

## 2ED2304S06F

### 650 V Half Bridge Gate Driver with Integrated Bootstrap Diode (BSD)

#### Features

- Infineon thin-film-SOI-technology
- Fully operational to +650 V
- Floating channel designed for bootstrap operation
- Output source/sink current capability +0.36 A/-0.7 A
- Integrated Ultra-fast, low  $R_{DS(ON)}$  Bootstrap Diode
- Tolerant to negative transient voltage up to -100 V (Pulse width is up 300 ns) given by SOI-technology
- 10 ns typ., 60 ns max. propagation delay matching
- dV/dt immune  $\pm 50$  V
- Gate drive supply range from 10 V to 20 V
- Undervoltage lockout for both channels
- 3.3 V, 5 V and 15 V input logic compatible
- RoHS compliant

#### Product summary

$V_{OFFSET}$	= 670 V max.
$I_{O+/-}$ (typ.)	= 0.36 A/0.7 A
$V_{OUT}$	= 10 V - 17.5 V
Delay Matching	= 60 ns max.
Internal deadtime	= 75 ns
$t_{on/off}$ (typ.)	= 310 ns/300 ns

#### Package

DSO-8



#### Potential applications

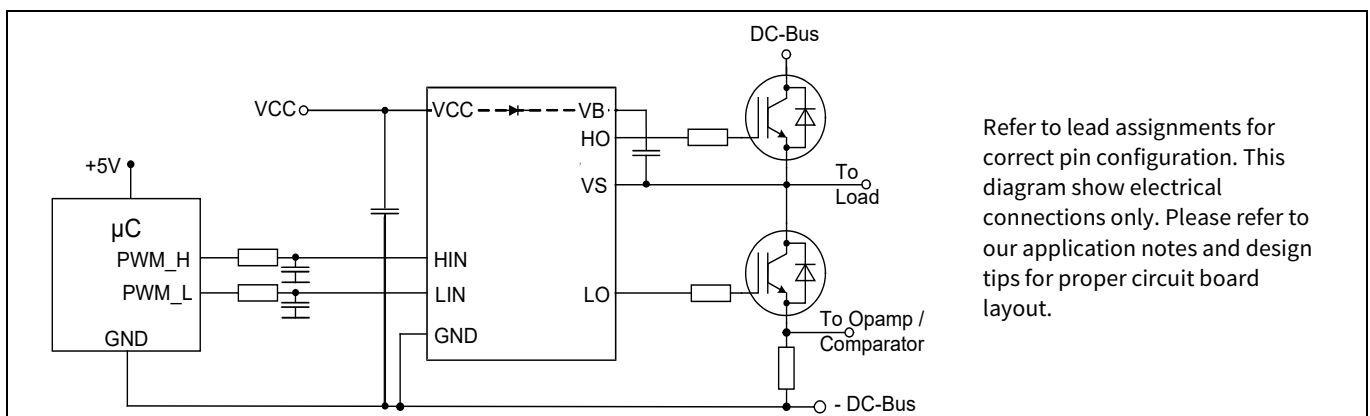
- Motor drives, General purpose inverters
- Refrigeration compressors
- Half-bridge and full-bridge converters in offline AC-DC power supplies for telecom and lighting

#### Product validation

Qualified for industrial applications according to the relevant tests of JEDEC47/20/22.

#### Description

The 2ED2304S06F is a 650-V half-bridge gate driver. Its Infineon thin-film-SOI technology provides excellent ruggedness and noise immunity. The Schmitt trigger logic inputs are compatible with standard CMOS or LSTTL logic down to 3.3 V. The output drivers features a high pulse current buffer stage designed for minimum driver cross-conduction. The floating channel can be used to drive an N-channel power MOSFET or IGBT in the high side configuration which operates up to 650 V. Additionally, the offline clamping function provides an inherent protection of the parasitic turn-on by floating gate conditions when IC is not supplied.



Refer to lead assignments for correct pin configuration. This diagram show electrical connections only. Please refer to our application notes and design tips for proper circuit board layout.

**Figure 1** Typical application diagram

Ordering information

Ordering information

Base Part Number	Package	Standard Pack		Orderable Part Number
		Form	Quantity	
<a href="#">2ED2304S06F</a>	PG-DSO-8	Tube/Bulk	95	2ED2304S06FXLSA1
		Tape and Reel	2500	2ED2304S06FXUMA1

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# 1 Block diagram

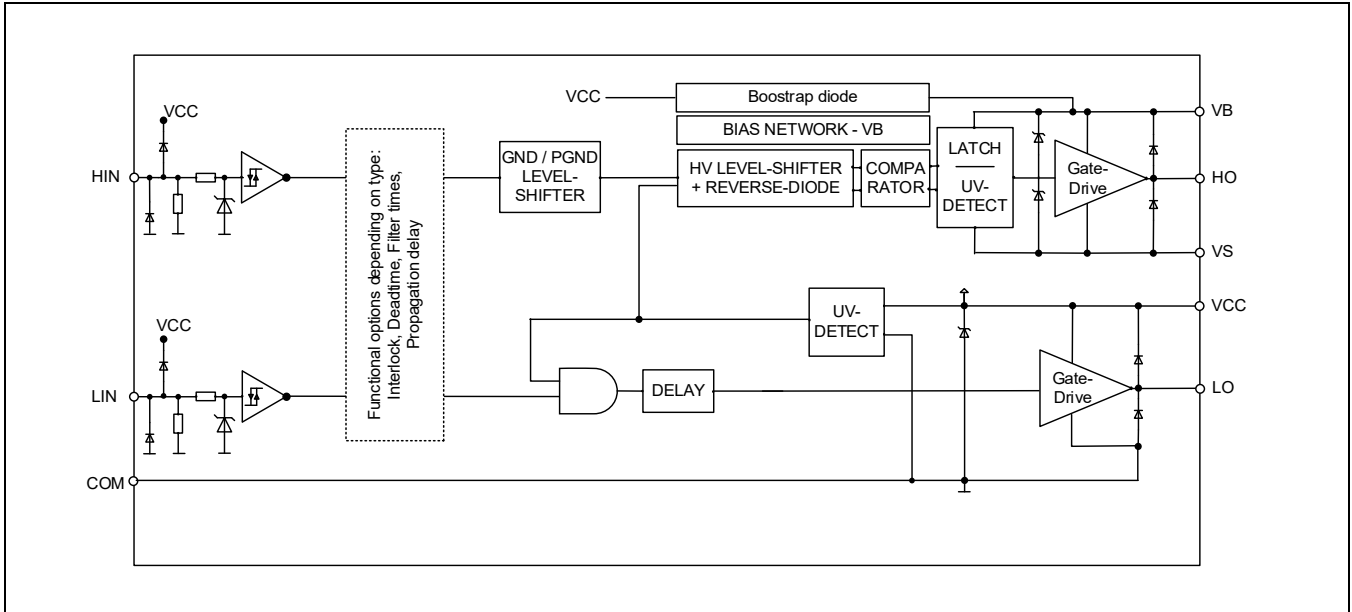


Figure 2 Functional block diagram

# 2 Lead definitions

Table 1 2ED2304S06F lead definitions

Pin no.	Name	Function
1	LIN	Logic input for low-side gate driver output (LO), in phase. Schmitt trigger inputs with hysteresis and pull down
2	HIN	Logic input for high-side gate driver output (HO), in phase. Schmitt trigger inputs with hysteresis and pull down
3	VCC	Low-side and logic supply voltage
4	COM	Low-side gate drive return
5	LO	Low-side driver output
6	VS	High voltage floating supply return
7	HO	High-side driver output
8	VB	High-side gate drive floating supply

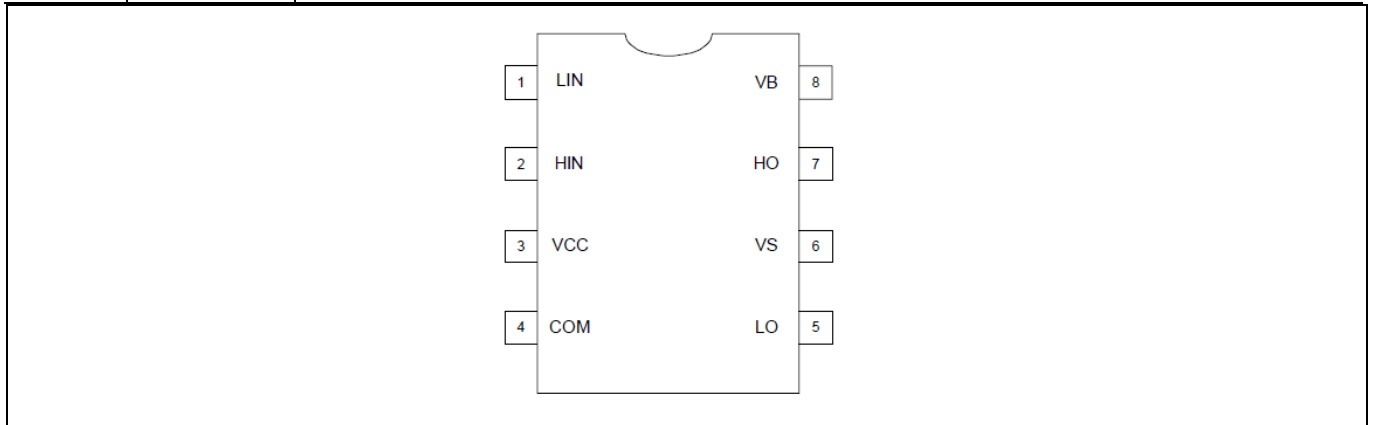


Figure 3 2ED2304S06F lead assignments PG-DSO-8 (top view)

### 3 Electrical parameters

#### 3.1 Absolute maximum ratings

Absolute maximum ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM unless otherwise stated in the table. The thermal resistance and power dissipation ratings are measured under board mounted and still air conditions.

**Table 2 Absolute maximum ratings**

Symbol	Definition	Min.	Max.	Units
$V_B$	High-side floating well supply voltage <sup>1</sup>	$V_{CC} - 6$	670	V
	High-side floating well supply voltage ( $t_p < 300$ ns) <sup>1</sup>	$V_{CC} - 100$	—	
$V_S$	High-side floating well supply return voltage	$V_{CC} - V_{BS} - 6$	650	
	High-side floating well supply return voltage ( $t_p < 300$ ns) <sup>1</sup>	$V_{CC} - V_{BS} - 100$	—	
$V_{HO}$	Floating gate drive output voltage	$V_S - 0.5$	$V_B + 0.5$	
$V_{BS}$	Floating gate drive voltage supply voltage	-1	20	
$V_{CC}$	Low side supply voltage	-1	20	
$V_{LO}$	Low-side output voltage	-0.5	$V_{CC} + 0.5$	
$V_{IN}$	Logic input voltage	-0.5	$V_{CC} + 0.5$	
$dV_S/dt$	Allowable $V_S$ offset supply transient relative to GND <sup>2</sup>	—	50	
$P_D$	Package power dissipation @ $T_A \leq +25$ °C	—	0.6	W
$R_{thJA}$	Thermal resistance, junction to ambient	—	195	°C/W
$T_J$	Junction temperature	—	150	°C
$T_S$	Storage temperature	-40	150	
$T_L$	Lead temperature (soldering, 10 seconds)	—	300	

#### 3.2 Recommended operating conditions

For proper operation, the device should be used within the recommended conditions. All voltage parameters are absolute voltages referenced to COM unless otherwise stated in the table. The offset rating is tested with supplies of  $(V_{CC} - COM) = (V_B - V_S) = 15$  V.

**Table 3 Recommended operating conditions**

Symbol	Definition	Min	Max	Units
$V_B$	High-side floating well supply voltage	$V_S + 10$	$V_S + 17.5$	V
$V_S$	High-side floating well supply offset voltage <sup>3</sup>	$V_{CC} - V_{BS} - 1$	650	
$V_{HO}$	Floating gate drive output voltage	10	$V_{BS}$	
$V_{BS}$	High-side supply voltage	10	17.5	
$V_{CC}$	Low-side supply voltage	10	17.5	
$V_{LO}$	Low-side output voltage	0	$V_{CC}$	
$V_{IN}$	Logic input voltage <sup>4</sup>	0	$V_{CC}$	
$T_A$	Ambient temperature	-40	125	°C
$t_{IN}$	Pulse width for ON and OFF <sup>5</sup>	0.3	—	µs

<sup>1</sup> In case  $V_{CC} > V_B$  there is an additional power dissipation in the internal bootstrap diode between pins  $V_{CC}$  and  $V_B$  in case of activated bootstrap diode. Insensitivity to negative transient not subject to production test. Verified by design/characterization.

<sup>2</sup> Not subject to production test, verified by characterization.

<sup>3</sup> Logic operation for  $V_S$  of -8 V to +600 V.

<sup>4</sup> All input pins (HIN, LIN) are internally clamped

<sup>5</sup> Input pulses may not be transmitted properly in case of LIN/HIN below 0.3 µs

### 3.3 Static electrical characteristics

$(V_{CC} - COM) = (V_B - V_S) = 15\text{ V}$ , and  $T_A = 25^\circ\text{C}$  unless otherwise specified. The  $V_{IL}$ ,  $V_{IH}$  and  $I_{IN}$  parameters are referenced to GND and are applicable to the respective input leads: HIN and LIN. The  $V_O$  and  $I_O$  parameters are referenced to COM/VS and are applicable to the respective output leads HO or LO. The  $V_{CCUV}$  parameters are referenced to COM. The  $V_{BSUV}$  parameters are referenced to  $V_S$ .

**Table 4 Static electrical characteristics**

Symbol	Definition	Min.	Typ.	Max.	Units	Test Conditions
$V_{BSUV+}$	$V_{BS}$ supply undervoltage positive going threshold	8.3	9.1	9.9	V	
$V_{BSUV-}$	$V_{BS}$ supply undervoltage negative going threshold	7.5	8.3	9.0		
$V_{BSUVHY}$	$V_{BS}$ supply undervoltage hysteresis	0.5	0.9	—		
$V_{CCUV+}$	$V_{CC}$ supply undervoltage positive going threshold	8.3	9.1	9.9		
$V_{CCUV-}$	$V_{CC}$ supply undervoltage negative going threshold	7.5	8.3	9.0		
$V_{CCUVHY}$	$V_{CC}$ supply undervoltage hysteresis	0.5	0.9	—		
$I_{LK}$	High-side floating well offset supply leakage	—	1	12.5	$\mu\text{A}$	$V_B = V_S = 600\text{ V}$
$I_{LK}$	High-side floating well offset supply leakage <sup>1</sup>	—	10	—		$T_J = 125^\circ\text{C}$ , $V_S = 600\text{ V}$
$I_{QBS}$	Quiescent $V_{BS}$ supply current	—	170	300		
$I_{QCC}$	Quiescent $V_{CC}$ supply current	—	300	600		
$V_{OH}$	High level output voltage drop, $V_{BIAS} - V_O$	—	0.45	1	V	$I_O = 20\text{ mA}$
$V_{OL}$	Low level output voltage drop, $V_O$	—	0.13	0.3		
$I_{O+}$	Peak output current turn-on <sup>1</sup>	—	360	—	mA	$V_O = 0\text{ V}$ $PW = 10\ \mu\text{s}$
$I_{O+mean}$	Mean output current from 3 V (20%) to 6 V (40%)	180	230	—		$C_L = 22\text{ nF}$
$I_{O-}$	Peak output current turn-off <sup>1</sup>	—	700	—		$V_O = 15\text{ V}$ $PW = 10\ \mu\text{s}$
$I_{O-mean}$	Mean output current from 12 V (80%) to 9 V (60%)	390	480	—		$C_L = 22\text{ nF}$
$V_{IH}$	Logic “1” input voltage	1.7	2.1	2.4	V	
$V_{IL}$	Logic “0” input voltage	0.7	0.9	1.1		
$I_{IN+}$	Input bias current (HO = High)	15	35	60	$\mu\text{A}$	$V_{IN} = 3.3\text{ V}$
$I_{IN-}$	Input bias current (HO = Low)	—	0	—		$V_{IN} = 0\text{ V}$
$V_{FBSD}$	Bootstrap diode forward voltage between $V_{CC}$ and $V_B$	—	1	1.2	V	$IF = 0.3\text{ mA}$
$I_{FBSD}$	Bootstrap diode forward current between $V_{CC}$ and $V_B$	30	55	80	mA	$V_{CC} - V_B = 4\text{ V}$
$R_{BSD}$	Bootstrap diode resistance	20	36	55	$\Omega$	$V_{F1} = 4\text{ V}$ , $V_{F2} = 5\text{ V}$

<sup>1</sup> Not subjected to production test, verified by characterization.

### 3.4 Dynamic electrical characteristics

$V_{CC} = V_{BS} = 15\text{ V}$ ,  $V_{SS} = \text{COM}$ ,  $T_A = 25^\circ\text{C}$  and  $C_L = 1000\text{ pF}$  unless otherwise specified.

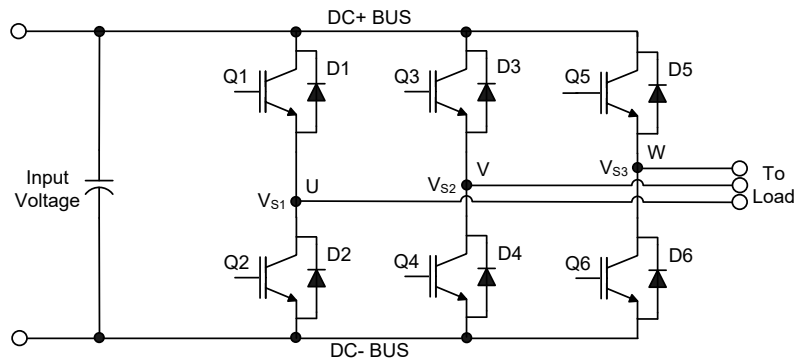
**Table 5** Dynamic electrical characteristics

Symbol	Definition	Min.	Typ.	Max.	Units	Test Conditions
$t_{ON}$	Turn-on propagation delay	210	310	460	ns	$V_{LIN/HIN} = 0\text{ or }3.3\text{ V}$
$t_{OFF}$	Turn-off propagation delay	200	300	440		$V_{LIN/HIN} = 0\text{ or }3.3\text{ V}$
$t_R$	Turn-on rise time	—	48	80		$C_L = 1\text{ nF}$
$t_F$	Turn-off fall time	—	24	40		$V_{LIN/HIN} = 0\text{ \& }3.3\text{ V}$
$t_{FILIN}$	Input filter time	100	150	250		external dead time > 500 ns
MT	Delay matching time (HS & LS turn-on/off)	—	10	60		$V_{LIN/HIN} = 0\text{ \& }3.3\text{ V}$
DT	Dead time	30	75	140		ext. dead time 0 ns
MDT	Dead time matching time	—	10	50		

## 4 Tolerant to negative transient voltage on VS pin (-VS)

A common problem in today’s high-power switching converters is the transient response of the switch node’s voltage as the power switches transition on and off quickly while carrying a large current. A typical three phase inverter circuit is shown in Figure 4; here we define the power switches and diodes of the inverter.

If the high-side switch (e.g., the IGBT Q1 in Figures 5 and 6) switches off, while the U phase current is flowing to an inductive load, a current commutation occurs from high-side switch (Q1) to the diode (D2) in parallel with the low-side switch of the same inverter leg. At the same instance, the voltage node  $V_{S1}$ , swings from the positive DC bus voltage to the negative DC bus voltage.



**Figure 4** Three phase inverter

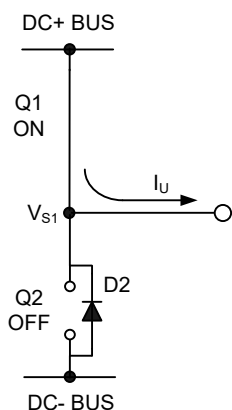


Figure 5 Q1 conducting

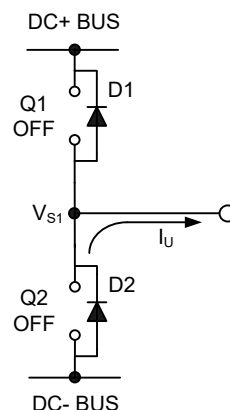


Figure 6 D2 conducting

Also when the V phase current flows from the inductive load back to the inverter (see Figures 7 and 8), and Q4 IGBT switches on, the current commutation occurs from D3 to Q4. At the same instance, the voltage node,  $V_{s2}$ , swings from the positive DC bus voltage to the negative DC bus voltage.

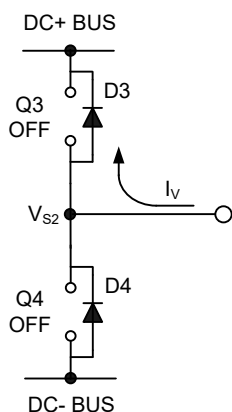


Figure 7 D3 conducting

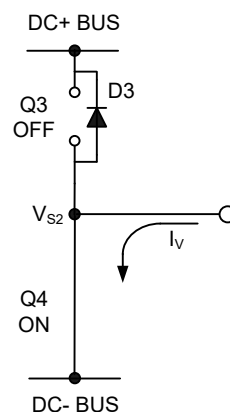


Figure 8 Q4 conducting

However, in a real inverter circuit the  $V_s$  voltage swing does not stop at the level of the negative DC bus but instead swings below the level of the negative DC bus. This undershoot voltage is called “negative transient voltage”.

The circuit shown in Figure 9 depicts one leg of the three phase inverter; Figures 10 and 11 show a simplified illustration of the commutation of the current between Q1 and D2. The parasitic inductances in the power circuit from the die bonding to the PCB tracks are lumped together in  $L_c$  and  $L_e$  for each IGBT. When the high-side switch is on,  $V_{s1}$  is below the DC+ voltage by the voltage drops associated with the power switch and the parasitic elements of the circuit. When the high-side power switch turns off, the load current momentarily flows in the low-side freewheeling diode due to the inductive load connected to  $V_{s1}$  (the load is not shown in these figures). This current flows from the DC- bus (which is connected to the COM pin of the HVIC) to the load and a negative voltage between  $V_{s1}$  and the DC- Bus is induced (i.e., the COM pin of the HVIC is at a higher potential than the VS pin).

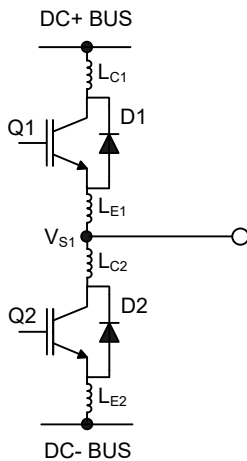


Figure 9 Parasitic Elements

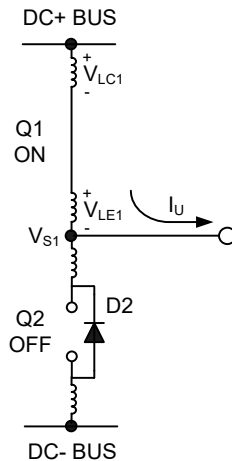


Figure 10 VS positive

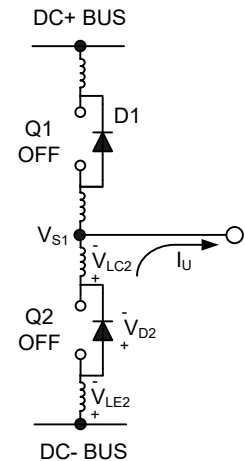


Figure 11 VS negative

In a typical motor drive system,  $dV/dt$  is typically designed to be in the range of 3-5 V/ns. The negative transient voltage can exceed this range during some events such as short circuit and over-current shutdown, when  $di/dt$  is greater than in normal operation.

Infineon’s HVICs have been designed for the robustness required in many of today’s demanding applications. An indication of the 2ED2304S06F’s robustness can be seen in Figure 12, where the 2ED2304S06F Safe Operating Area is shown at  $V_{BS}=15$  V based on repetitive negative transient voltage spikes. A negative transient voltage falling in the grey area (outside SOA) may lead to IC permanent damage; viceversa unwanted functional anomalies or permanent damage to the IC do not appear if negative VS transients fall inside the SOA.

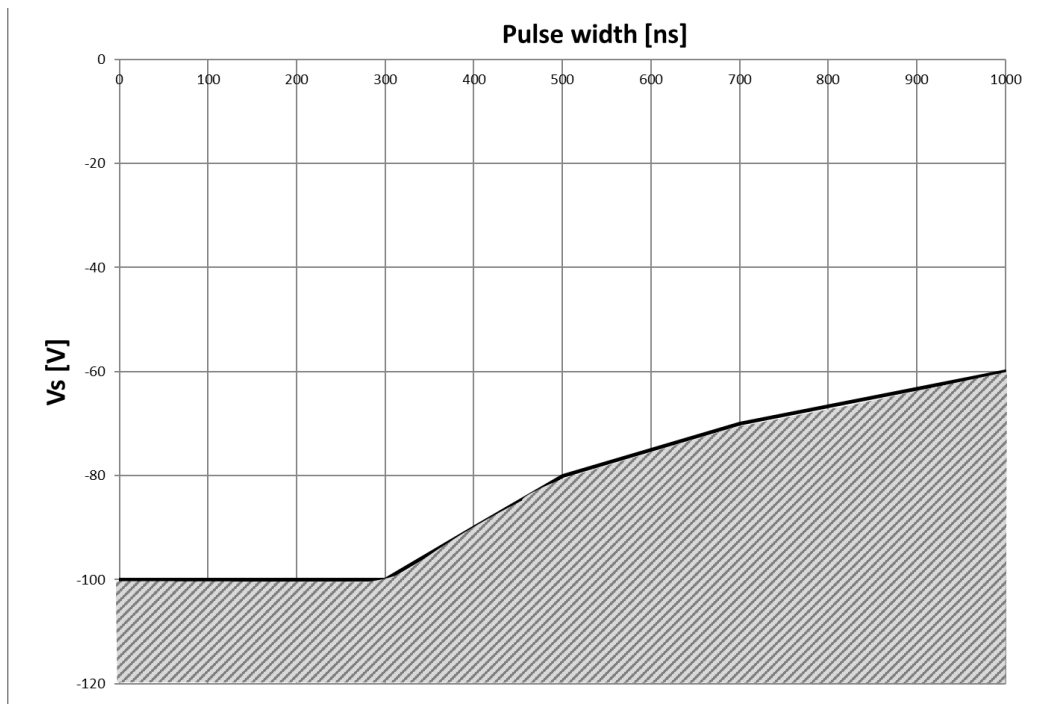


Figure 12 Negative transient voltage SOA on VS pin for 2ED2304S06F @  $V_{BS}=15$  V

Even though the 2ED2304S06F has been shown to be able to handle these large negative transient voltage conditions, it is highly recommended that the circuit designer always limit the negative transient voltage on VS pin as much as possible by careful PCB layout and component use.



### 5 Package information DSO-8

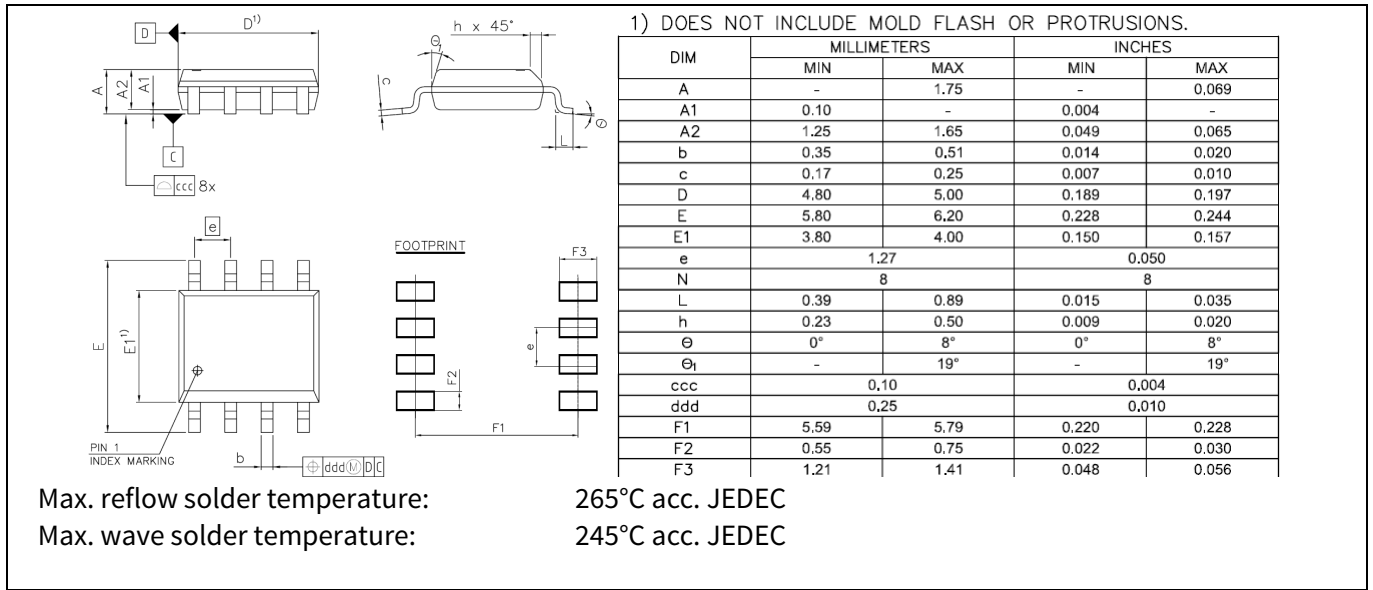


Figure 13 Package outline PG-DSO-8

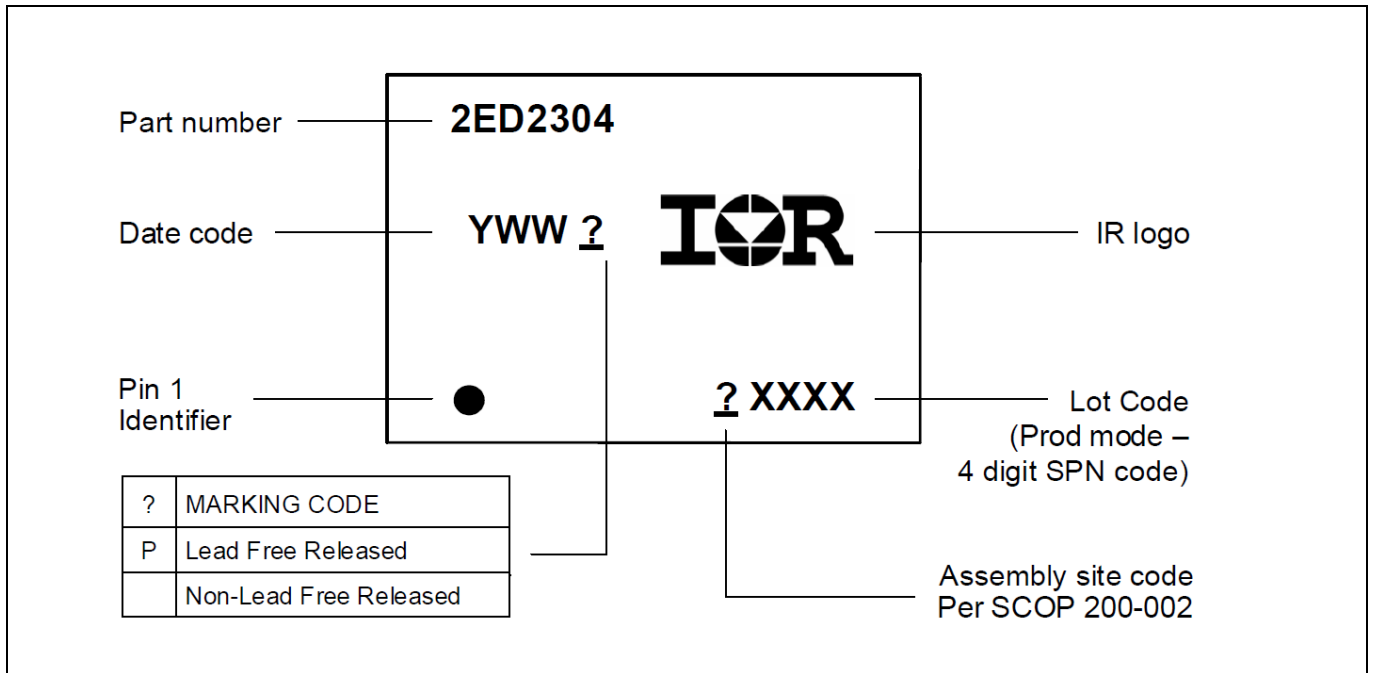


Figure 14 Marking information PG-DSO-8 (2ED2304S06F)

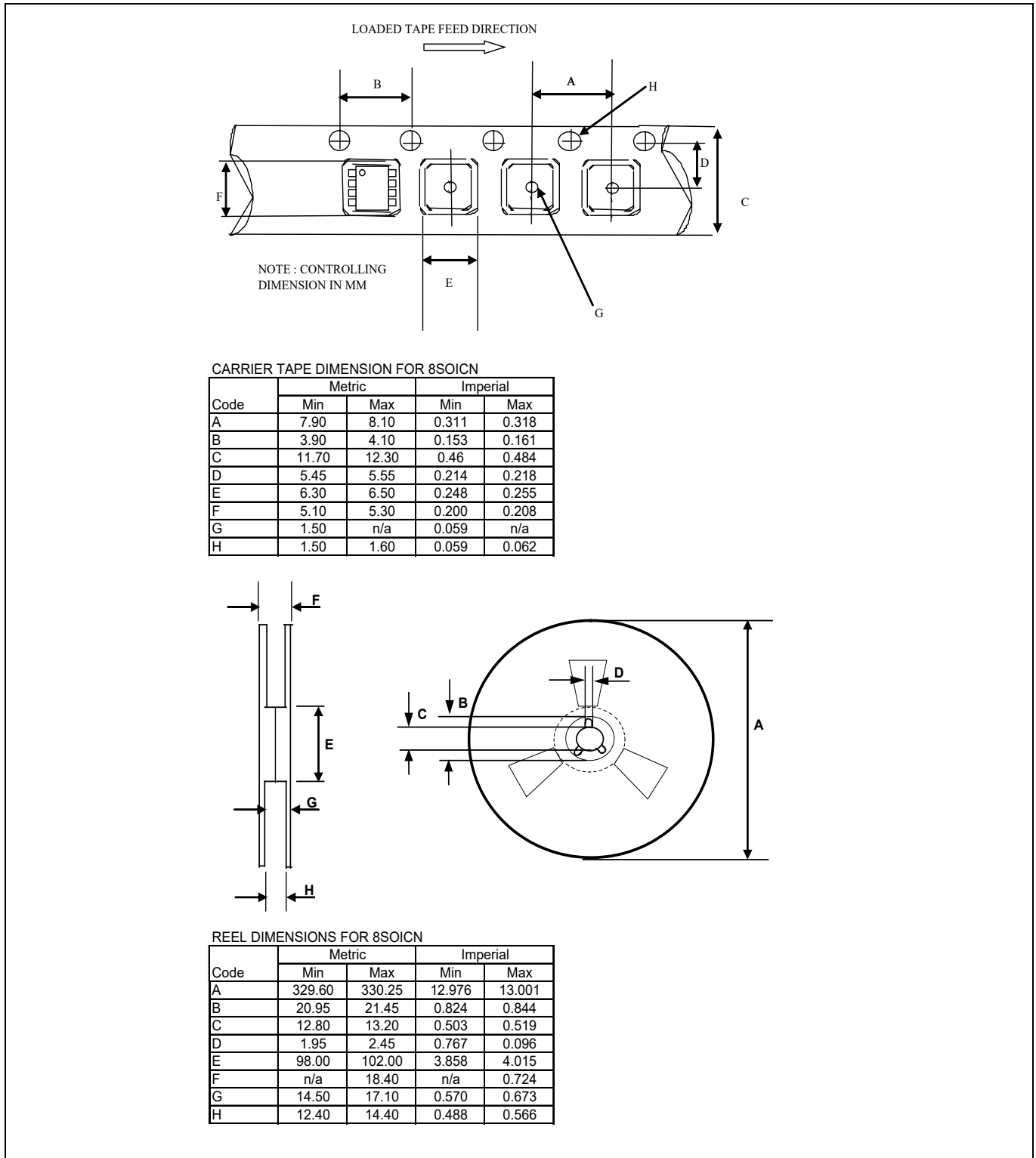


Figure 15 Tape and reel details PG-DSO-8

## 6 Qualification information<sup>1</sup>

**Table 6 Qualification information**

Qualification level		Industrial <sup>2</sup>	
		Note: This family of ICs has passed JEDEC's Industrial qualification. Consumer qualification level is granted by extension of the higher Industrial level.	
Moisture sensitivity level		DSO-8	MSL3 <sup>3</sup> , 260°C (per IPC/JEDEC J-STD-020)
ESD	Charged device model	Class C3 (> 1.0 kV) (per JESD22-C101)	
	Human body model	Class 2 (per JEDEC standard JESD22-A114)	
IC latch-up test		Class II Level A (per JESD85)	
RoHS compliant		Yes	

## 7 Related products

**Table 7**

Product	Description
<b>Gate Driver ICs</b>	
<a href="#">6EDL04I06 / 6EDL04N06</a>	600 V, 3 phase level shift thin-film SOI gate driver with integrated high speed, low $R_{DS(ON)}$ bootstrap diodes with over-current protection (OCP), 240/420 mA source/sink current drive, Fault reporting, and Enable for MOSFET or IGBT switches.
<a href="#">2EDL23I06 / 2EDL23N06</a>	600 V, Half-bridge thin-film SOI level shift gate driver with integrated high speed, low $R_{DS(ON)}$ bootstrap diode, with over-current protection (OCP), 2.3/2.8 A source/sink current driver, and one pin Enable/Fault function for MOSFET or IGBT switches.
<b>Power Switches</b>	
<a href="#">IKD04N60R / RE</a>	600 V TRENCHSTOP™ IGBT with integrated diode in PG-TO252-3 package
IKD06N65ET6	650 V TRENCHSTOP™ IGBT with integrated diode in DPAK
<a href="#">IPD65R950CFD</a>	650 V CoolMOS CFD2 with integrated fast body diode in DPAK
<a href="#">IPN50R950CE</a>	500 V CoolMOS CE Superjunction MOSFET in PG-SOT223 package
<b>iMOTION™ Controllers</b>	
<a href="#">IRMCK099</a>	iMOTION™ Motor control IC for variable speed drives utilizing sensor-less Field Oriented Control (FOC) for Permanent Magnet Synchronous Motors (PMSM).
<a href="#">IMC101T</a>	High performance Motor Control IC for variable speed drives based on field oriented control (FOC) of permanent magnet synchronous motors (PMSM).

<sup>1</sup> Qualification standards can be found at Infineon's web site [www.infineon.com](http://www.infineon.com)

<sup>2</sup> Higher qualification ratings may be available should the user have such requirements. Please contact your Infineon sales representative for further information.

<sup>3</sup> Higher MSL ratings may be available for the specific package types listed here. Please contact your Infineon sales representative for further information.

## Revision history

Document version	Date of release	Description of changes
1.0	2016-07-12	Preliminary datasheet
2.0	2018-02-07	First Release Version
2.1	2018-07-13	Updated the marking information
2.11	2018-09-12	Deleting typo
2.2	2018-10-26	Adding negative $V_s$ information
2.3	2018-11-19	Updated ESD HBM information
2.4	2019-01-24	Updated Chapter 4 Tolerant to negative transient voltage on VS pin

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- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



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