notation):

(W4, W2, W1) = (0, 0, 0) → (0, 0, 1) → ... → (1, 1, 0) → (0, 0, 0)

- * Correspondences between days of the week and the day-of-week digits are user-definable (e.g. Sunday = 0, 0, 0)
- * The writing of (1, 1, 1) to (W4, W2, W1) is prohibited except when days of the week are unused.

• Calendar Counter (Address 4-6h)

Day-of-month Counter (Address 4h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----|----|----------------|----------------|----------------|----------------|----------------|----------------|
| - | - | D20 | D10 | D8 | D4 | D2 | D1 |
| 0 | 0 | D20 | D10 | D8 | D4 | D2 | D1 |
| 0 | 0 | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

Month Counter + Century Bit (Address 5h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------------|----|----|----------------|----------------|----------------|----------------|----------------|
| 19 /20 | - | - | MO10 | MO8 | MO4 | MO2 | MO1 |
| 19 /20 | 0 | 0 | MO10 | MO8 | MO4 | MO2 | MO1 |
| Indefi nite | 0 | 0 | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

Year Counter (Address 6h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----------------|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| Y80 | Y40 | Y20 | Y10 | Y8 | Y4 | Y2 | Y1 |
| Y80 | Y40 | Y20 | Y10 | Y8 | Y4 | Y2 | Y1 |
| Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

- *) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

* The calendar counters are configured to display the calendar digits in BCD format by using the automatic calendar function as follows:

The day-of-month digits (D20 to D1) range from 1 to 31 for January, March, May, July, August, October, and December; from 1 to 30 for April, June, September, and November; from 1 to 29 for February in leap years; from 1 to 28 for February in ordinary years. The day-of-month digits are carried to the month digits in reversion from the last day of the month to 1. The month digits (MO10 to MO1) range from 1 to 12 and are carried to the year digits in reversion from 12 to 1.

The year digits (Y80 to Y1) range from 00 to 99 (00, 04, 08, ..., 92, and 96 in leap years) and are carried to the 19 /20 digits in reversion from 99 to 00.

The 19 /20 digits cycle between 0 and 1 in reversion from 99 to 00 in the year digits.

* Any carry from lower digits with the writing of non-existent calendar data may cause the calendar counters to malfunction. Therefore, such incorrect writing should be replaced with the writing of existent calendar data.

• Oscillation Adjustment Register (Address 7h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|-----|----|----|----|----|----|----|----|---------------------|
| DEV | F6 | F5 | F4 | F3 | F2 | F1 | F0 | (For Writing) |
| DEV | F6 | F5 | F4 | F3 | F2 | F1 | F0 | (For Reading) |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

DEV bit

When DEV is set to 0, the Oscillation Adjustment Circuit operates 00, 20, 40 seconds.

When DEV is set to 1, the Oscillation Adjustment Circuit operates 00 seconds.

F6 to F0 bits

The Oscillation Adjustment Circuit is configured to change time counts of 1 second on the basis of the settings of the Oscillation Adjustment Register at the timing set by DEV.

* The Oscillation Adjustment Circuit will not operate with the same timing (00, 20, or 40 seconds) as the timing of writing to the Oscillation Adjustment Register.

* The F6 bit setting of 0 causes an increment of time counts by $((F5, F4, F3, F2, F1, F0) - 1) \times 2$.

The F6 bit setting of 1 causes a decrement of time counts by $((\overline{F5}, \overline{F4}, \overline{F3}, \overline{F2}, \overline{F1}, \overline{F0}) + 1) \times 2$.

The settings of "*", 0, 0, 0, 0, 0, "*" ("*" representing either "0" or "1") in the F6, F5, F4, F3, F2, F1, and F0 bits cause neither an increment nor decrement of time counts.

Example:

If (DEV, F6, F5, F4, F3, F2, F1, F0) is set to (0, 0, 0, 0, 0, 1, 1, 1), when the second digits read 00, 20, or 40, an increment of the current time counts of $32768 + (7 - 1) \times 2$ to 32780 (a current time count loss).

If (DEV, F6, F5, F4, F3, F2, F1, F0) is set to (0, 0, 0, 0, 0, 0, 0, 1), when the second digits read 00, 20, 40, neither an increment nor a decrement of the current time counts of 32768.

If (DEV, F6, F5, F4, F3, F2, F1, F0) is set to (1, 1, 1, 1, 1, 1, 1, 0), when the second digits read 00, a decrement of the current time counts of $32768 + (-2) \times 2$ to 32764 (a current time count gain).

An increase of two clock pulses once per 20 seconds causes a time count loss of approximately 3 ppm ($2 / (32768 \times 20) = 3.051$ ppm). Conversely, a decrease of two clock pulses once per 20 seconds causes a time count gain of 3 ppm. Consequently, when DEV is set to "0", deviations in time counts can be corrected with a precision of ± 1.5 ppm. In the same way, when DEV is set to "1", deviations in time counts can be corrected with a precision of ± 0.5 ppm. Note that the oscillation adjustment circuit is configured to correct deviations in time counts and not the oscillation frequency of the 32.768-kHz clock pulses. For further details, see "P38 Configuration of Oscillation Circuit and Correction of Time Count Deviations • Oscillation Adjustment Circuit".

Alarm_W Registers (Address 8-Ah)

Alarm_W Minute Register (Address 8h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| - | WM40 | WM20 | WM10 | WM8 | WM4 | WM2 | WM1 |
| 0 | WM40 | WM20 | WM10 | WM8 | WM4 | WM2 | WM1 |
| 0 | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

Alarm_W Hour Register (Address 9h)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----|----|----------------------------|----------------|----------------|----------------|----------------|----------------|
| - | - | WH20 WP/ \overline{A} | WH10 | WH8 | WH4 | WH2 | WH1 |
| 0 | 0 | WH20 WP/ \overline{A} | WH10 | WH8 | WH4 | WH2 | WH1 |
| 0 | 0 | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

Alarm_W Day-of-week Register (Address Ah)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|----|----------------|----------------|----------------|----------------|----------------|----------------|----------------|
| - | WW6 | WW5 | WW4 | WW3 | WW2 | WW1 | WW0 |
| 0 | WW6 | WW5 | WW4 | WW3 | WW2 | WW1 | WW0 |
| 0 | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite | Indefi nite |

(For Writing)
(For Reading)
Default Settings *)

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

* The D5 bit of the Alarm_W Hour Register represents WP/ \overline{A} when the 12-hour mode is selected (0 for a.m. and 1 for p.m.) and WH20 when the 24-hour mode is selected (tens in the hour digits).

* The Alarm_W Registers should not have any non-existent alarm time settings.

(Note that any mismatch between current time and preset alarm time specified by the Alarm_W registers may disable the alarm interrupt circuit.)

* When the 12-hour mode is selected, the hour digits read 12 and 32 for 0 a.m. and 0 p.m., respectively. (See "P18 •Control Register 1 (ADDRESS Eh) (2) $\overline{12}$ /24: $\overline{12}$ /24-hour Mode Selection Bit")

* WW0 to WW6 correspond to W4, W2, and W1 of the day-of-week counter with settings ranging from (0, 0, 0) to (1, 1, 0).

* WW0 to WW6 with respective settings of 0 disable the outputs of the Alarm_W Registers.

Example of Alarm Time Setting

| Alarm Preset alarm time | Day-of-week | | | | | | | 12-hour mode | | | | 24-hour mode | | | |
|------------------------------------|-------------|------|------|------|-----|------|------|--------------|---|---|---|--------------|---|---|---|
| | Sun. | Mon. | Tue. | Wed. | Th. | Fri. | Sat. | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| | WW | WW | WW | WW | WW | WW | WW | h | h | m | m | h | h | m | m |
| 00:00 a.m. on all days | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 01:30 a.m. on all days | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 3 | 0 | 0 | 1 | 3 | 0 |
| 11:59 a.m. on all days | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 5 | 9 | 1 | 1 | 5 | 9 |
| 00:00 p.m. on Mon. to Fri. | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 3 | 2 | 0 | 0 | 1 | 2 | 0 | 0 |
| 01:30 p.m. on Sun. | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 2 | 1 | 3 | 0 | 1 | 3 | 3 | 0 |
| 11:59 p.m. on Mon., Wed., and Fri. | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 3 | 1 | 5 | 9 | 2 | 3 | 5 | 9 |

Note that the correspondence between WW0 to WW6 and the days of the week shown in the above table is only an example and not mandatory.

• Alarm_D Register (Address B-Ch)

Alarm_D Minute Register (Address Bh)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|------------|------------|------------|------------|------------|------------|------------|---------------------|
| - | DM40 | DM20 | DM10 | DM8 | DM4 | DM2 | DM1 | (For Writing) |
| 0 | DM40 | DM20 | DM10 | DM8 | DM4 | DM2 | DM1 | (For Reading) |
| 0 | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Default Settings *) |

Alarm_D Hour Register (Address Ch)

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | |
|----|----|-----------------------|------------|------------|------------|------------|------------|---------------------|
| - | - | DH20 DP/ \bar{A} | DH10 | DH8 | DH4 | DH2 | DH1 | (For Writing) |
| 0 | 0 | DH20 DP/ \bar{A} | DH10 | DH8 | DH4 | DH2 | DH1 | (For Reading) |
| 0 | 0 | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Indefinite | Default Settings *) |

*) Default settings: Default value means read / written values when the PON bit is set to "1" due to VDD power-on from 0 volts.

* The D5 bit represents DP/ \bar{A} when the 12-hour mode is selected (0 for a.m. and 1 for p.m.) and DH20 when the 24-hour mode is selected (tens in the hour digits).

* The Alarm_D registers should not have any non-existent alarm time settings.

(Note that any mismatch between current time and preset alarm time specified by the Alarm_D registers may disable the alarm interrupt circuit.)

* When the 12-hour mode is selected, the hour digits read 12 and 32 for 0a.m. and 0p.m., respectively.

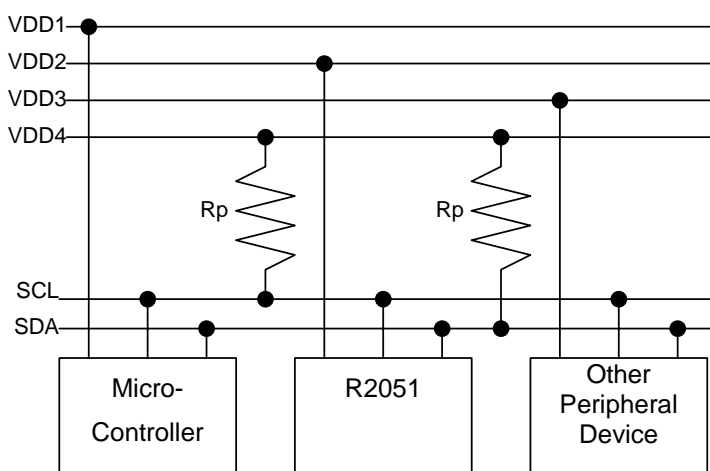
(See "P18 •Control Register 1 (Address Eh) (2) 12 /24: 12 /24-hour Mode Selection Bit")

Interfacing with the CPU

The R2051 employs the I²C-Bus system to be connected to the CPU via 2-wires. Connection and system of I²C-Bus are described in the following sections.

• Connection of I²C-Bus

2-wires, SCL and SDA pins that are connected to I²C-Bus are used for transmit clock pulses and data respectively. All ICs that are connected to these lines are designed that will not be clamped when a voltage beyond supply voltage is applied to input or output pins. Open drain pins are used for output. This construction allows communication of signals between ICs with different supply voltages by adding a pull-up resistor to each signal line as shown in the figure below. Each IC is designed not to affect SCL and SDA signal lines when power to each of these is turned off separately.



* For data interface, the following conditions must be met:

$$VCC4 \geq VCC1$$

$$VCC4 \geq VCC2$$

$$VCC4 \geq VCC3$$

* When the master is one, the micro-controller is ready for driving SCL to "H" and Rp of SCL may not be required.

Cautions on determining Rp resistance,

- (1) Dropping voltage at Rp due to sum of input current or output current at off conditions on each IC pin connected to the I²C-Bus shall be adequately small.
- (2) Rising time of each signal be kept short even when all capacity of the bus is driven.
- (3) Current consumed in I²C-Bus is small compared to the consumption current permitted for the entire system.

When all ICs connected to I²C-Bus are CMOS type, condition (1) may usually be ignored since input current and off-state output current is extremely small for the many CMOS type ICs. Thus the maximum resistance of Rp may be determined based on (2), while the minimum on (3) in most cases.

In actual cases a resistor may be placed between the bus and input/output pins of each IC to improve noise margins in which case the Rp minimum value may be determined by the resistance.

Consumption current in the bus to review (3) above may be expressed by the formula below:

Bus consumption current \approx

$$\frac{(\text{Sum of input current and off state output current of all devices in standby mode}) \times \text{Bus standby duration}}{\text{Bus stand-by duration} + \text{the Bus operation duration}}$$

$$+ \frac{\text{Supply voltage} \times \text{Bus operation duration} \times 2}{\text{Rp resistance} \times 2 \times (\text{Bus stand-by duration} + \text{bus operation duration})}$$

$$+ \text{Supply voltage} \times \text{Bus capacity} \times \text{Charging/Discharging times per unit time}$$

Operation of "× 2" in the second member denominator in the above formula is derived from assumption that "L"

duration of SDA and SCL pins are the half of bus operation duration. “× 2” in the numerator of the same member is because there are two pins of SDA and SCL. The third member, (charging/discharging times per unit time) means number of transition from “H” to “L” of the signal line.

Calculation example is shown below:

Pull-up resistor (Rp) = 10kΩ, Bus capacity = 50pF(both for SCL, SDA), V_{CC}=3V,

In a system with sum of input current and off-state output current of each pin = 0.1μA,

I²C-Bus is used for 10ms every second while the rest of 990ms in the stand-by mode,

In this mode, number of transitions of the SCL pin from “H” to “L” state is 100 while SDA 50, every second.

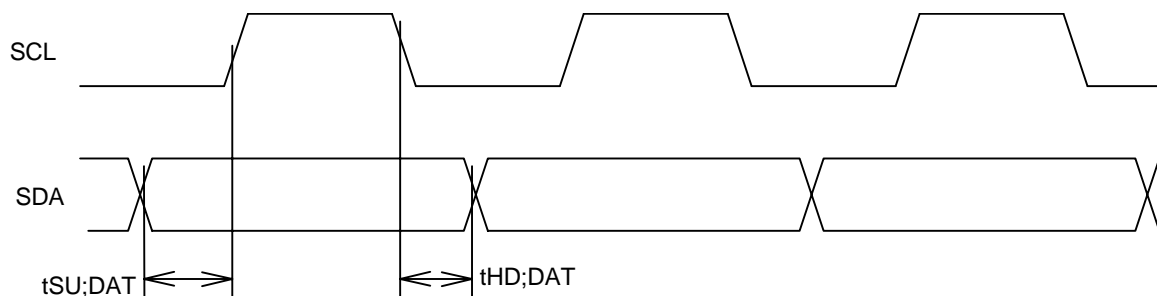
$$\begin{aligned}\text{Bus consumption current} &\approx \frac{0.1\mu\text{A} \times 990\text{msec}}{990\text{msec} + 10\text{msec}} \\ &+ \frac{3\text{V} \times 10\text{msec} \times 2}{10\text{k}\Omega \times 2 \times (990\text{msec} + 10\text{msec})} \\ &+ 3\text{V} \times 50\text{pF} \times (100 + 50) \\ &\approx 0.099\mu\text{A} + 3.0\mu\text{A} + 0.0225\mu\text{A} \approx 3.12\mu\text{A}\end{aligned}$$

Generally, the second member of the above formula is larger enough than the first and the third members bus consumption current may be determined by the second member in many cases.

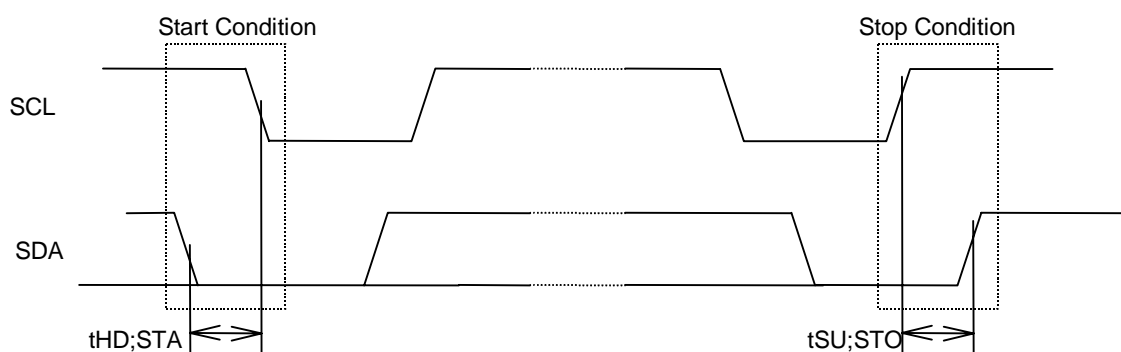
• Transmission System of I²C-Bus

(1) Start Condition and Stop Condition

In I²C-Bus, SDA must be kept at a certain state while SCL is at the “H” state during data transmission as shown below.

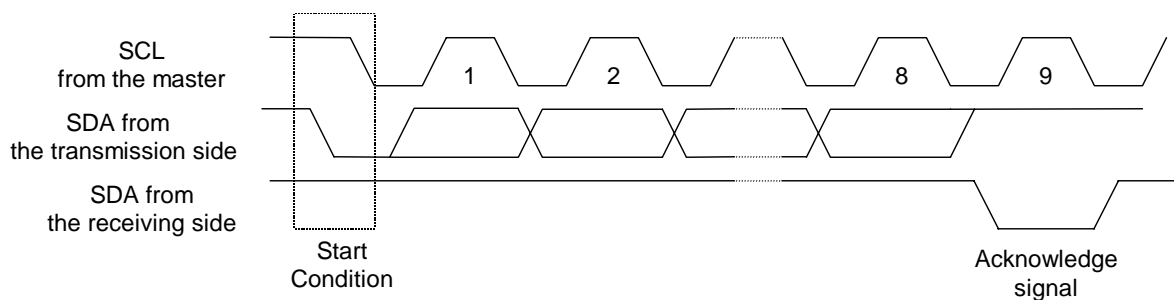


The SCL and SDA pins are at the “H” level when no data transmission is made. Changing the SDA from “H” to “L” when the SCL and the SDA are “H” activates the Start Condition and access is started. Changing the SDA from “L” to “H” when the SCL is “H” activates Stop Condition and accessing stopped. Generation of Start and Stop Conditions are always made by the master (see the figure below).



• (2) Data transmission and its acknowledge

After Start condition is entered, data is transmitted by 1byte (8bits). Any bytes of data may be serially transmitted. The receiving side will send an acknowledge signal to the transmission side each time 8bit data is transmitted. The acknowledge signal is sent immediately after falling to “L” of SCL 8bit clock pulses of data is transmitted, by releasing the SDA by the transmission side that has asserted the bus at that time and by turning SDA to “L” by receiving side. When transmission of 1byte data next to preceding 1byte of data is received the receiving side releases the SDA pin at falling edge of the SCL 9bit of clock pulses or when the receiving side switches to the transmission side it starts data transmission. When the master is receiving side, it generates no acknowledge signal after last 1byte of data from the slave to tell the transmitter that data transmission has completed. The slave side (transmission side) continues to release the SDA pin so that the master will be able to generate Stop Condition, after falling edge of the SCL 9bit of clock pulses.



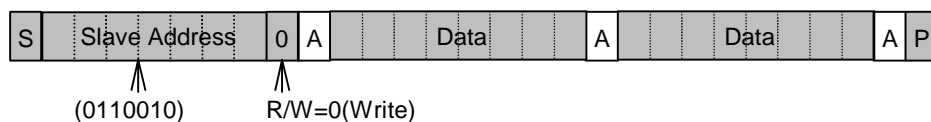
(3) Data Transmission Format in I²C-Bus

I²C-Bus has no chip enable signal line. In place of it, each device has a 7bit Slave Address allocated. The first 1byte is allocated to this 7bit address and to the command (R/W) for which data transmission direction is designated by the data transmission thereafter. 7bit address is sequentially transmitted from the MSB and 2 and after bytes are read, when 8bit is “H” and when write “L”.

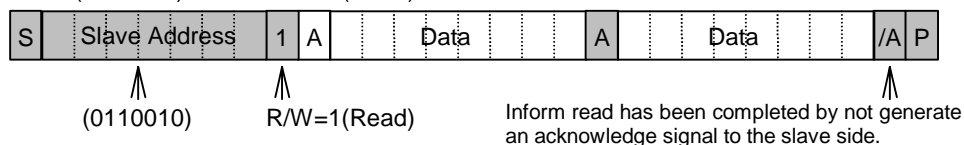
The Slave Address of the R2051 is specified at (0110010).

At the end of data transmission / receiving, Stop Condition is generated to complete transmission. However, if start condition is generated without generating Stop Condition, Repeated Start Condition is met and transmission / receiving data may be continue by setting the Slave Address again. Use this procedure when the transmission direction needs to be change during one transmission.

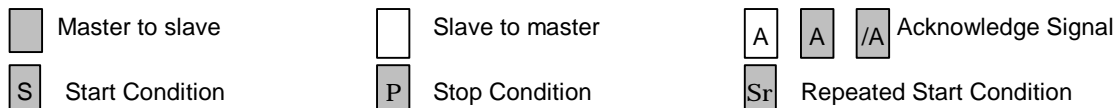
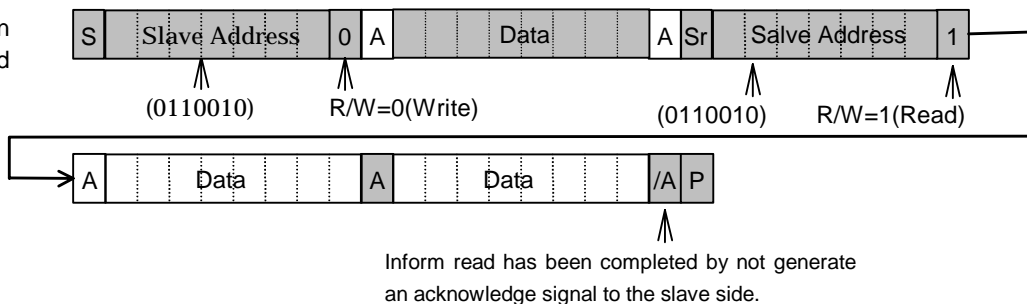
Data is written to the slave from the master



When data is read from the slave immediately after 7bit addressing from the master



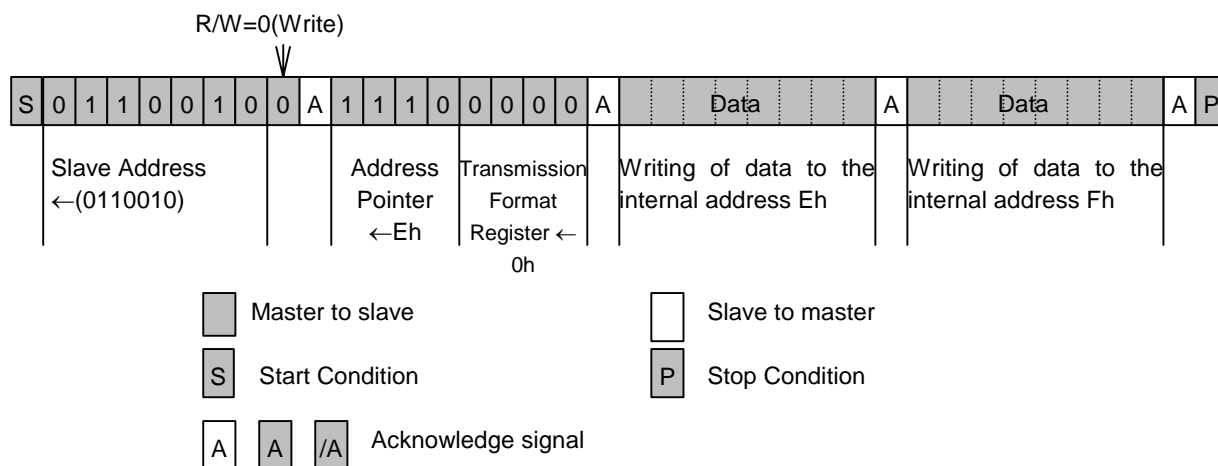
When the transmission direction is to be changed during transmission.



(4) Data Transmission Write Format in the R2051

Although the I²C-Bus standard defines a transmission format for the slave allocated for each IC, transmission method of address information in IC is not defined. The R2051 transmits data the internal address pointer (4bit) and the Transmission Format Register (4bit) at the 1byte next to one which transmitted a Slave Address and a write command. For write operation only one transmission format is available and (0000) is set to the Transmission Format Register. The 3byte transmits data to the address specified by the internal address pointer written to the 2byte. Internal address pointer setting are automatically incremented for 4byte and after. Note that when the internal address pointer is Fh, it will change to 0h on transmitting the next byte.

Example of data writing (When writing to internal address Eh to Fh)

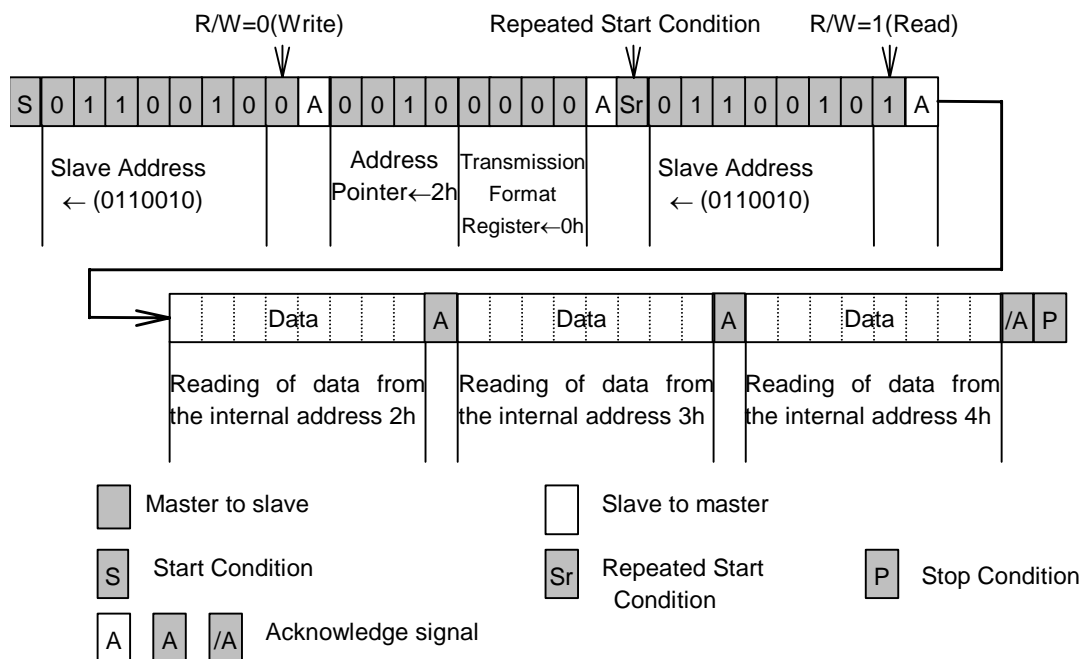


(5) Data transmission read format of the R2051

The R2051 allows the following three read out method of data an internal register.

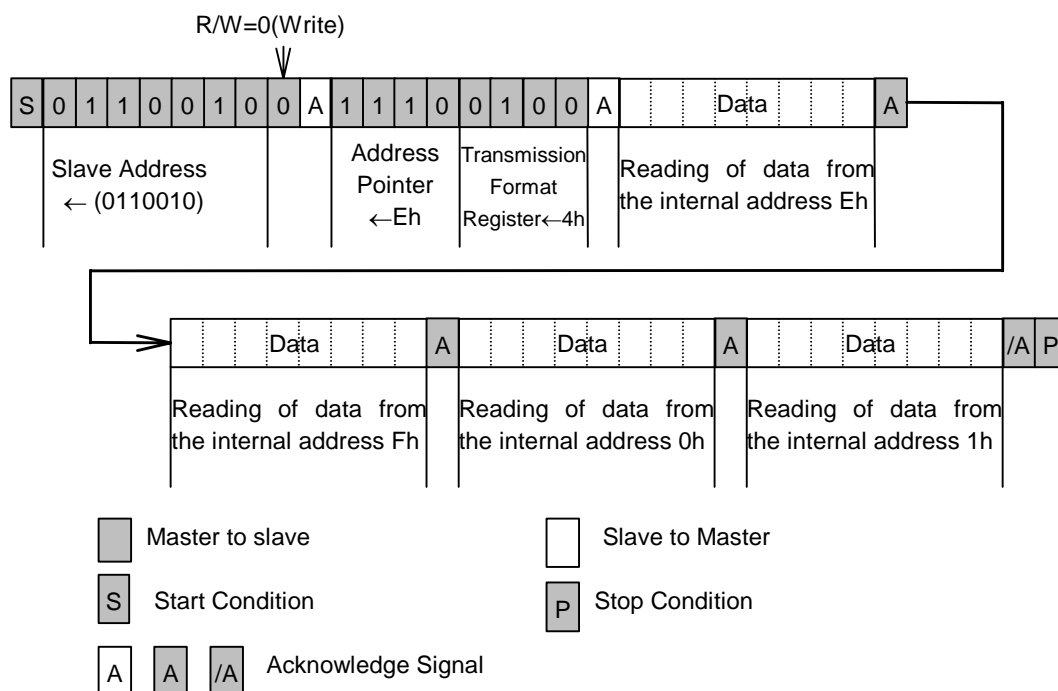
The first method to reading data from the internal register is to specify an internal address by setting the internal address pointer and the transmission format register described P31 (4), generate the Repeated Start Condition (See P30 (3)) to change the data transmission direction to perform reading. The internal address pointer is set to Fh when the Stop Condition is met. Therefore, this method of reading allows no insertion of Stop Condition before the Repeated Start Condition. Set 0h to the Transmission Format Register when this method used.

Example 1 of Data Read (when data is read from 2h to 4h)



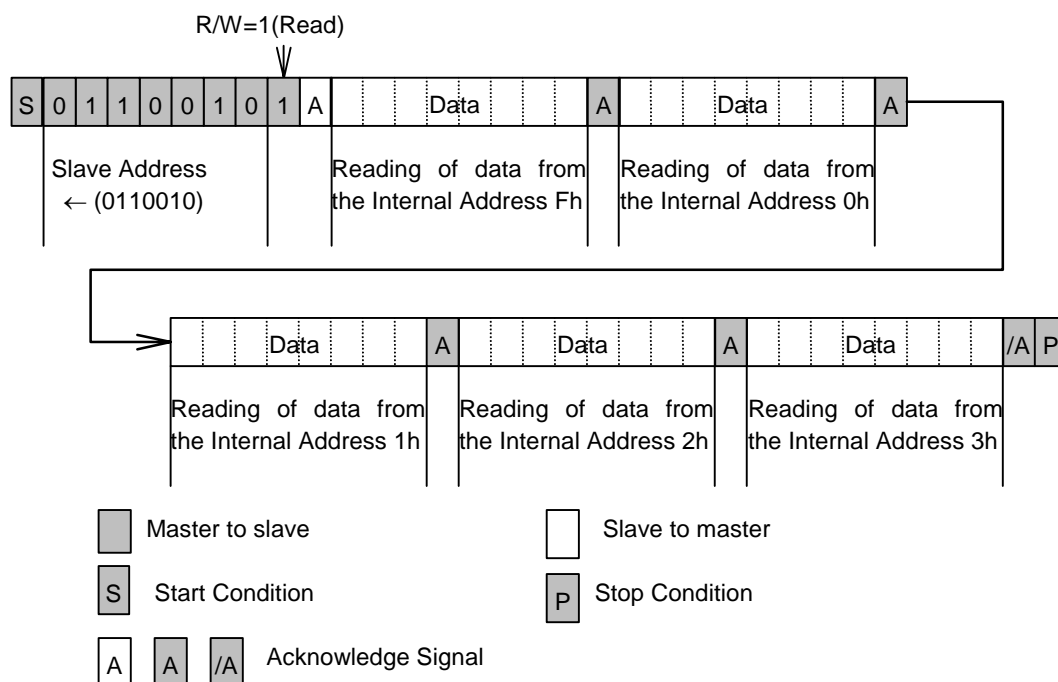
The second method to reading data from the internal register is to start reading immediately after writing to the Internal Address Pointer and the Transmission Format Register. Although this method is not based on I²C-Bus standard in a strict sense it still effective to shorten read time to ease load to the master. Set 4h to the transmission format register when this method used.

Example 2 of data read (when data is read from internal addresses Eh to 1h)



The third method to reading data from the internal register is to start reading immediately after writing to the Slave Address and R/W bit. Since the Internal Address Pointer is set to Fh by default as described in the first method, this method is only effective when reading is started from the Internal Address Fh.

Example 3 of data read (when data is read from internal addresses Fh to 3h)



• Data Transmission under Special Condition

The R2051 holds the clock tentatively for duration from Start Condition to avoid invalid read or write clock on carrying clock. When clock carried during this period, which will be adjusted within approx. 61μs from Stop Condition. To prevent invalid read or write, clock and calendar data shall be made during one transmission operation (from Start Condition to Stop Condition). When 0.5 to 1.0 second elapses after Start Condition, any access to the R2051 is automatically released to release tentative hold of the clock, and access from the CPU is forced to be terminated (The same action as made Stop Condition is received: automatic resume function from I²C-Bus interface). Therefore, one access must be complete within 0.5 seconds. The automatic resume function prevents delay in clock even if SCL is stopped from sudden failure of the system during clock read operation.

Also a second Start Condition after the first Start Condition and before the Stop Condition is regarded "Repeated Start Condition". Therefore, when 0.5 to 1.0 seconds passed after the first Start Condition, an access to the R2051 is automatically released.

If access is tried after automatic resume function is activated, no acknowledge signal will be output for writing while FFh will be output for reading.

The user shall always be able to access the real-time clock as long as three conditions are met.

No Stop Condition shall be generated until clock and calendar data read/write is started and completed.

One cycle read/write operation shall be complete within 0.5 seconds.

Do not make Start Condition within 61μs from Stop Condition. When clock is carried during the access, which will be adjusted within approx. 61μs from Stop Condition.

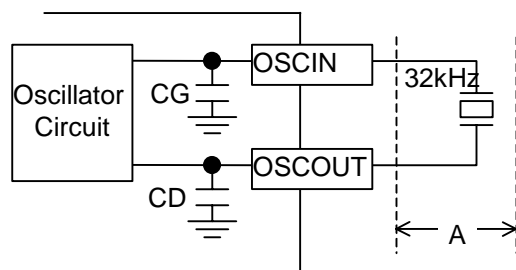
Bad example of reading from seconds to hours (invalid read)

(Start Condition) → (Read of seconds) → (Read of minutes) → (Stop Condition) → (Start Condition) → (Read of hour) → (Stop Condition)

Assuming read was started at 05:59:59 P.M. and while reading seconds and minutes the time advanced to 06:00:00 P.M. At this time second digit is hold so read the read as 05:59:59. Then the R2051 confirms (Stop Condition) and carries second digit being hold and the time change to 06:00:00 P.M. Then, when the hour digit is read, it changes to 6. The wrong results of 06:59:59 will be read.

Configuration of Oscillation Circuit and Correction of Time Count Deviations

• Configuration of Oscillation Circuit



Typical externally-equipped element

X'tal : 32.768kHz

(R1=30kΩ typ)

(CL=6pF to 8pF)

Standard values of internal elements

CG,CD 10pF typ

The oscillation circuit is driven at a constant voltage of approximately 1.2 volts relative to the level of the VSS pin input. As such, it is configured to generate an oscillating waveform with a peak-to-peak voltage on the order of 1.1 volts on the positive side of the VSS pin input.

< Considerations in Handling quartz crystal unit >

Generally, quartz crystal units have basic characteristics including an equivalent series resistance (R1) indicating the ease of their oscillation and a load capacitance (CL) indicating the degree of their center frequency. Particularly, quartz crystal units intended for use in the R2051 are recommended to have a typical R1 value of 30kΩ and a typical CL value of 6 to 8pF. To confirm these recommended values, contact the manufacturers of quartz crystal units intended for use in these particular models.

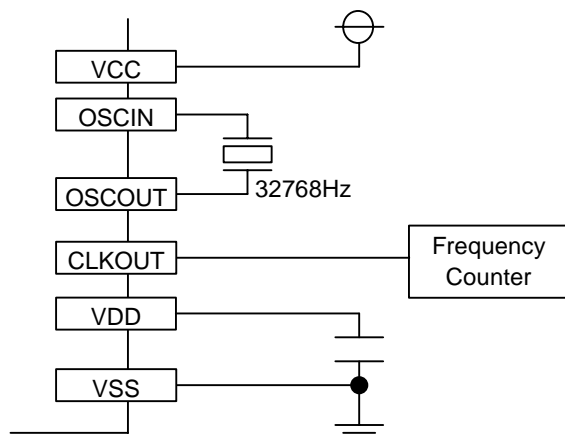
< Considerations in Installing Components around the Oscillation Circuit >

- 1) Install the quartz crystal unit in the closest possible vicinity to the real-time clock ICs.
- 2) Avoid laying any signal lines or power lines in the vicinity of the oscillation circuit (particularly in the area marked "A" in the above figure).
- 3) Apply the highest possible insulation resistance between the OSCIN and OSCOUT pins and the printed circuit board.
- 4) Avoid using any long parallel lines to wire the OSCIN and OSCOUT pins.
- 5) Take extreme care not to cause condensation, which leads to various problems such as oscillation halt.

< Other Relevant Considerations >

- 1) We cannot recommend connecting the external input of 32.768-kHz clock pulses to the OSCIN pin.
- 2) To maintain stable characteristics of the quartz crystal unit, avoid driving any other IC through 32.768-kHz clock pulses output from the OSCOUT pin.

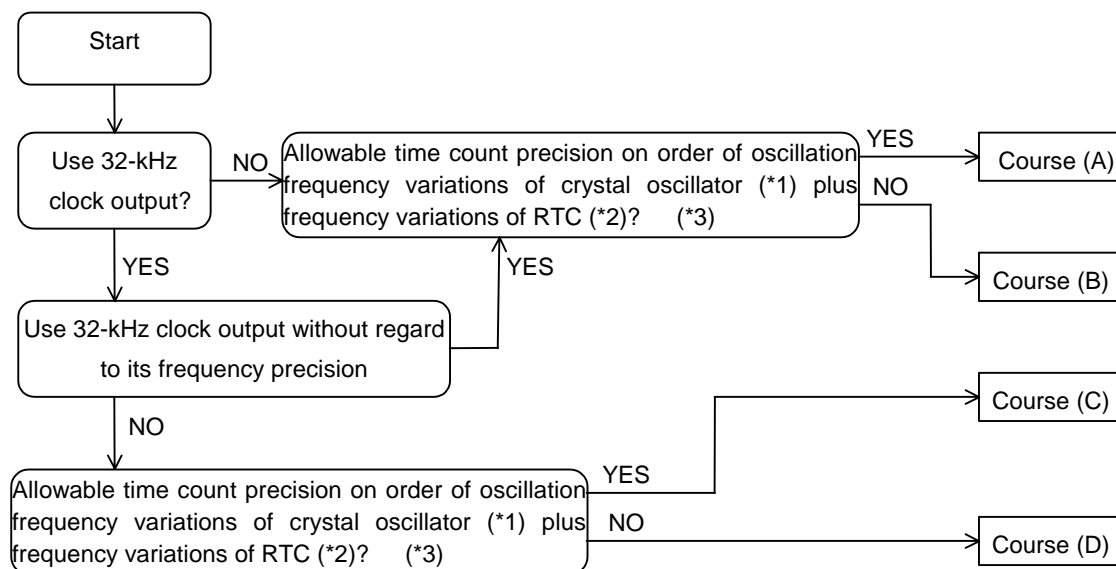
• Measurement of Oscillation Frequency



- * 1) The R2051 is configured to generate 32.768-kHz clock pulses for output from the CLKOUT pin.
- * 2) A frequency counter with 6 (more preferably 7) or more digits on the order of 1ppm is recommended for use in the measurement of the oscillation frequency of the oscillation circuit.

• Adjustment of Oscillation frequency

The oscillation frequency of the oscillation circuit can be adjusted by varying procedures depending on the usage of Model R2051 in the system into which they are to be built and on the allowable degree of time count errors. The flow chart below serves as a guide to selecting an optimum oscillation frequency adjustment procedure for the relevant system.



* 1) Generally, quartz crystal units for commercial use are classified in terms of their center frequency depending on their load capacitance (CL) and further divided into ranks on the order of ± 10 , ± 20 , and ± 50 ppm depending on the degree of their oscillation frequency variations.

* 2) Basically, Model R2051 is configured to cause frequency variations on the order of ± 5 to ± 10 ppm at 25°C.

* 3) Time count precision as referred to in the above flow chart is applicable to normal temperature and actually affected by the temperature characteristics and other properties of quartz crystal units.

Course (A)

When the time count precision of each RTC is not to be adjusted, the quartz crystal unit intended for use in that RTC may have any CL value requiring no presetting. The quartz crystal unit may be subject to frequency variations which are selectable within the allowable range of time count precision. Several quartz crystal units and RTCs should be used to find the center frequency of the quartz crystal units by the method described in "P36 • Measurement of Oscillation Frequency" and then calculate an appropriate oscillation adjustment value by the method described in "P38 • Oscillation Adjustment Circuit" for writing this value to the R2051.

Course (B)

When the time count precision of each RTC is to be adjusted within the oscillation frequency variations of the quartz crystal unit plus the frequency variations of the real-time clock ICs, it becomes necessary to correct deviations in the time count of each RTC by the method described in " P38 • Oscillation Adjustment Circuit". Such oscillation adjustment provides quartz crystal units with a wider range of allowable settings of their oscillation frequency variations and their CL values. The real-time clock IC and the quartz crystal unit intended for use in that real-time clock IC should be used to find the center frequency of the quartz crystal unit by the method described in " P36 • Measurement of Oscillation Frequency" and then confirm the center frequency thus found to fall within the range adjustable by the oscillation adjustment circuit before adjusting the oscillation frequency of the oscillation circuit. At normal temperature, the oscillation frequency of the oscillator circuit can be adjusted by up to approximately $\pm 0.5\text{ppm}$.

Course (C)

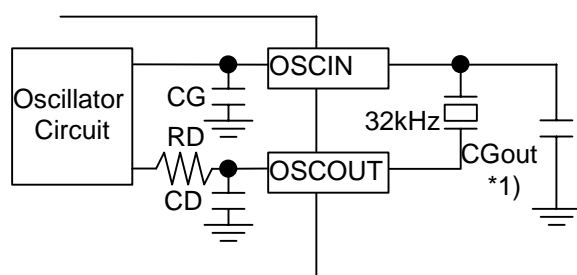
Course (C) together with Course (D) requires adjusting the time count precision of each RTC as well as the frequency of 32.768-kHz clock pulses output from the CLKOUT pin. Normally, the oscillation frequency of the quartz crystal unit intended for use in the RTCs should be adjusted by adjusting the oscillation stabilizing capacitors CG and CD connected to both ends of the quartz crystal unit. The R2051, which incorporate the CG and the CD, require adjusting the oscillation frequency of the quartz crystal unit through its CL value.

Generally, the relationship between the CL value and the CG and CD values can be represented by the following equation:

$$CL = (CG \times CD) / (CG + CD) + CS$$
 where "CS" represents the floating capacity of the printed circuit board.

The quartz crystal unit intended for use in the R2051 is recommended to have the CL value on the order of 6 to 8pF. Its oscillation frequency should be measured by the method described in " P36 • Measurement of Oscillation Frequency". Any quartz crystal unit found to have an excessively high or low oscillation frequency (causing a time count gain or loss, respectively) should be replaced with another one having a smaller and greater CL value, respectively until another one having an optimum CL value is selected. In this case, the bit settings disabling the oscillation adjustment circuit (see " P38 • Oscillation Adjustment Circuit ") should be written to the oscillation adjustment register.

Incidentally, the high oscillation frequency of the quartz crystal unit can also be adjusted by adding an external oscillation stabilization capacitor CGout as illustrated in the diagram below.



*1) The CGout should have a capacitance ranging from 0 to 15 pF.

Course (D)

It is necessary to select the quartz crystal unit in the same manner as in Course (C) as well as correct errors in the time count of each RTC in the same manner as in Course (B) by the method described in " P38 • Oscillation Adjustment Circuit ".

• Oscillation Adjustment Circuit

The oscillation adjustment circuit can be used to correct a time count gain or loss with high precision by varying the number of 1-second clock pulses once per 20 seconds or 60 seconds. When DEV bit in the Oscillation Adjustment Register is set to 0, R2051 varies number of 1-second clock pulses once per 20 seconds. When DEV bit is set to 1, R2051 varies number of 1-second clock pulses once per 60 seconds. The oscillation adjustment circuit can be disabled by writing the settings of "*", 0, 0, 0, 0, 0, 0, "*" ("*" representing "0" or "1") to the F6, F5, F4, F3, F2, F1, and F0 bits in the oscillation adjustment circuit. Conversely, when such oscillation adjustment is to be made, an appropriate oscillation adjustment value can be calculated by the equation below for writing to the oscillation adjustment circuit.

(1) When Oscillation Frequency (* 1) Is Higher Than Target Frequency (* 2) (Causing Time Count Gain)

When DEV=0:

$$\begin{aligned}\text{Oscillation adjustment value (*3)} &= \frac{(\text{Oscillation frequency} - \text{Target Frequency} + 0.1)}{\text{Oscillation frequency} \times 3.051 \times 10^{-6}} \\ &\approx (\text{Oscillation Frequency} - \text{Target Frequency}) \times 10 + 1\end{aligned}$$

When DEV=1:

$$\begin{aligned}\text{Oscillation adjustment value (*3)} &= \frac{(\text{Oscillation frequency} - \text{Target Frequency} + 0.0333)}{\text{Oscillation frequency} \times 1.017 \times 10^{-6}} \\ &\approx (\text{Oscillation Frequency} - \text{Target Frequency}) \times 30 + 1\end{aligned}$$

* 1) Oscillation frequency:

Frequency of clock pulse output from the CLKOUT pin at normal temperature in the manner described in " P36 • Measurement of Oscillation Frequency".

* 2) Target frequency:

Desired frequency to be set. Generally, a 32.768-kHz quartz crystal unit has such temperature characteristics as to have the highest oscillation frequency at normal temperature. Consequently, the quartz crystal unit is recommended to have target frequency settings on the order of 32.768 to 32.76810 kHz (+3.05ppm relative to 32.768 kHz). Note that the target frequency differs depending on the environment or location where the equipment incorporating the RTC is expected to be operated.

* 3) Oscillation adjustment value:

Value that is to be finally written to the F0 to F6 bits in the Oscillation Adjustment Register and is represented in 7-bit coded decimal notation.

(2) When Oscillation Frequency Is Equal To Target Frequency (Causing Time Count neither Gain nor Loss)

Oscillation adjustment value = 0, +1, -64, or -63

(3) When Oscillation Frequency Is Lower Than Target Frequency (Causing Time Count Loss)

When DEV=0:

$$\text{Oscillation adjustment value} = \frac{(\text{Oscillation frequency} - \text{Target Frequency})}{\text{Oscillation frequency} \times 3.051 \times 10^{-6}}$$

$$\begin{aligned} & \text{Oscillation frequency} \times 3.051 \times 10^{-6} \\ & \approx (\text{Oscillation Frequency} - \text{Target Frequency}) \times 10 \end{aligned}$$

When DEV=1:

$$\begin{aligned} \text{Oscillation adjustment value} &= \frac{(\text{Oscillation frequency} - \text{Target Frequency})}{\text{Oscillation frequency} \times 1.017 \times 10^{-6}} \\ &\approx (\text{Oscillation Frequency} - \text{Target Frequency}) \times 30 \end{aligned}$$

Oscillation adjustment value calculations are exemplified below

(A) For an oscillation frequency = 32768.85Hz and a target frequency = 32768.05Hz

When setting DEV bit to 0:

$$\begin{aligned} \text{Oscillation adjustment value} &= (32768.85 - 32768.05 + 0.1) / (32768.85 \times 3.051 \times 10^{-6}) \\ &\approx (32768.85 - 32768.05) \times 10 + 1 \\ &= 9.001 \approx 9 \end{aligned}$$

In this instance, write the settings (DEV,F6,F5,F4,F3,F2,F1,F0)=(0,0,0,0,1,0,0,1) in the oscillation adjustment register. Thus, an appropriate oscillation adjustment value in the presence of any time count gain represents a distance from 01h.

When setting DEV bit to 1:

$$\begin{aligned} \text{Oscillation adjustment value} &= (32768.85 - 32768.05 + 0.0333) / (32768.85 \times 1.017 \times 10^{-6}) \\ &\approx (32768.85 - 32768.05) \times 30 + 1 \\ &= 25.00 \approx 25 \end{aligned}$$

In this instance, write the settings (DEV,F6,F5,F4,F3,F2,F1,F0)=(1,0,0,1,1,0,0,1) in the oscillation adjustment register.

(B) For an oscillation frequency = 32762.22Hz and a target frequency = 32768.05Hz

When setting DEV bit to 0:

$$\begin{aligned} \text{Oscillation adjustment value} &= (32762.22 - 32768.05) / (32762.22 \times 3.051 \times 10^{-6}) \\ &\approx (32762.22 - 32768.05) \times 10 \\ &= -58.325 \approx -58 \end{aligned}$$

To represent an oscillation adjustment value of - 58 in 7-bit coded decimal notation, subtract 58 (3Ah) from 128 (80h) to obtain 46h. In this instance, write the settings of (DEV,F6,F5,F4,F3,F2,F1,F0) = (0,1,0,0,0,1,1,0) in the oscillation adjustment register. Thus, an appropriate oscillation adjustment value in the presence of any time count loss represents a distance from 80h.

When setting DEV bit to 1:

$$\begin{aligned} \text{Oscillation adjustment value} &= (32762.22 - 32768.05) / (32762.22 \times 1.017 \times 10^{-6}) \\ &\approx (32762.22 - 32768.05) \times 30 \\ &= -174.97 \approx -175 \end{aligned}$$

Oscillation adjustment value can be set from -62 to 63. Then, in this case, Oscillation adjustment value is out of range.

(4) Difference between DEV=0 and DEV=1

Difference between DEV=0 and DEV=1 is following,

| | DEV=0 | DEV=1 |
|---------------------|------------------------|------------------|
| Maximum value range | -189.2ppm to +189.2ppm | -62ppm to +63ppm |
| Minimum resolution | 3ppm | 1ppm |

Notes:

1) Oscillation adjustment circuit does not affect the frequency of 32.768-kHz clock pulses output from the CLKOUT pin.

2) If following 3 conditions are completed, actual clock adjustment value could be different from target adjustment value that set by oscillator adjustment function.

1. Using oscillator adjustment function

2. Access to R2051 at random, or synchronized with external clock that has no relation to R2051, or synchronized with periodic interrupt in pulse mode.

3. Access to R2051 more than 2 times per each second on average.

For more details, please contact to Ricoh.

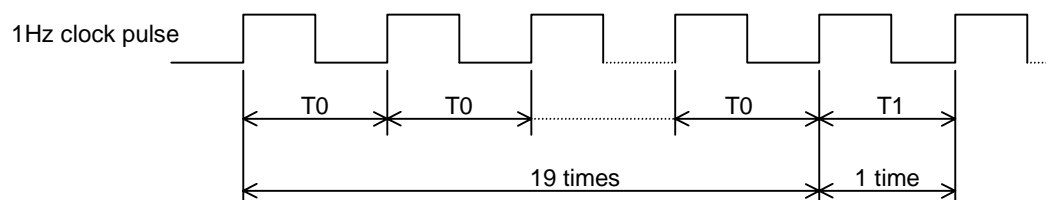
● How to evaluate the clock gain or loss

The oscillator adjustment circuit is configured to change time counts of 1 second on the basis of the settings of the oscillation adjustment register once in 20 seconds or 60 seconds. The oscillation adjustment circuit does not effect the frequency of 32768Hz-clock pulse output from the CLKOUT pin. Therefore, after writing the oscillation adjustment register, we cannot measure the clock error with probing CLKOUT clock pulses. The way to measure the clock error as follows (except R2051Txx):

(1) Output a 1Hz clock pulse of Pulse Mode with interrupt pin

Set (0,0,x,x,0,0,1,1) to Control Register 1 at address Eh.

(2) After setting the oscillation adjustment register, 1Hz clock period changes every 20seconds (or every 60 seconds) like next page figure.



Measure the interval of T0 and T1 with frequency counter. A frequency counter with 7 or more digits is recommended for the measurement.

(3) Calculate the typical period from T0 and T1

$$T = (19 \times T0 + 1 \times T1) / 20$$

Calculate the time error from T.

Power-on Reset, Oscillation Halt Sensing, and Supply Voltage Monitoring

• PON, \overline{XST} , and VDET

The power-on reset circuit is configured to reset control register1, 2, and clock adjustment register when VDD power up from 0v. The oscillation halt sensing circuit is configured to record a halt on oscillation by 32.768-kHz clock pulses. The supply voltage monitoring circuit is configured to record a drop in supply voltage below a threshold voltage of 2.1 or 1.35v.

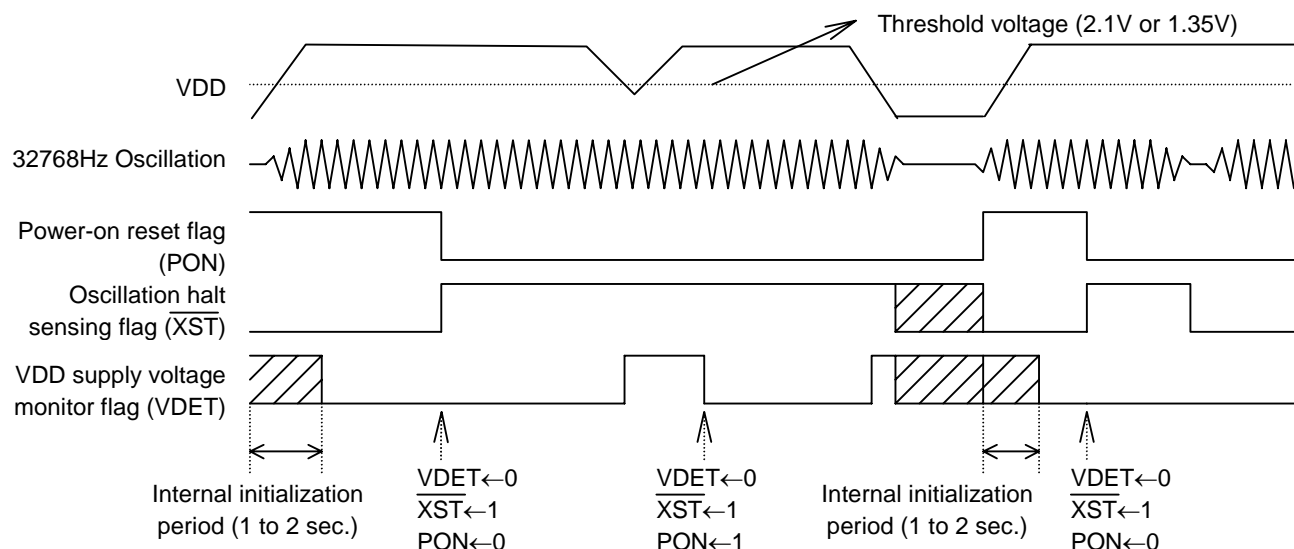
Each function has a monitor bit. I.e. the PON bit is for the power-on reset circuit, and \overline{XST} bit is for the oscillation halt sensing circuit, and VDET is for the supply voltage monitoring circuit. PON and VDET bits are activated to "H". However, \overline{XST} bit is activated to "L". The PON and VDET accept only the writing of 0, but \overline{XST} accepts the writing of 0 and 1. The PON bit is set to 1, when VDD power-up from 0V, but VDET is set to 0, and \overline{XST} is indefinite.

The functions of these three monitor bits are shown in the table below.

| | PON | \overline{XST} | VDET |
|---------------------------|--|--|--|
| Function | Monitoring for the power-on reset function | Monitoring for the oscillation halt sensing function | a drop in supply voltage below a threshold voltage of 2.1 or 1.35v |
| Address | D4 in Address Fh | D5 in Address Fh | D6 in Address Fh |
| Activated | High | Low | High |
| When VDD power up from 0v | 1 | Indefinite | 0 |
| accept the writing | 0 only | Both 0 and 1 | 0 only |

The relationship between the PON, \overline{XST} , and VDET is shown in the table below.

| PON | \overline{XST} | VDET | Conditions of supply voltage and oscillation | Condition of oscillator, and back-up status |
|-----|------------------|------|---|--|
| 0 | 0 | 0 | Halt on oscillation, but no drop in VDD supply voltage below threshold voltage | Halt on oscillation cause of condensation etc. |
| 0 | 0 | 1 | Halt on oscillation and drop in VDD supply voltage below threshold voltage, but no drop to 0V | Halt on oscillation cause of drop in back-up battery voltage |
| 0 | 1 | 0 | No drop in VDD supply voltage below threshold voltage and no halt in oscillation | Normal condition |
| 0 | 1 | 1 | Drop in VDD supply voltage below threshold voltage and no halt on oscillation | No halt on oscillation, but drop in back-up battery voltage |
| 1 | * | * | Drop in supply voltage to 0v | Power-up from 0v, |



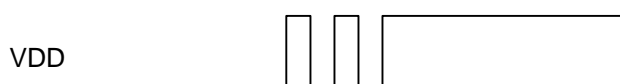
When the PON bit is set to 1 in the control register 2, the DEV, F6 to F0, WALE, DALE, $\overline{12}/24$, SCRATCH2, TEST, CT2, CT1, CT0, VDSL, VDET, SCRATCH1, CTFG, WAFG, and DAFG bits are reset to 0 in the oscillation adjustment register, the control register 1, and the control register 2. The PON bit is also set to 1 at power-on from 0 volts.

< Considerations in Using Oscillation Halt Sensing Circuit >

Be sure to prevent the oscillation halt sensing circuit from malfunctioning by preventing the following:

- 1) Instantaneous power-down on the VDD
- 2) Condensation on the quartz crystal unit
- 3) On-board noise to the quartz crystal unit
- 4) Applying to individual pins voltage exceeding their respective maximum ratings

In particular, note that the \overline{XST} bit may fail to be set to 0 in the presence of any applied supply voltage as illustrated below in such events as backup battery installation. Further, give special considerations to prevent excessive chattering in the oscillation halt sensing circuit.

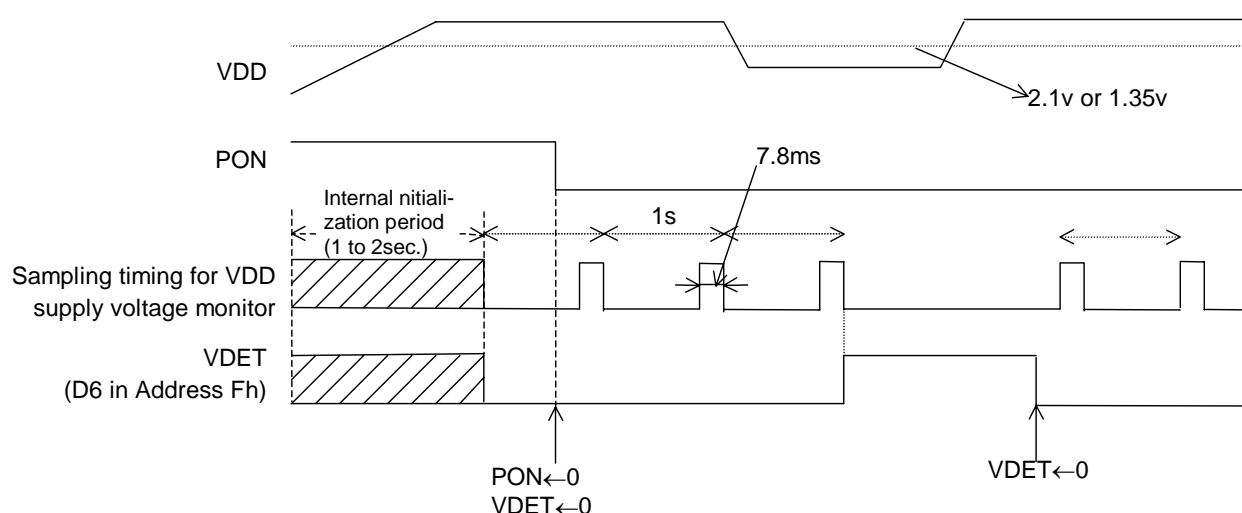


• Voltage Monitoring Circuit

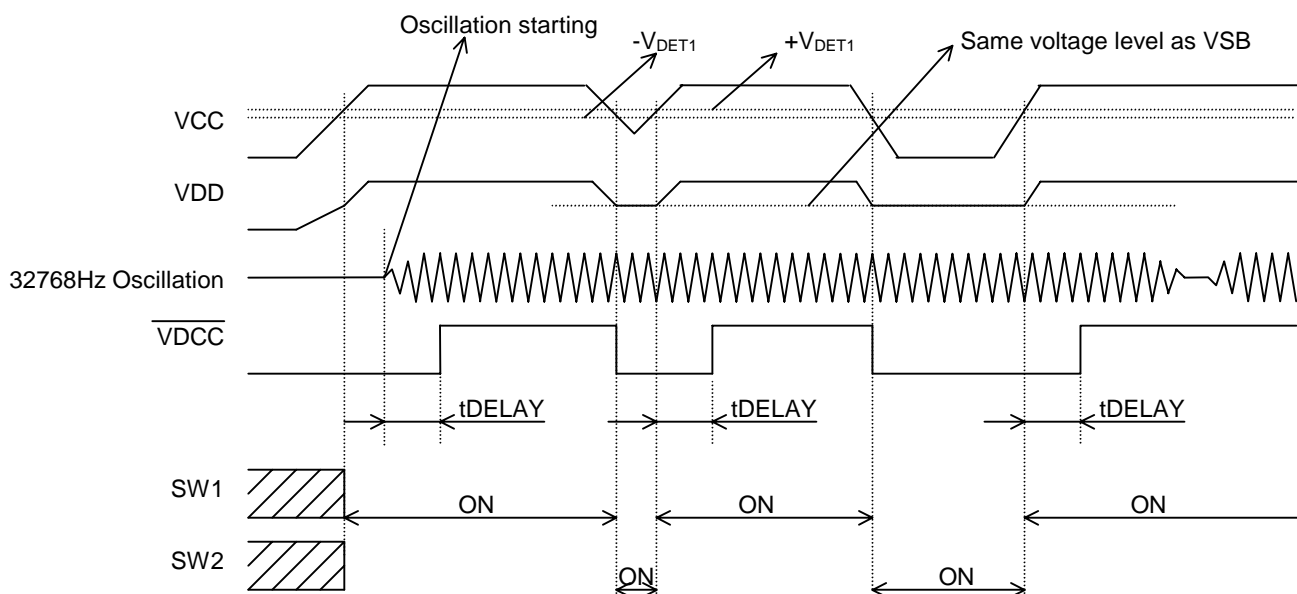
R2051S/Kxx incorporates two kinds of voltage monitoring function. (R2051Txx incorporates one kind only.) These are shown in the table below.

| | VCC Voltage Monitoring Circuit (except R2051Txx) | VDD Voltage Monitoring Circuit (VDET) |
|---|---|---|
| Purpose | CPU reset output | Back-up battery checker |
| Monitoring supply voltage | VCC pin | VDD pin (supply voltage for the internal RTC circuit) |
| Output for result | $\overline{\text{VDCC}}$ pin | Store in the Control Register 2 (D6 in Address Fh) |
| Function | After falling VCC, $\overline{\text{VDCC}}$ outputs "L". $\overline{\text{VDCC}}$ outputs "H" after rising VCC, $\overline{\text{VDCC}}$ outputs "H" (OFF) Below the threshold voltage, SW1 turns off and SW2 turns on. Over the threshold voltage, SW1 turns on and SW2 turns off. | |
| Detector Threshold (falling edge of power supply voltage) | $-V_{\text{DET1}}$ | Selecting from V_{DETH} or V_{DETL} by writing to the register (D7 in Address Fh) |
| Detector Released Voltage (rising edge of power supply voltage) | $+V_{\text{DET1}}$ | Same as falling edge (No hysteresis) |
| The way to monitor | Always | One time every second |

The VDD supply voltage monitoring circuit is configured to conduct a sampling operation during an interval of 7.8ms per second to check for a drop in supply voltage below a threshold voltage of 2.1 or 1.35v for the VDSL bit setting of 0 (the default setting) or 1, respectively, in the Control Register 2, thus minimizing supply current requirements as illustrated in the timing chart below. This circuit suspends a sampling operation once the VDET bit is set to 1 in the Control Register 2. The VDD supply voltage monitor is useful for back-up battery checking.

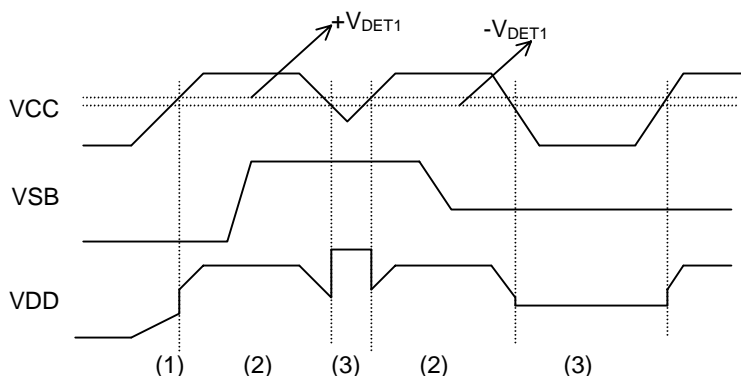


The VCC supply voltage monitor circuit operates always. When VCC rising over $+V_{DET1}$, SW1 turns on, and SW2 turns off. And t_{DELAY} after rising VCC, \overline{VDCC} outputs OFF(H). But when oscillation is halt, VCC outputs OFF(H) t_{DELAY} after oscillation starting. When VCC falling beyond $-V_{DET1}$, SW1 turns off, and SW2 turns on. And \overline{VDCC} outputs "L". R2051Tx does not have \overline{VDCC} output pin.



Battery Switch Over Circuit

R2051 incorporates three power supply pins, VDD, VCC, and VSB. VDD pin is the power supply pin for internal real time clock circuit. When VCC voltage is lower than $\pm V_{DET1}$, VSB supplies the power to VDD, and when higher than $\pm V_{DET1}$, VCC supplies the power to VDD. The timing chart for VCC, VDD, and VSB is shown following.



- (1) When VSB is 0v and VCC is rising from 0v, VDD follows half of VCC voltage level. After VCC rising over $+V_{DET1}$, VDD follows VCC voltage level.
- (2) When VCC is higher than $+V_{DET1}$, VDD level is equal to VCC.
- (3) After VCC falling beyond $-V_{DET1}$, VDD level is equal to VSB.

Alarm and Periodic Interrupt

The R2051 incorporates the alarm interrupt circuit and the periodic interrupt circuit that are configured to generate alarm signals and periodic interrupt signals for output from the $\overline{\text{INTR}}$ pin as described below. R2051Txx has these functions registers, but does not have the $\overline{\text{INTR}}$ output pin.

(1) Alarm Interrupt Circuit

The alarm interrupt circuit is configured to generate alarm signals for output from the $\overline{\text{INTR}}$, which is driven low (enabled) upon the occurrence of a match between current time read by the time counters (the day-of-week, hour, and minute counters) and alarm time preset by the alarm registers (the Alarm_W registers intended for the day-of-week, hour, and minute digit settings and the Alarm_D registers intended for the hour and minute digit settings).

(2) Periodic Interrupt Circuit

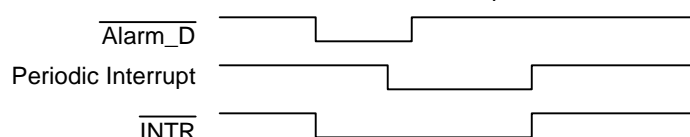
The periodic interrupt circuit is configured to generate either clock pulses in the pulse mode or interrupt signals in the level mode for output from the $\overline{\text{INTR}}$ pin depending on the CT2, CT1, and CT0 bit settings in the control register 1.

The above two types of interrupt signals are monitored by the flag bits (i.e. the WAFG, DAFG, and CTFG bits in the Control Register 2) and enabled or disabled by the enable bits (i.e. the WALE, DALE, CT2, CT1, and CT0 bits in the Control Register 1) as listed in the table below.

| | Flag bits | Enable bits |
|--------------------|----------------------------|--|
| Alarm_W | WAFG (D1 at Address Fh) | WALE (D7 at Address Eh) |
| Alarm_D | DAFG (D0 at Address Fh) | DALE (D6 at Address Eh) |
| Periodic interrupt | CTFG (D2 at Address Fh) | CT2=CT1=CT0=0 (These bit setting of "0" disable the Periodic Interrupt) (D2 to D0 at Address Eh) |

- * At power-on, when the WALE, DALE, CT2, CT1, and CT0 bits are set to 0 in the Control Register 1, the $\overline{\text{INTR}}$ pin is driven high (disabled).
- * When two types of interrupt signals are output simultaneously from the $\overline{\text{INTR}}$ pin, the output from the $\overline{\text{INTR}}$ pin becomes an OR waveform of their negative logic.

Example: Combined Output to $\overline{\text{INTR}}$ Pin Under Control of
ALARM_D and Periodic Interrupt



In this event, which type of interrupt signal is output from the $\overline{\text{INTR}}$ pin can be confirmed by reading the DAFG, and CTFG bit settings in the Control Register 2.

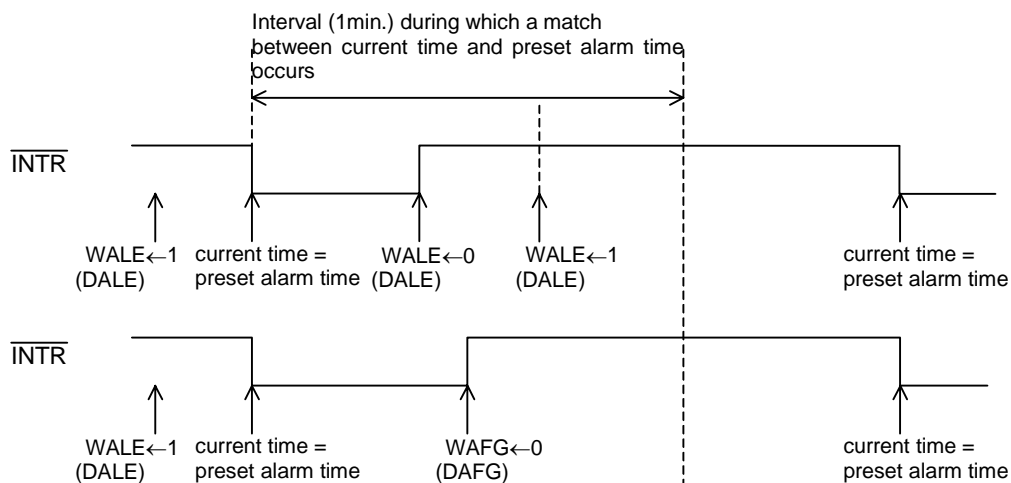
• Alarm Interrupt

The alarm interrupt circuit is controlled by the enable bits (i.e. the WALE and DALE bits in the Control Register 1) and the flag bits (i.e. the WAFG and DAFG bits in the Control Register 2). The enable bits can be used to enable this circuit when set to 1 and to disable it when set to 0. When intended for reading, the flag bits can be used to monitor alarm interrupt signals. When intended for writing, the flag bits will cause no event when set to 1 and will

drive high (disable) the alarm interrupt circuit when set to 0.

The enable bits will not be affected even when the flag bits are set to 0. In this event, therefore, the alarm interrupt circuit will continue to function until it is driven low (enabled) upon the next occurrence of a match between current time and preset alarm time.

The alarm function can be set by presetting desired alarm time in the alarm registers (the Alarm_W Registers for the day-of-week digit settings and both the Alarm_W Registers and the Alarm_D Registers for the hour and minute digit settings) with the WALE and DALE bits once set to 0 and then to 1 in the Control Register 1. Note that the WALE and DALE bits should be once set to 0 in order to disable the alarm interrupt circuit upon the coincidental occurrence of a match between current time and preset alarm time in the process of setting the alarm function.



After setting WALE(DALE) to 0, Alarm registers is set to current time, and WALE(DALE) is set to 1, $\overline{\text{INTR}}$ will be not driven to "L" immediately, $\overline{\text{INTR}}$ will be driven to "L" at next alarm setting time.

• Periodic Interrupt

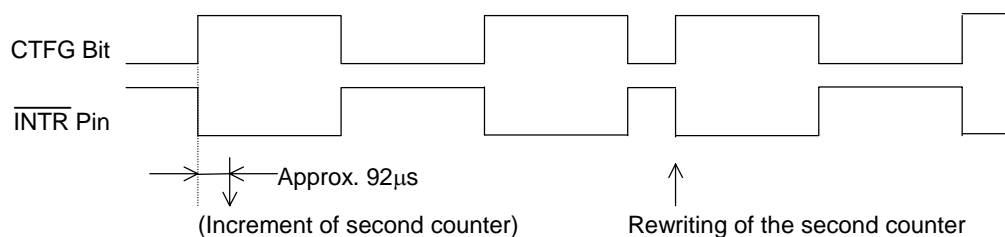
Setting of the periodic selection bits (CT2 to CT0) enables periodic interrupt to the CPU. There are two waveform modes: pulse mode and level mode. In the pulse mode, the output has a waveform duty cycle of around 50%. In the level mode, the output is cyclically driven low and, when the CTFG bit is set to 0, the output is return to High (OFF).

| CT2 | CT1 | CT0 | Description | |
|-----|-----|-----|----------------|--|
| | | | Wave form mode | Interrupt Cycle and Falling Timing |
| 0 | 0 | 0 | - | OFF(H) |
| 0 | 0 | 1 | - | Fixed at "L" |
| 0 | 1 | 0 | Pulse Mode *1) | 2Hz(Duty50%) |
| 0 | 1 | 1 | Pulse Mode *1) | 1Hz(Duty50%) |
| 1 | 0 | 0 | Level Mode *2) | Once per 1 second (Synchronized with Second counter increment) |
| 1 | 0 | 1 | Level Mode *2) | Once per 1 minute (at 00 seconds of every Minute) |
| 1 | 1 | 0 | Level Mode *2) | Once per hour (at 00 minutes and 00 Seconds of every hour) |
| 1 | 1 | 1 | Level Mode *2) | Once per month (at 00 hours, 00 minutes, and 00 seconds of first day of every month) |

(Default)

***1) Pulse Mode:**

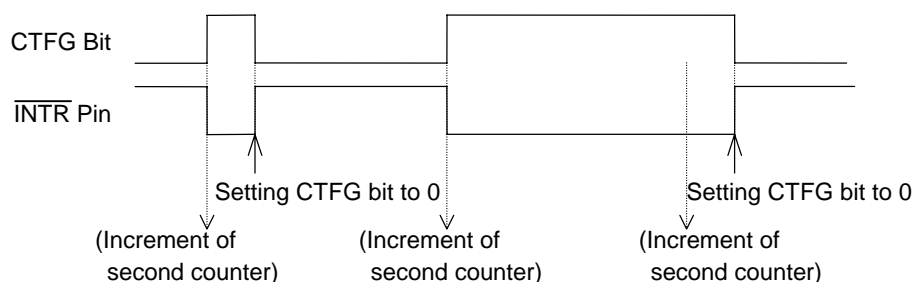
2-Hz and 1-Hz clock pulses are output in synchronization with the increment of the second counter as illustrated in the timing chart below.



In the pulse mode, the increment of the second counter is delayed by approximately 92 μ s from the falling edge of clock pulses. Consequently, time readings immediately after the falling edge of clock pulses may appear to lag behind the time counts of the real-time clocks by approximately 1 second. Rewriting the second counter will reset the other time counters of less than 1 second, driving the $\overline{\text{INTR}}$ pin low.

***2) Level Mode:**

Periodic interrupt signals are output with selectable interrupt cycle settings of 1 second, 1 minute, 1 hour, and 1 month. The increment of the second counter is synchronized with the falling edge of periodic interrupt signals. For example, periodic interrupt signals with an interrupt cycle setting of 1 second are output in synchronization with the increment of the second counter as illustrated in the timing chart below.



*1), *2) When the oscillation adjustment circuit is used, the interrupt cycle will fluctuate once per 20sec. as follows:

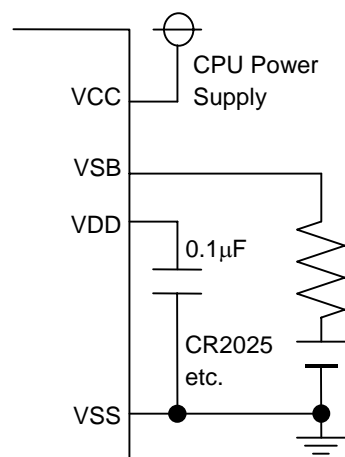
Pulse Mode: The "L" period of output pulses will increment or decrement by a maximum of ± 3.784 ms. For example, 1-Hz clock pulses will have a duty cycle of $50 \pm 0.3784\%$.

Level Mode: A periodic interrupt cycle of 1 second will increment or decrement by a maximum of ± 3.784 ms.

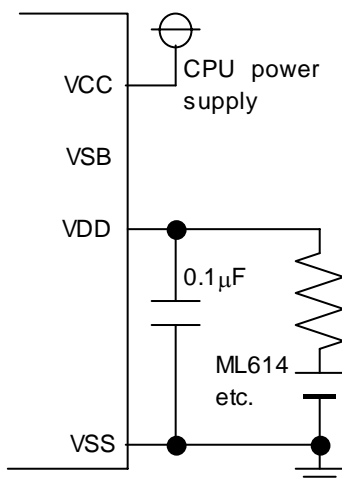
Typical Applications

• Typical Power Circuit Configurations

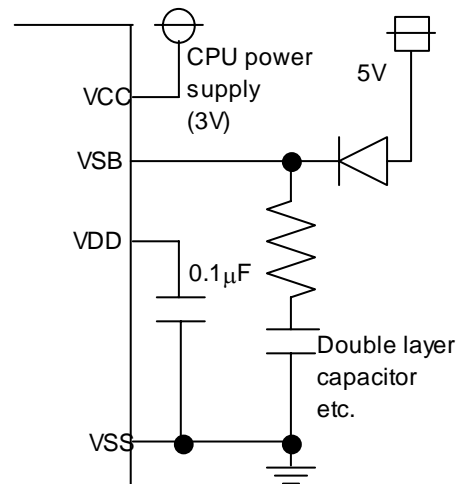
The case of back-up by primary battery



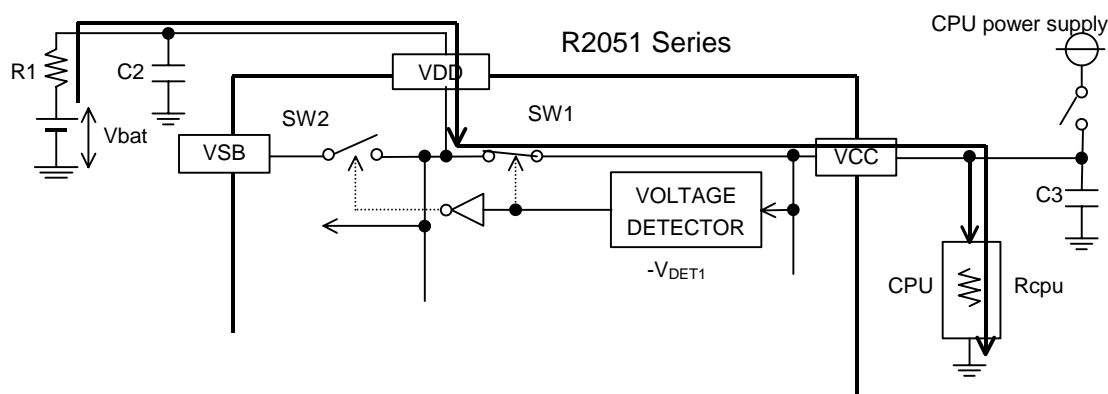
The case of back-up by capacitor or secondary battery
(Charging voltage is equal to CPU power supply voltage)



The case of back-up by capacitor or secondary battery
(Charging voltage is not equal to CPU power supply voltage)



VDD pin cannot be connected to any additional heavy load components such as SRAM. And VDD pin must be connected C2, and C2 should be over 0.1μF.



When secondary battery or double layer capacitor connects to VDD pin, after CPU power supply turning off, secondary battery discharges through the root above figure. If R1 is much smaller than CPU impedance (Rcpu), VCC voltage keeps higher than -VDET1, and SW1 keeps on. Therefore R1 must be specified by following formula.

$$R1 > R_{cpu} \times (V_{bat} - (-V_{DET1})) / (-V_{DET1})$$

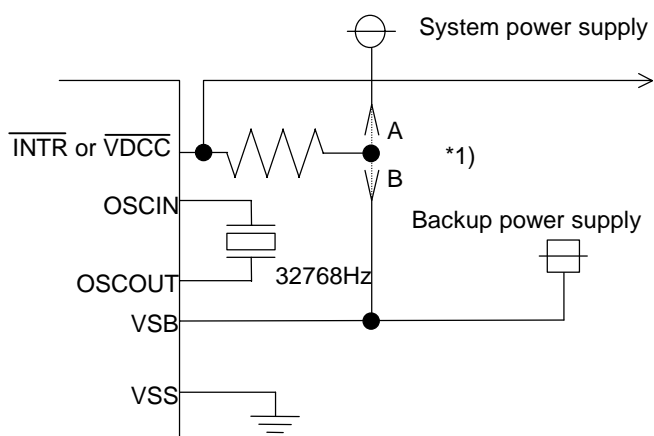
R1 is specified by back-up battery or double layer capacitor, too. Please check the data sheet for back-up devices.

• Connection of CIN pin

Please connect capacitor over 0.1μF between CIN and VSS pin.

• Connection of $\overline{\text{INTR}}$ and $\overline{\text{VDCC}}$ Pin (except R2051Txx)

The $\overline{\text{INTR}}$ and $\overline{\text{VDCC}}$ pins follow the N-channel open drain output logic and contains no protective diode on the power supply side. As such, it can be connected to a pull-up resistor of up to 5.5 volts regardless of supply voltage.



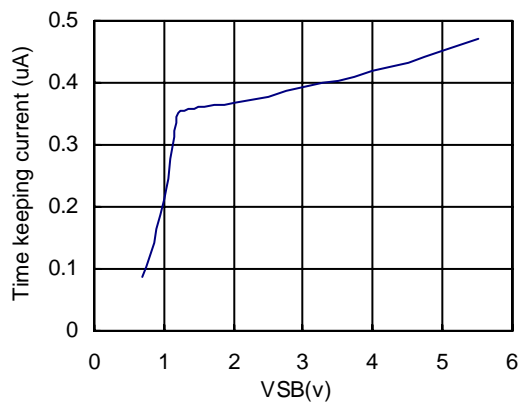
*1) Depending on whether the $\overline{\text{INTR}}$ and $\overline{\text{VDCC}}$ pins are to be used during battery backup, it should be connected to a pull-up resistor at the following different positions:

- (1) Position A in the left diagram when it is not to be used during battery backup.
- (2) Position B in the left diagram when it is to be used during battery backup.

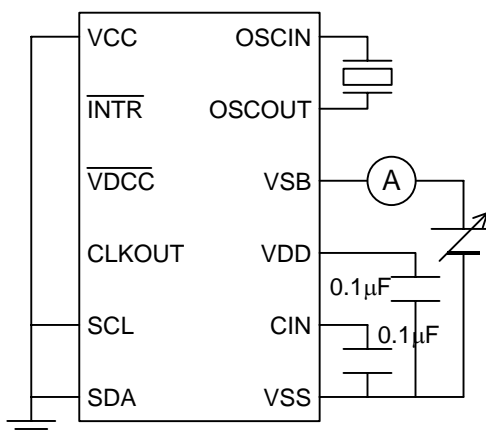
Typical Characteristics

• Time keeping current (I_{SB}) vs. Supply voltage (V_{SB})

($T_{opt}=25^{\circ}\text{C}$)

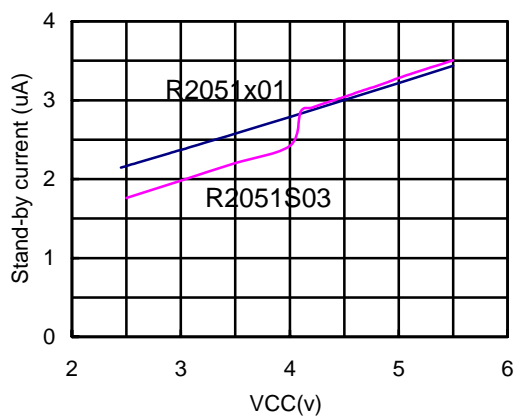


Test Circuit

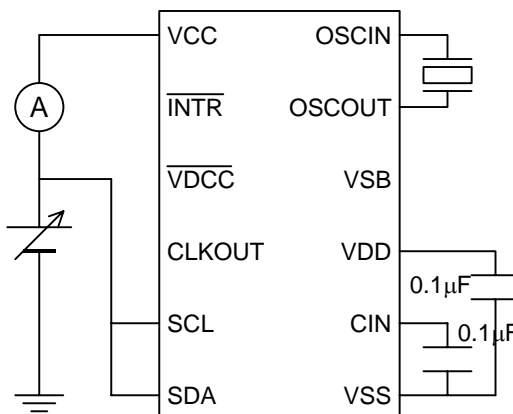


• Stand-by current (I_{CC}) vs. Supply voltage (V_{CC})

($T_{opt}=25^{\circ}\text{C}$)

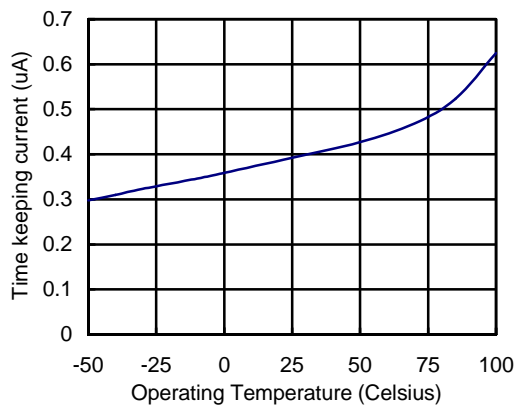


Test circuit

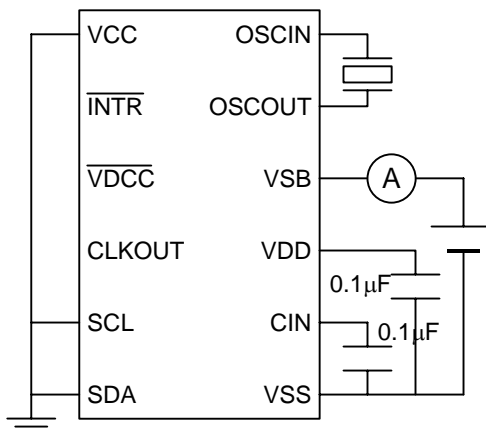


• Time keeping current (I_{SB}) vs. Operating Temperature (T_{opt})

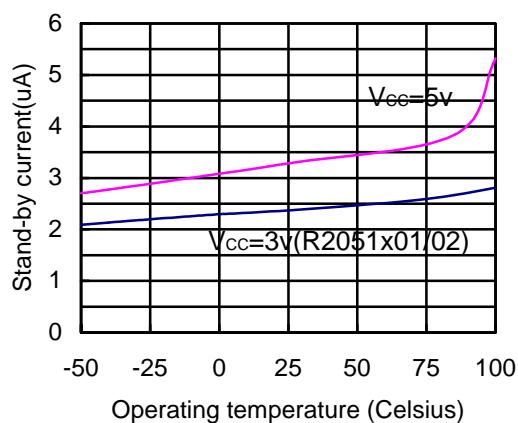
($V_{SB}=3\text{V}$)



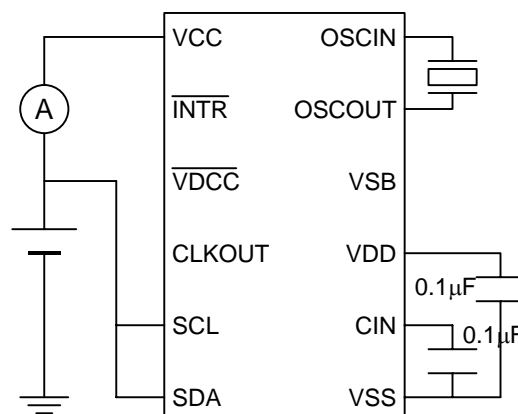
Test circuit



• Stand-by current (I_{CC}) vs. Operating Temperature (T_{opt})

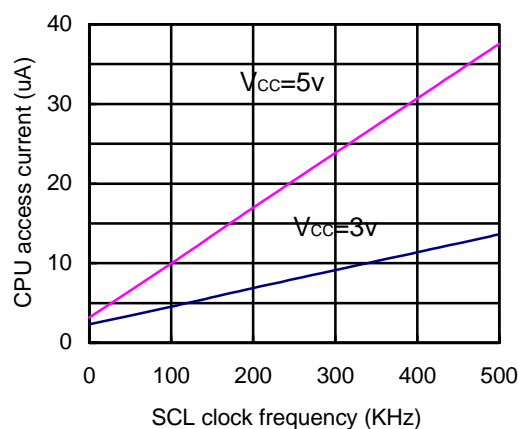


Test circuit



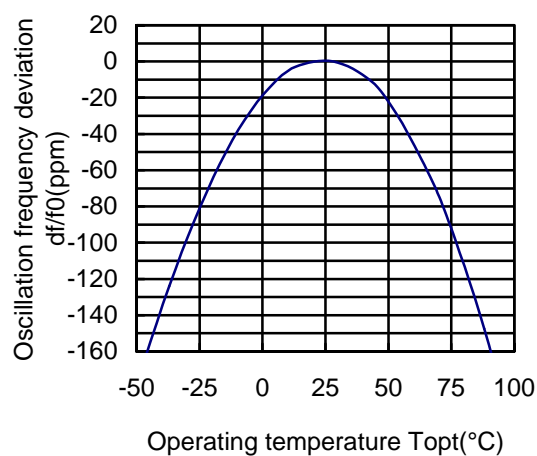
• CPU access current vs. SCL clock frequency (kHz)

($T_{opt}=25^{\circ}C$)

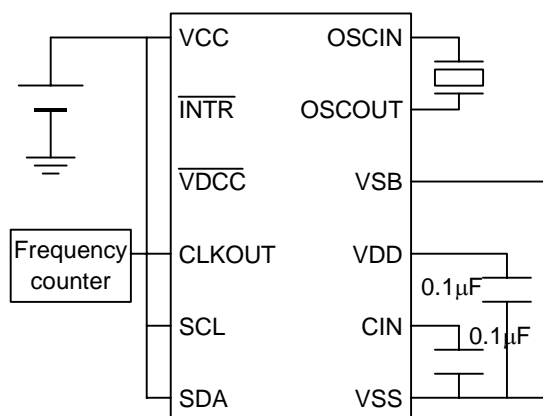


• Oscillation frequency deviation ($\Delta f/f_0$) vs. Operating temperature (T_{opt})

($V_{CC}=3V$) $T_{opt}=25^{\circ}C$ as standard

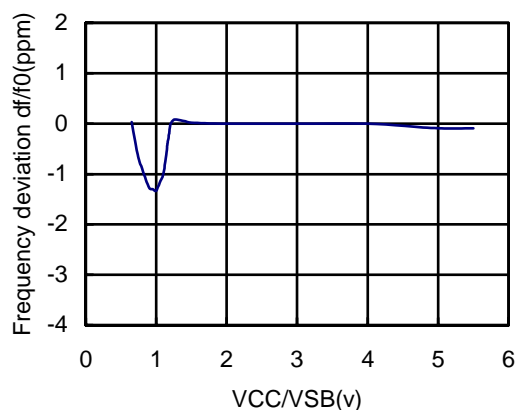


Test circuit

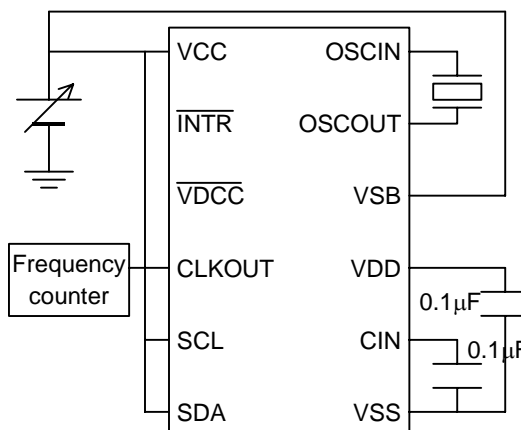


• Frequency deviation ($\Delta f/f_0$) vs. Supply voltage (V_{SB}/V_{CC})

($T_{opt}=25^\circ\text{C}$) $V_{CC}/V_{SB}=3\text{V}$ as standard

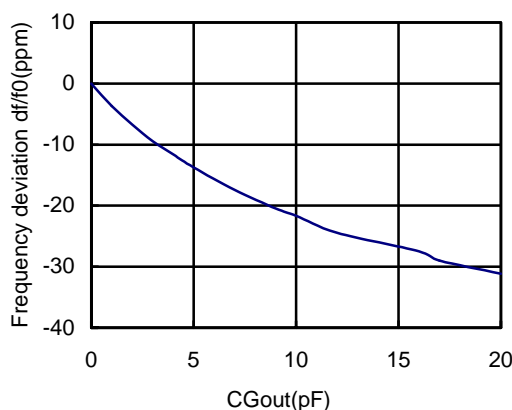


Test circuit

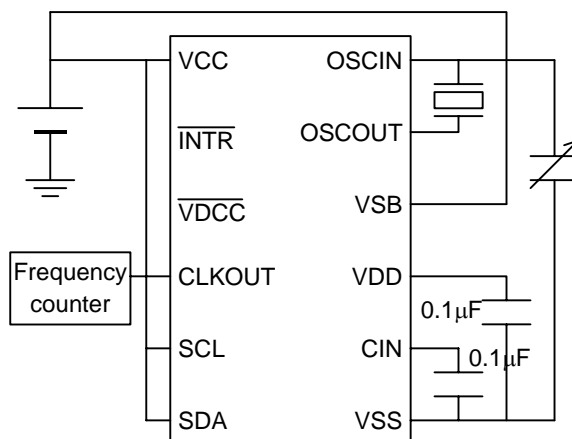


• Frequency deviation ($\Delta f/f_0$) vs. CGout

($T_{opt}=25^\circ\text{C}$, $V_{CC}=3\text{V}$) $CG_{out}=0\text{pF}$ as standard

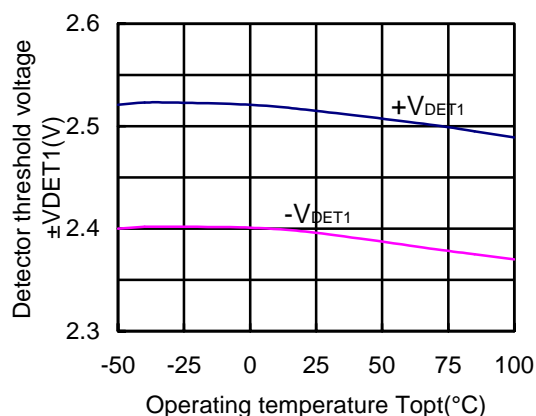


Test circuit

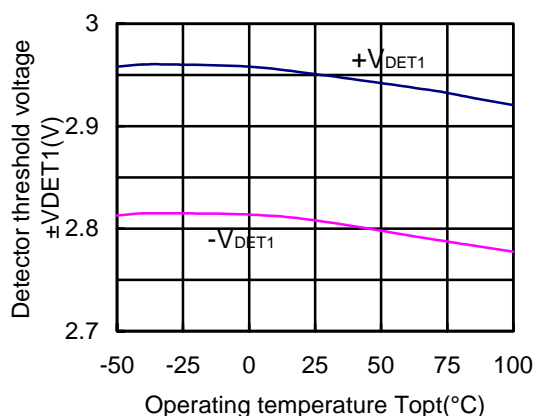


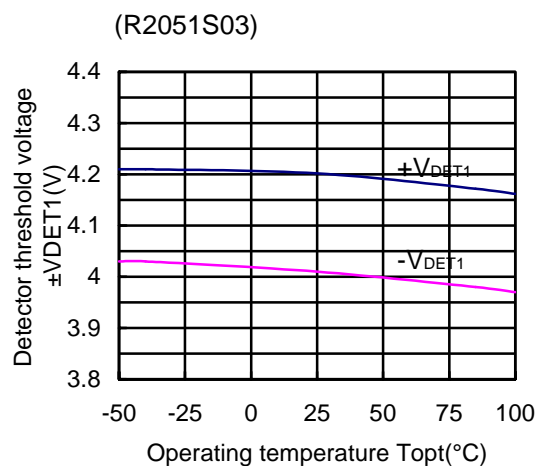
• Detector threshold voltage ($+V_{DET1}/-V_{DET1}$) vs. Operating temperature (T_{opt}) ($V_{SB}=3\text{V}$)

(R2051x01)

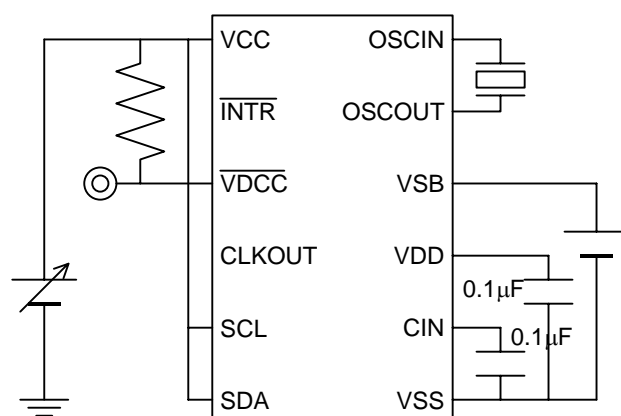


(R2051x02)



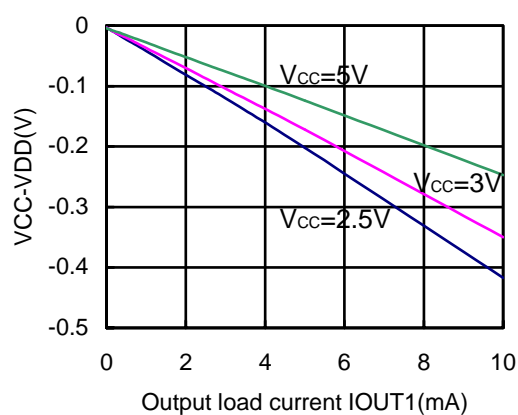


Test circuit

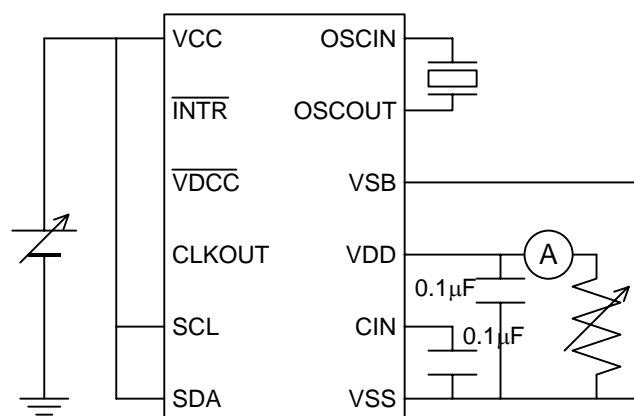


• $V_{CC}-V_{DD}(V_{DDOUT1})$ vs. Output load current (I_{OUT1})

($T_{opt}=25^{\circ}\text{C}$)

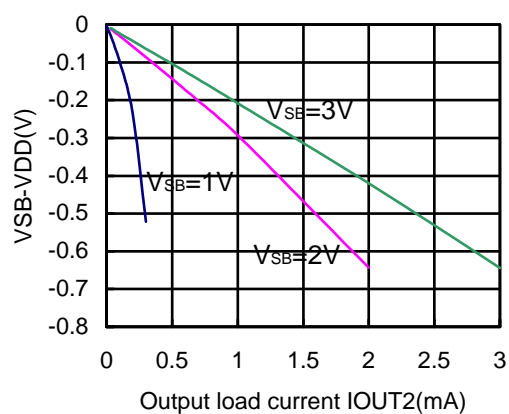


Test circuit

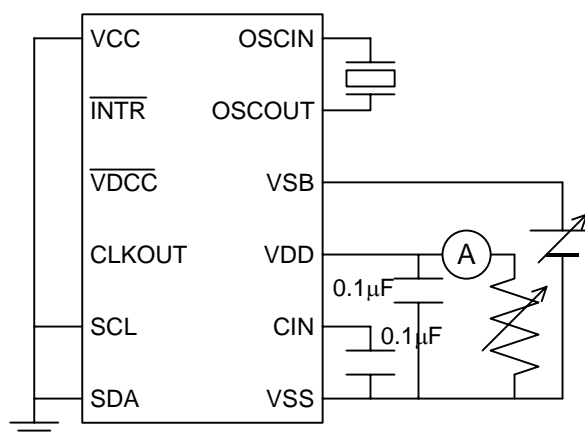


• $V_{SB}-V_{DD}(V_{DDOUT2})$ vs. Output load current (I_{OUT2})

($T_{opt}=25^{\circ}\text{C}$)

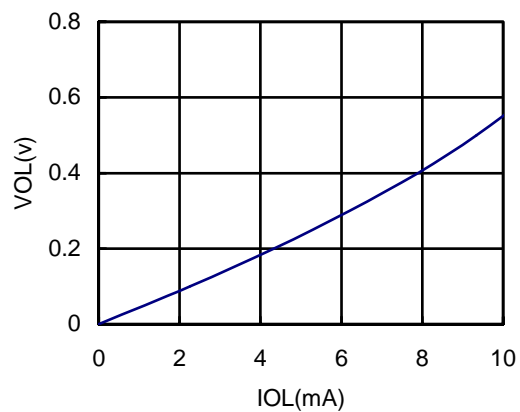


Test circuit



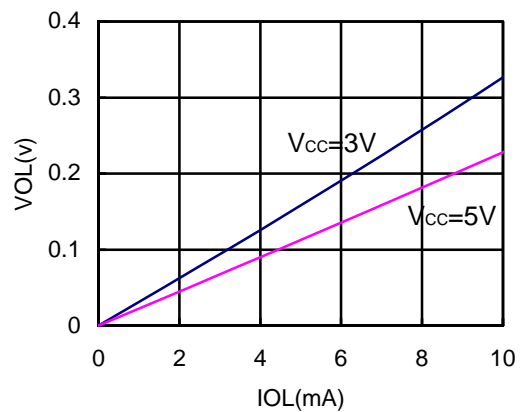
• V_{OL} vs. I_{OL} (\overline{VDCC} pin) (Except R2051Txx)

($T_{opt}=25^{\circ}C$, $V_{SB}=V_{CC}=2V$)



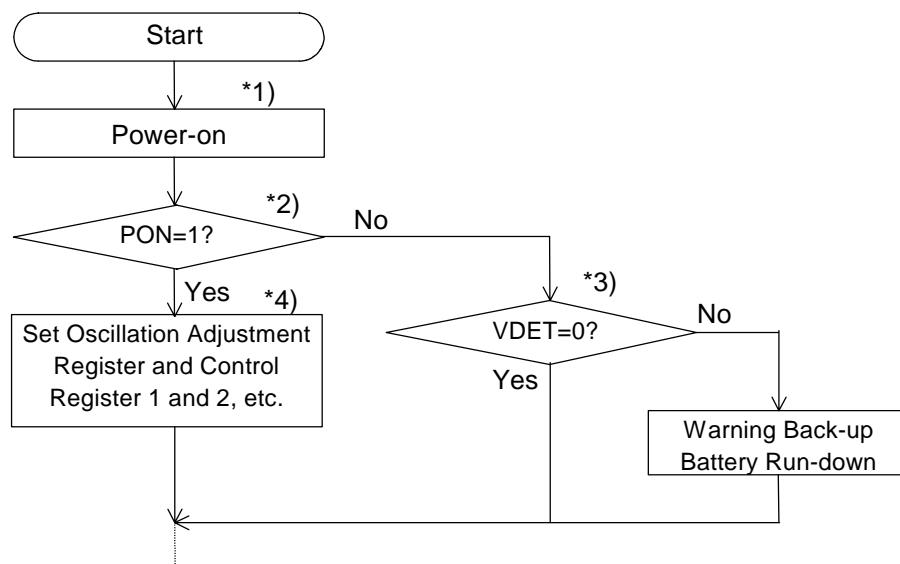
• V_{OL} vs. I_{OL} (\overline{INTR} pin) (Except R2051Txx)

($T_{opt}=25^{\circ}C$)



Typical Software-based Operations

• Initialization at Power-on



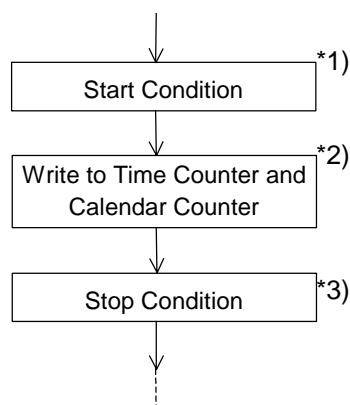
*1) After power-on from 0 volt, the process of internal initialization require a time span on 1sec, so that access should be done after $\overline{\text{VDCC}}$ turning to OFF(H).

*2) The PON bit setting of 0 in the Control Register 1 indicates power-on from backup battery and not from 0v. For further details, see "P.41 ■Power-on Reset, Oscillation Halt Sensing, and Supply Voltage Monitoring •PON, $\overline{\text{XST}}$, and VDET".

*3) This step is not required when the supply voltage monitoring circuit is not used.

*4) This step involves ordinary initialization including the Oscillation Adjustment Register and interrupt cycle settings, etc.

• Writing of Time and Calendar Data



*1) When writing to clock and calendar counters, do not insert Stop Condition until all times from second to year have been written to prevent error in writing time. (Detailed in "P.34 Data Transmission under Special Condition".

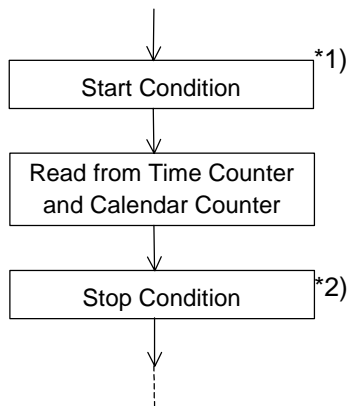
*2) Any writing to the second counter will reset divider units lower than the second digits.

*3) Take care so that process from Start Condition to Stop Condition will be complete within 0.5sec. (Detailed in "P.34 Data Transmission under Special Condition".

The R2051 may also be initialized not at power-on but in the process of writing time and calendar data.

• Reading Time and Calendar Data

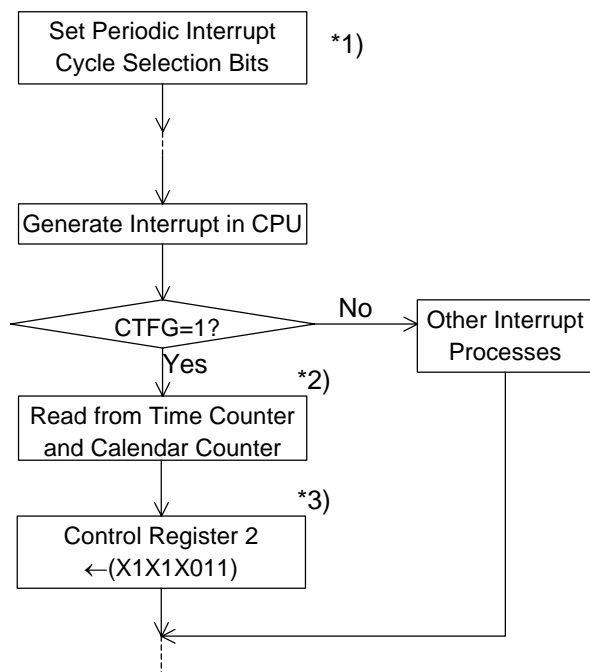
(1) Ordinary Process of Reading Time and Calendar Data



*1) When reading to clock and calendar counters, do not insert Stop Condition until all times from second to year have been written to prevent error in writing time. (Detailed in "P.34 Data Transmission under Special Condition".

*2) Take care so that process from Start Condition to Stop Condition will be complete within 0.5sec. (Detailed in "P.34 Data Transmission under Special Condition".

(2) Basic Process of Reading Time and Calendar Data with Periodic Interrupt Function



*1) This step is intended to select the level mode as a waveform mode for the periodic interrupt function.

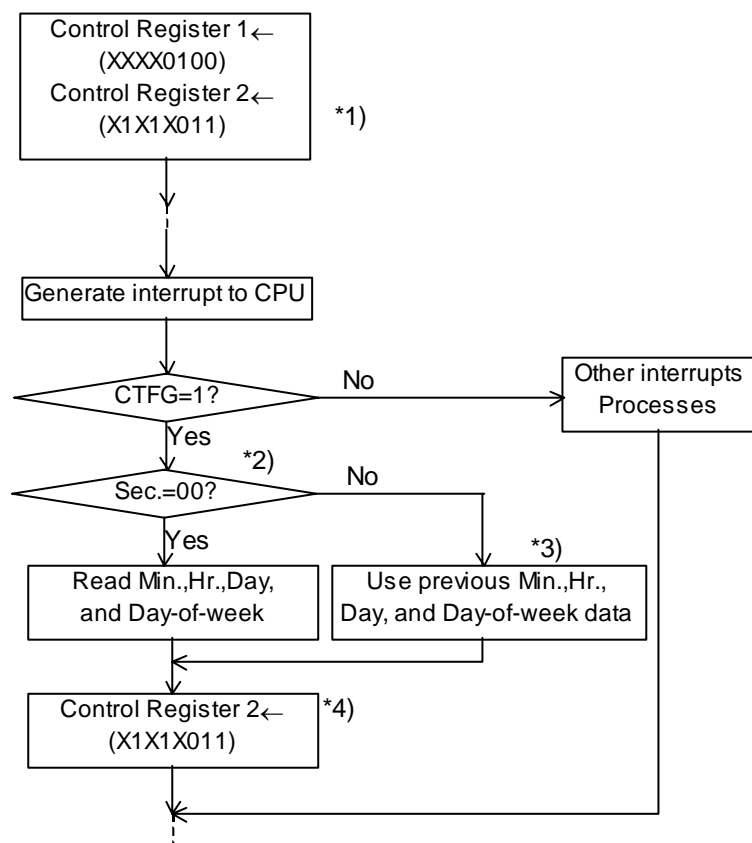
*2) This step must be completed within 0.5 second.

*3) This step is intended to set the CTFG bit to 0 in the Control Register 2 to cancel an interrupt to the CPU.

(3) Applied Process of Reading Time and Calendar Data with Periodic Interrupt Function (Except R2051Txx)

Time data need not be read from all the time counters when used for such ordinary purposes as time count indication. This applied process can be used to read time and calendar data with substantial reductions in the load involved in such reading.

For Time Indication in "Day-of-Month, Day-of-week, Hour, Minute, and Second" Format:



*1) This step is intended to select the level mode as a waveform mode for the periodic interrupt function.

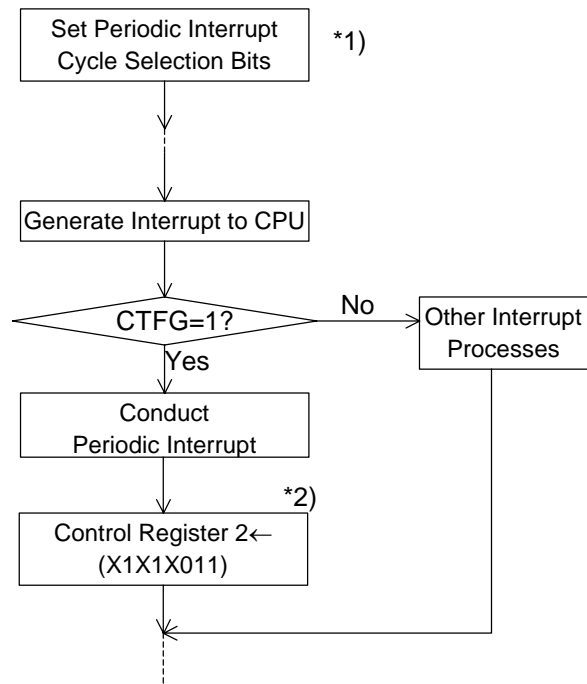
*2) This step must be completed within 0.5 sec.

*3) This step is intended to read time data from all the time counters only in the first session of reading time data after writing time data.

*4) This step is intended to set the CTFG bit to 0 in the Control Register 2 to cancel an interrupt to the CPU.

• Interrupt Process

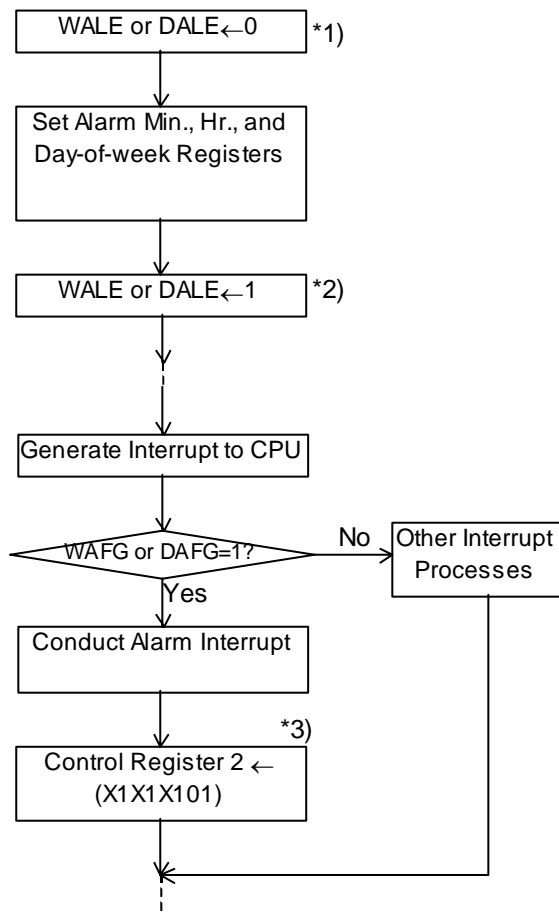
(1) Periodic Interrupt (except R2051Txx)



*1) This step is intended to select the level mode as a waveform mode for the periodic interrupt function.

*2) This step is intended to set the CTFG bit to 0 in the Control Register 2 to cancel an interrupt to the CPU.

(2) Alarm Interrupt (except R2051Txx)



*1) This step is intended to once disable the alarm interrupt circuit by setting the WALE or DALE bits to 0 in anticipation of the coincidental occurrence of a match between current time and preset alarm time in the process of setting the alarm interrupt function.

*2) This step is intended to enable the alarm interrupt function after completion of all alarm interrupt settings.

*3) This step is intended to once cancel the alarm interrupt function by writing the settings of "X,1,X,1,X,1,0,1" and "X,1,X,1,X,1,1,0" to the Alarm_W Registers and the Alarm_D Registers, respectively.



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



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