
2 A PWM/VFM Step-down DC/DC Converter with Synchronous Rectifier

NO.EA-296-160523

OUTLINE

The RP506K is a low supply current CMOS-based PWM/VFM step-down DC/DC converter with synchronous rectifier featuring 2 A^{*1} output current. Internally, a single converter consists of an oscillator, a reference voltage unit, an error amplifier, a switching control circuit, a mode control circuit, a soft start circuit, a latch type protection circuit, an under-voltage lockout (UVLO) circuit, a thermal shutdown circuit, and switching transistors.

The RP506K is employing synchronous rectification for improving the efficiency of rectification by replacing diodes with built-in switching transistors. Using synchronous rectification not only increases circuit performance but also allows a design to reduce parts count.

Power controlling method can be selected from forced PWM control type or PWM/VFM auto switching control type by inputting a signal to the MODE pin. In low output current, forced PWM control switches at fixed frequency rate in order to reduce noise. Likewise, in low output current, PWM/VFM auto switching control automatically switches from PWM mode to VFM mode in order to achieve high efficiency.

Output voltage type can be selected from an internally fixed output voltage type (RP506Kxx1A/B/D/E) or an externally adjustable output voltage type (RP506K001C/F). The output voltage of the RP506Kxx1A/B/D/E can be set by 0.1 V step and the output voltage accuracy is as high as $\pm 1.5\%$ or ± 18 mV. The output voltage of the RP506K001C/F can be set by using the external resistors.

Oscillator frequency can be selected from 2.25 MHz (RP506Kxx1A/B/C) or 1.2 MHz (RP506Kxx1D/E/F). Soft-start time is Typ. 0.15 ms, and by connecting an external capacitor to the TSS pin, soft-start time is adjustable.

Power good (PG) function monitors the V_{OUT} pin voltage or the feedback pin voltage (V_{FB}), and switches the PG pin to low if any abnormal condition is detected.

Protection circuits included in the RP506K are over current protection circuit, latch type protection circuit and thermal shutdown circuit. Over current protection circuit supervises the inductor peak current in each switching cycle, and if the current exceeds the L_X current limit (I_{LXLIM}), it turns off Pch Tr. Latch type protection circuit latches the built-in driver to the OFF state and stops the operation of the step-down DC/DC converter if the over current status continues or V_{OUT} continues being the half of the setting voltage for equal or longer than protection delay time (t_{prot}). Thermal shutdown circuit detects overheating of the converter if the output pin is shorted to the ground pin (GND) etc. and stops the converter operation to protect it from damage if the junction temperature exceeds the specified temperature.

The RP506K is available in DFN(PLP)2527-10 which achieves high-density mounting on boards.

^{*1} This is an approximate value. The output current is dependent on conditions and external components.

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FEATURES

- Supply Current.....Typ. 48 μ A (VFM mode, Lx at no load)
- Standby Current.....Typ. 0 μ A
- Input Voltage Range 2.5 V to 5.5 V (Absolute Maximum Ratings: 6.5 V)
- Output Voltage Range^{*1}

Version	Forced PWM Control	PWM/VFM Auto Switching Control
RP506Kxx1A/B	1.1 V to 3.3 V	0.8 V to 3.3 V
RP506K001C	1.1 V to 4.0 V	0.8 V to 4.0 V
RP506Kxx1D/E	0.6 V to 3.3 V	
RP506K001F	0.6 V to 4.0 V	

- Output Voltage Accuracy..... $\pm 1.5\%$ ($V_{SET}^{*2} \geq 1.2$ V),
..... ± 18 mV ($V_{SET} < 1.2$ V) (RP506Kxx1A/B/D/E)
- Feedback Voltage Accuracy..... ± 9 mV ($V_{FB} = 0.6$ V) (RP506K001C/F)
- Output Voltage/Feedback Voltage
Temperature Coefficient ± 100 ppm/ $^{\circ}$ C
- Oscillator Frequency Typ. 2.25 MHz (RP506Kxx1A/B/C)
.....Typ. 1.2 MHz (RP506Kxx1D/E/F)
- Oscillator Maximum Duty Min. 100%
- Built-in Driver ON Resistance.....Typ. Pch. 0.130 Ω , Nch. 0.125 Ω ($V_{IN} = 3.6$ V)
- UVLO Detector Threshold Typ. 2.2 V
- Inductor Current Limit Circuit.....Current limit Typ. 2.8 A
- Latch Type Protection Circuit Typ. 1.5 ms
- Package.....DFN(PLP)2527-10

^{*1} Refer to *Selection Guide* for detailed information.

Fixed output voltage type (RP506Kxx1A/B/D/E) can be set up to 3.3 V by 0.1 V step.

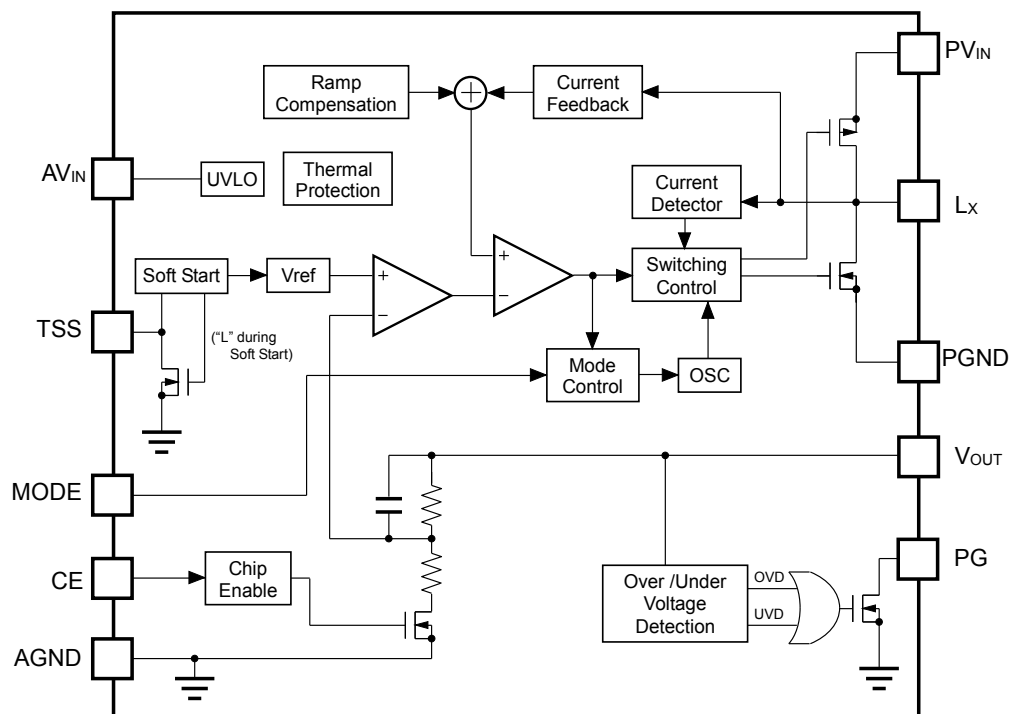
Adjustable output voltage type (RP506K001C/F) can be set up to 4.0 V.

^{*2} V_{SET} = Set Output Voltage

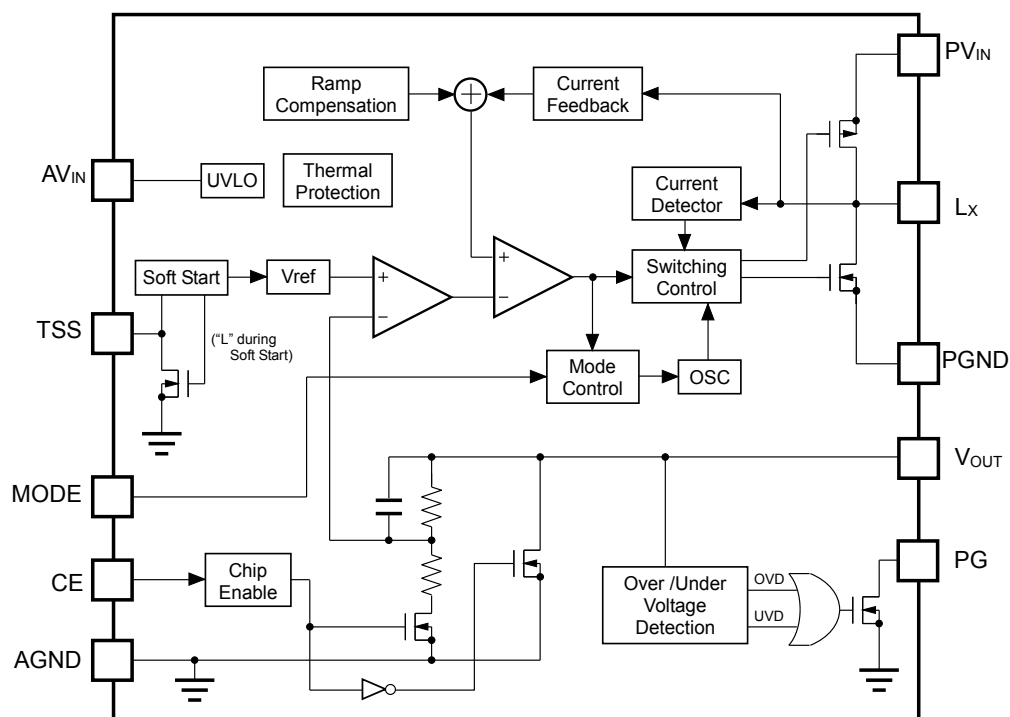
APPLICATION

- Power source for Li-ion battery-used equipment
- Power source for portable communication equipment, camcorder, DSC, Notebook PC
- Power source for HDD, WLAN

BLOCK DIAGRAM



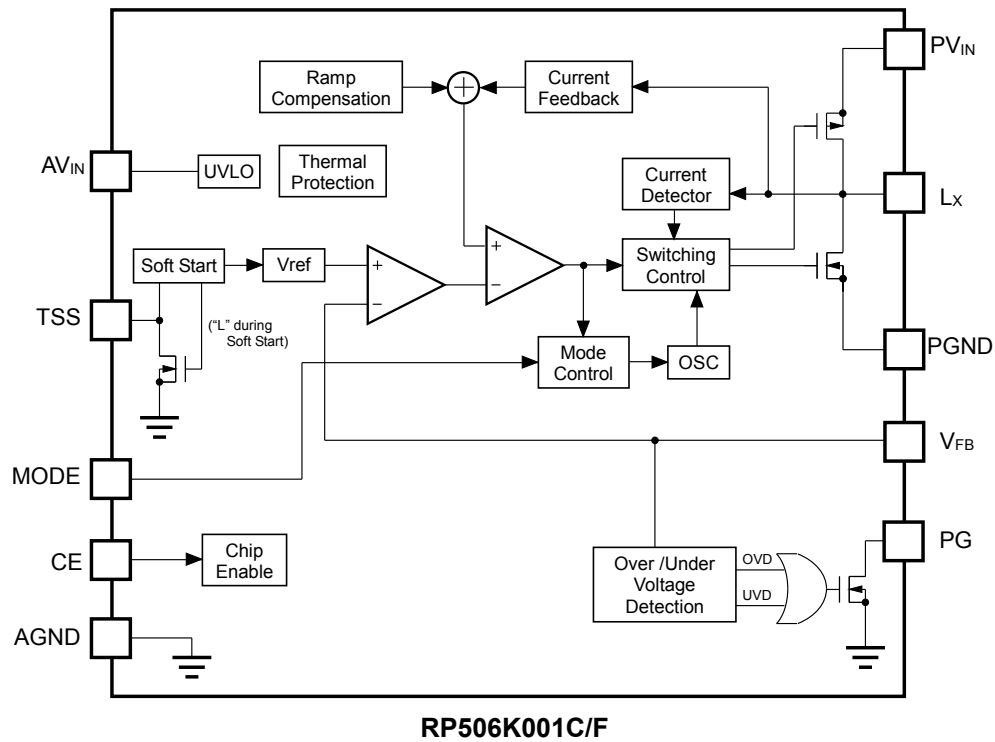
RP506Kxx1A/D



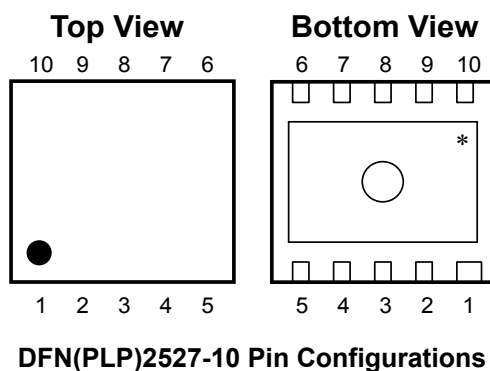
RP506Kxx1B/E

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PIN DESCRIPTION

**DFN(PLP)2527-10 Pin Description**

Pin No.	Symbol	Description
1	PV _{IN}	PV _{IN} Input Voltage Pin ^{*1}
2	AV _{IN}	AV _{IN} Input Voltage Pin ^{*1}
3	PG	Power Good Pin
4	CE	Chip Enable Pin (Active “H”)
5	MODE	Mode Control Pin (“H”: forced PWM control, “L”: PWM/VFM auto switching control)
6	TSS	Soft-start Pin
7	V _{OUT} / V _{FB}	Output/ Feedback Voltage Pin
8	AGND	Analog Ground Pin ^{*2}
9	L _X	Switching Pin
10	PGND	Power Ground Pin ^{*2}

* The tab on the bottom of the package enhances thermal performance and is electrically connected to GND (substrate level). It is recommended that the tab be connected to the ground plane on the board, or otherwise be left floating.

^{*1} No.1 pin and No.2 pin must be wired to the V_{IN} plane when mounting on boards.

^{*2} No.8 pin and No.10 pin must be wired to the GND plane when mounting on boards.

ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings

(AGND = PGND = 0 V)

Symbol	Item		Rating	Unit
V_{IN}	A/PV _{IN} Pin Voltage		-0.3 to 6.5	V
V_{LX}	L _X Pin Voltage		-0.3 to A/PV _{IN} +0.3	V
V_{CE}	CE Pin Voltage		-0.3 to 6.5	V
$V_{OUT/V_{FB}}$	V _{OUT} /V _{FB} Pin Voltage		-0.3 to 6.5	V
V_{MODE}	MODE Pin Voltage		-0.3 to 6.5	V
V_{PG}	PG Pin Voltage		-0.3 to 6.5	V
V_{TSS}	TSS Pin Voltage		-0.3 to AV _{IN} +0.3	V
I_{LX}	L _X Pin Output Current		2.8	A
P_D	Power Dissipation*1	Standard Land Pattern	910	mW
		High Wattage Land Pattern	1400	mW
T_j	Junction Temperature		-40 to 125	°C
T_{stg}	Storage Temperature Range		-55 to 125	°C

*1 Refer to *PACKAGE INFORMATION* for detailed information.

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the lifetime and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

RECOMMENDED OPERATING CONDITIONS

Recommended Operating Conditions

Symbol	Item	Rating	Unit
V_{IN}	Input Voltage	2.5 to 5.5	V
T_a	Operating Temperature Range	-40 to 85	°C

RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

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ELECTRICAL CHARACTERISTICS**RP506Kxx1 Electrical Characteristics**

(Ta = 25°C)

Symbol	Item	Conditions	Min.	Typ.	Max.	Unit
Istandby	Standby Current	A/PVIN = 5.5 V, VCE = 0 V		0	5	μA
ICEH	CE “H” Input Current	A/PVIN = VCE = 5.5 V	−1	0	1	μA
ICEL	CE “L” Input Current	A/PVIN = 5.5 V, VCE = 0 V	−1	0	1	μA
IMODEH	MODE “H” Input Current	A/PVIN = VMODE = 5.5 V, VCE = 0 V	−1	0	1	μA
IMODEL	MODE “L” Input Current	A/PVIN = 5.5 V, VCE = VMODE = 0 V	−1	0	1	μA
ILXLEAKH	Lx Leakage Current “H”	A/PVIN = VLX = 5.5 V, VCE = 0 V	−1	0	6	μA
ILXLEAKL	Lx Leakage Current “L”	A/PVIN = 5.5 V, VCE = VLX = 0 V	−6	0	1	μA
VCEH	CE “H” Input Voltage	A/PVIN = 5.5 V	1.0			V
VCEL	CE “L” Input Voltage	A/PVIN = 2.5 V			0.4	V
VMODEH	MODE “H” Input Voltage	A/PVIN = 5.5 V	1.0			V
VMODEL	MODE “L” Input Voltage	A/PVIN = 2.5 V			0.4	V
RONP	On Resistance of Pch Transistor	A/PVIN = 3.6 V, ILX = −100 mA		0.130		Ω
RONN	On Resistance of Nch Transistor	A/PVIN = 3.6 V, ILX = −100 mA		0.125		Ω
Maxduty	Maximum Duty Cycle		100			%
tstart1	Soft-start Time 1	A/PVIN = VCE = 3.6 V or VSET +1 V, TSS = OPEN		150	300	μs
tstart2	Soft-start Time 2	A/PVIN = VCE = 3.6 V or VSET +1 V, CSS = 0.1 μF	15	30	45	ms
ILXLIM	Lx Current Limit	A/PVIN = VCE = 3.6 V or VSET +1 V	2300	2800		mA
tprot	Protection Delay Time	A/PVIN = VCE = 3.6 V or VSET +1 V	0.5	1.5	5	ms
VUVLO1	UVLO Detector Threshold	A/PVIN = VCE	2.1	2.2	2.3	V
VUVLO2	UVLO Released Voltage	A/PVIN = VCE	2.2	2.3	2.4	V
T _{TSD}	Thermal Shutdown Temperature	Junction Temperature		150		°C
T _{TSR}	Thermal Shutdown Released Temperature	Junction Temperature		100		°C
R _{PG}	On Resistance of PG Pin When Low Output	A/PVIN = 3.6 V, VOUT = 0 V or VFB = 0 V		45		Ω

ELECTRICAL CHARACTERISTICS (continued)

RP506Kxx1A/B, RP506K001C (Oscillator Frequency: 2.25 MHz) Electrical Characteristics (Ta = 25°C)

V_{IN}	When MODE = H Operating Input Voltage*1	$1.1\text{ V} \leq V_{SET} < 1.2\text{ V}$	2.5		4.5	V
		$1.2\text{ V} \leq V_{SET}$	2.5		5.5	
	When MODE = L Operating Input Voltage*2	$0.8\text{ V} \leq V_{SET} < 1.0\text{ V}$	2.5		4.5	
		$1.0\text{ V} \leq V_{SET}$	2.5		5.5	
fosc	Oscillator Frequency	$A/PV_{IN} = V_{CE} = 3.6\text{ V}$ or $V_{SET} + 1\text{ V}$	2.00	2.25	2.50	MHz

RP506Kxx1D/E, RP506K001F (Oscillator Frequency: 1.2 MHz) Electrical Characteristics

V_{IN}	When MODE = H Operating Input Voltage	$0.6\text{ V} \leq V_{SET} < 0.7\text{ V}$	2.5		4.5	V
		$0.7\text{ V} \leq V_{SET}$	2.5		5.5	
	When MODE = L Operating Input Voltage	$0.6\text{ V} \leq V_{SET}$	2.5		5.5	
fosc	Oscillator Frequency	$A/PV_{IN} = V_{CE} = 3.6\text{ V}$ or $V_{SET} + 1\text{ V}$	1.00	1.20	1.40	MHz

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ELECTRICAL CHARACTERISTICS (continued)

RP506Kxx1A/B/D/E (Fixed Output Voltage Type) Electrical Characteristics

(Ta = 25°C)

Symbol	Item	Conditions	Min.	Typ.	Max.	Unit
V _{OUT}	Output Voltage	A/PV _{IN} = V _{CE} = 3.6 V or V _{SET} + 1 V	V _{SET} ≥ 1.2 V	x0.985	x1.015	V
			V _{SET} < 1.2 V	-0.018	+0.018	
ΔV _{OUT} /ΔTa	Output Voltage Temperature Coefficient	-40°C ≤ Ta ≤ 85°C		±100		ppm/°C
I _{DD1}	Supply Current 1	A/PV _{IN} = V _{CE} = 5.5 V, V _{OUT} = V _{SET} × 0.8		600		μA
I _{DD2}	Supply Current 2	A/PV _{IN} = V _{CE} = V _{OUT} = 5.5 V	V _{MODE} = 0 V	48	72	μA
			V _{MODE} = 5.5 V	600		μA
I _{VOU_{TL}}	V _{OUT} "L" Current	A/PV _{IN} = 5.5 V, V _{CE} = V _{OUT} = 0 V	-1	0	1	μA
V _{OVD}	OVD Voltage	A/PV _{IN} = 3.6 V		V _{SET} × 1.2		V
V _{UVD}	UVD Voltage	A/PV _{IN} = 3.6 V		V _{SET} × 0.8		V

RP506Kxx1A/D (Fixed Output Voltage Type without Auto-discharge Function)

I _{VOU_{TH}}	V _{OUT} "H" Current	A/PV _{IN} = V _{OUT} = 5.5 V, V _{CE} = 0 V	-1	0	1	μA
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RP506Kxx1B/E (Fixed Output Voltage Type with Auto-discharge Function)

R _{LOW}	On Resistance of Low Output	A/PV _{IN} = 3.6 V, V _{CE} = 0 V		45		Ω
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RP506K001C/F (Adjustable Output Voltage Type) Electrical Characteristics

V _{FB}	Feedback Voltage	A/PV _{IN} = V _{CE} = 3.6 V	0.591	0.600	0.609	V
ΔV _{FB} /ΔTa	Feedback Voltage Temperature Coefficient	-40°C ≤ Ta ≤ 85°C		±100		ppm/°C
I _{DD1}	Supply Current 1	A/PV _{IN} = V _{CE} = 5.5 V, V _{FB} = 0.48 V		600		μA
I _{DD2}	Supply Current 2	A/PV _{IN} = V _{CE} = V _{FB} = 5.5 V	V _{MODE} = 0 V	48	72	μA
			V _{MODE} = 5.5 V	600		μA
I _{VFBH}	V _{FB} "H" Current	A/PV _{IN} = V _{FB} = 5.5 V, V _{CE} = 0 V	-1	0	1	μA
I _{VFBL}	V _{FB} "L" Current	A/PV _{IN} = 5.5 V, V _{CE} = V _{FB} = 0 V	-1	0	1	μA
V _{OVD}	OVD Voltage	A/PV _{IN} = 3.6 V		0.72		V
V _{UVD}	UVD Voltage	A/PV _{IN} = 3.6 V		0.48		V

All test items listed under *Electrical Characteristics* are done under the pulse load condition (T_j ≈ Ta = 25°C) except Output Voltage Temperature Coefficient and Feedback Voltage Temperature Coefficient.

*1 As for RP506Kxx1A/B/C (MODE = H), V_{SET} can be set from 1.1 V.

*2 As for RP506Kxx1A/B/C (MODE = L), V_{SET} can be set from 0.8 V.

ELECTRICAL CHARACTERISTICS (continued)**RP506K Electrical Characteristics by Different Output Voltage**

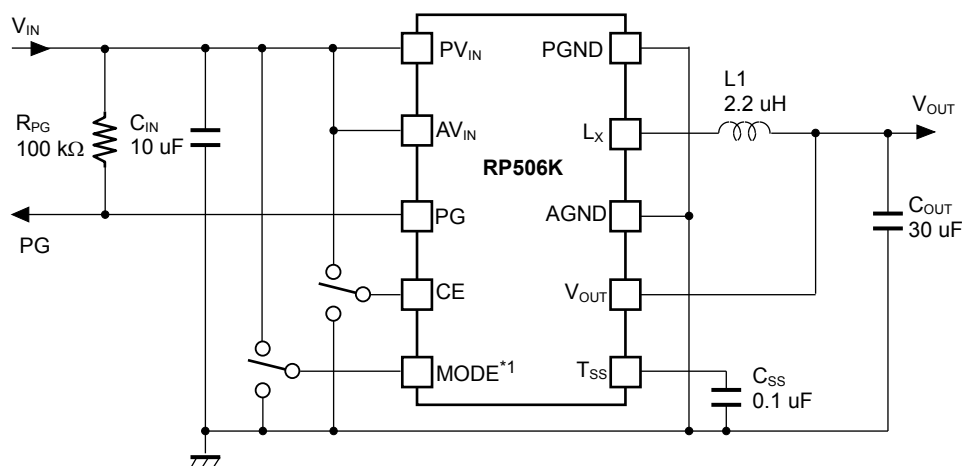
(Ta = 25°C)

Product Name	Output Voltage (V _{OUT}) [V]		
	Min.	Typ.	Max.
RP506K061x	0.582	0.600	0.618
RP506K071x	0.682	0.700	0.718
RP506K081x	0.782	0.800	0.818
RP506K091x	0.882	0.900	0.918
RP506K101x	0.982	1.000	1.018
RP506K111x	1.082	1.100	1.118
RP506K121x	1.182	1.200	1.218
RP506K131x	1.281	1.300	1.319
RP506K141x	1.379	1.400	1.421
RP506K151x	1.478	1.500	1.522
RP506K161x	1.576	1.600	1.624
RP506K171x	1.675	1.700	1.725
RP506K181x	1.773	1.800	1.827
RP506K191x	1.872	1.900	1.928
RP506K201x	1.97	2.000	2.03
RP506K211x	2.069	2.100	2.131
RP506K221x	2.167	2.200	2.233
RP506K231x	2.266	2.300	2.334
RP506K241x	2.364	2.400	2.436
RP506K251x	2.463	2.500	2.537
RP506K261x	2.561	2.600	2.639
RP506K271x	2.66	2.700	2.74
RP506K281x	2.758	2.800	2.842
RP506K291x	2.857	2.900	2.943
RP506K301x	2.955	3.000	3.045
RP506K311x	3.054	3.100	3.146
RP506K321x	3.152	3.200	3.248
RP506K331x	3.251	3.300	3.349

TYPICAL APPLICATION

PG function is used, 30 ms Soft-start Time

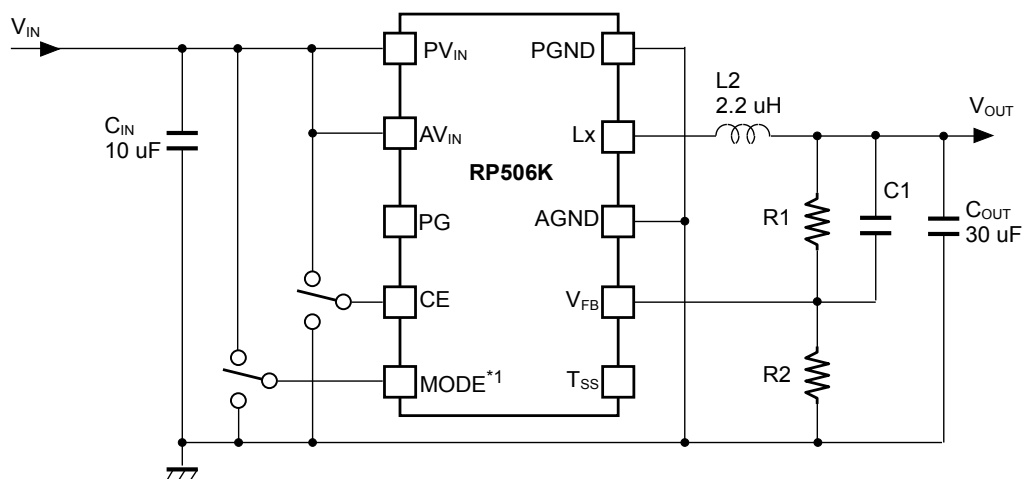
***1 MODE = "H": forced PWM control, MODE = "L": PWM/VFM auto switching control**



RP506Kxx1A/B/D/E (Fixed Output Voltage Type)

PG function is not used, 150 μ s Soft-start Time

***1 MODE = "H": forced PWM control, MODE = "L": PWM/VFM auto switching control**



RP506K001C/F (Adjustable Output Voltage Type)

Table 1. Recommended External Components

Symbol	Size	Part Description	Model
C _{IN}	10 μ F	Ceramic Capacitor	C1608JB0J106M (TDK)
			JMK107BJ106MA (Taiyo Yuden)
C _{OUT}	22 μ F x 2	Ceramic Capacitor	C2012JB0J226M (TDK)
	10 μ F x 3	Ceramic Capacitor	C1608JB0J106M (TDK)
L (V _{SET} $\leq$ 3.3V)	2.2 μ H	Inductor	JMK107BJ106MA (Taiyo Yuden)
			SLF6045T-2R2N3R3 (TDK)
			CLF7045T-2R2N (TDK)
			FDSD0415-2R2M (TOKO)
L (V _{SET} $>$ 3.3V ^{*1})	4.7 μ H	Inductor	RLF7030T-2R2M5R4 (TDK)
			SLF6045T-4R7N2R4 (TDK)
			CLF7045T-4R7N (TDK)
			FDSD0415-4R7M (TOKO)
			RLF7030T-4R7M3R4 (TDK)

^{*1} Only for RP506K001C/F

Small and Low Profile External Components

Symbol	Size	Part Description	Model
L (V _{SET} $\leq$ 1.5V)	1.0 μ H	Inductor	DFE252010R-H-1R0M (TOKO)
			VLS252010HBX-1R0M (TDK)
L (V _{SET} $\leq$ 2.3V)	1.5 μ H	Inductor	DFE252010R-H-1R5M (TOKO)
			VLS252010HBX-1R5M (TDK)
L	2.2 μ H	Inductor	DFE252010R-H-2R2M (TOKO)
			VLS252010HBX-2R2M (TDK)

TECHNICAL NOTES

The performance of power source circuits using this IC largely depends on peripheral circuits. When selecting the peripheral components, please consider the conditions of use. Do not allow each component, PCB pattern or the IC to exceed their respected rated values (voltage, current, and power) when designing the peripheral circuits.

- AGND and PGND must be wired to the GND plane when mounting on boards.
- AV_{IN} and PV_{IN} must be wired to the V_{IN} plane when mounting on boards.
- Ensure the A/PV_{IN} and A/PGND lines are sufficiently robust. A large switching current flows through the A/ PGND line, the V_{DD} line, the V_{OUT} line, an inductor, and L_x. If their impedance is too high, noise pickup or unstable operation may result. Set the external components as close as possible to the IC and minimize the wiring between the components and the IC. Especially, place a capacitor (C_{IN}) as close as possible to the PV_{IN} pin and PGND. For the RP506Kxx1A/B/D/E, separate the wiring between the V_{OUT} pin and an inductor (L1) from the wiring between L1 and Load. Likewise, for the RP506K001C/F, separate the wiring between a resistor for setting output voltage (R1) and an inductor (L2) from the wiring between L2 and Load.
- Choose a low ESR ceramic capacitor. The ceramic capacitance of C_{IN} should be more than or equal to 10 μ F. For a ceramic capacitor (C_{OUT}), it is recommended that three paralleled 10 μ F ceramic capacitors or two paralleled 22 μ F ceramic capacitors be used.
- When V_{SET} $\leq$ 3.3 V, a 2.2 μ H inductor is recommended for RP506Kxx1A/B/C/D/E/F. When V_{SET} $\leq$ 2.3 V, a 1.5 μ H inductor can be used for RP506Kxx1A/B/C. When V_{SET} $\leq$ 1.5 V, a 1 μ H inductor can be used for RP506Kxx1A/B/C. When V_{SET} > 3.3 V, a 4.7 μ H inductor is recommended for RP506K001C/F. The phase compensation of this IC is designed according to the C_{OUT} and L values. Choose an inductor that has small DC resistance, has enough allowable current and is hard to cause magnetic saturation. If the inductance value of an inductor is extremely small, the peak current of L_x may increase along with the load current. As a result, over current protection circuit may start to operate when the peak current of L_x reaches to “L_x limit current”.

Set Output Voltage Range vs. Inductance Range

Version	RP506Kxx1A/B			RP506Kxx1D/E
V _{SET} (V)	L = 1.0 μ H	L = 1.5 μ H	L = 2.2 μ H	L = 2.2 μ H
up to 1.5	Acceptable	Acceptable	Recommended	Recommended
1.6 to 2.3	-	Acceptable	Recommended	Recommended
2.4 to 3.3	-	-	Recommended	Recommended

Version	RP506K001C				RP506K001F	
V _{SET} (V)	L = 1.0 μ H	L = 1.5 μ H	L = 2.2 μ H	L = 4.7 μ H	L = 2.2 μ H	L = 4.7 μ H
up to 1.5	Acceptable	Acceptable	Recommended	-	Recommended	-
1.6 to 2.3	-	Acceptable	Recommended	-	Recommended	-
2.4 to 3.3	-	-	Recommended	-	Recommended	-
3.4 or more	-	-	-	Recommended	-	Recommended

- Over current protection circuit and latch type protection circuit may be affected by self-heating or power dissipation environment.
- The output voltage (V_{SET}) is adjustable by changing the resistance values of resistors ($R1$, $R2$) as follows.

$$V_{SET} = V_{FB} \times (R1 + R2) / R2 \text{ (Recommended } V_{OUT} \text{ range for RP506K001F: } 0.6 \text{ V} \leq V_{SET} \leq 4.0 \text{ V)}$$

$$\text{(Recommended } V_{OUT} \text{ range for RP506K001C: } 0.8 \text{ V} \leq V_{SET} \leq 4.0 \text{ V)}$$

If $R1$ and $R2$ are too large, the impedances of V_{FB} also become large, as a result, the IC could be easily affected by noise. For this reason, $R2$ should be 220 k Ω or less. If the operation becomes unstable due to the high impedances, the impedances should be decreased.

$C1$ can be calculated by the following equations. Please use the value close to the calculation result.

If the output voltage is lower than or equal to 3.3 V:

$$C1 = 4.84 \times 10^{-6} / R2 \text{ [F]}$$

If the output voltage exceeds 3.3 V:

$$C1 = 1.50 \times 10^{-6} / R2 \text{ [F]}$$

The recommended resistance values for $R1$ and $C1$ when $R2 = 220 \text{ k}\Omega$ or $100 \text{ k}\Omega$ are as follows.

Set Output Voltage (V_{SET}) vs. Resistors ($R1$, $R2$) and Capacitor ($C1$)

V_{SET} [V]	0.6	0.7	0.8	1.2	1.8	2.5	3.3	3.8	4.0
$R1$ [k Ω]	0	36.7	73.3	220	440	697	990	533	567
$R2$ [k Ω]	220	220	220	220	220	220	220	100	100
$C1$ [pF]	-	22	22	22	22	22	22	15	15

- Soft-start Time (t_{start}) is adjustable by connecting a capacitor (C_{SS}) between the TSS pin and GND. The capacitance value for C_{SS} that is suitable for t_{start} can be calculated by the following equation.

$$C_{SS} \text{ (nF)} = 3.5 \times t_{start} \text{ (ms)}$$

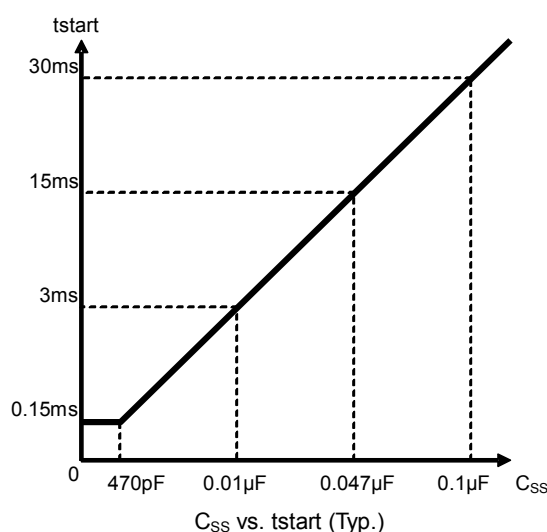
The TSS pin must be open if Soft-start time function is not used. Soft-start time is set to typically 150 μ s when the TSS pin is open.

- When using the power good function, the resistance value of a resistor (R_{PG}) should be between 10 k Ω to 100 k Ω . The PG pin must be open or connected to GND if the power good function is not used.

SOFT START TIME ADJUSTMENT AND POWER GOOD FUNCTIONS

Soft-start Time Adjustment Function

Soft-start time (t_{start}) of the RP506K is adjustable by connecting a soft-start time adjustment capacitor (C_{SS}) between the TSS pin and GND. t_{start} can be set from Typ. 0.15 ms. As Figure 6 shows, if 0.1 μ F C_{SS} is connected, t_{start} will be 30 ms. The TSS pin must be open if the soft-start time function is not used. t_{start} is set to 0.15 ms (Typ.) when the TSS pin is open.



Soft-start Time (t_{start}) vs. Soft-start Time Adjustment Capacitor (C_{SS})

Power Good Function

The RP506K contains a power good function using Nch open drain. If any abnormal condition is detected, the power good function turns Nch transistor on and switches the PG pin to low. If the cause of the abnormal condition is removed, the power good function turns Nch transistor off and switches the PG pin back to high. After the recovery from abnormal condition, it takes typically 0.05 ms for the IC to turn Nch transistor off. The followings are the abnormal conditions that the power good function can detect.

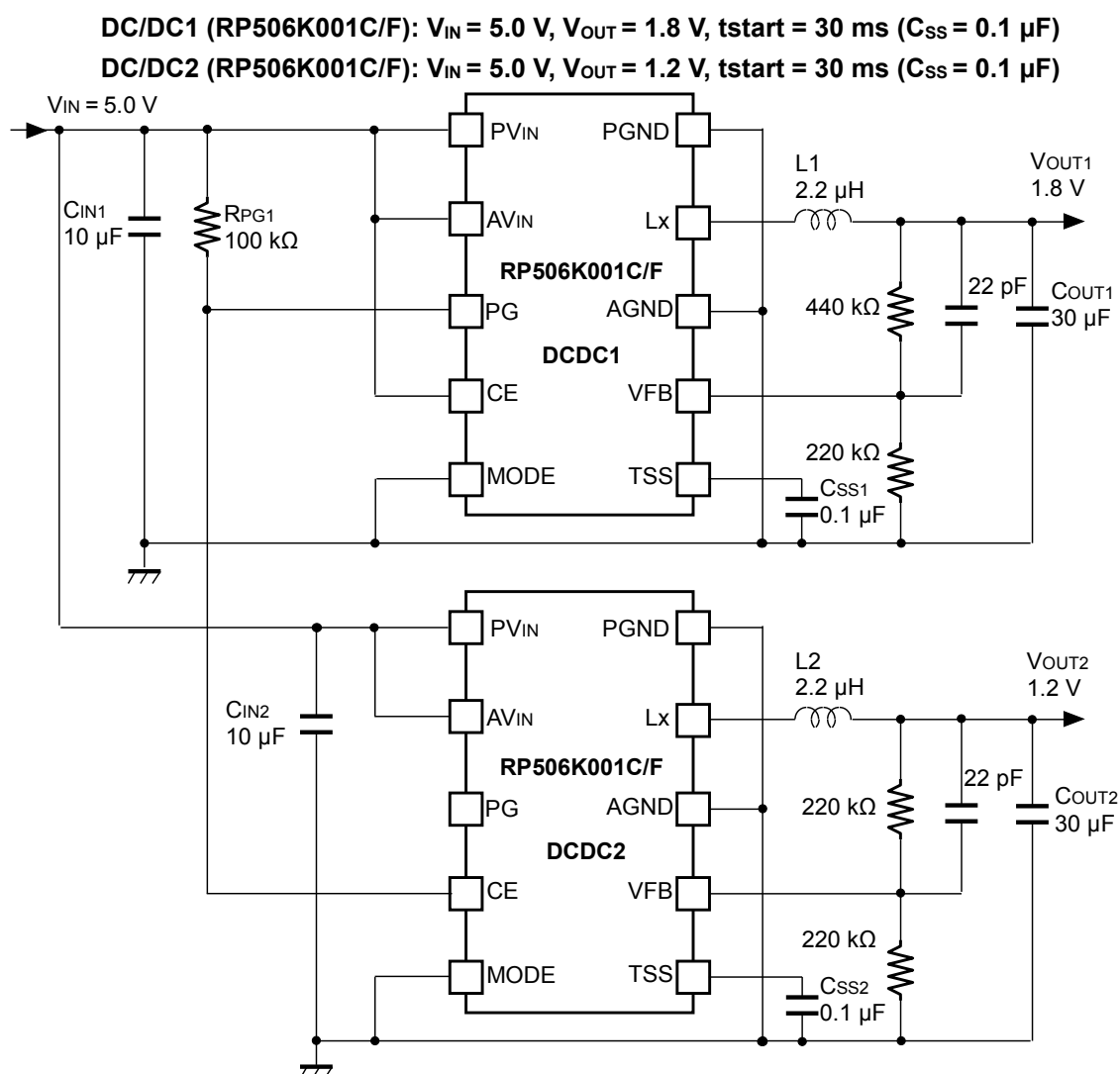
- CE = "L" (Shut down)
- UVLO (Shut down)
- Thermal Shutdown
- Over Voltage Detection: Typ. $V_{SET} \times 1.2$ V (RP506Kxx1A/B/D/E) or 0.72 V (RP506K001C/F)
- Under Voltage Detection: Typ. $V_{SET} \times 0.8$ V (RP506Kxx1A/B/D/E) or 0.48 V (RP506K001C/F)
- Latch Type Protection

Notes: When using the power good function, the resistance of PG pin (R_{PG}) should be between 10 k Ω to 100 k Ω . The PG pin must be open or connected to GND if the power good function is not used.

SEQUENTIAL START-UP USING SOFT-START TIME ADJUSTMENT AND POWER GOOD FUNCTIONS

Sequential startup circuits can be built by using soft-start time adjustment and power good functions of the RP506K. The figure below is an example of sequential startup circuits using DC/DC1 and DC/DC2.

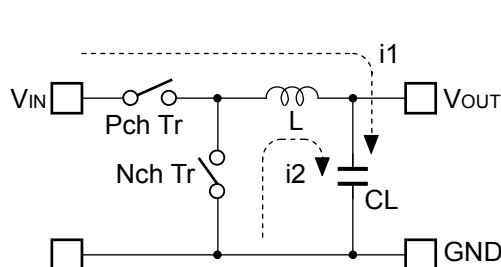
The DC/DC1 starts up first followed by the DC/DC2: the output of DC/DC1 reaches 1.44 V ($V_{SET} \times 0.8$), the PG pin of DC/DC1 sends a high signal to the CE pin of DC/DC2, and then the DC/DC2 starts soft-start.



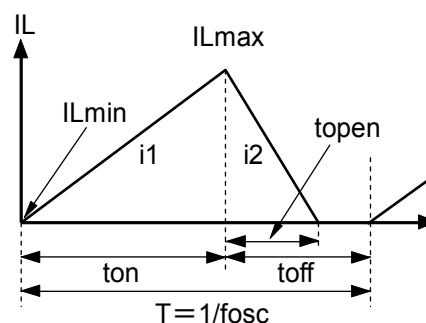
Circuits Example using Sequential Startup

OPERATION OF STEP-DOWN DC/DC CONVERTER AND OUTPUT CURRENT

The step-down DC/DC converter charges energy in the inductor when L_X Tr. turns "ON", and discharges the energy from the inductor when L_X Tr. turns "OFF" and controls with less energy loss, so that a lower output voltage (V_{OUT}) than the input voltage (V_{IN}) can be obtained. The operation of the step-down DC/DC converter is explained in the following figures.



Basic Circuit



Inductor Current (I_L) flowing through Inductor (L)

- Step1.** Pch Tr. turns "ON" and I_L (i_1) flows, L is charged with energy. At this moment, i_1 increases from the minimum inductor current (I_{Lmin}), which is 0 A, and reaches the maximum inductor current (I_{Lmax}) in proportion to the on-time period (ton) of Pch Tr.
- Step2.** When Pch Tr. turns "OFF", L tries to maintain I_L at I_{Lmax} , so L turns Nch Tr. "ON" and I_L (i_2) flows into L .
- Step3.** i_2 decreases gradually and reaches I_{Lmin} after the open-time period ($topen$) of Nch Tr., and then Nch Tr. turns "OFF". This is called discontinuous current mode.
- As the output current (I_{OUT}) increases, the off-time period ($toff$) of Pch Tr. runs out before I_L reaches I_{Lmin} . The next cycle starts, and Pch Tr. turns "ON" and Nch Tr. turns "OFF", which means I_L starts increasing from I_{Lmin} . This is called continuous current mode.

In the case of PWM mode, V_{OUT} is maintained by controlling ton . During PWM mode, the oscillator frequency ($fosc$) is being maintained constant.

When the step-down DC/DC operation is constant, I_{Lmin} and I_{Lmax} during ton of Pch Tr. would be same as during $toff$ of Pch Tr. The current differential between I_{Lmax} and I_{Lmin} is described as ΔI .

$$\Delta I = I_{Lmax} - I_{Lmin} = V_{OUT} \times topen / L = (V_{IN} - V_{OUT}) \times ton / L \dots \dots \dots \text{Equation 1}$$

However,

$$T = 1 / fosc = ton + toff$$

$$\text{duty (\%)} = ton / T \times 100 = ton \times fosc \times 100$$

$$topen \leq toff$$

In Equation 1, " $V_{OUT} \times topen / L$ " shows the amount of current change in "ON" state. Also, " $(V_{IN} - V_{OUT}) \times ton / L$ " shows the amount of current change at "OFF" state.

Discontinuous Mode and Continuous Mode

As illustrated in Figure A, when I_{OUT} is relatively small, $t_{open} < t_{off}$. In this case, the energy charged into L during t_{on} will be completely discharged during t_{off} , as a result, $I_{Lmin} = 0$. This is called discontinuous mode. When I_{OUT} is gradually increased, eventually $t_{open} = t_{off}$ and when I_{OUT} is increased further, eventually $I_{Lmin} > 0$, as illustrated in Figure B. This is called continuous mode.

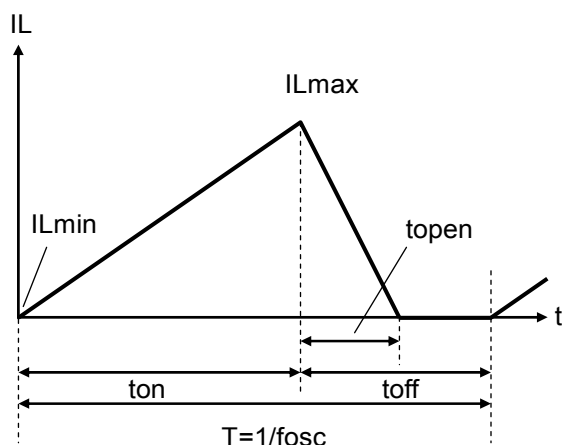


Figure A. Discontinuous Mode

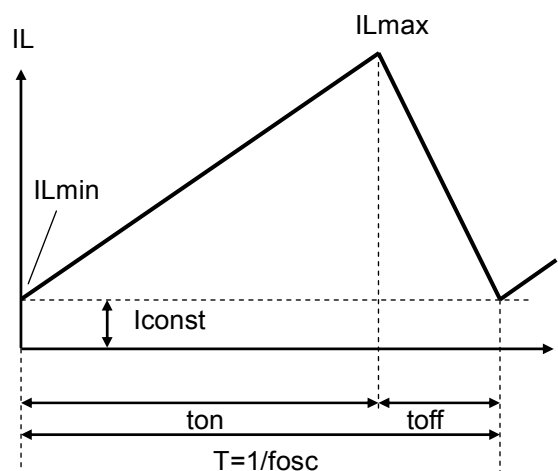


Figure B. Continuous Mode

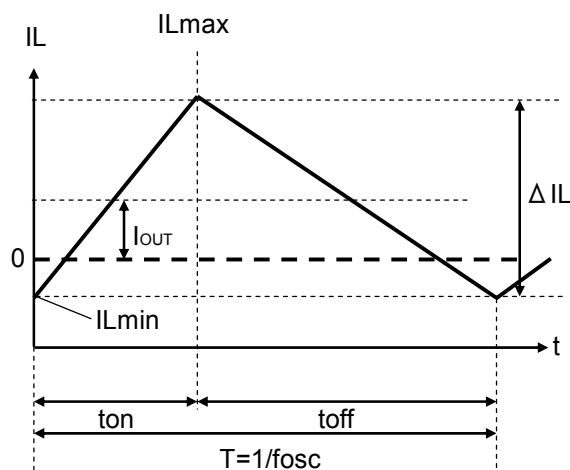
In the continuous mode, the solution of Equation 1 is described as t_{onc} .

$$t_{onc} = T_a \times V_{OUT} / V_{IN} \dots \dots \dots \text{Equation 2}$$

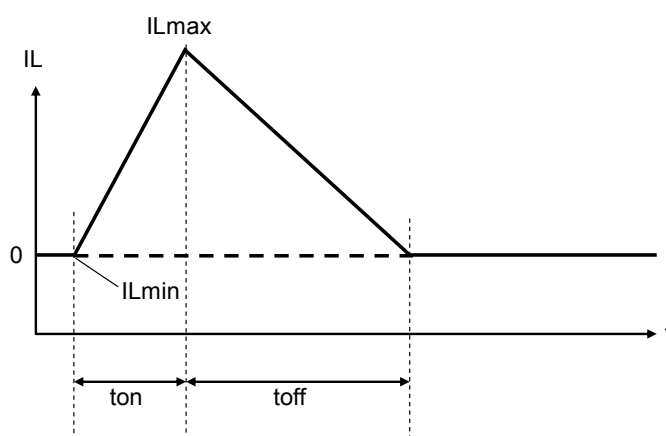
When $t_{on} < t_{onc}$, it is discontinuous mode, and when $t_{on} = t_{onc}$, it is continuous mode.

Forced PWM Mode

By setting the MODE pin to “H”, the IC switches the frequency at the fixed rate to reduce noise even when the output load is light. Therefore, when I_{OUT} is $\Delta I_L/2$ or less, I_{Lmin} becomes less than 0. That is, the accumulated electricity in CL is discharged through the IC side while I_L is increasing from I_{Lmin} to 0 during t_{on} , and also while I_L is decreasing from 0 to I_{Lmin} during t_{off} .

**Forced PWM Mode****VFM Mode**

By setting the MODE pin to “L”, in low output current, the IC automatically switches into VFM mode in order to achieve high efficiency. In VFM mode, t_{on} is forced to end when the inductor current reaches the pre-set I_{Lmax} . In the VFM mode, I_{Lmax} is typically set to 400 mA for the RP506Kxx1A/B/C, and 550 mA for the RP506Kxx1D/E/F. When t_{on} reaches 1.5 times of $T = 1 / f_{osc}$, t_{on} will be forced to end even if the inductor current is not reached I_{Lmax} .

**VFM Mode**

OUTPUT CURRENT AND SELECTION OF EXTERNAL COMPONENTS

The following equations explain the relationship between output current and peripheral components that are listed in *Table1. Recommended External Components in TYPICAL APPLICATION*.

Ripple Current P-P value is described as I_{RP} , ON resistance of Pch Tr. is described as R_{ONP} , ON resistance of Nch Tr. is described as R_{ONN} , and DC resistor of the inductor is described as R_L .

First, when Pch Tr. is "ON", the following equation is satisfied.

$$V_{IN} = V_{OUT} + (R_{ONP} + R_L) \times I_{OUT} + L \times I_{RP} / t_{on} \quad \text{Equation 3}$$

Second, when Pch Tr. is "OFF" (Nch Tr. is "ON"), the following equation is satisfied.

$$L \times I_{RP} / t_{off} = R_{ONN} \times I_{OUT} + V_{OUT} + R_L \times I_{OUT} \quad \text{Equation 4}$$

Put Equation 4 into Equation 3 to solve ON duty of Pch Tr. ($D_{ON} = t_{on} / (t_{off} + t_{on})$):

$$D_{ON} = (V_{OUT} + R_{ONN} \times I_{OUT} + R_L \times I_{OUT}) / (V_{IN} + R_{ONN} \times I_{OUT} - R_{ONP} \times I_{OUT}) \quad \text{Equation 5}$$

Ripple Current is described as follows:

$$I_{RP} = (V_{IN} - V_{OUT} - R_{ONP} \times I_{OUT} - R_L \times I_{OUT}) \times D_{ON} / f_{osc} / L \quad \text{Equation 6}$$

Peak current that flows through L, and L_X Tr. is described as follows:

$$I_{LXmax} = I_{OUT} + I_{RP} / 2 \quad \text{Equation 7}$$

Notes: Please consider I_{LXmax} when setting conditions of input and output, as well as selecting the external components. The above calculation formulas are based on the ideal operation of the ICs in continuous mode.

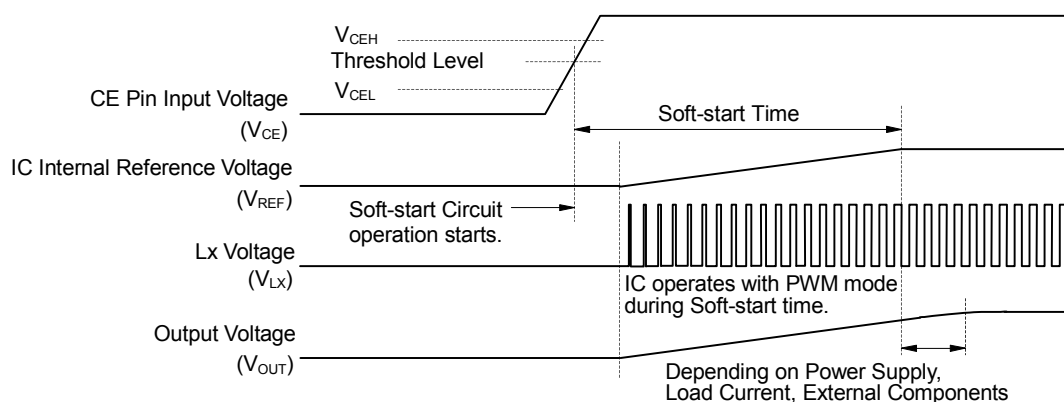
TIMING CHART

(1) Soft-start Time

Starting-up with CE Pin

The IC starts to operate when the CE pin voltage (V_{CE}) exceeds the threshold voltage. The threshold voltage is preset between CE “H” input voltage (V_{CEH}) and CE “L” input voltage (V_{CEL}).

After the start-of the start-up of the IC, soft-start circuit starts to operate. Then, after a certain period of time, the reference voltage (V_{REF}) in the IC gradually increases up to the specified value.



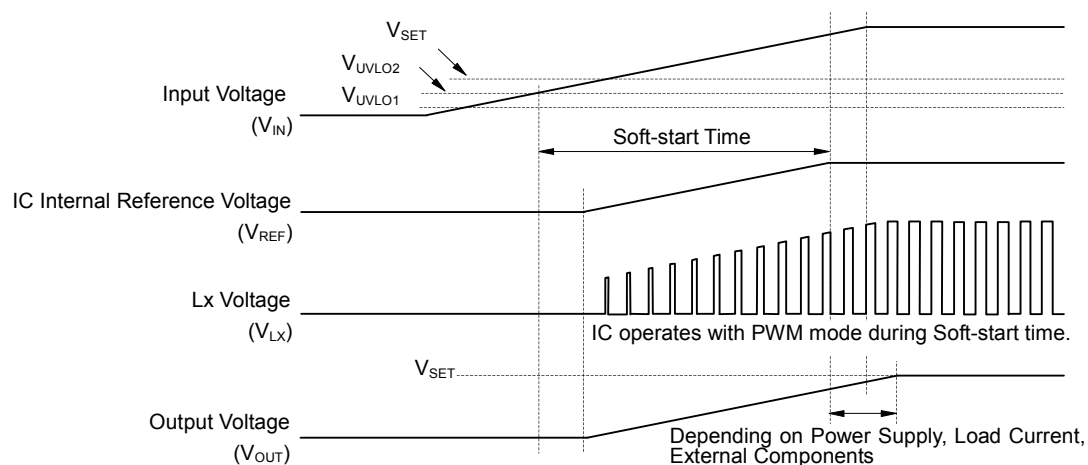
Timing Chart

Soft-start time starts when soft-start circuit is activated, and ends when the reference voltage reaches the specified voltage.

Notes: Soft start time is not always equal to the turn-on speed of the step-down DC/DC converter. Please note that the turn-on speed could be affected by the power supply capacity, the output current, the inductance value and the C_{OUT} value.

Starting-up with Power Supply

After the power-on, when V_{IN} exceeds the UVLO released voltage (V_{UVLO2}), the IC starts to operate. Then, soft-start circuit starts to operate and after a certain period of time, V_{REF} gradually increases up to the specified value. Soft-start time starts when soft-start circuit is activated, and ends when V_{REF} reaches the specified value.



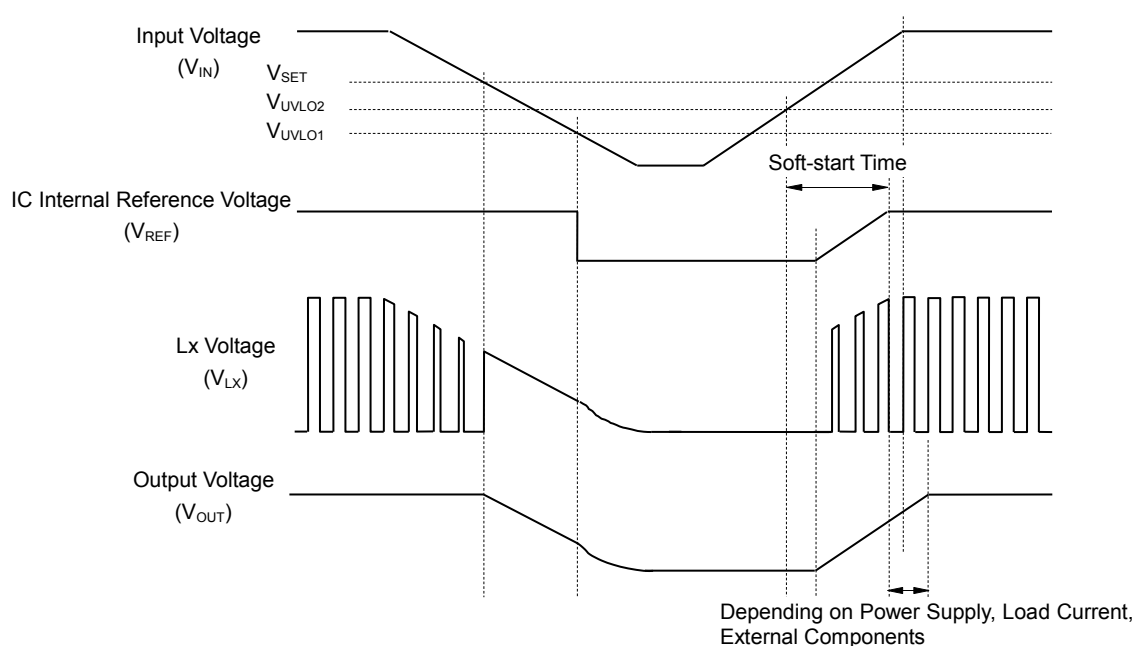
Notes: Please note that the turn-on speed of V_{OUT} could be affected by the power supply capacity, the output current, the inductance value, the C_{OUT} value and the turn-on speed of V_{IN} determined by C_{IN} .

(2) Under Voltage Lockout (UVLO) Circuit

If V_{IN} becomes lower than V_{SET} , the step-down DC/DC converter stops the switching operation and ON duty becomes 100%, and then V_{OUT} gradually drops according to V_{IN} .

If the V_{IN} drops more and becomes lower than the UVLO detector threshold (V_{UVLO1}), the UVLO circuit starts to operate, V_{REF} stops, and Pch and Nch built-in switch transistors turn "OFF". As a result, V_{OUT} drops according to the C_{OUT} capacitance value and the load.

To restart the operation, V_{IN} needs to be higher than V_{UVLO2} . The timing chart below shows the voltage shifts of V_{REF} , V_{LX} and V_{OUT} when V_{IN} value is varied.



Timing Chart

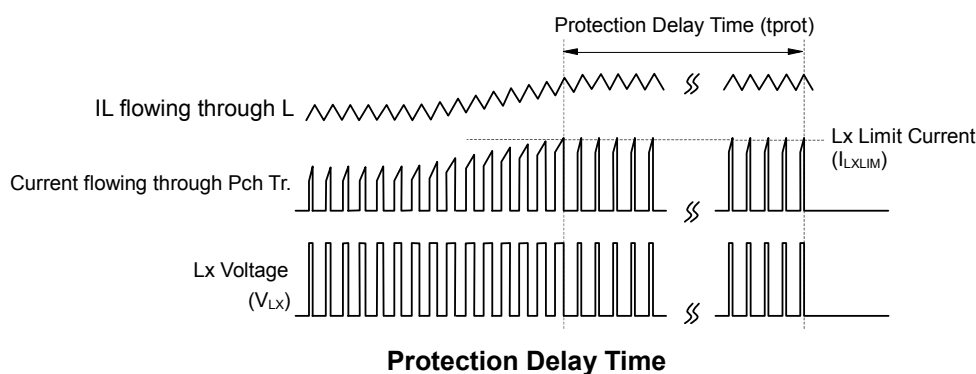
Notes: Falling edge (operating) and rising edge (releasing) waveforms of V_{OUT} could be affected by the initial voltage of C_{OUT} and the output current of V_{OUT} .

(3) Over Current Protection Circuit, Latch Type Protection Circuit

Over current protection circuit supervises the inductor peak current (the peak current flowing through Pch Tr.) in each switching cycle, and if the current exceeds the Lx current limit (I_{LXLIM}), it turns off Pch Tr. I_{LXLIM} of the RP506K is set to Typ.2800 mA.

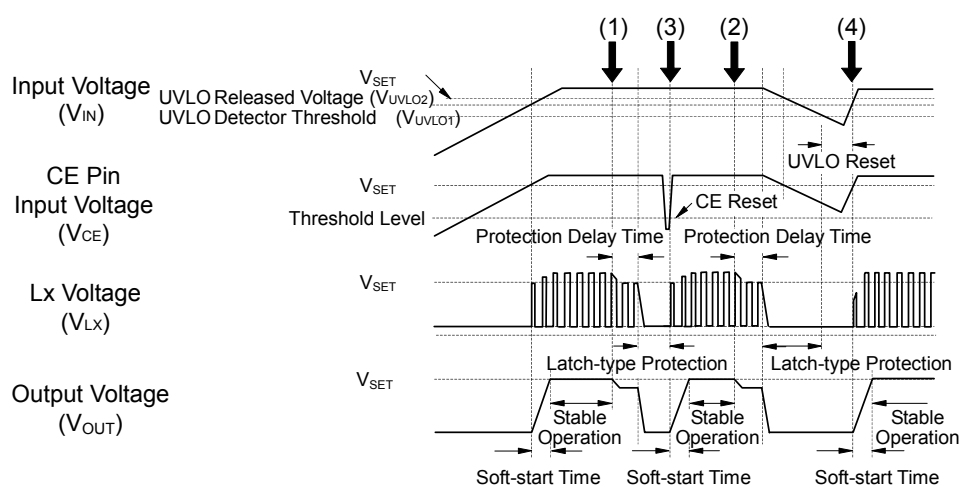
Latch type protection circuit latches the built-in driver to the OFF state and stops the operation of the step-down DC/DC converter if the over current status continues or V_{OUT} continues being the half of the setting voltage for equal or longer than protection delay time (t_{prot}). To release the latch type protection circuit, restart the IC by inputting "L" signal to the CE pin, or restart the IC with power-on or make the supply voltage lower than V_{UVLO1} .

Notes: I_{LXLIM} and t_{prot} could be easily affected by self-heating or ambient environment. If the V_{IN} drops dramatically or becomes unstable due to short-circuit, protection operation and t_{prot} could be affected.



The timing chart below shows the voltage shift of V_{CE} , V_{LX} and V_{OUT} when the IC status is changed by the following orders: V_{IN} rising → stable operation → high load → CE reset → stable operation → V_{IN} falling → V_{IN} recovering (UVLO reset) → stable operation.

- (1)(2) If the large current flows through the circuit or the IC goes into low V_{OUT} condition due to short-circuit or other reasons, the latch type protection circuit latches the built-in driver to "OFF" state after t_{prot} . Then, V_{LX} becomes "L" and V_{OUT} turns "OFF".
- (3) The latch type protection circuit is released by CE reset, which puts the IC into "L" once with the CE pin and back into "H".
- (4) The latch type protection circuit is released by UVLO reset, which makes V_{IN} lower than V_{UVLO1} .

**Timing Chart**

PACKAGE INFORMATION

POWER DISSIPATION (DFN(PLP)2527-10)

Power Dissipation (P_D) of the package is dependent on PCB material, layout, and environmental conditions. The following conditions are used in this measurement.

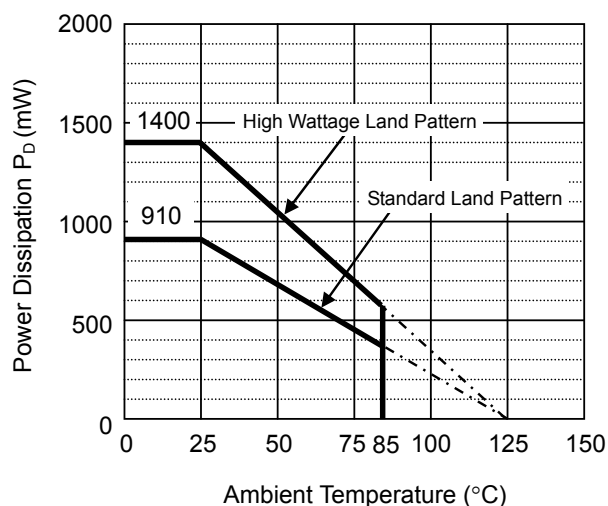
Measurement Conditions

	High Wattage Land Pattern	Standard Test Land Pattern
Environment	Mounting on Board (Wind Velocity = 0 m/s)	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (4-Layers)	Glass Cloth Epoxy Plastic (Double-sided)
Board Dimensions	35 mm x 90 mm x 0.8 mm	40 mm x 40 mm x 1.6 mm
Copper Ratio	Each layers: Approx. 15%	Top side: Approx. 50%, Back side: Approx. 50%
Through-holes	ϕ 0.30 mm x 9 pcs ϕ 0.50 mm x 10 pcs	ϕ 0.54 mm x 30 pcs

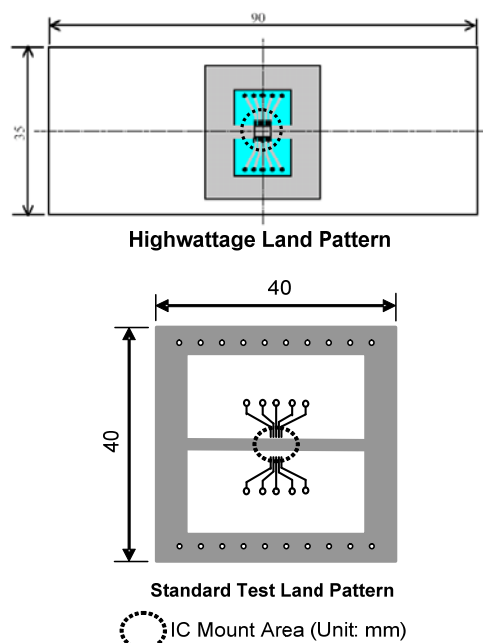
Measurement Result

($T_a = 25^\circ\text{C}$, $T_{j\text{max}} = 125^\circ\text{C}$)

	High Wattage Land Pattern	Standard Test Land Pattern
Power Dissipation	1400 mW ($T_{j\text{max}} = 125^\circ\text{C}$)	910 mW ($T_{j\text{max}} = 125^\circ\text{C}$)
Thermal Resistance	$\theta_{ja} = (125 - 25^\circ\text{C}) / 1.4\text{W} = 71^\circ\text{C/W}$	$\theta_{ja} = (125 - 25^\circ\text{C}) / 0.91\text{W} = 110^\circ\text{C/W}$

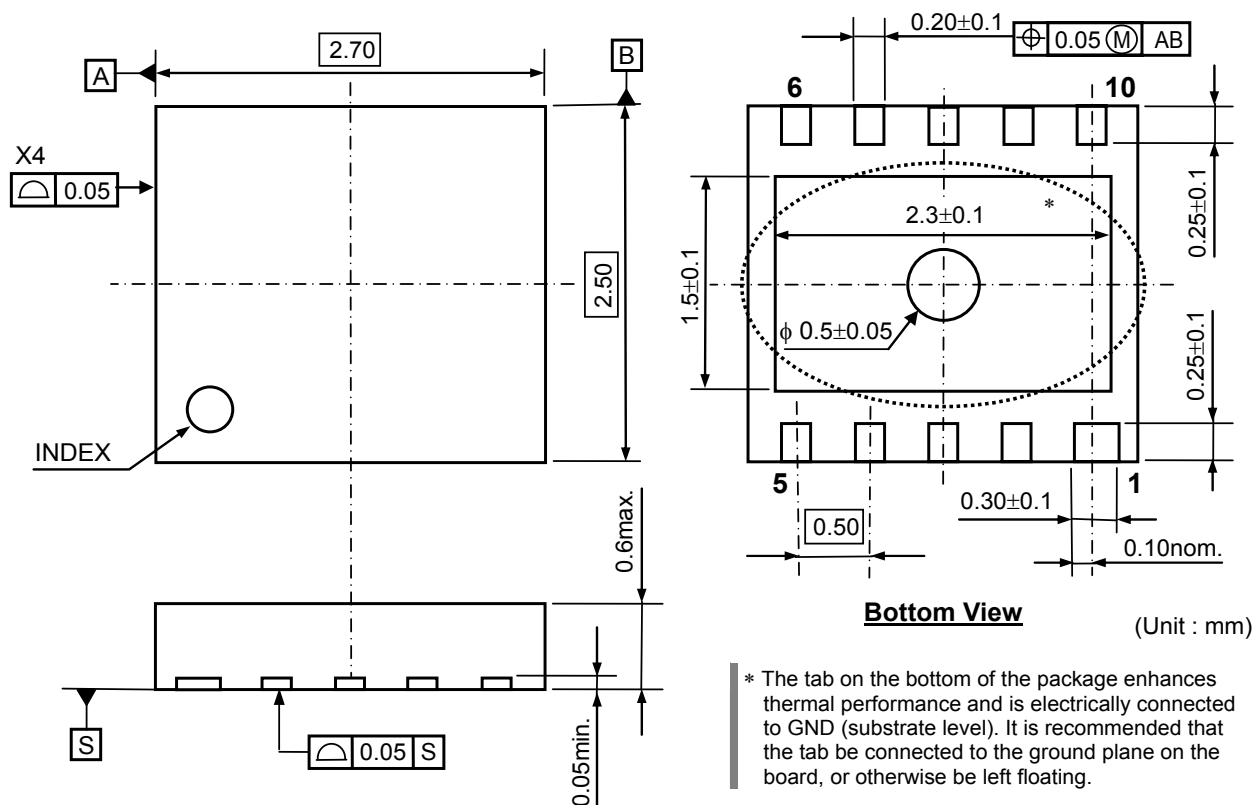


Power Dissipation vs. Ambient Temperature



Measurement Board Pattern

PACKAGE DIMENSIONS (DFN(PLP)2527-10)

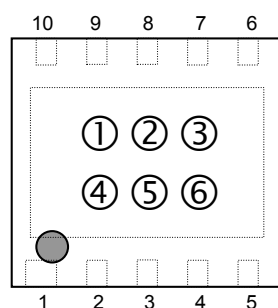


DFN (PLP) 2527-10 Package Dimensions

MARK SPECIFICATION (DFN(PLP)2527-10)

①②③④: Product Code ... **Refer to MARK SPECIFICATION TABLE (DFN(PLP)2527-10)**

⑤⑥: Lot Number ... Alphanumeric Serial Number



DFN (PLP) 2527-10 Mark Specification

MARK SPECIFICATION TABLE (DFN (PLP) 2527-10)

Product Name	①②③④	⑤⑥	Product Name	①②③④	⑤⑥
RP506K081A	D 0 8 A	Lot No.	RP506K061D	D 0 6 D	Lot No.
RP506K091A	D 0 9 A	Lot No.	RP506K071D	D 0 7 D	Lot No.
RP506K101A	D 1 0 A	Lot No.	RP506K081D	D 0 8 D	Lot No.
RP506K111A	D 1 1 A	Lot No.	RP506K091D	D 0 9 D	Lot No.
RP506K121A	D 1 2 A	Lot No.	RP506K101D	D 1 0 D	Lot No.
RP506K131A	D 1 3 A	Lot No.	RP506K111D	D 1 1 D	Lot No.
RP506K141A	D 1 4 A	Lot No.	RP506K121D	D 1 2 D	Lot No.
RP506K151A	D 1 5 A	Lot No.	RP506K131D	D 1 3 D	Lot No.
RP506K161A	D 1 6 A	Lot No.	RP506K141D	D 1 4 D	Lot No.
RP506K171A	D 1 7 A	Lot No.	RP506K151D	D 1 5 D	Lot No.
RP506K181A	D 1 8 A	Lot No.	RP506K161D	D 1 6 D	Lot No.
RP506K191A	D 1 9 A	Lot No.	RP506K171D	D 1 7 D	Lot No.
RP506K201A	D 2 0 A	Lot No.	RP506K181D	D 1 8 D	Lot No.
RP506K211A	D 2 1 A	Lot No.	RP506K191D	D 1 9 D	Lot No.
RP506K221A	D 2 2 A	Lot No.	RP506K201D	D 2 0 D	Lot No.
RP506K231A	D 2 3 A	Lot No.	RP506K211D	D 2 1 D	Lot No.
RP506K241A	D 2 4 A	Lot No.	RP506K221D	D 2 2 D	Lot No.
RP506K251A	D 2 5 A	Lot No.	RP506K231D	D 2 3 D	Lot No.
RP506K261A	D 2 6 A	Lot No.	RP506K241D	D 2 4 D	Lot No.
RP506K271A	D 2 7 A	Lot No.	RP506K251D	D 2 5 D	Lot No.
RP506K281A	D 2 8 A	Lot No.	RP506K261D	D 2 6 D	Lot No.
RP506K291A	D 2 9 A	Lot No.	RP506K271D	D 2 7 D	Lot No.
RP506K301A	D 3 0 A	Lot No.	RP506K281D	D 2 8 D	Lot No.
RP506K311A	D 3 1 A	Lot No.	RP506K291D	D 2 9 D	Lot No.
RP506K321A	D 3 2 A	Lot No.	RP506K301D	D 3 0 D	Lot No.
RP506K331A	D 3 3 A	Lot No.	RP506K311D	D 3 1 D	Lot No.
RP506K121A5	D 0 0 A	Lot No.	RP506K321D	D 3 2 D	Lot No.
RP506K081B	D 0 8 B	Lot No.	RP506K331D	D 3 3 D	Lot No.
RP506K091B	D 0 9 B	Lot No.	RP506K121D5	D 0 0 D	Lot No.
RP506K101B	D 1 0 B	Lot No.	RP506K061E	D 0 6 E	Lot No.
RP506K111B	D 1 1 B	Lot No.	RP506K071E	D 0 7 E	Lot No.
RP506K121B	D 1 2 B	Lot No.	RP506K081E	D 0 8 E	Lot No.
RP506K131B	D 1 3 B	Lot No.	RP506K091E	D 0 9 E	Lot No.
RP506K141B	D 1 4 B	Lot No.	RP506K101E	D 1 0 E	Lot No.
RP506K151B	D 1 5 B	Lot No.	RP506K111E	D 1 1 E	Lot No.
RP506K161B	D 1 6 B	Lot No.	RP506K121E	D 1 2 E	Lot No.
RP506K171B	D 1 7 B	Lot No.	RP506K131E	D 1 3 E	Lot No.
RP506K181B	D 1 8 B	Lot No.	RP506K141E	D 1 4 E	Lot No.
RP506K191B	D 1 9 B	Lot No.	RP506K151E	D 1 5 E	Lot No.
RP506K201B	D 2 0 B	Lot No.	RP506K161E	D 1 6 E	Lot No.
RP506K211B	D 2 1 B	Lot No.	RP506K171E	D 1 7 E	Lot No.
RP506K221B	D 2 2 B	Lot No.	RP506K181E	D 1 8 E	Lot No.
RP506K231B	D 2 3 B	Lot No.	RP506K191E	D 1 9 E	Lot No.
RP506K241B	D 2 4 B	Lot No.	RP506K201E	D 2 0 E	Lot No.
RP506K251B	D 2 5 B	Lot No.	RP506K211E	D 2 1 E	Lot No.
RP506K261B	D 2 6 B	Lot No.	RP506K221E	D 2 2 E	Lot No.
RP506K271B	D 2 7 B	Lot No.	RP506K231E	D 2 3 E	Lot No.
RP506K281B	D 2 8 B	Lot No.	RP506K241E	D 2 4 E	Lot No.
RP506K291B	D 2 9 B	Lot No.	RP506K251E	D 2 5 E	Lot No.
RP506K301B	D 3 0 B	Lot No.	RP506K261E	D 2 6 E	Lot No.
RP506K311B	D 3 1 B	Lot No.	RP506K271E	D 2 7 E	Lot No.
RP506K321B	D 3 2 B	Lot No.	RP506K281E	D 2 8 E	Lot No.
RP506K331B	D 3 3 B	Lot No.	RP506K291E	D 2 9 E	Lot No.
RP506K121B5	D 0 0 B	Lot No.	RP506K301E	D 3 0 E	Lot No.
RP506K001C	D 0 0 C	Lot No.	RP506K311E	D 3 1 E	Lot No.
			RP506K321E	D 3 2 E	Lot No.
			RP506K331E	D 3 3 E	Lot No.
			RP506K121E5	D 0 0 E	Lot No.
			RP506K001F	D 0 0 F	Lot No.

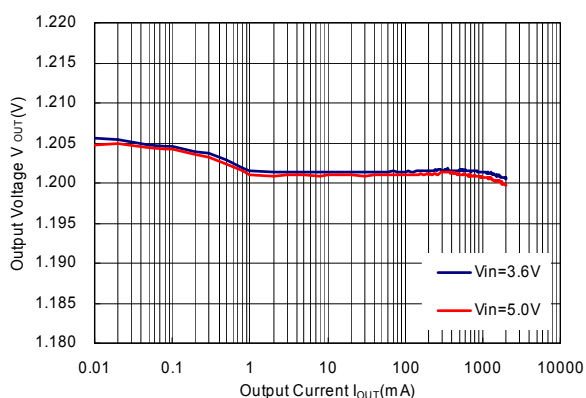
TYPICAL PERFORMANCE CHARACTERISTICS

Note: Typical Characteristics are intended to be used as reference data; they are not guaranteed.

1) Output Voltage vs. Output Current

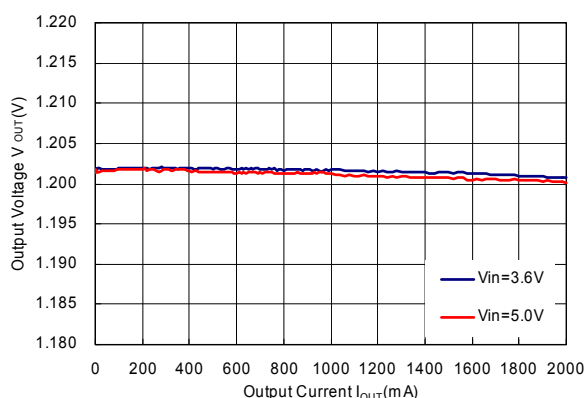
RP506Kxx1A/B/C $V_{OUT} = 1.2\text{ V}$

MODE = "L" PWM/VFM Auto Switching Control



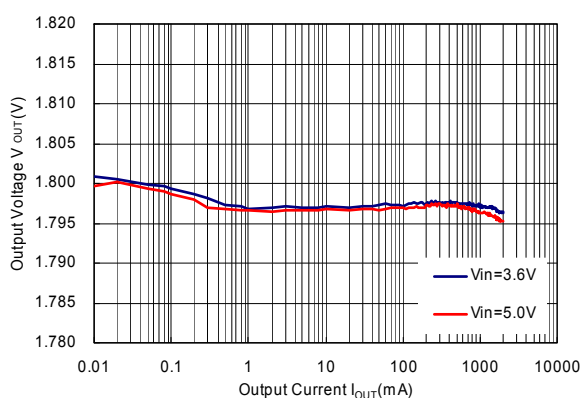
RP506Kxx1A/B/C $V_{OUT} = 1.2\text{ V}$

MODE = "H" Forced PWM Control



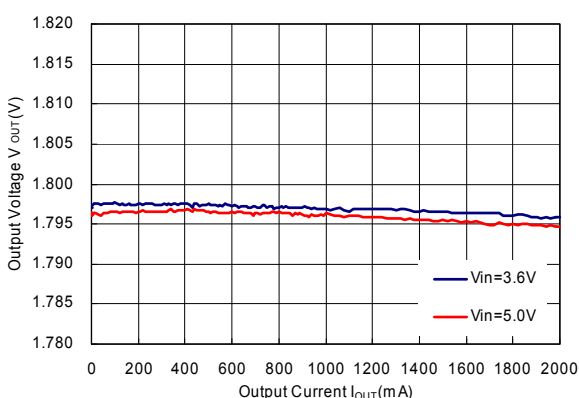
RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$

MODE = "L" PWM/VFM Auto Switching Control



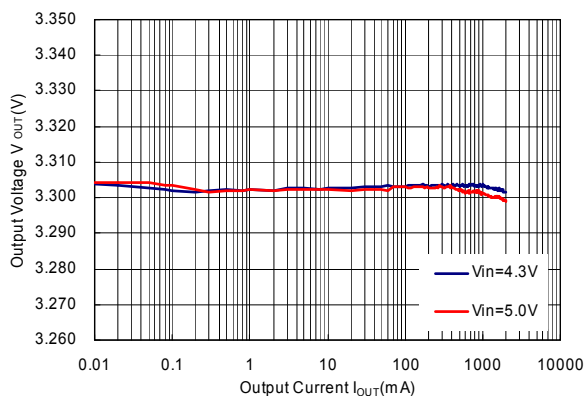
RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$

MODE = "H" Forced PWM Control



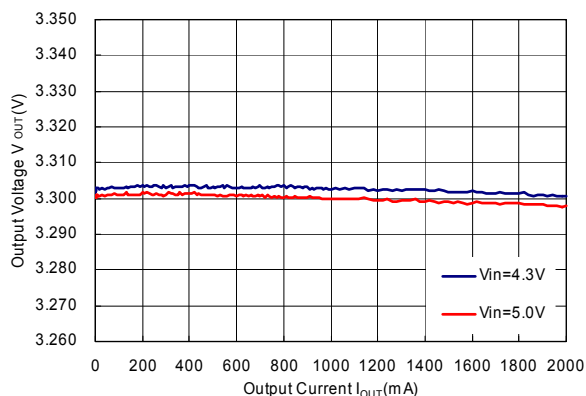
RP506Kxx1A/B/C $V_{OUT} = 3.3\text{ V}$

MODE = "L" PWM/VFM Auto Switching Control

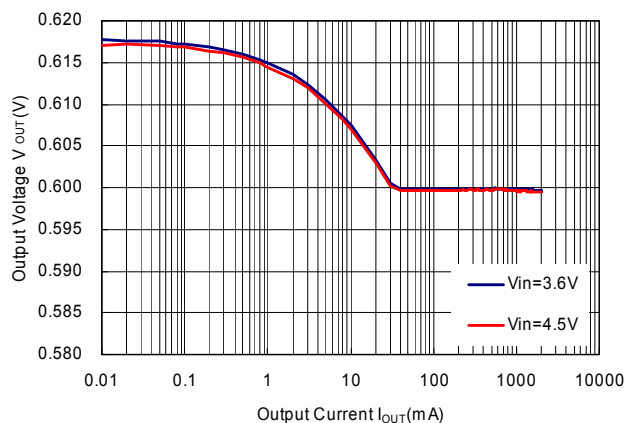


RP506Kxx1A/B/C $V_{OUT} = 3.3\text{ V}$

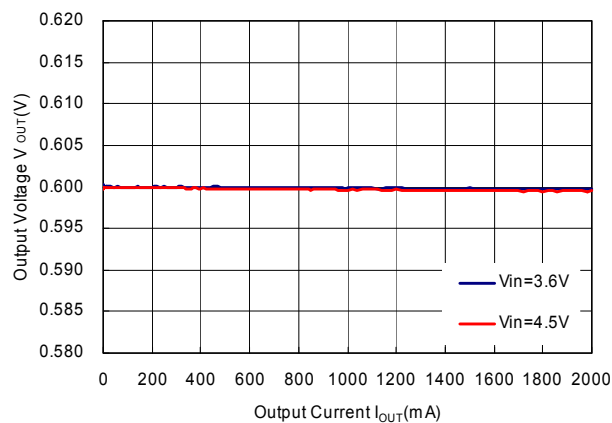
MODE = "H" Forced PWM Control



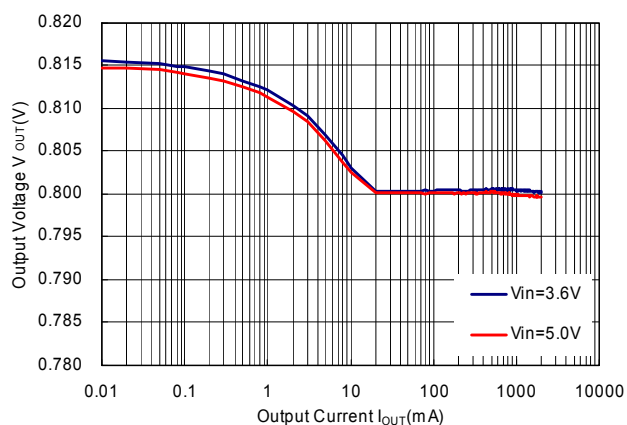
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$
MODE = "L" PWM/VFM Auto Switching Control



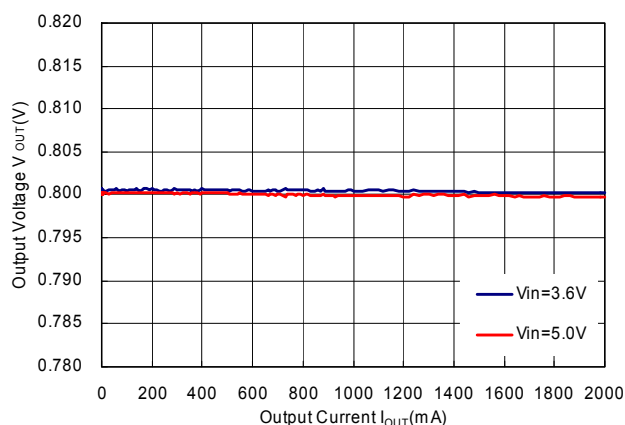
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$
MODE = "H" Forced PWM Control



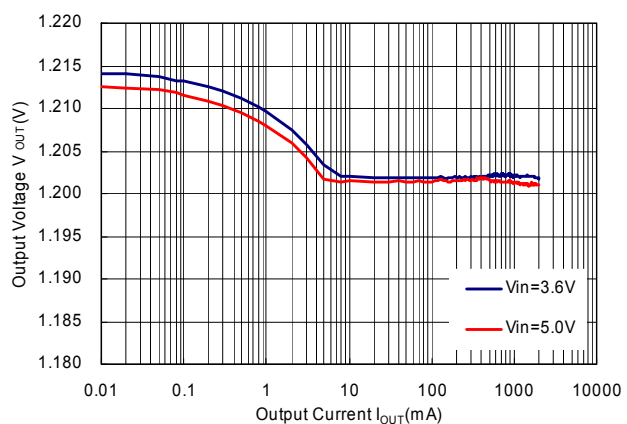
RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$
MODE = "L" PWM/VFM Auto Switching Control



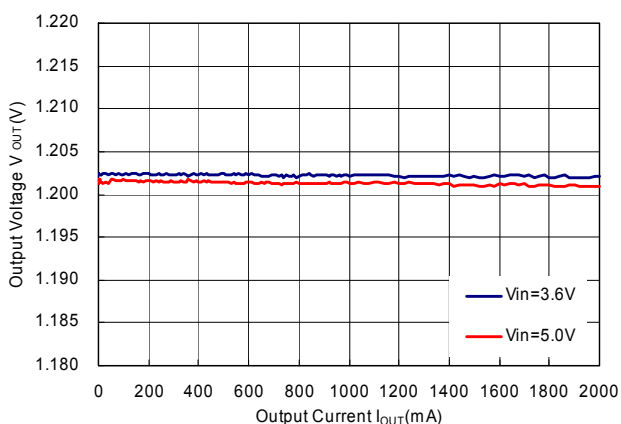
RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$
MODE = "H" Forced PWM Control



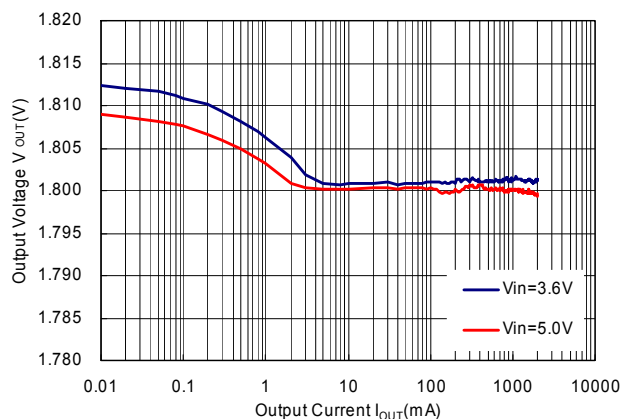
RP506Kxx1D/E/F $V_{OUT} = 1.2\text{ V}$
MODE = "L" PWM/VFM Auto Switching Control



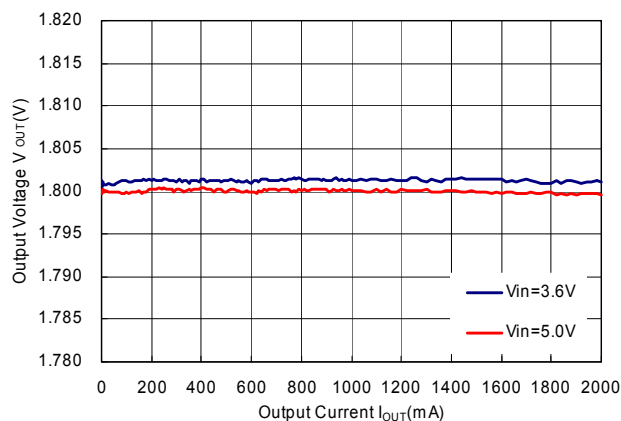
RP506Kxx1D/E/F $V_{OUT} = 1.2\text{ V}$
MODE = "H" Forced PWM Control



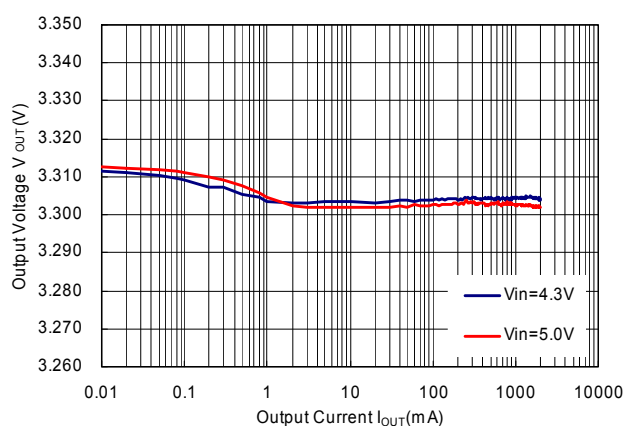
RP506Kxx1D/E/F $V_{OUT} = 1.8\text{ V}$
MODE = "L" PWM/VFM Auto Switching Control



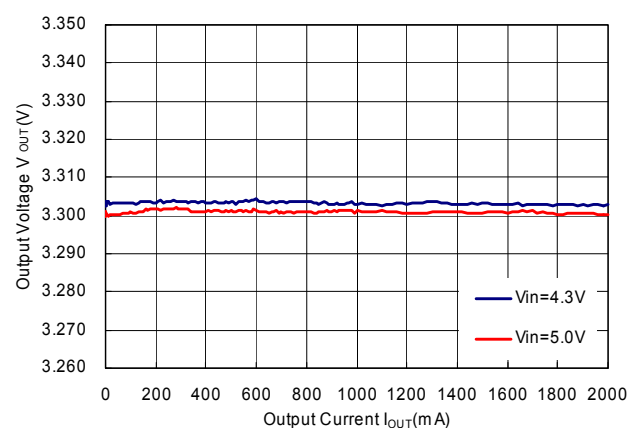
RP506Kxx1D/E/F $V_{OUT} = 1.8\text{ V}$
MODE = "H" Forced PWM Control



RP506Kxx1D/E/F $V_{OUT} = 3.3\text{ V}$
MODE = "L" PWM/VFM Auto Switching Control

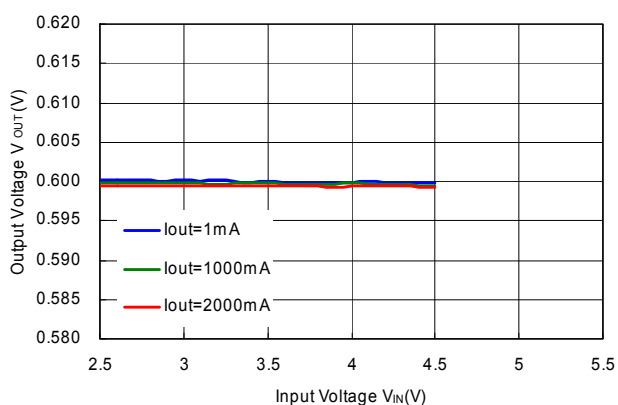


RP506Kxx1D/E/F $V_{OUT} = 3.3\text{ V}$
MODE = "H" Forced PWM Control

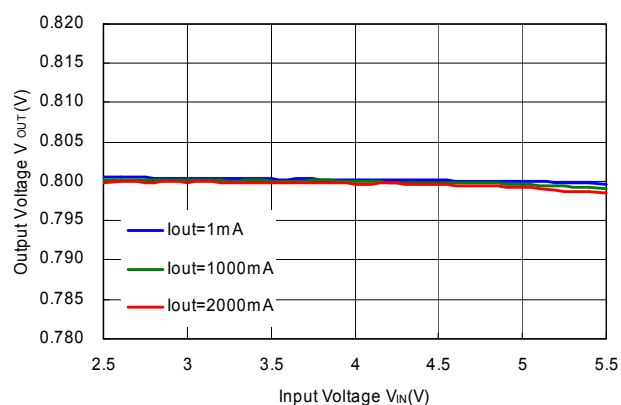


2) Output Voltage vs. Input Voltage

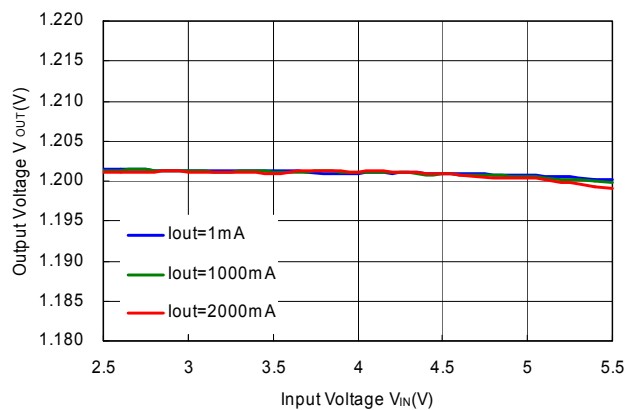
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$
MODE = "H" Forced PWM Control



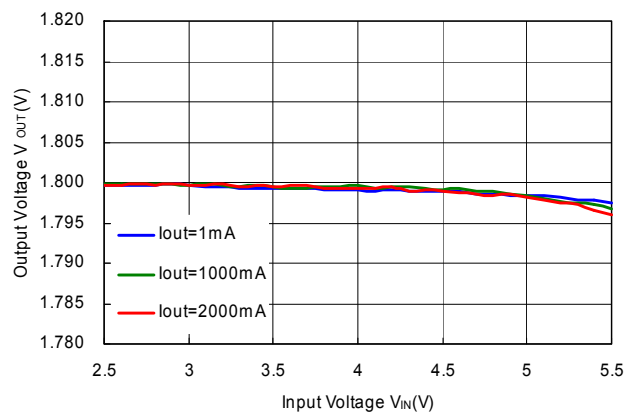
RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$
MODE = "H" Forced PWM Control



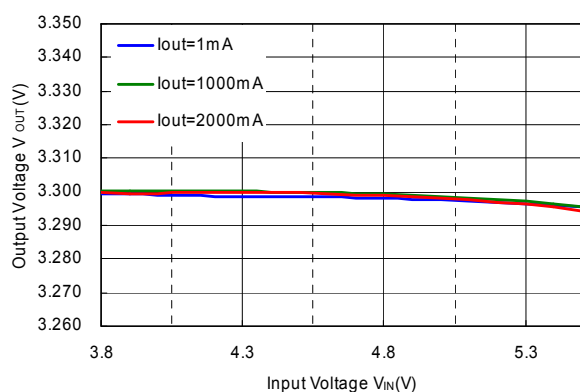
RP506K $V_{OUT} = 1.2\text{ V}$
MODE = "H" Forced PWM Control



RP506K $V_{OUT} = 1.8\text{ V}$
MODE = "H" Forced PWM Control

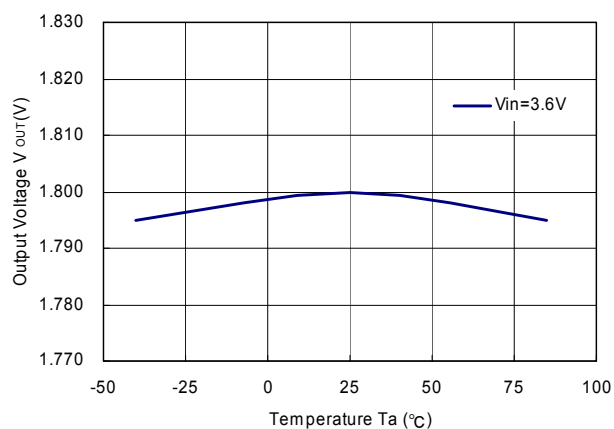


RP506K $V_{OUT} = 3.3\text{ V}$
MODE = "H" Forced PWM Control



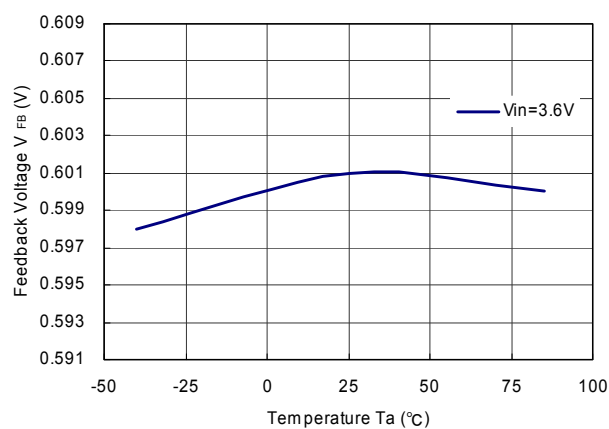
3) Output Voltage vs. Ambient Temp.

RP506K181A/B/D/E $V_{OUT} = 1.8\text{ V}$



4) Feedback Voltage vs. Ambient Temp.

RP506K001C/F

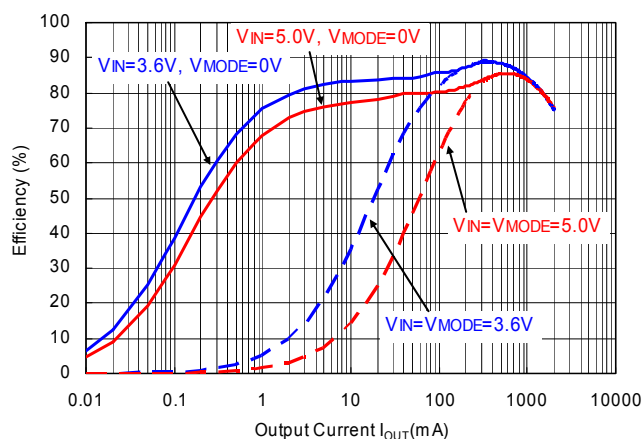


RP506K

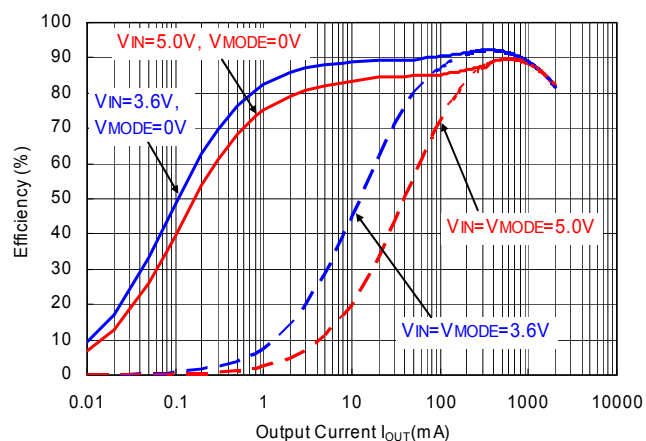
NO.EA-296-160523

5) Efficiency vs. Output Current

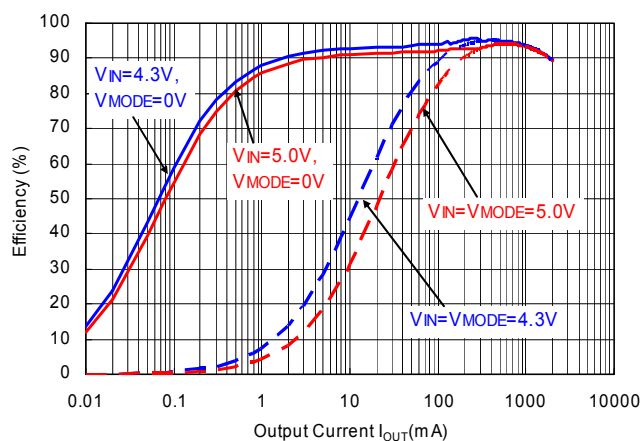
RP506Kxx1A/B/C $V_{OUT} = 1.2\text{ V}$



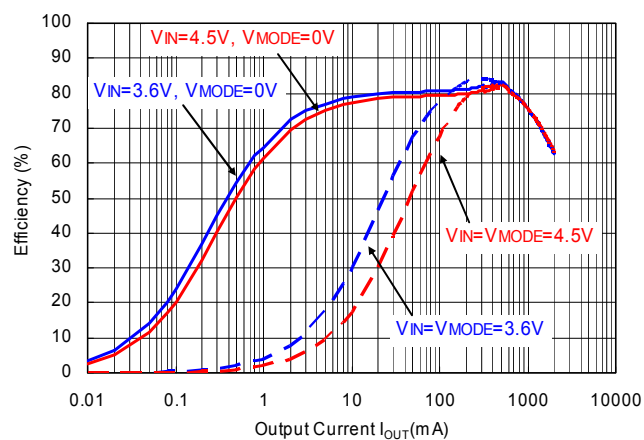
RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$



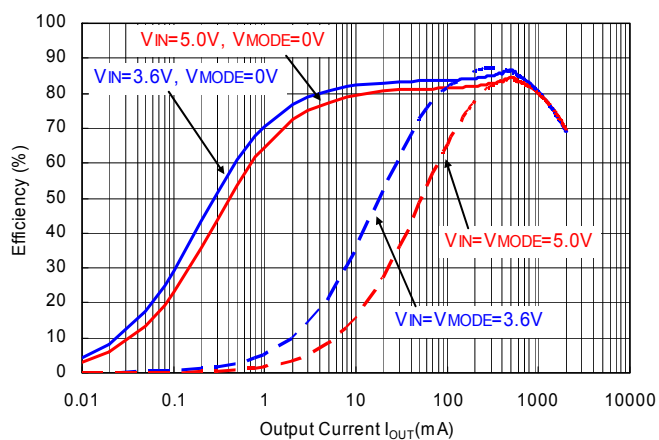
RP506Kxx1A/B/C $V_{OUT} = 3.3\text{ V}$



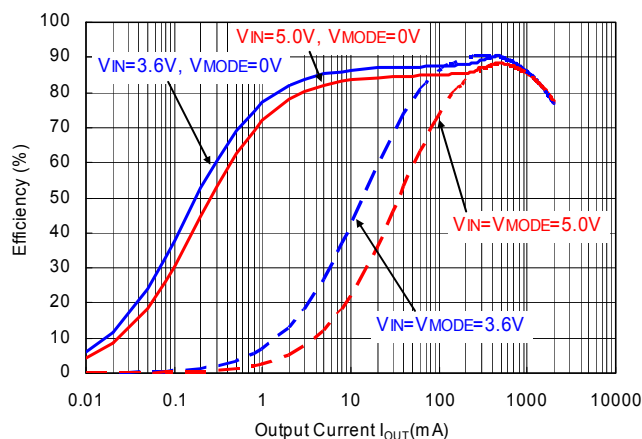
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$



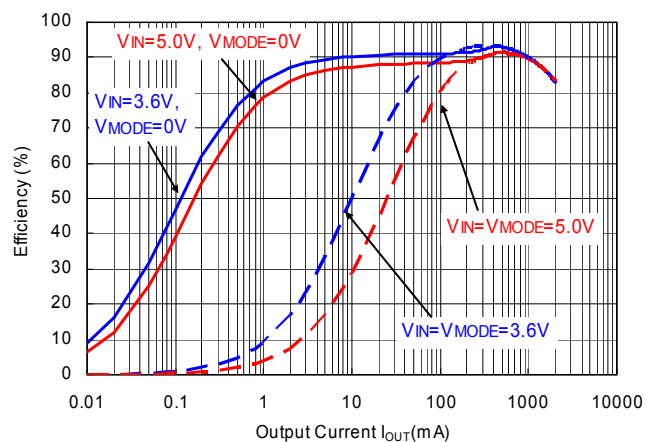
RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$



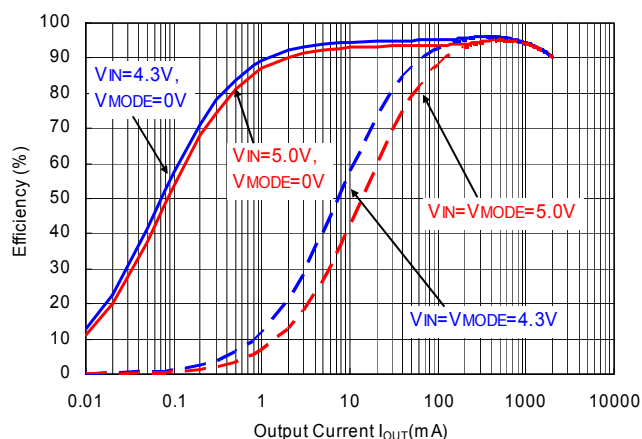
RP506Kxx1D/E/F $V_{OUT} = 1.2\text{ V}$



RP506Kxx1D/E/F $V_{OUT} = 1.8\text{ V}$



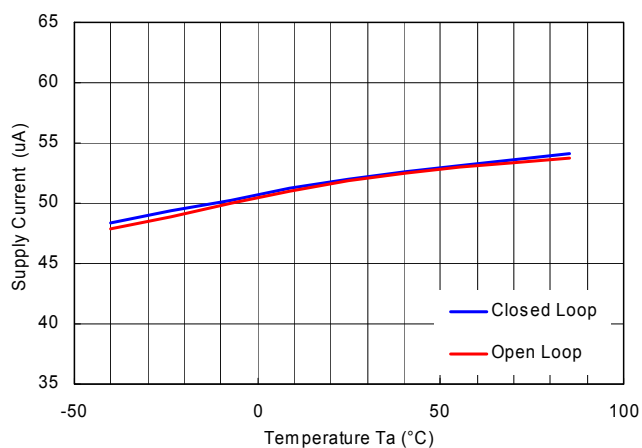
RP506Kxx1D/E/F $V_{OUT} = 3.3\text{ V}$



6) Supply Current vs. Ambient Temp.

RP506K $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 5.5\text{ V}$)

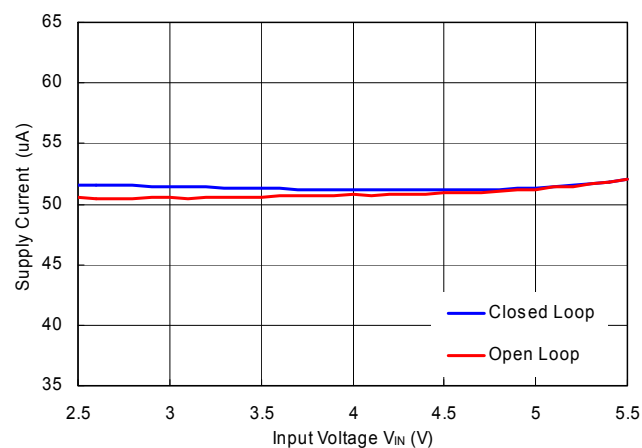
MODE = "L" PWM/VFM Auto Switching Control



7) Supply Current vs. Input Voltage

RP506K $V_{OUT} = 1.8\text{ V}$

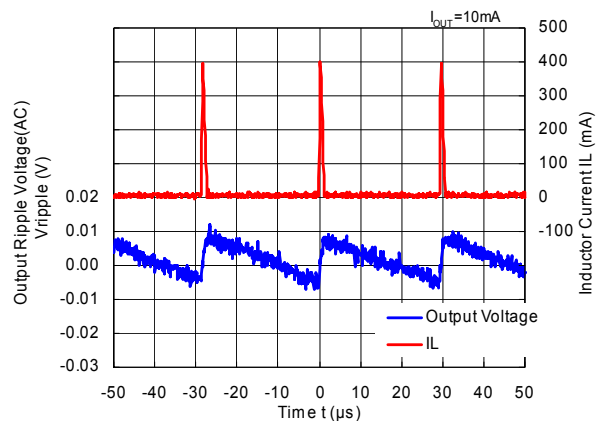
MODE = "L" PWM/VFM Auto Switching Control



8) Output Voltage Waveform

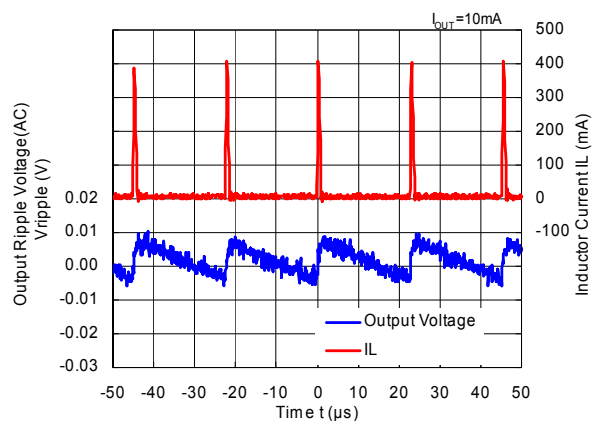
RP506Kxx1A/B/C $V_{OUT} = 0.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



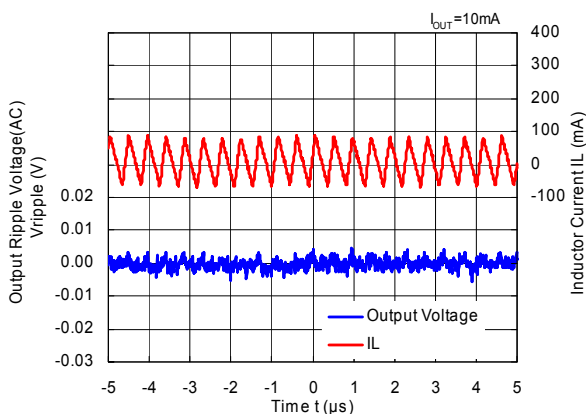
RP506Kxx1A/B/C $V_{OUT} = 1.2\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



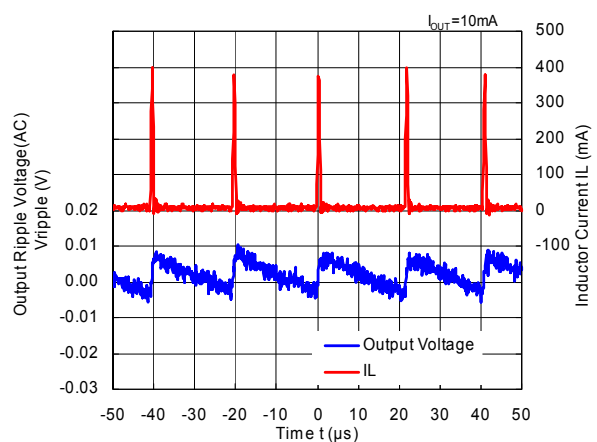
RP506Kxx1A/B/C $V_{OUT} = 1.2\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "H" Forced PWM Control



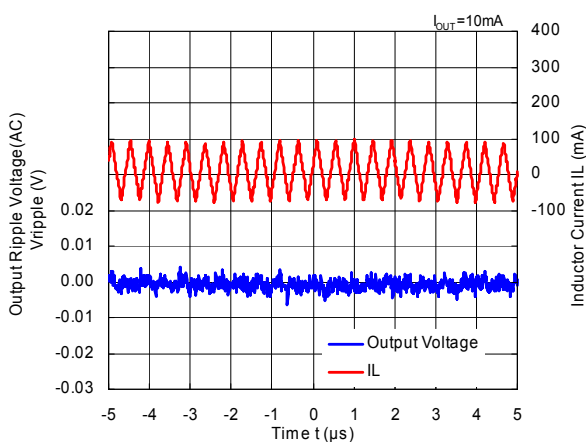
RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control

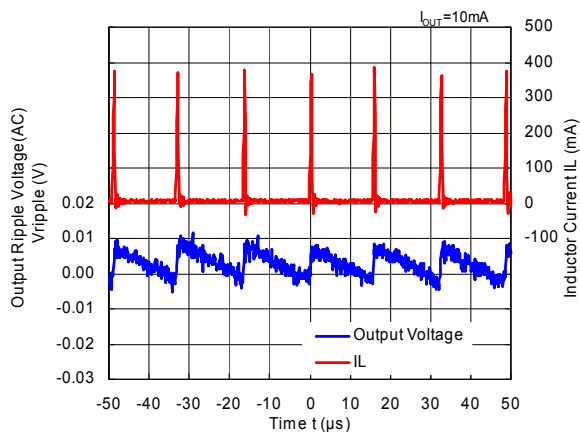


RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

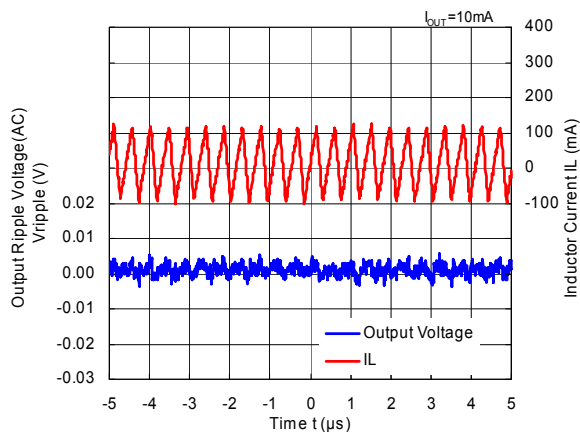
MODE = "H" Forced PWM Control



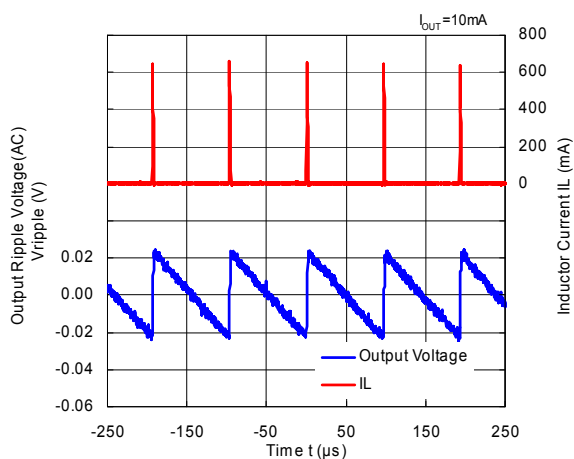
RP506Kxx1A/B/C $V_{OUT} = 3.3\text{ V}$ ($V_{IN} = 5.0\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



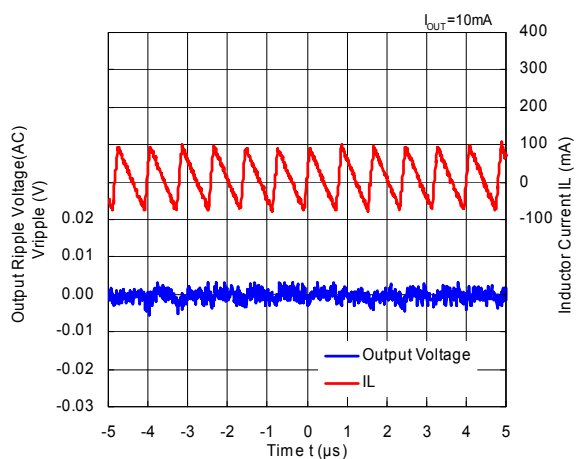
RP506Kxx1A/B/C $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 5.0\text{ V}$)
MODE = "H" Forced PWM Control



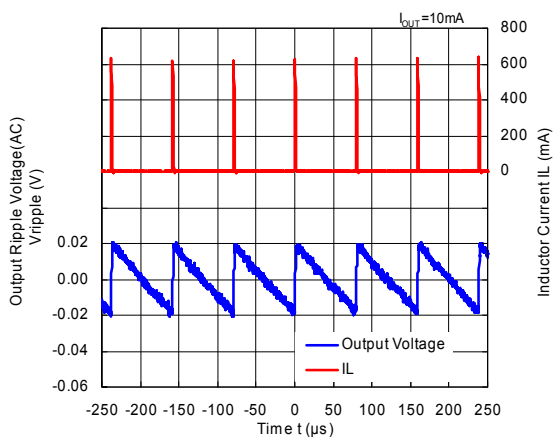
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$ ($V_{IN} = 3.6\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



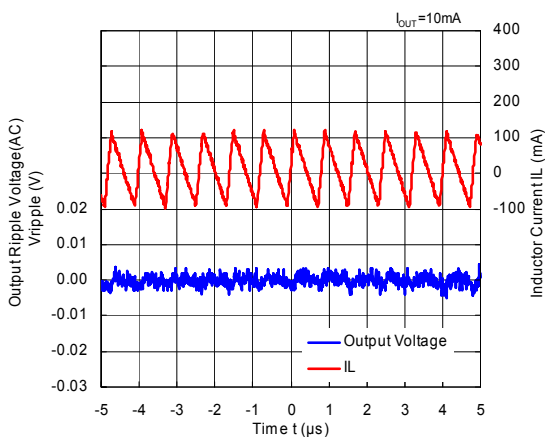
RP506Kxx1D/E/F $V_{OUT} = 0.6\text{ V}$ ($V_{IN} = 3.6\text{ V}$)
MODE = "H" Forced PWM Control



RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control

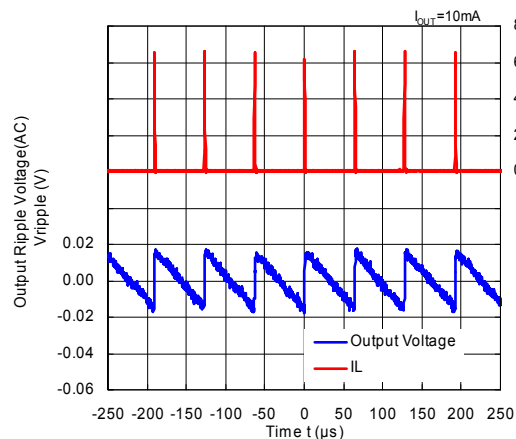


RP506Kxx1D/E/F $V_{OUT} = 0.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)
MODE = "H" Forced PWM Control



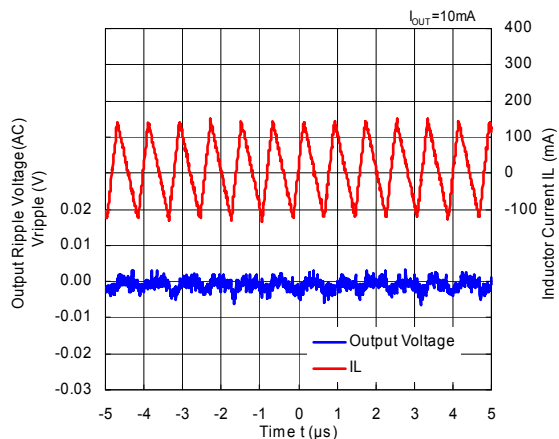
RP506Kxx1D/E/F $V_{OUT} = 1.2\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



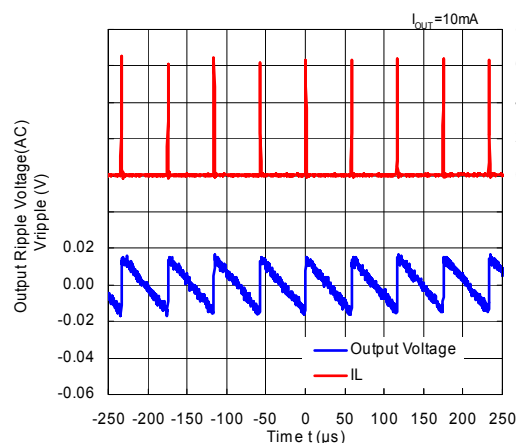
RP506Kxx1D/E/F $V_{OUT} = 1.2\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "H" Forced PWM Control



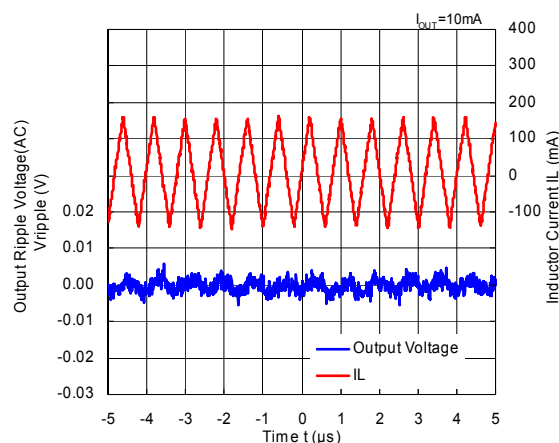
RP506Kxx1D/E/F $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



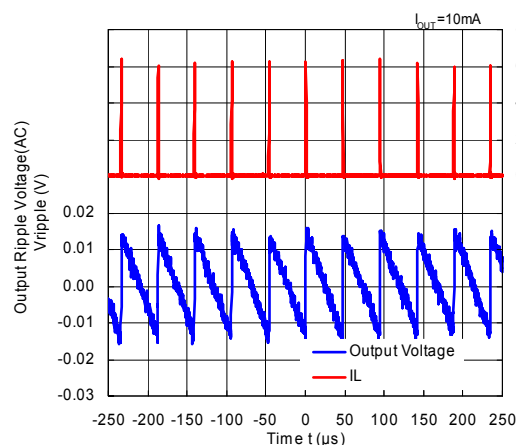
RP506Kxx1D/E/F $V_{OUT} = 1.8\text{ V}$ ($V_{IN} = 3.6\text{ V}$)

MODE = "H" Forced PWM Control



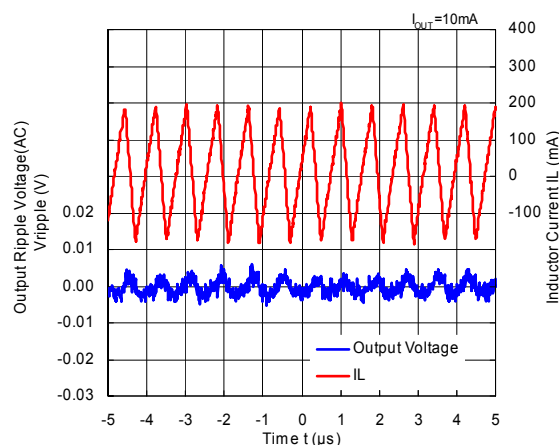
RP506Kxx1D/E/F $V_{OUT} = 3.3\text{ V}$ ($V_{IN} = 5.0\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



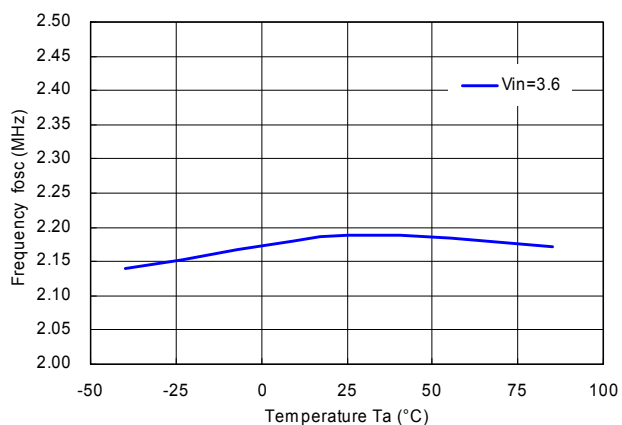
RP506Kxx1D/E/F $V_{OUT} = 3.3\text{ V}$ ($V_{IN} = 5.0\text{ V}$)

MODE = "H" Forced PWM Control

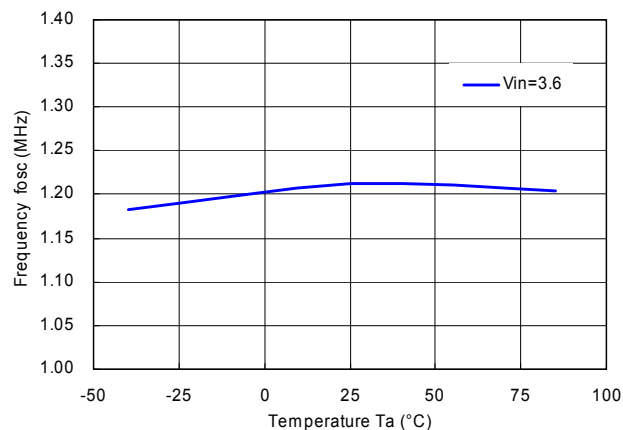


9) Oscillator Frequency vs. Ambient Temp.

RP506Kxx1A/B/C

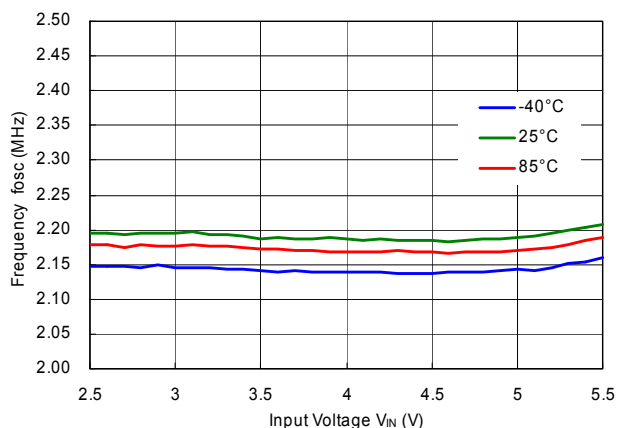


RP506Kxx1D/E/F

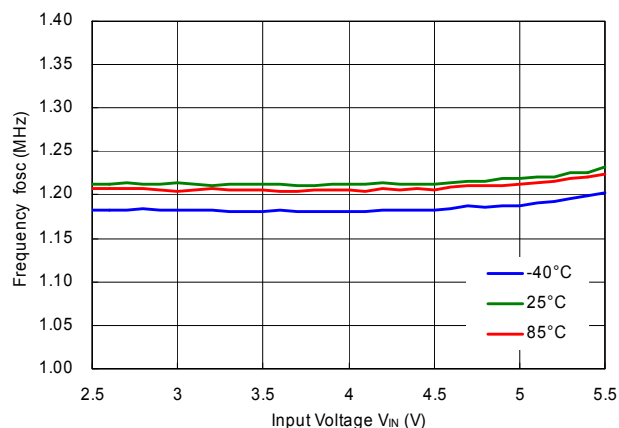


10) Oscillator Frequency vs. Input Voltage

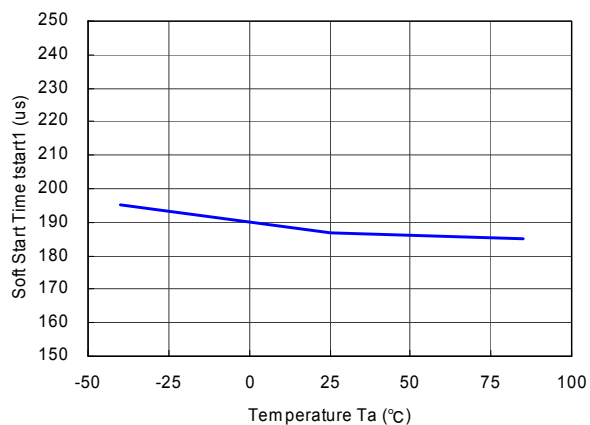
RP506Kxx1A/B/C



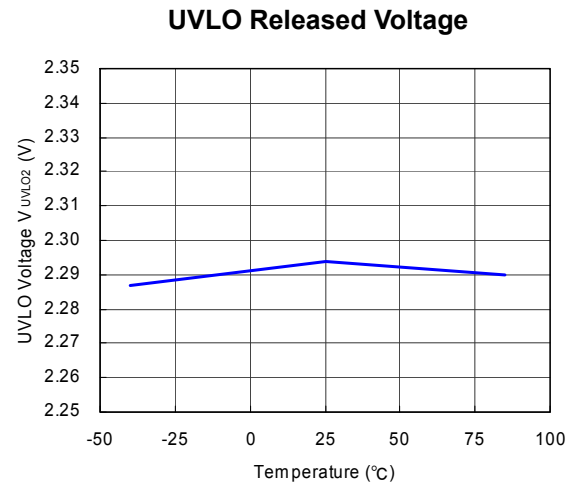
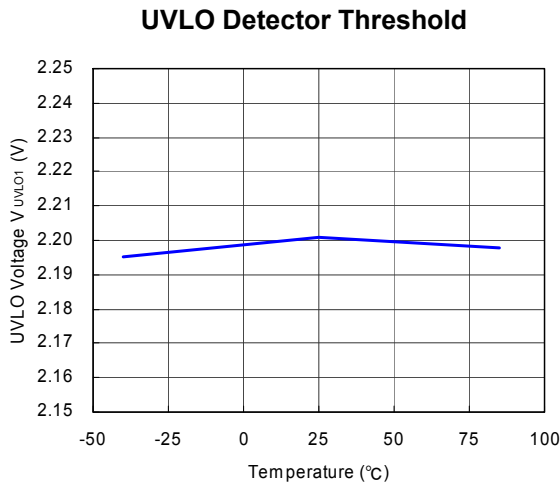
RP506Kxx1D/E/F



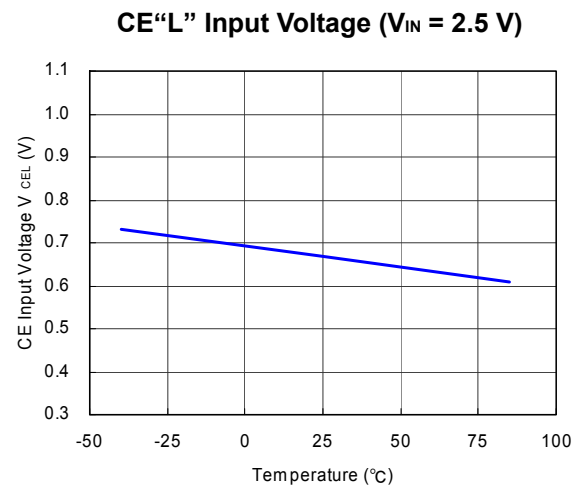
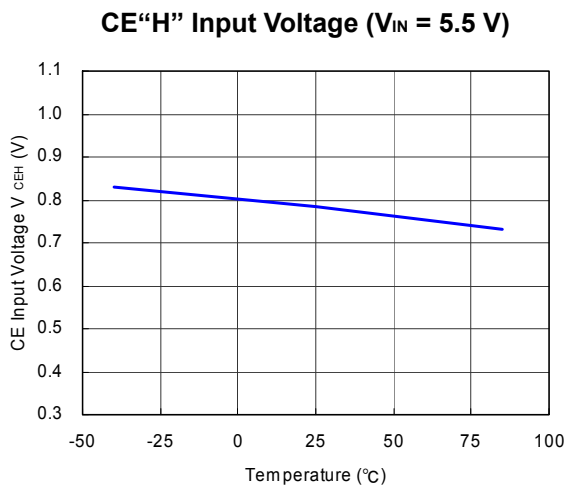
11) Soft-start Time vs. Ambient Temp.



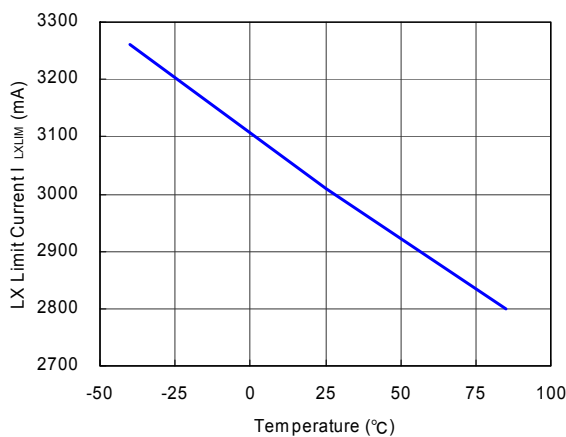
12) UVLO Detector Threshold/ Released Voltage vs. Ambient Temp.



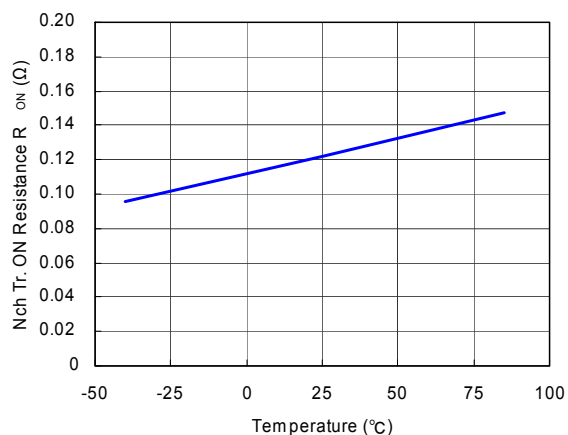
13) CE Input Voltage vs. Ambient Temp.



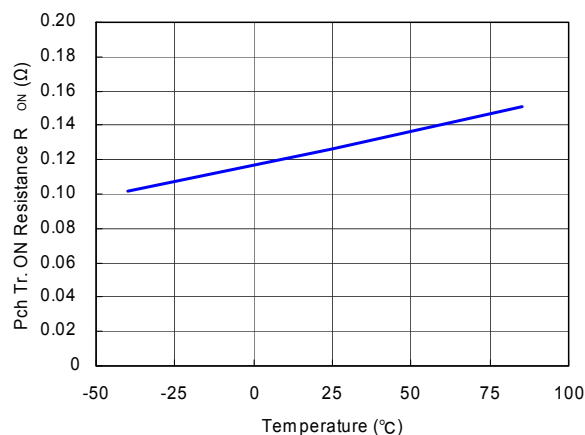
14) Lx Limit Current vs. Ambient Temp.



15) Nch Tr. ON Resistance vs. Ambient Temp.

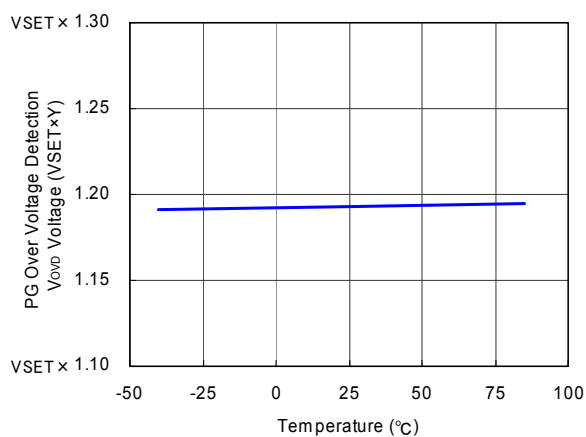


16) Pch Tr. ON Resistance vs. Ambient Temp.

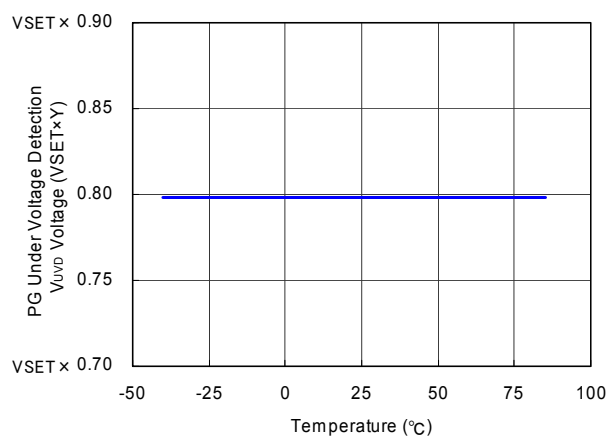


17) PG Detector Threshold vs. Ambient Temp.

Over Voltage Detection (V_{OVD})

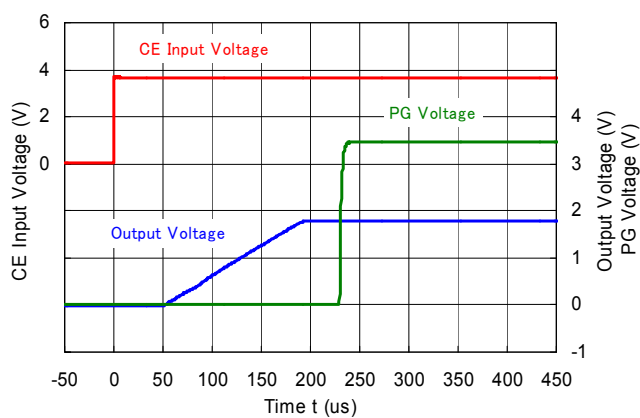


Under Voltage Detection (V_{UVD})

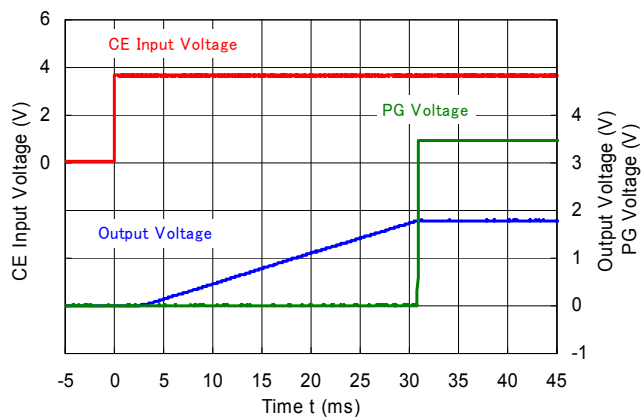


18) Soft-start Waveform

RP506K $V_{OUT} = 1.8$ V TSS = Open



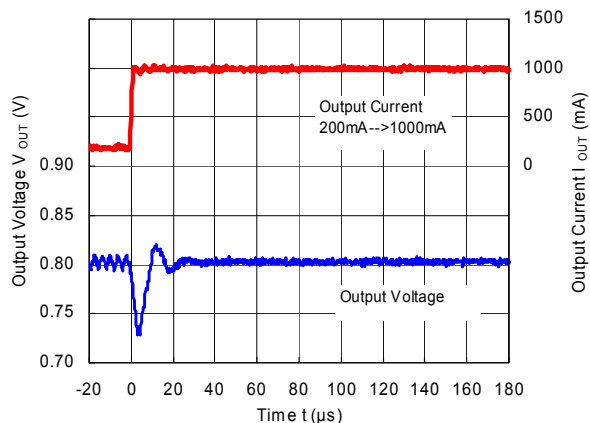
RP506K $V_{OUT} = 1.8$ V TSS = 0.1 μF



19) Load Transient Response

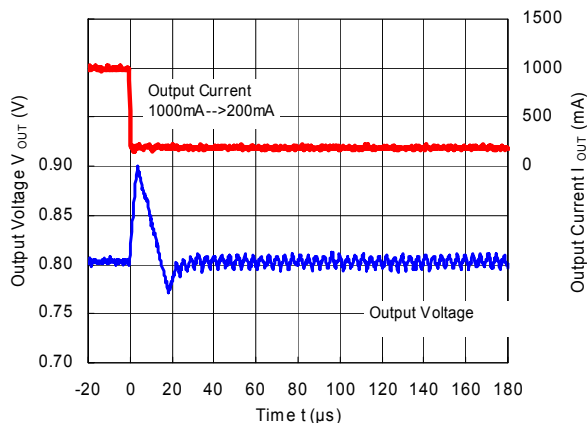
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



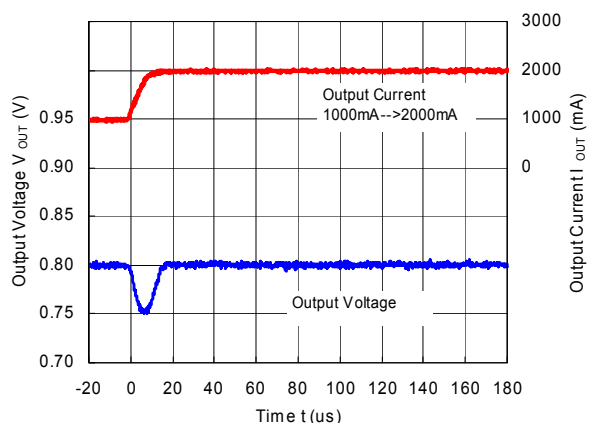
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



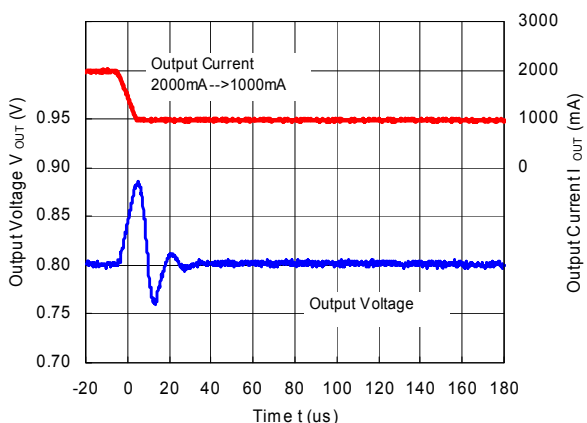
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



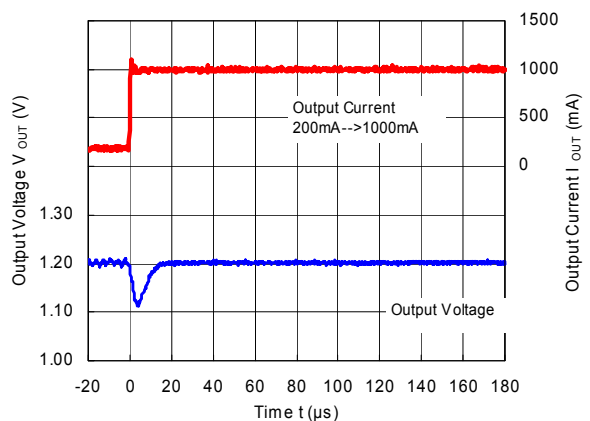
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



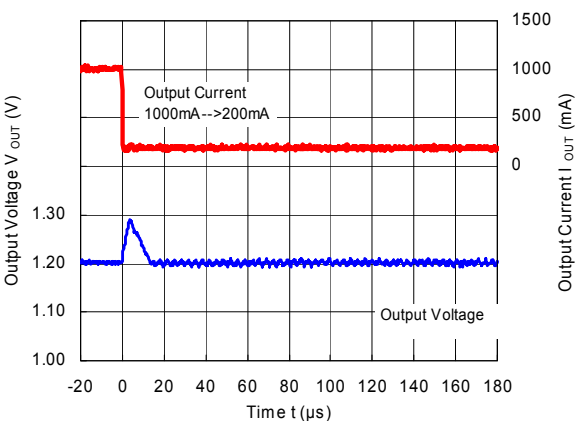
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



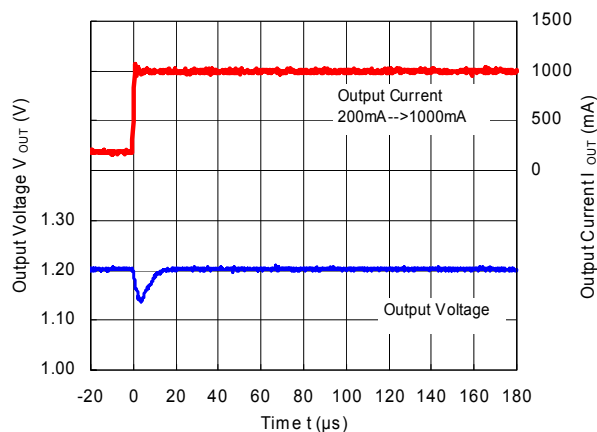
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control



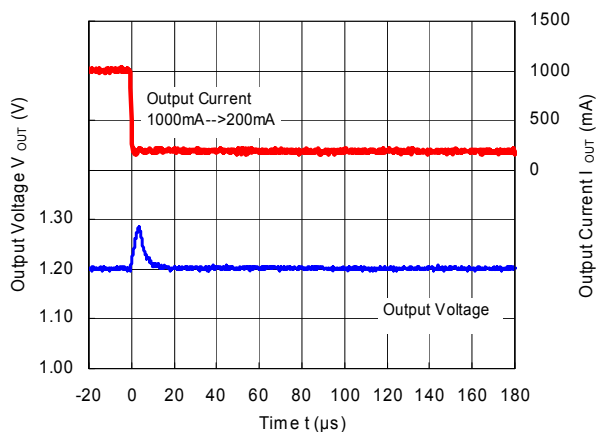
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

MODE = "H" Forced PWM Control

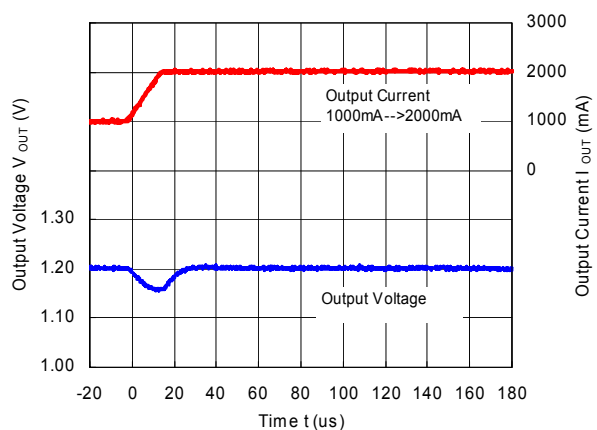


RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

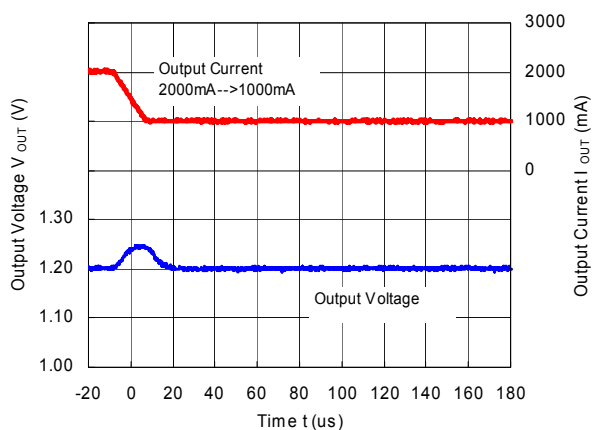
MODE = "H" Forced PWM Control



RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

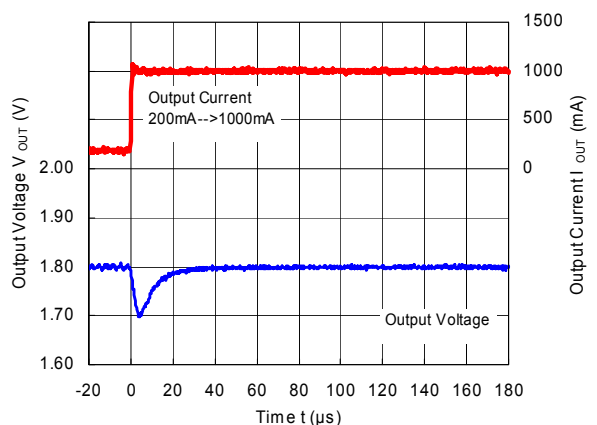


RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)



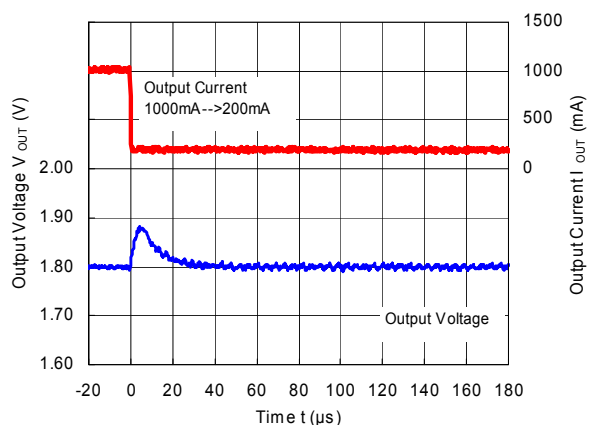
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control

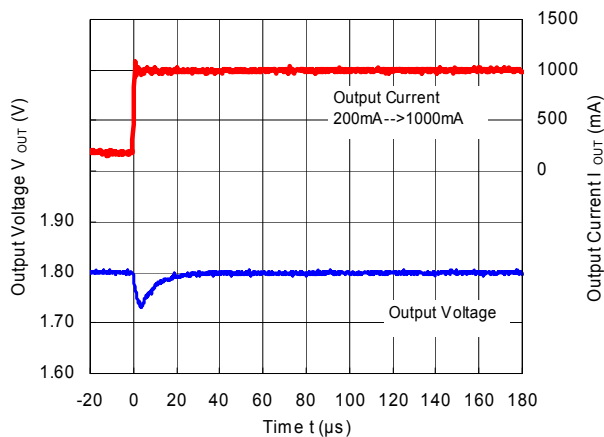


RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)

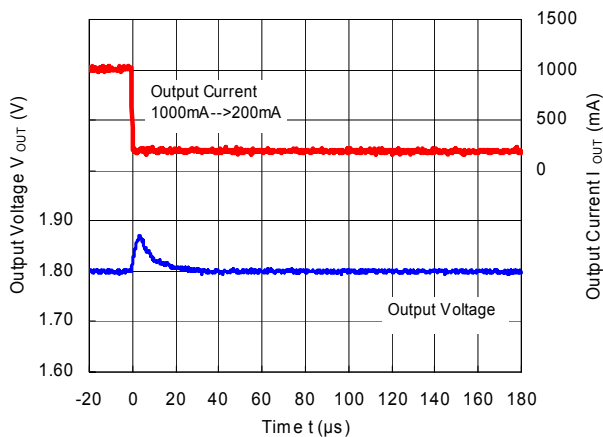
MODE = "L" PWM/VFM Auto Switching Control



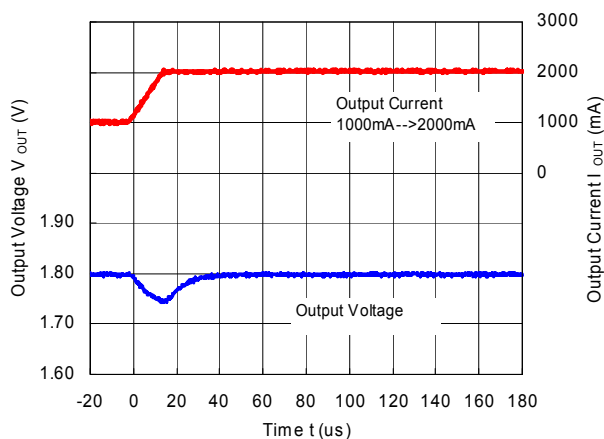
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)
MODE = "H" Forced PWM Control



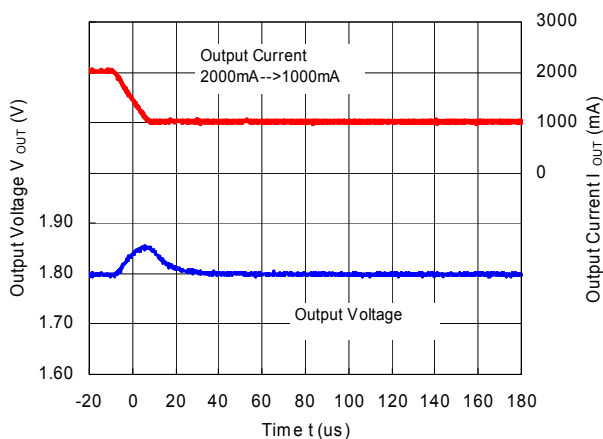
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)
MODE = "H" Forced PWM Control



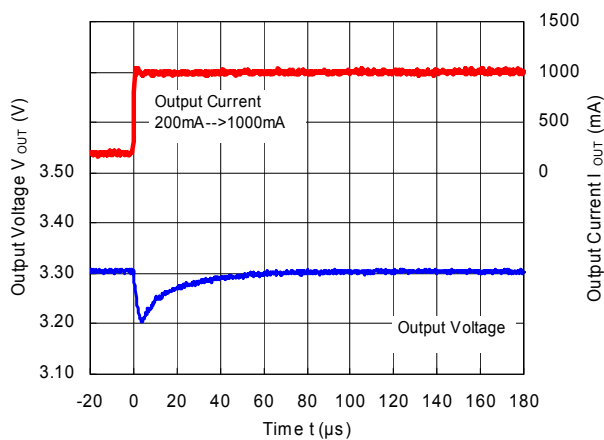
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)



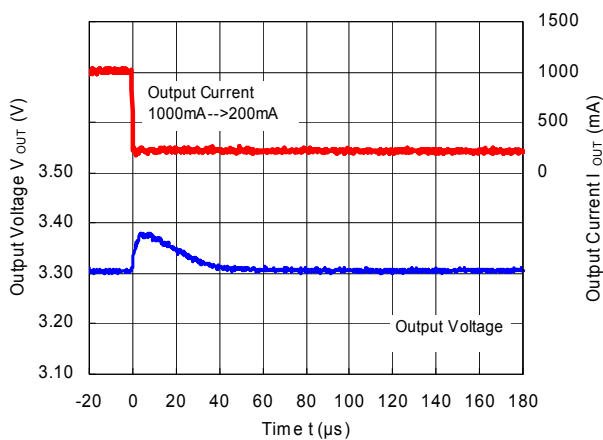
RP506Kxx1A/B/C ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)



RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control

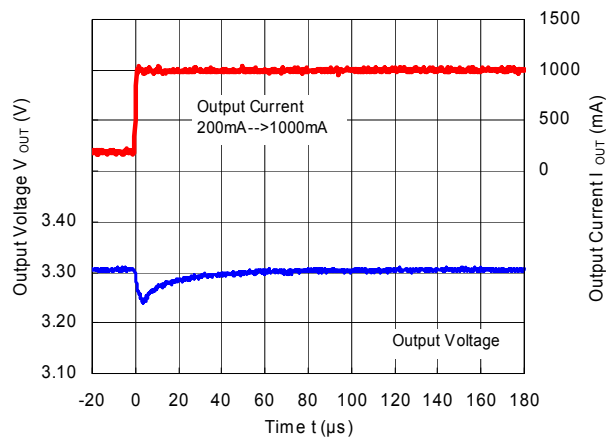


RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



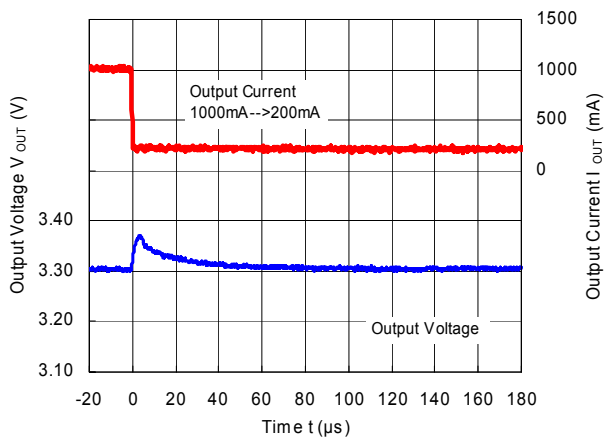
RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)

MODE = "H" Forced PWM Control

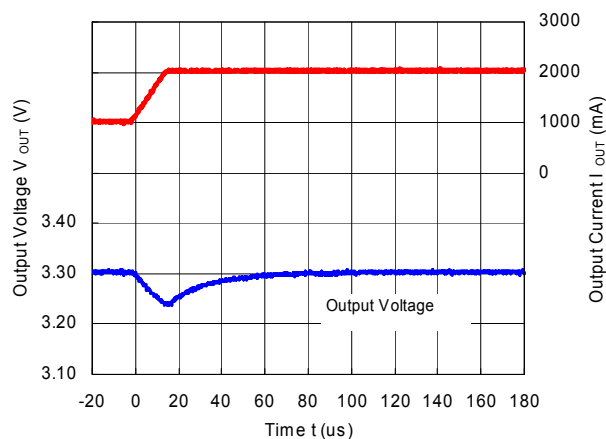


RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)

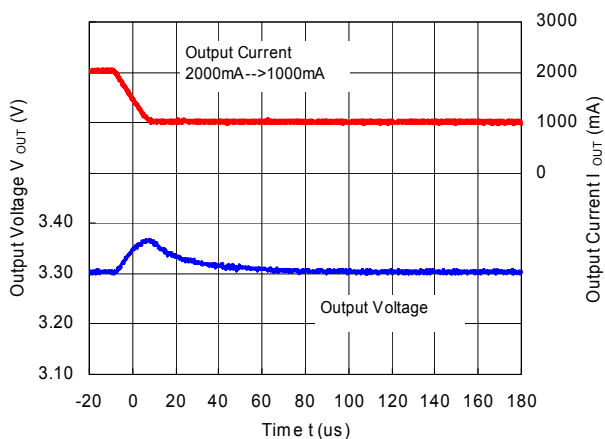
MODE = "H" Forced PWM Control



RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)

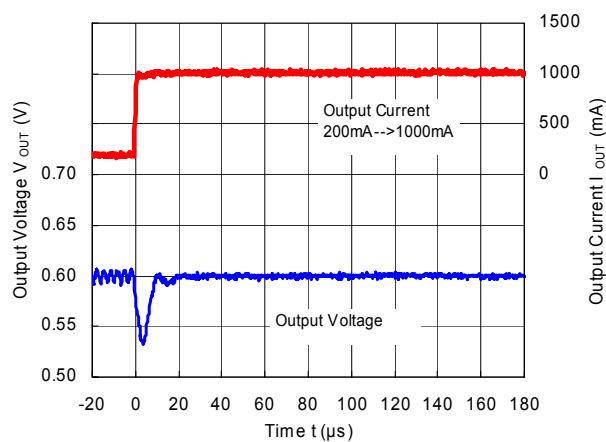


RP506Kxx1A/B/C ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)



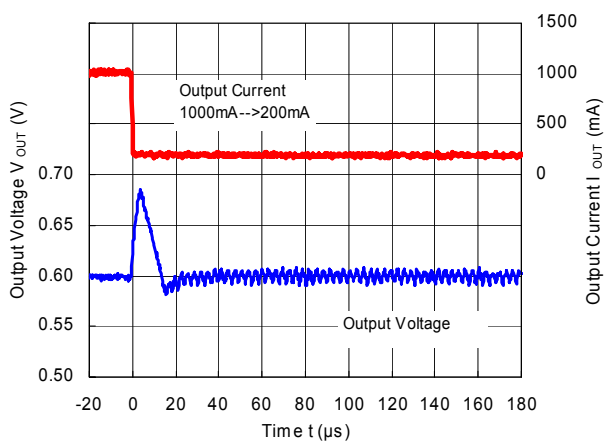
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.6\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control

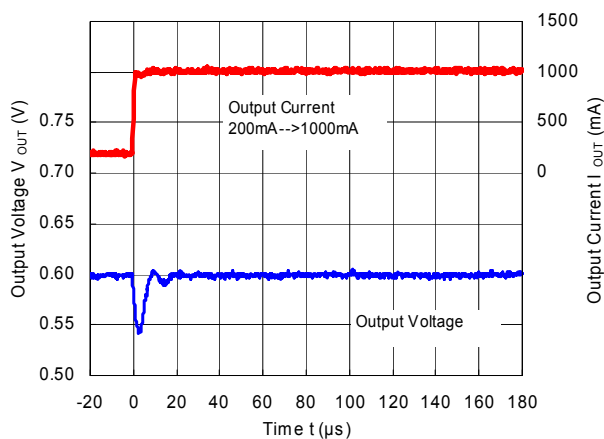


RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.6\text{ V}$)

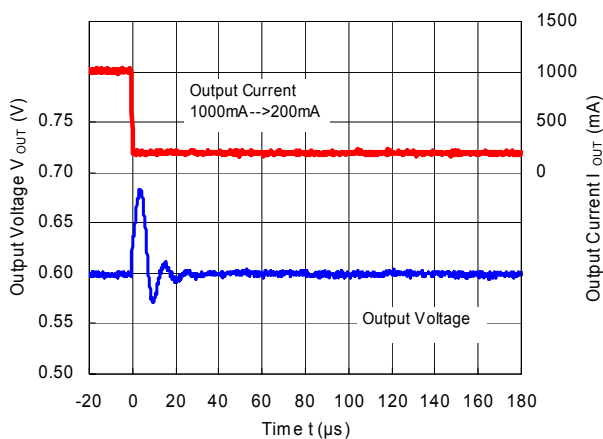
MODE = "L" PWM/VFM Auto Switching Control



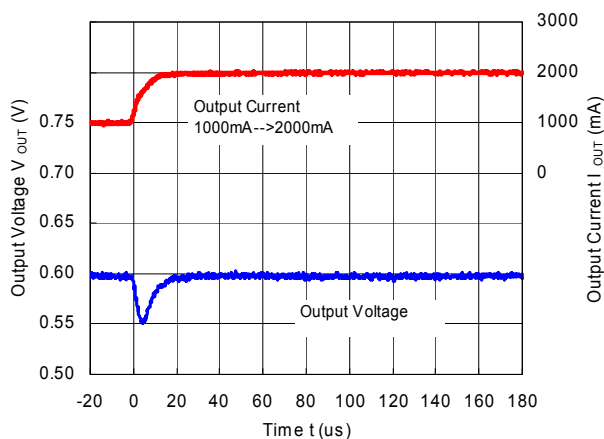
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.6\text{ V}$)
MODE = "H" Forced PWM Control



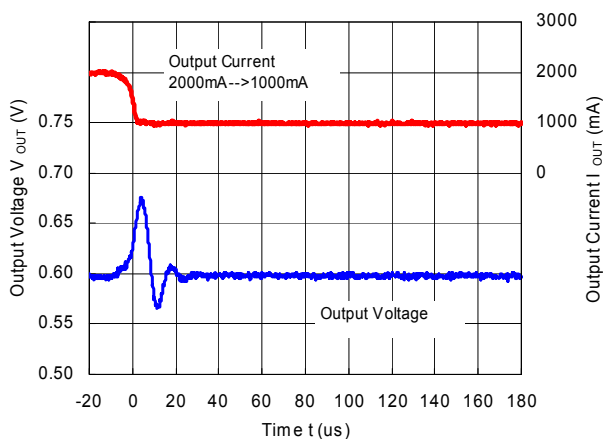
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.6\text{ V}$)
MODE = "H" Forced PWM Control



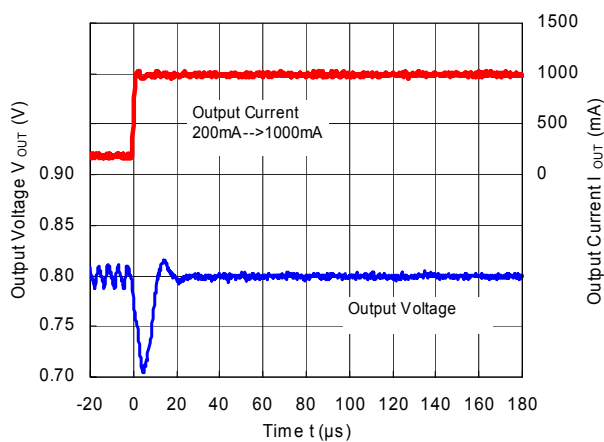
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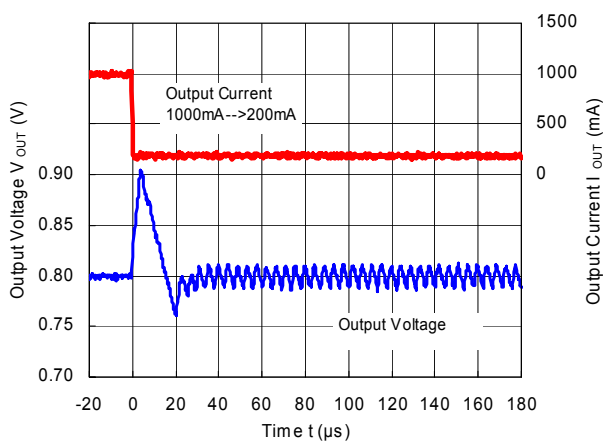
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.6\text{ V}$)



RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control

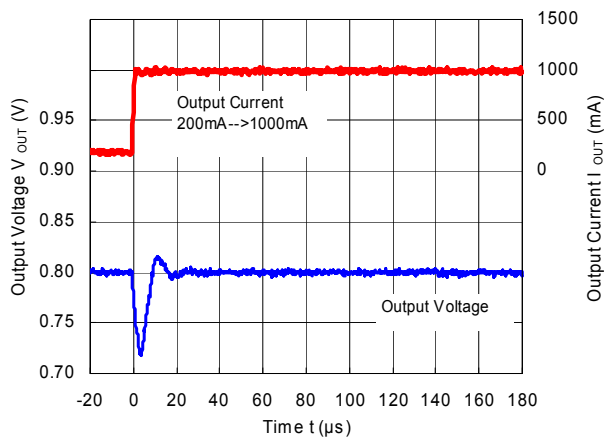


RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



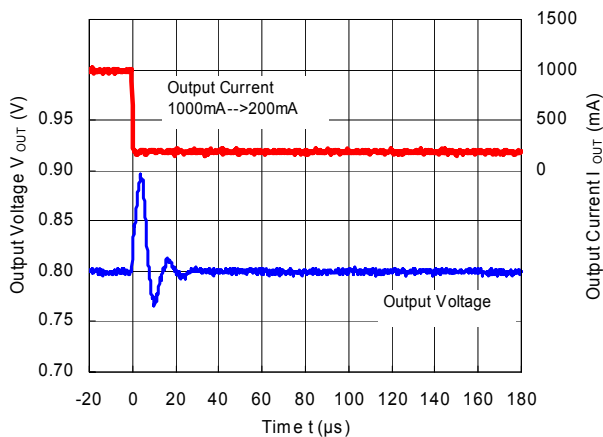
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MODE = "H" Forced PWM Control

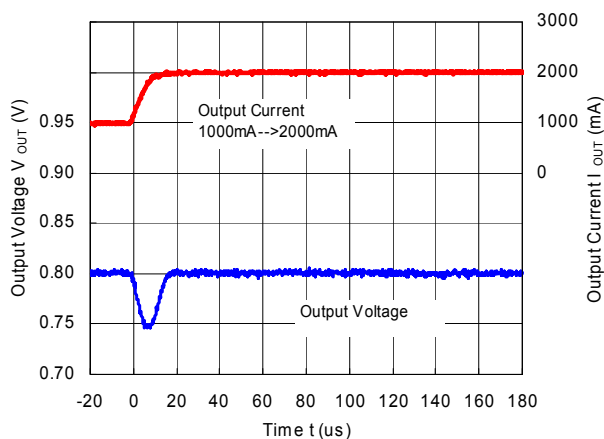


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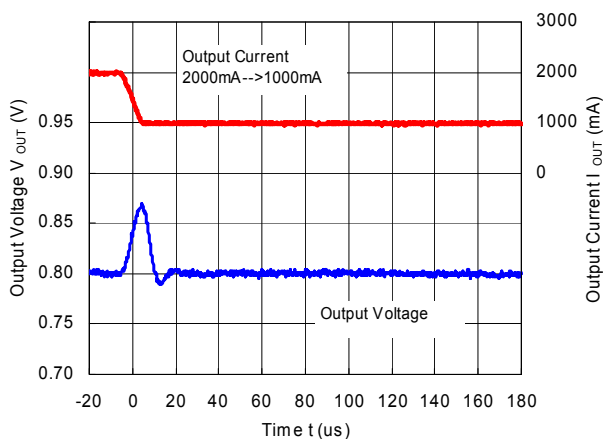
MODE = "H" Forced PWM Control



RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)

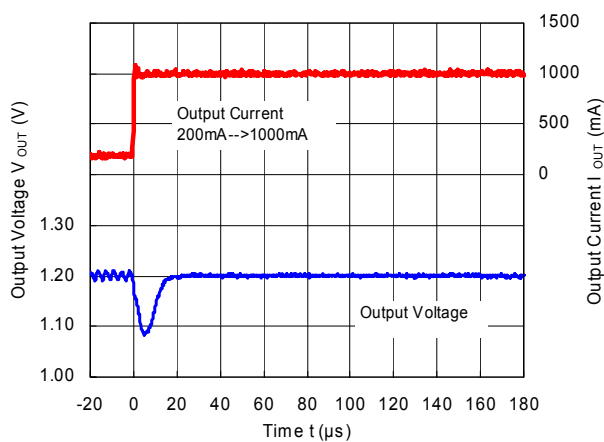


RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 0.8\text{ V}$)



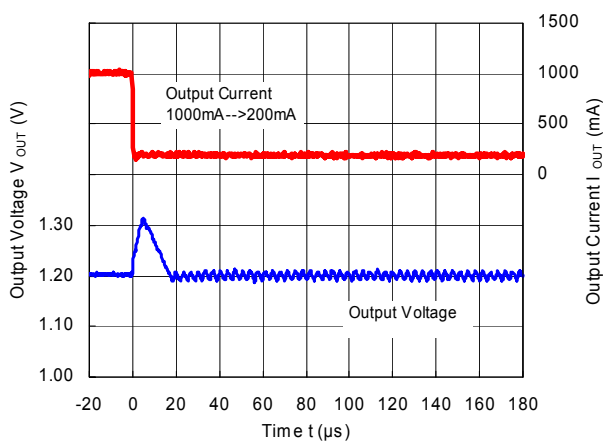
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

MODE = "L" PWM/VFM Auto Switching Control

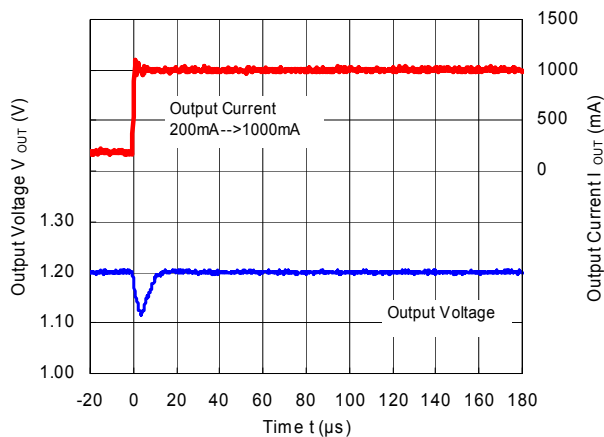


RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)

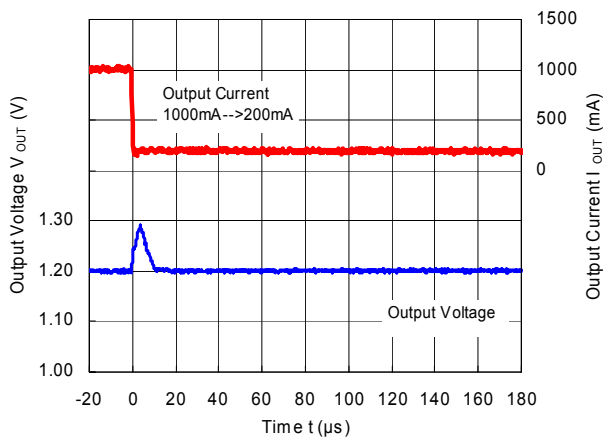
MODE = "L" PWM/VFM Auto Switching Control



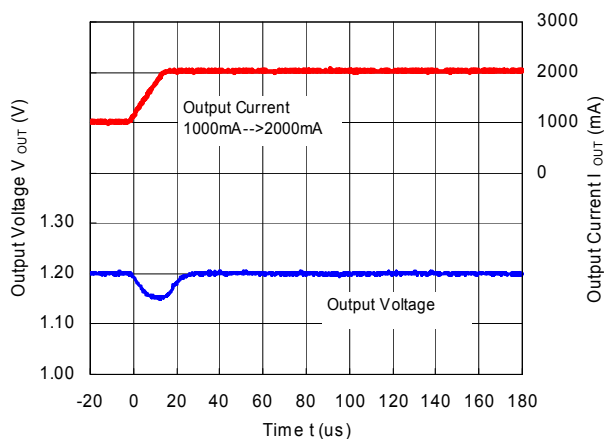
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)
MODE = "H" Forced PWM Control



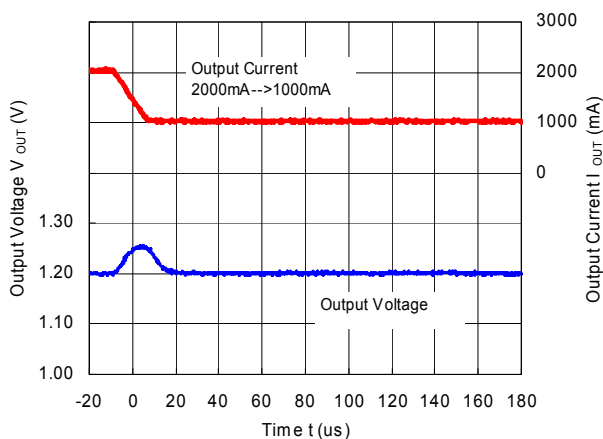
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$)
MODE = "H" Forced PWM Control



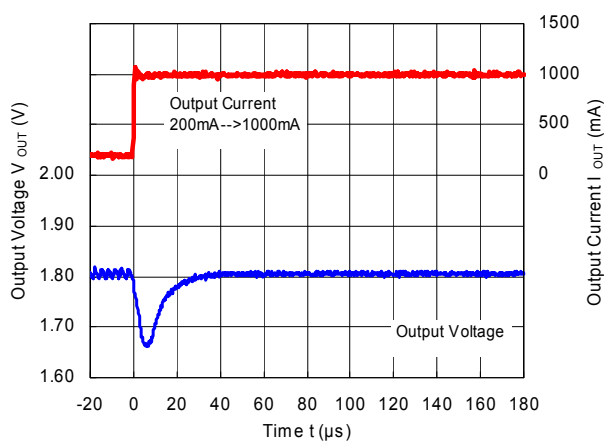
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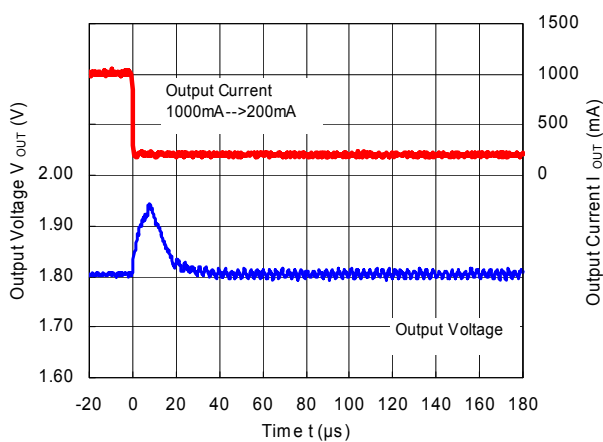
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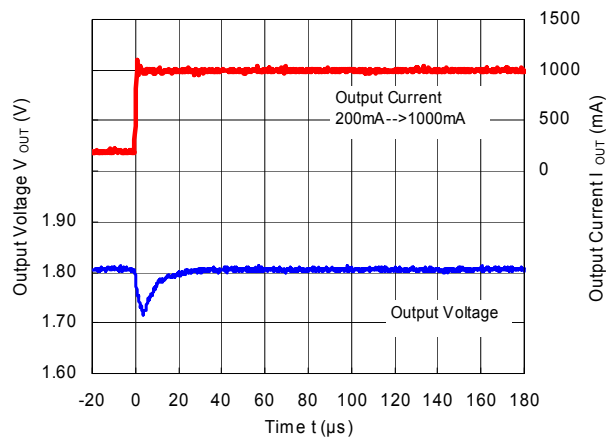
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MODE = "L" PWM/VFM Auto Switching Control



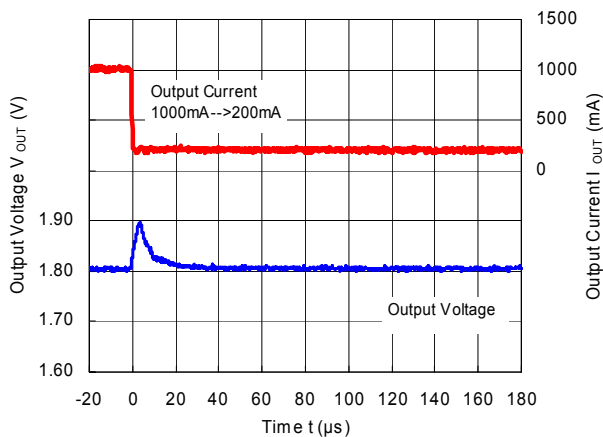
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



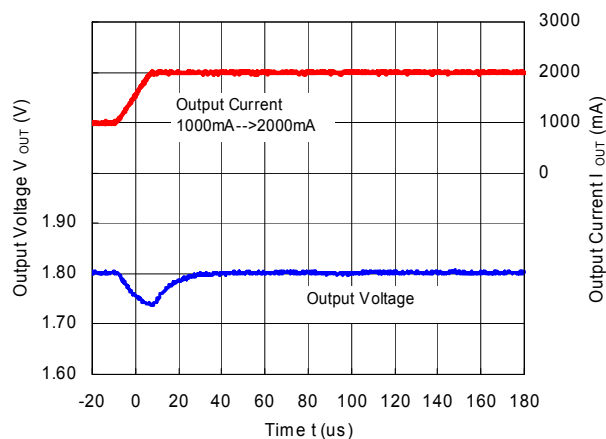
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)
MODE = "H" Forced PWM Control



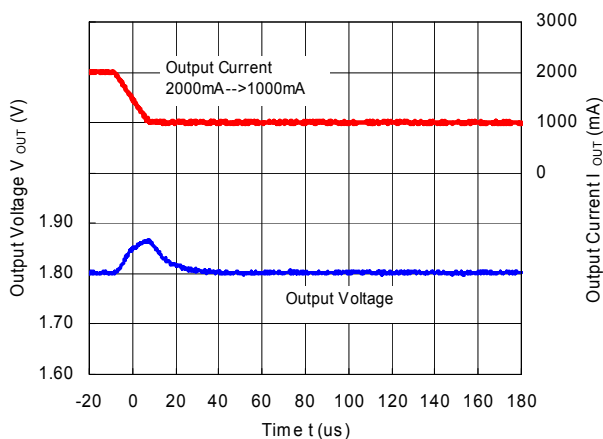
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)
MODE = "H" Forced PWM Control



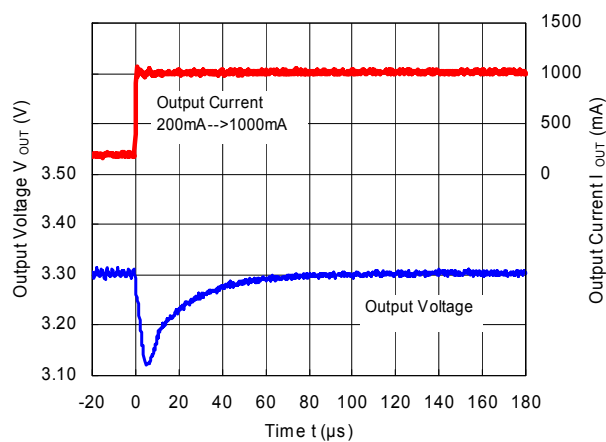
RP506Kxx1D/E/F ($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$)



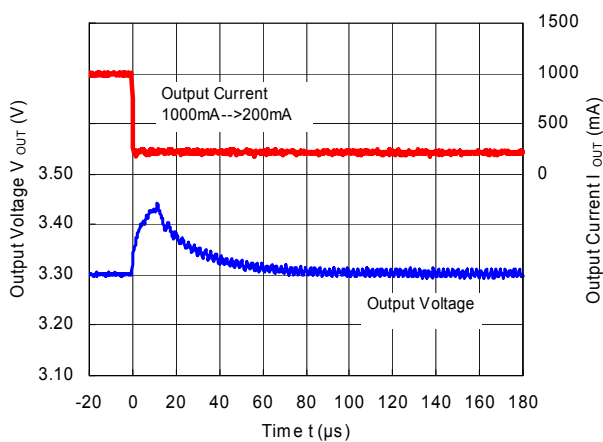
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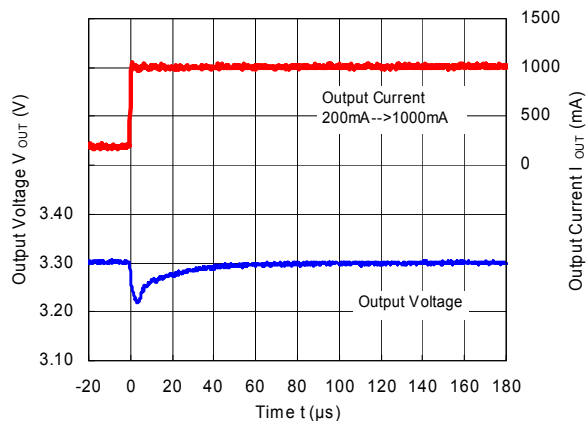
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MODE = "L" PWM/VFM Auto Switching Control



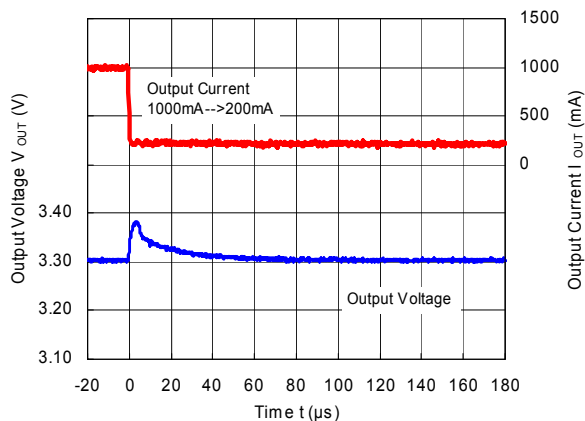
RP506Kxx1D/E/F ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)
MODE = "L" PWM/VFM Auto Switching Control



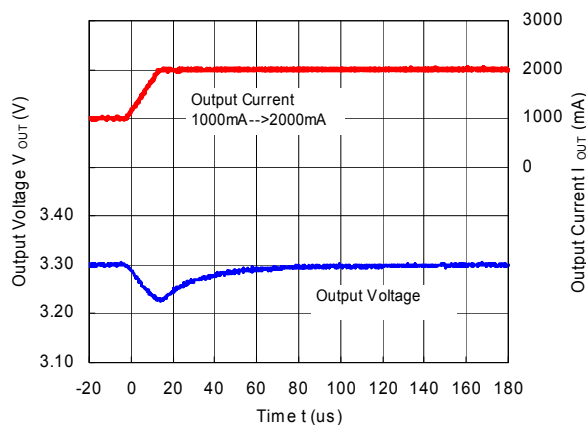
RP506Kxx1D/E/F ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)
MODE = "H" Forced PWM Control



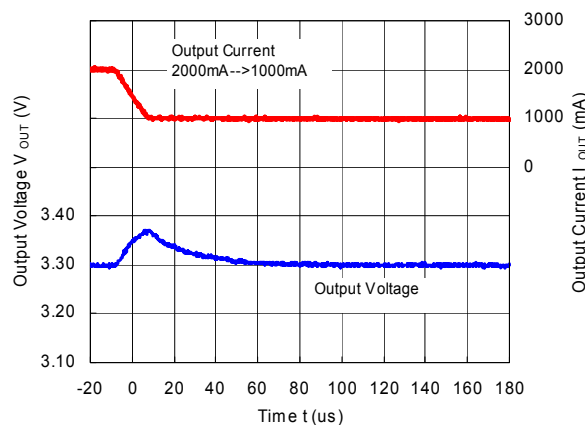
RP506Kxx1D/E/F ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)
MODE = "H" Forced PWM Control



RP506Kxx1D/E/F ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)



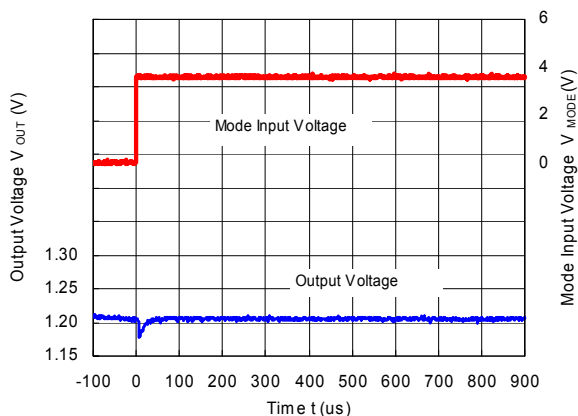
RP506Kxx1D/E/F ($V_{IN} = 5.0\text{ V}$, $V_{OUT} = 3.3\text{ V}$)



20) Auto Switching Control Waveform

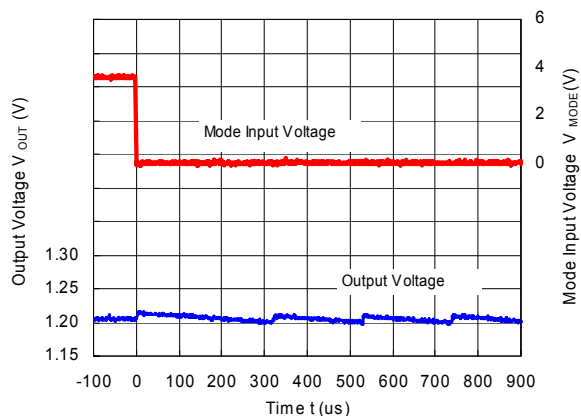
RP506Kxx1A/B/C

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$, $I_{OUT} = 1\text{ mA}$)
MODE = "L" --> MODE = "H"



RP506Kxx1A/B/C

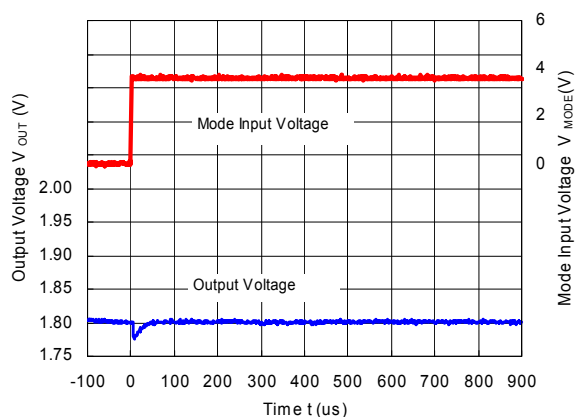
($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$, $I_{OUT} = 1\text{ mA}$)
MODE = "H" --> MODE = "L"



RP506Kxx1A/B/C

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$, $I_{OUT} = 1\text{ mA}$)

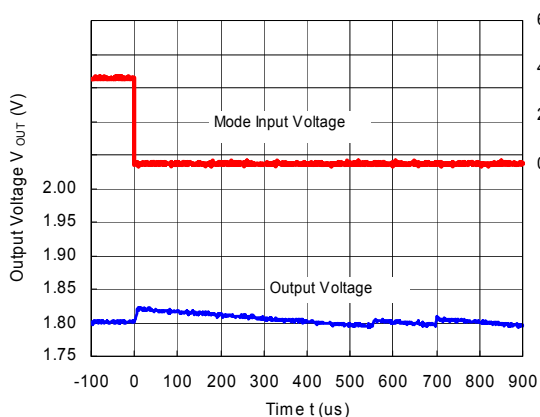
MODE = "L" --> MODE = "H"



RP506Kxx1A/B/C

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$, $I_{OUT} = 1\text{ mA}$)

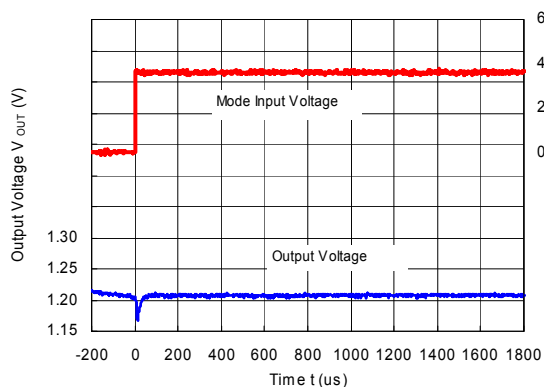
MODE = "H" --> MODE = "L"



RP506Kxx1D/E/F

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$, $I_{OUT} = 1\text{ mA}$)

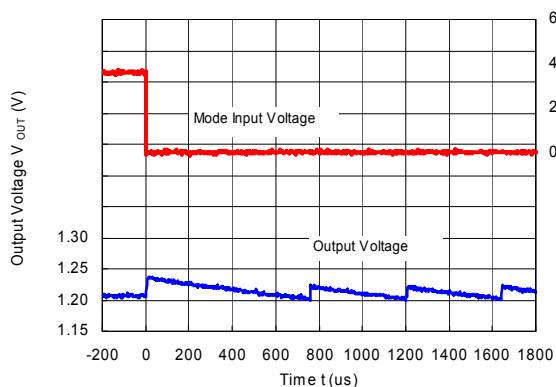
MODE = "L" --> MODE = "H"



RP506Kxx1D/E/F

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.2\text{ V}$, $I_{OUT} = 1\text{ mA}$)

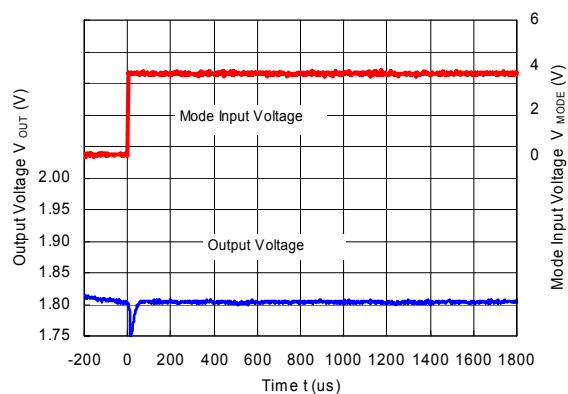
MODE = "H" --> MODE = "L"



RP506Kxx1D/E/F

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$, $I_{OUT} = 1\text{ mA}$)

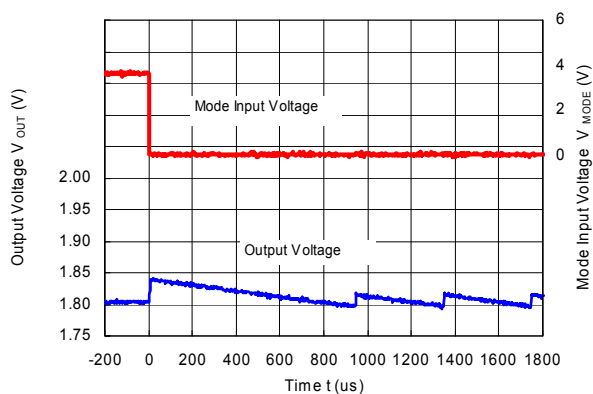
MODE = "L" --> MODE = "H"



RP506Kxx1D/E/F

($V_{IN} = 3.6\text{ V}$, $V_{OUT} = 1.8\text{ V}$, $I_{OUT} = 1\text{ mA}$)

MODE = "H" --> MODE = "L"





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Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
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- Оперативные сроки поставки под заказ (от 5 рабочих дней);
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- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
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Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
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- Поставка образцов и прототипов;
- Техническая поддержка проекта;
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Как с нами связаться

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