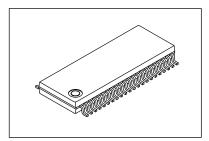
# LV8702V

## **PWM Current Control High-efficient** Stepper Motor Driver



#### Overview

The LV8702V is a 2-channel Full-bridge driver IC that can drive a stepper motor driver, which is capable of micro-step drive and supports quarter step. Current is controlled according to motor load and rotational speed at half step, half step full-torque and quarter step excitation, thereby highly efficient drive is realized. Consequently, the reduction of power consumption, heat generation, vibration and noise is achieved.



#### Feature

- Built-in 1ch PWM current control stepper motor driver (bipolar type)
- Ron (High-side Ron:  $0.3\Omega$ , Low-side Ron:  $0.25\Omega$ , total:  $0.55\Omega$ , Ta =  $25^{\circ}$ C, I $_{\Omega}$  = 2.5A)
- Micro-step mode is configurable as follows: full step/half step full-torque/half step/quarter step
- Excitation step moves forward only with step signal input
- Built-in output short protection circuit (latch method)
- Control power supply is unnecessary
- Built-in high-efficient drive function (supports half step full-torque/half step/quarter step excitation mode)
- Built-in step-out detection function (Step-out detection may not be accurate during high speed rotation)
- BiCDMOS process IC
- IO max=2.5A
- Built-in thermal shut down circuit

## **Typical Applications**

- Printer
- Scanner
- Surveillance camera (CCTV)
- Textile machine

#### **ORDERING INFORMATION**

See detailed ordering and shipping information on page 27 of this data sheet.

SSOP44J (275mil)

#### **Specifications Absolute Maximum Ratings** at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	
Power supply voltage	V <sub>M</sub> max	VM , VM1 , VM2	/M1 , VM2 36	
Output peak current	I <sub>O</sub> peak	tw $\leq$ 10ms , duty 20% , Per 1ch 3		А
Output current	I <sub>O</sub> max	Per 1ch 2.5		А
Logic input voltage	V <sub>IN</sub>	GMG1, GMG2 , GAD , FR , STEP , ST , RST , MD1 , MD2 , OE , GST1 , GST2	–0.3 to +6	V
DST1, DST2, MONI,	Vdst1, Vdst2,		–0.3 to +6	V
Allowable power dissipation	Pd max	*	5.5	W
Operating temperature	Topr		-40 to +85	°C
Storage temperature	Tstg		–55 to +150	°C

\* Specified board : 90.0mm  $\times$  90.0mm  $\times$  1.6mm, glass epoxy 4-layer board, with backside mounting.

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.

Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

#### **Recommended Operating Range** at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Range of power supply voltage	VM	VM , VM1 , VM2	9 to 32	V
Logic input voltage	V <sub>IN</sub>	GMG1 , GMG2 , GAD , FR , STEP , ST , RST , MD1 , MD2 , OE , GST1 , GST2	0 to 5.5	V
Range of VREF input voltage	VREF		0 to 3	V

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

## Electrical Characteristics at Ta = 25°C, $V_M$ = 24V, VREF = 1.5V

P.		0 sets st		0		Ratings		11.21
Parameter		Symbol	Symbol Conditions		min	typ	max	Unit
Consumption c standby	urrent during	IMstn	ST = "L" , I(VN	1)+I(VM1)+I(VM2)		110	400	μA
Consumption c	urrent	IM	ST = "H", OE = I(VM)+I(VM1)+	= "L", STEP = "L", non-load H(VM2)		4.5	6.5	mA
VREG5 output	voltage	VREG5	I <sub>O</sub> = -1mA		4.5	5	5.5	V
Thermal shutdo	own temperature	TSD	Design certific	ation	150	180	210	°C
Thermal hyster	resis width	∆TSD	SD Design certification			40		°C
Motor driver		•	•			•		
Output on resis	Output on resistor		I <sub>O</sub> = 2.5A, Source-side Ron			0.3	0.4	Ω
		Rond	I <sub>O</sub> = 2.5A, Sink-side Ron			0.25	0.33	Ω
Output leak cur	rrent	l <sub>O</sub> leak	VM = 32V				50	μA
Forward diode	voltage	VD	ID = -2.5A			1.2	1.4	V
Logic pin input	current	I <sub>IN</sub> L	V <sub>IN</sub> = 0.8V	GMG1, GMG2, GAD, FR,	4	8	12	μA
		IINH	V <sub>IN</sub> = 5V	STEP, ST, RST, MD1, MD2, OE, GST1, GST2	30	50	70	μA
ADIN pin input voltage		Vadin	Ra2 = 100kΩ:	refer to 15-4)	0		12	V
Logic input	High	V <sub>IN</sub> H	GMG1, GMG2	2 , GAD , FR , STEP , ST ,	2.0		5.5	V
voltage	Low	VINL	RST , MD1 , M	ID2 , OE , GST1 , GST2	0		0.8	V

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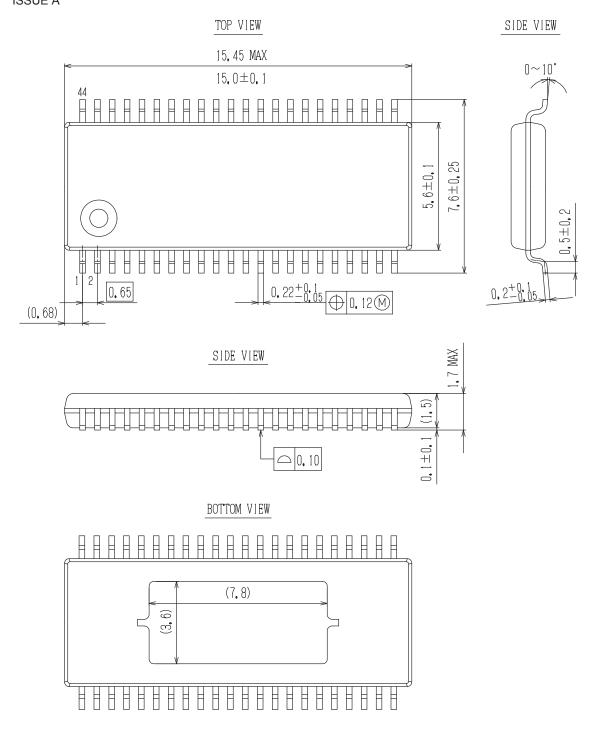
D.		0 mbal			Ratings		
Parameter		Symbol	Symbol Conditions		min typ max		Unit
Current quarter step		Vtdac0_W	Step0 (initial status, 1ch comparator level)	290	300	310	mV
selection		Vtdac1_W	Step1 (initial + 1)	264	276	288	mV
reference		Vtdac2_W	Step2 (initial + 2)	199	210	221	mV
voltage level		Vtdac3_W	Step3 (initial + 3)	106	114	122	mV
	half step	Vtdac0_H	Step0 (initial status, 1ch comparator level)	290	300	310	mV
		Vtdac2_H	Step2 (initial + 1)	199	210	221	mV
(fu	half step	Vtdac0_HF	Step0 (initial status, 1ch comparator level)	290	300	310	mV
	(full-torque)	Vtdac2'_HF	Step2' (initial + 1)	290	300	310	mV
	full step	Vtdac2'_F	Step2' (initial status, 1ch comparator level)	290	300	310	mV
Chopping frequency		Fchop	Cchop = 200pF	35	50	65	kHz
CHOP pin charge/discharge		Ichop		7	10	13	μA
Chopping oscill	ator circuit	Vtup		0.8	1	1.2	V
threshold voltage	ge	Vtdown		0.4	0.5	0.6	V
VREF pin input	current	Iref	VREF = 1.5V	-0.5			μA
DST1, DST2, MONI, SST pin saturation voltage			ldst1 = ldst2 = Imoni = lsst = 1mA			400	mV
Charge pump							
VG output voltage		VG		28	28.7	29.8	V
Rise time		tONG	VG = $0.1\mu$ F , Between CP1-CP2 0.1uF ST="H" $\rightarrow$ VG=VM+4V			0.5	mS
Oscillator freque	ency	Fosc		90	125	160	kHz

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

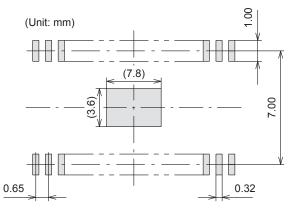
#### **Package Dimensions**

unit : mm

SSOP44J (275mil) Exposed Pad CASE 940AG ISSUE A



#### **SOLDERING FOOTPRINT\***



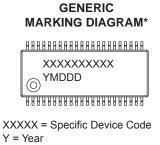
#### NOTES:

- 1. The measurements are for reference only, and unable to guarantee.
- 2. Please take appropriate action to design the actual Exposed Die Pad and Fin portion.
- 3. After setting, verification on the product must be done.

(Although there are no recommended design for Exposed Die Pad and Fin portion Metal mask and shape for Through–Hole pitch (Pitch & Via etc), checking the soldered joint condition and reliability verification of soldered joint will be needed. Void 

gradient 
insufficient thickness of soldered joint or bond degradation could lead IC destruction because thermal conduction to substrate becomes poor.)

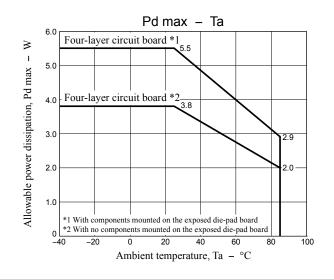
\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.





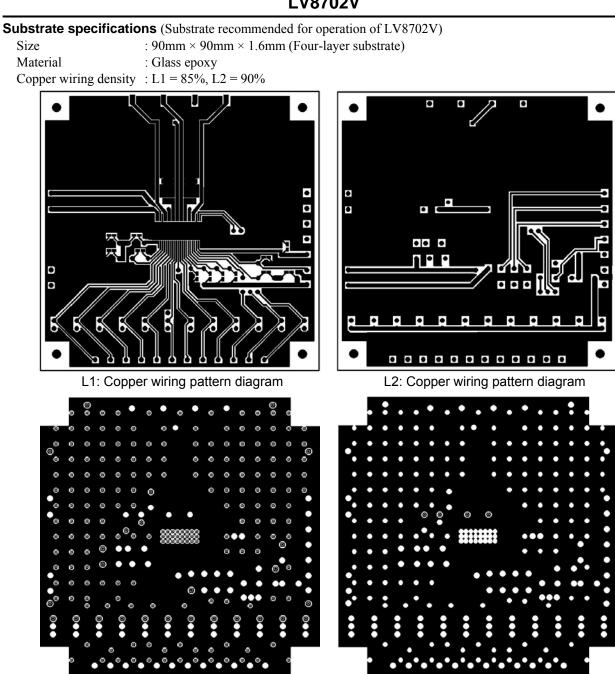
DDD = Additional Traceability Data

\*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot " •", may or may not be present.



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## LV8702V



L3: GND laver

Power supply layer <u>4</u>.

#### Cautions

1) The data for the case with the Exposed Die-Pad substrate mounted shows the values when 90% or more of the Exposed Die-Pad is wet.

2) For the set design, employ the derating design with sufficient margin.

Stresses to be derated include the voltage, current, junction temperature, power loss, and mechanical stress such as vibration, impact, and tension.

Accordingly, the design must ensure these stresses to be as low or small as possible.

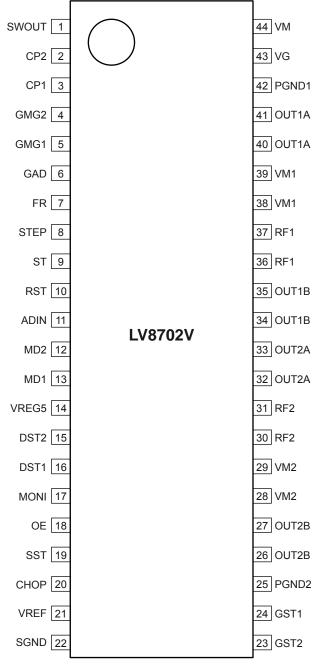
The guideline for ordinary derating is shown below:

- (1)Maximum value 80% or less for the voltage rating
- (2)Maximum value 80% or less for the current rating
- (However this does not apply to high efficiency drive because operating current is lower than the setting current.) (3)Maximum value 80% or less for the temperature rating

3) After the set design, be sure to verify the design with the actual product.

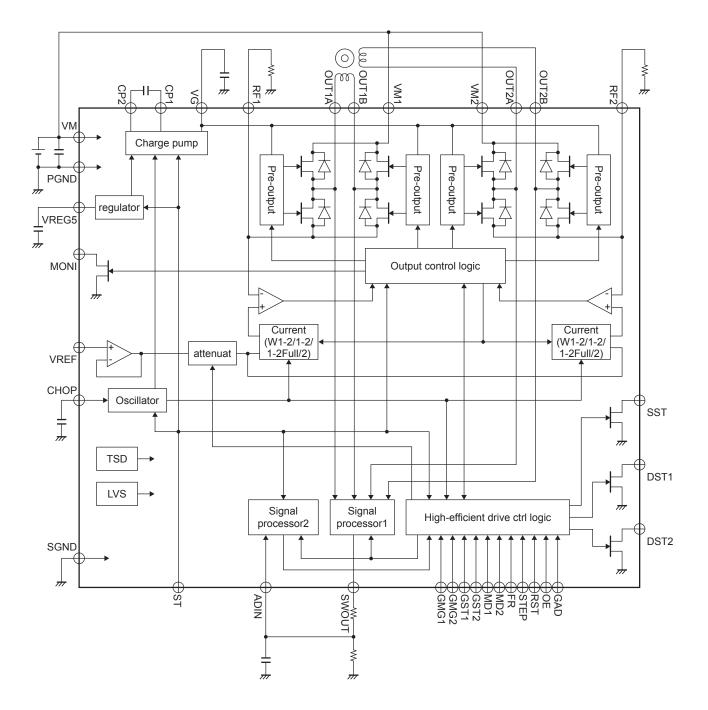
Confirm the solder joint state and verify also the reliability of solder joint for the Exposed Die-Pad, etc. Any void or deterioration, if observed in the solder joint of these parts, causes deteriorated thermal conduction, possibly resulting in thermal destruction of IC.

## Pin Assignment



Top view

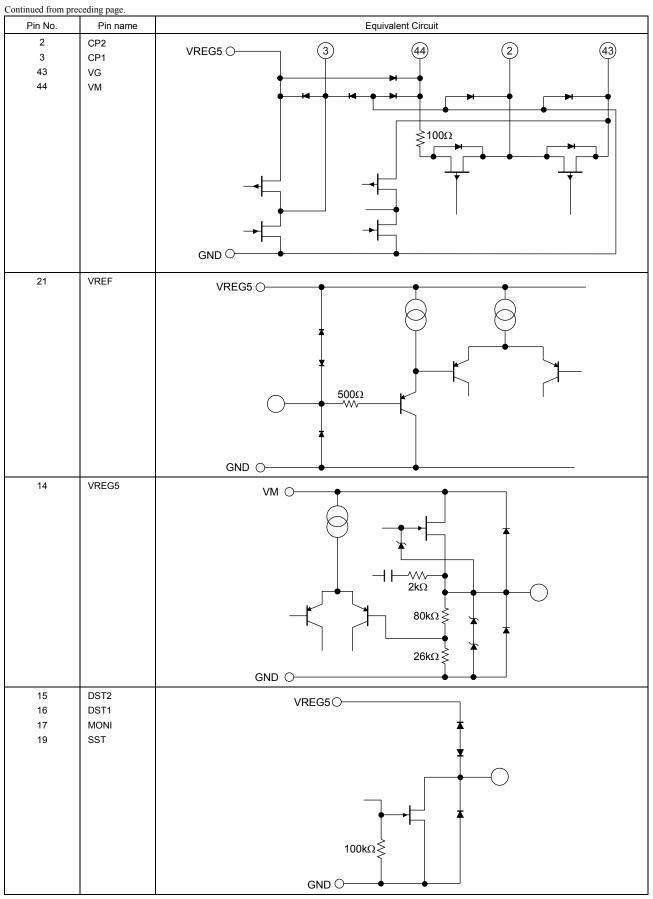
## Block Diagram



Pin No.	Pin name	Description
1	SWOUT	Control signal output pin
2	CP2	Capacitor connection pin for charge pump
3	CP1	Capacitor connection pin for charge pump
4	GMG2	Driving capability margin adjuster pin
5	GMG1	Driving capability margin adjuster pin
6	GAD	High-efficient drive switching pin
7	FR	Forward/ reverse signal input pin
8	STEP	STEP signal input pin
9	ST	Chip enable pin
10	RST	RESET signal input pin
11	ADIN	Control signal input pin
12	MD2	Excitation mode switching pin
13	MD1	Excitation mode switching pin
14	VREG5	Capacitor connection pin for internal power supply
15	DST2	Drive status warning output pin
16	DST1	Drive status warning output pin
17	MONI	Position detection monitor pin
18	OE	Output enable signal input pin
19	SST	Motor stop detection output pin
20	CHOP	Capacitor connection pin for chopping frequency setting
21	VREF	Constant current control reference voltage input pin
22	SGND	Signal GND
23	GST2	Boost-up adjuster pin
24	GST1	Boost-up adjuster pin
25	PGND2	2ch power GND
26, 27	OUT2B	2ch OUTB output pin
28, 29	VM2	2ch motor power supply connection pin
30, 31	RF2	2ch current sense resistor connection pin
32, 33	OUT2A	2ch OUTA output pin
34, 35	OUT1B	1ch OUTB output pin
36, 37	RF1	1ch current sense resistor connection pin
38, 39	VM1	1ch motor power supply connection pin
40, 41	OUT1A	1ch OUTA output pin
42	PGND1	1ch power GND
43	VG	Capacitor connection pin for charge pump

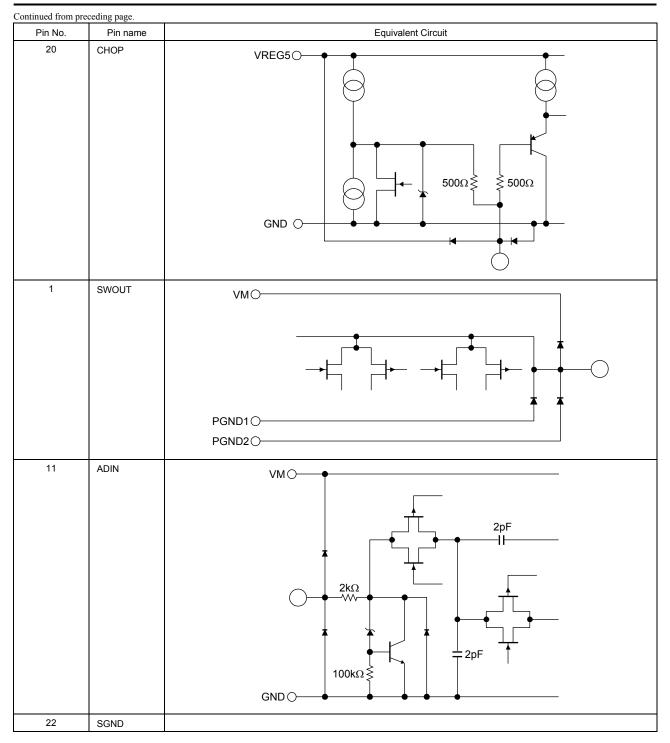
Pin Descrip	tion	
Pin No.	Pin name	Equivalent Circuit
4	GMG2	
		VREG5O
5	GMG1	
6	GAD	
7	FR	
8	STEP	
10	RST	<b>★ ★↓</b>
12	MD2	
13	MD1	
18	OE	
23	GST2	
24	GST1	
	0011	
		<b>★</b> <b>♦</b> 100kΩ <b>★ ♦ • • • • • • • • • •</b>
9	ST	
-	-	VREG5
		▲
		¥
		<b>ξ</b> 20kΩ
		10κΩ
		δ 80kΩ
25	PGND2	(38)(39)
26, 27	OUT2B	28720
28, 29	VM2	Keykar
30, 31	RF2	
32, 33	OUT2A	
34, 35	OUT1B	
36, 37	RF1	
38, 39	VM1	
40, 41	OUT1A	
42	PGND1	
		│
1	1	(30)(31)

Continued on next page.



Continued on next page.

## LV8702V



#### **Operation description**

#### Input Pin Function

Each input terminal has the function to prevent the flow of the current from an input to a power supply. Therefore, even if a power supply(VM) is turned off in the state that applied voltage to an input terminal, the electric current does not flow into the power supply.

#### 1. Chip enable function

The mode of the IC is switched with ST pin between standby and operation mode. In standby mode, the IC is set to power saving mode and all the logics are reset. During standby mode, the operation of the internal regulator circuit and the charge pump circuit are stopped.

ST	mode Internal regulator		Charge pump
"L" or OPEN	Standby mode	standby	standby
"H"	Operation mode	operation	operation

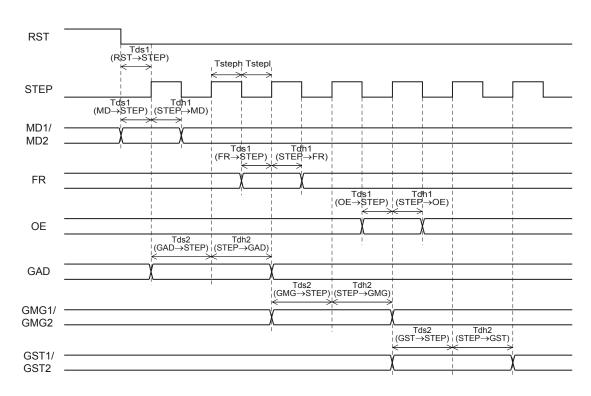
#### 2. STEP pin function

The excitation step progresses by inputting the step signal to the STP pin.

Input		Operation mode
ST	STEP	
L or OPEN	X*	Standby mode
Н		Excitation step forward
Н		Excitation step keep

\* Don't care

#### 3. Input timing



TstepH/TstepL : Clock H/L pulse width (min 12.5µs)

- Tds1 : Data set-up time (min 12.5µs)
- Tdh1 : Data hold time (min 12.5µs)
- Tds2 : Data set-up time (min 25µs)
- Tdh2 : Data hold time (min 25µs)

#### 4. Position detection monitor function

The MONI position detection monitoring pin is of an open drain type. When the excitation position is in the initial position, the MONI output is placed in the ON state. (Refer to "Examples of current waveforms in each micro-step mode.")

#### 5. Setting constant-current control reference current

This IC is designed to automatically exercise PWM constant-current chopping control for the motor current by setting the output current. Based on the voltage input to the VREF pin and the resistance connected between RF and GND, the output current that is subject to the constant-current control is set using the calculation formula below:

 $I_{OUT} = (VREF/5)/RF$  resistance

The above setting is the output current at 100% of each excitation mode.

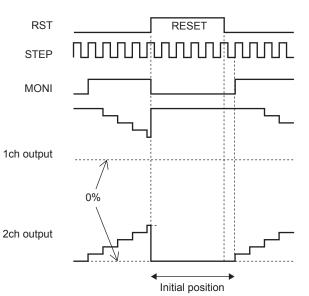
For example, where VREF=1.5V and RF resistance 0.2Ω, we obtain output current as follows.

 $I_{OUT} = 1.5 V/5/0.2 \Omega = 1.5 A$ 

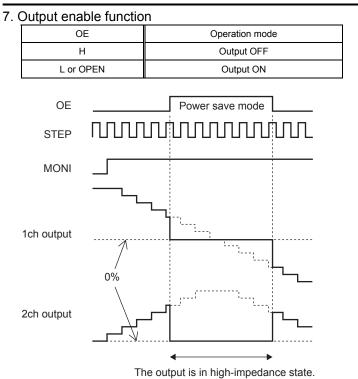
When high-efficient drive function is on, I<sub>OUT</sub> is adjusted automatically within the range of the current value set by VREF.

#### 6. Reset function

RST	Operation mode
L or OPEN	Normal operation
Н	RESET status



When RST pin = "H", the excitation position of the output is set to the initial position forcibly and MONI output is turned on. And then by setting RST = "L", the excitation position moves forward with the next step signal.



When OE pin = "H", the output is turned off forcibly and becomes a high-impedance output. However, since the internal logic circuit is in operation, an excitation position moves forward if step signal is input to STEP pin. Therefore, by setting back to OE = "L", the output pin outputs signal based on the excitation position by step signal.

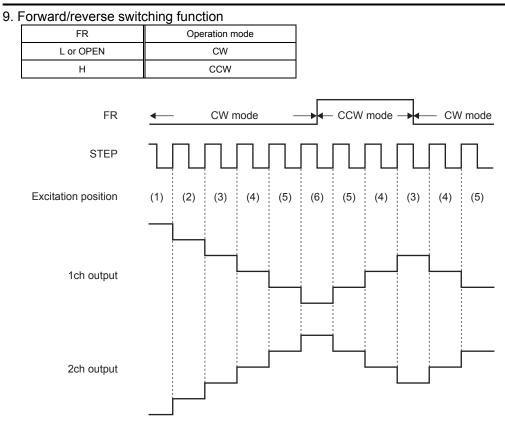
#### 8. Excitation mode setting function

MD1 and MD2 pin set excitation mode of the stepper motor as follows.

MD4	MDO	E situita sada	Initial p	l position	
MD1	MD2	Excitation mode	1ch	2ch	
L or OPEN	L or OPEN	full step excitation	100%	-100%	
Н	L or OPEN	half step excitation	100%	0%	
L or OPEN	н	quarter step excitation	100%	0%	
Н	Н	half step excitation (full-torque)	100%	0%	

The position of excitation mode is set to the initial position when: 1) a power is supplied and 2) counter is reset in each excitation mode.

During full step excitation mode, high-efficient drive function is turned off even when GAD = "H".



The built-in DA converter moves forward by 1bit with the rise of step signal that is input to STEP pin. Also a mode is switched between CW and CCW by setting FR pin.

In CW mode, the phase of 2ch current delays by 90° compared to that of 1ch current.

In CCW mode the phase of 2ch current moves forward by 90° compared to 1ch current.

#### 10. Chopping frequency setting

When you control constant current of this IC, chopping is performed using the frequency defined in the capacitor (Cchop) connected between CHOP pin and GND.

The calculation for the value of chopping frequency is:

Fchop = Ichop/ (Cchop×Vtchop×2) (Hz)

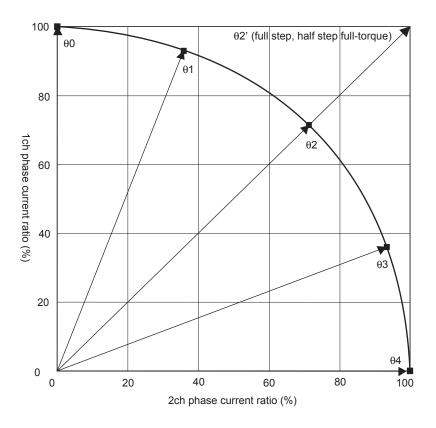
Ichop: Capacitor charge and discharge current typ: 10µA Vtchop: Charge and discharge hysteresis voltage (Vtup-Vtdown) typ: 0.5V

For example, where Cchop = 200pF, we obtain Fchop as follows: Fchop =  $10\mu A/(200pF \times 0.5V \times 2) = 50 \text{kHz}$ 

#### 11. Blanking time

If you attempt to control PWM constant current chopping of the motor current, when the mode shifts from DECAY to CHARGE, noise is generated in sense resistor pin due to the recovery current of parasitic diode flowing into current sense resistor, and this may cause error detection. The blanking time avoids noise at mode switch. During the blanking time, even if noise is generated in sense resistor, a mode does not switch from CHARGE to DECAY. In this IC, the blanking time is fixed to approximately 1µs.

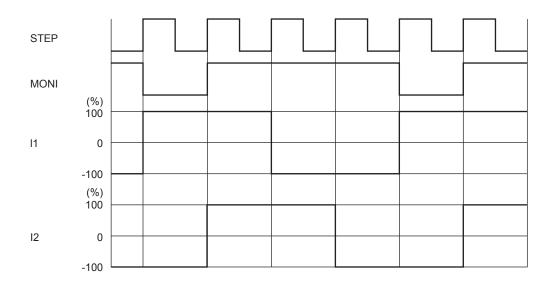
#### 12. Output current vector locus (1step is normalized to 90°)



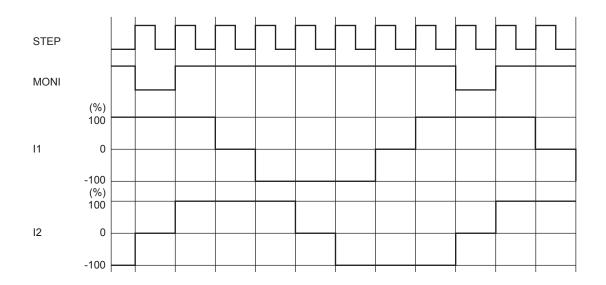
#### Setting current ration in each excitation mode

STEP	quarter s	step (%)	half ste	ep (%)	half step full	-torque (%)	full ste	p (%)
	1ch	2ch	1ch	2ch	1ch	2ch	1ch	2ch
θ0	100	0	100	0	100	0		
θ1	92	38						
θ2	70	70	70	70	100	100	100	100
θ <b>3</b>	38	92						
θ <b>4</b>	0	100	0	100	0	100		

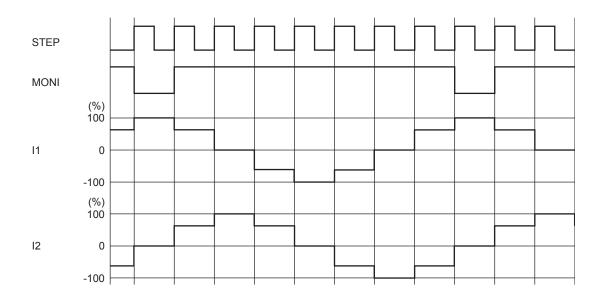
13. The example of current waveform in each micro-step mode full step (CW mode)



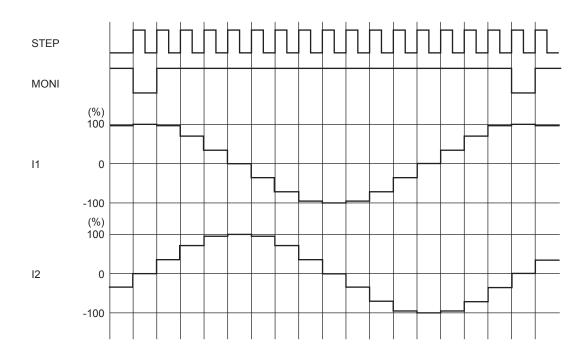
half step full-torque (CW mode)

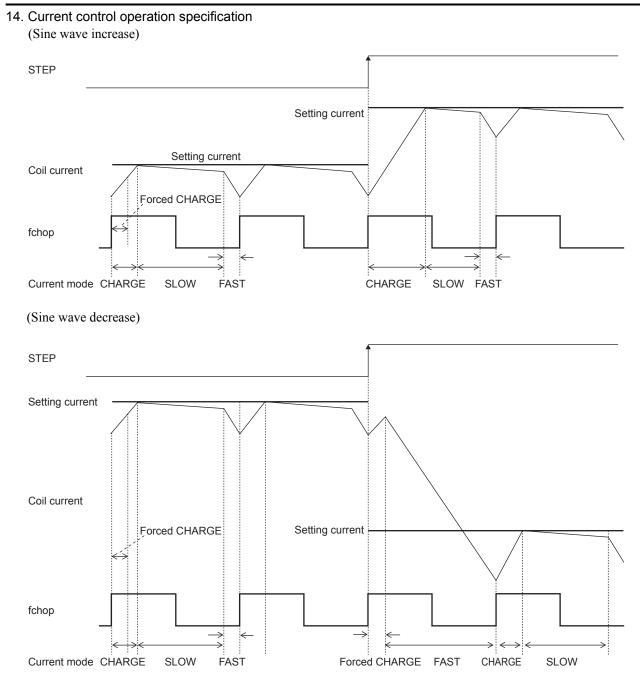


half step (CW mode)



quarter step (CW mode)





Each current mode is operated according to the following sequence.

- At rise of chopping frequency, the CHARGE mode begins. (In the time defined as the "blanking time," the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF).)
- The coil current (ICOIL) and set current (IREF) are compared in this blanking time.
  - When (ICOIL<IREF) state exists ;

The CHARGE mode up to ICOIL  $\geq$  IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for approximately 1µs.

When (ICOIL<IREF) state does not exist ;

The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.

Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

#### 15. High-efficient drive function

This IC includes high-efficient drive function. When high-efficient drive function is turned on, I<sub>OUT</sub> is adjusted automatically within the current value set with VREF pin. When high-efficient drive function is turned off, the current value of I<sub>OUT</sub> becomes the maximum value set by REF pin.

1) High-efficient drive enable function

High-efficient drive function is switched on and off with GAD pin.

However, in the case of full step excitation mode (MD1 = MD2 = "L"), even when GAD = "H", high-efficient drive function is turned off.

Even if you adjust the GMG1, GMG2 of 15-2) and GST1, GST2 of 15-3), in the case of abrupt motor acceleration or load variation to the extent that auto adjuster cannot follow up and eventually leads to the rotation stepping-out, it is recommended that you turn off the high-efficient drive function temporally. As high-efficient control may become unstable due to the control signal from the motor is unstable during low speed rotation, it is also recommended to turn off this function as well.

GAD	Operation mode
L or OPEN	Normal mode
Н	High-efficient mode
	(except for full step excitation mode)

#### Recommended speed of high-efficient drive

excitation	Operating conditions	Speed
half step	HB motor/no-load	over 1500pps
half step full-torque	PM motor/no-load	over 1000pps
quarter step	HB motor/no-load	over 3000pps
	PM motor/no-load	over 2500pps

When there is a load, the high-efficient drive is enabled at slower speed.

#### 2) High-efficient drive margin adjuster function

By setting GMG1 and GMG2 pin, margin for step-out is adjusted.

Where GMG1 = GMG2 = "L",  $I_{OUT}$  and consumption current are at the lowest. In some case, as the  $I_{OUT}$  becomes lower, the number of boost-up process\* may increase triggered by slight change of load. With insufficient driving capability, you need to increase the margin setting. One way to set GMG1 and GMG2 is to minimize boost-up level, then lower the margin from high to low to optimize the margin where motor rotates stably. In the application where load variation is excessive, you need to have a larger margin.

In the applicatio	n where load va	ariation is excessive, you need to have	a larger margin.	
CMC1	CMC2	Sotting	Current consumption	1.

GMG1	GMG2	Setting	Current consumption	Load following capability
L or OPEN	L or OPEN	Margin: small	Smallest	Ordinary
н	L or OPEN	Margin: middle	Smaller	Good
L or OPEN	Н	Margin: large	Small	Better
н	Н	Setting is inhibited	-	-

\*: This is a function to increase I<sub>OUT</sub> rapidly as soon as a possible stepping out is detected due to load variation during high efficiency drive.

#### 3) Boost-up adjuster function

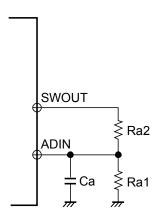
During high-efficient drive, boost-up adjuster function detects a possibility of step-out caused by such factors as abrupt load variation and then boosts up I<sub>OUT</sub> at once (Boost-up process). You can set a level of boost-up by setting GST1 and GST2 pins. One way to set GST1 and GST2 is to increase boost-up level from minimum to maximum within the maximum load condition and select the optimum boost-up setting where motor rotates without stepping out. Also, boost-up level varies depends on reference current defined by VREF. Therefore, you can increase load following capability by increasing VREF voltage.

The higher the boost-up level is, the more the IC becomes tolerant for abrupt load variation. However, rotation stability may become poor (vibration and rotation fluctuation may occur) because excessively high boost-up level leads to rapid increase of I<sub>OUT</sub> at load variation. You may be able to improve poor rotation stability with high boost-up level by increasing high-efficient drive margin.

GST1	GST2	Setting	Increase of lout	load following capability	Rotation stability
L or OPEN	L or OPEN	Boost-up level minimum	{(VREF/5)/RF resistance} × 1/128	Ordinary	Best
Н	L or OPEN	Boost-up level low	{(VREF/5)/RF resistance} × 4/128	Good	Better
L or OPEN	Н	Boost-up level high	{(VREF/5)/RF resistance} × 16/128	Better	Good
н	Н	Boost-up level maximum	{(VREF/5)/RF resistance} × 64/128	Best	Ordinary

#### 4) External component

The resistance value of Ra1, Ra2 (control signal resistors) is adjusted in such a way as to set the maximum SWOUT output voltage during motor rotation to 12V in ADIN pin. Preferably, resistance values of Ra1 and Ra2 are as high as possible to the extent that does not influence waveform. (Recommendation for Ra1:  $15k\Omega$ , Ra2:  $100k\Omega$ ). In some motor where boost-up process occurs at a high speed rotation of 7000pps to 8000pps or higher (HB motor: Half step excitation), you can suppress boost-up by lowering Ra1. Moreover, you can achieve high efficiency at lower speed of 1500pps or lower by increasing resistance for Ra1 (HB motor: Half step excitation). Although it depends on a usage motor, step-out is detectable at higher speed rotation by attaching smaller resistor for Ra1.



#### 5) Drive status warning function

DST1 and DST2 are open-drain output. The driving status can be monitored through a status of DST1 and DST2 pins. When step-out status is detected, DST1 is on for a period of 1 step. Likewise, when small step-out margin status is detected, DST2 turns on for the period of 1 step. In the case of output short status or overheat status, DST1 and DST2 stay on until ST = "L".

Step-out status and small step-out margin status are detectable during high-efficient drive only. In some cases, step-out status may not be detected properly. Hence, make sure to verify the operation with the usage application. If step-out or small step-out margin status occur frequently, make sure to set a large high-efficient drive margin or higher boost-up level.

DST1	DST2	Status	
OFF	OFF	Normal status	
ON	OFF	Step-out status *1(this function is enabled only in high-efficient drive)	
OFF	ON	Small step-out margin status *2(this function is enabled only in high-efficient drive)	
ON	ON	Output short status or overheat status	

\*1: Although it depends on a usage motor, step-out is detectable at higher speed rotation by attaching smaller resistor for Ra1.

\*2: If DST2 alone is turned on, boost-up processing is performed.

#### 16. Output short protection circuit

Output short protection circuit is included in this IC which sets an output to standby mode and turns on warning output. This protection circuit prevents IC destruction when the output is short due to power short or ground short.

1) Operation overview

When output short is detected, short detection circuit operates. If the short status continues for the period of internal timer ( $\approx 2\mu s$ ), the output of 1ch/ 2ch is turned off. If the short status exceeds the timer latch time ( $\approx 32\mu s$ ) set in the internal timer, the output is turned on again and detects short status again. If short is detected again, all the output of 1ch/ 2ch are switched to standby mode and the status is kept. To cancel the standby status, set ST = "L".

#### 2) Error status warning output pin (DST2, DST1)

When the IC detects error status and protection circuit operates, DST2 pin and DST1 pin outputs the error status to CPU side.

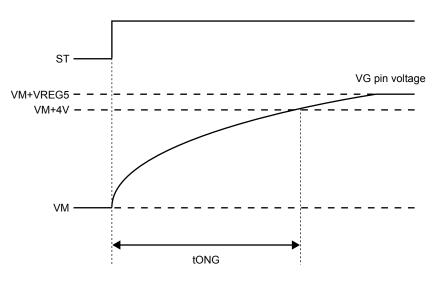
This pin is open-drain output. When error status is detected, DST2 and DST1 output turn on (DST2 = DST1 = "L").

DS12/DS11 plus are turned on in the following statuses.			
Error status	DST2	DST1	
Short is detected in 1ch side.	ON	ON	
Short is detected in 2ch side.	ON	ON	
When overheat is detected.	ON	ON	

#### DST2/DST1 pins are turned on in the following statuses:

#### 17. Charge pump circuit

When ST pin is set to "H", charge pump circuit operates and VG pin voltage increases from VM voltage to VM + VREG5 voltage. If the VG pin voltage is not boosted to VM+4V or more, the output pin cannot be turned on. Therefore it is recommended that the drive of motor is started after the time has passed tONG or more.

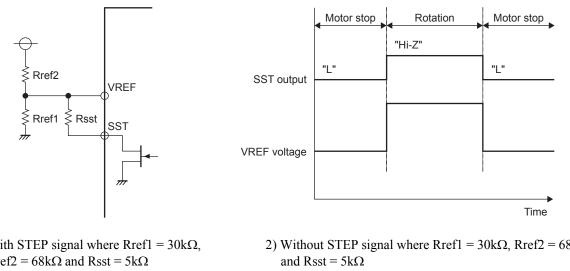


Fg. VG pin voltage

#### 18. Current save function when motor is stopped

SST pin is the open-drain output. When STEP signal is not input for about 16mS, (min: 13mS, max: 23mS), SST pin detects that the rotation of the motor is stopped and SST pin is turned on. At this time, high-efficient drive function is turned off automatically and full current value is set for I<sub>OUT</sub> by VREF pin. And then after signal is input to STEP pin, SST pin is turned off and high-efficient control function is enabled.

In this driver, the circuit constituent is as follows. By decreasing VREF voltage when the motor is stopped, I<sub>OUT</sub> current can be saved. However, this function is unusable when you rotate motor at which input cycle of STEP pulse signal is 16mS or longer.



1) With STEP signal where Rref1 =  $30k\Omega$ , Rref2 =  $68k\Omega$  and Rsst =  $5k\Omega$ VREF1 =  $5V \times 30k\Omega/(68k+30k\Omega) \approx 1.53V$ Where VREF1 = 1.53V, IOUT = VREF/5/0.22 $\Omega \approx 1.39A$  2) Without STEP signal where Rref1 =  $30k\Omega$ , Rref2 =  $68k\Omega$ , and Rsst =  $5k\Omega$ VREF2 =  $5V \times 4.3k\Omega/(68k+4.3k\Omega) \approx 0.3V$ Where VREF2 = 0.3VIOUT = VREF/5/ $0.22\Omega \approx 0.27A$ 

#### 19. Thermal shutdown function

The thermal shutdown circuit is included, and the output is turned off when junction temperature Tj exceeds 180°C and the abnormal state warning output is turned on at the same time.

When the temperature falls hysteresis level, output is driven again (automatic restoration).

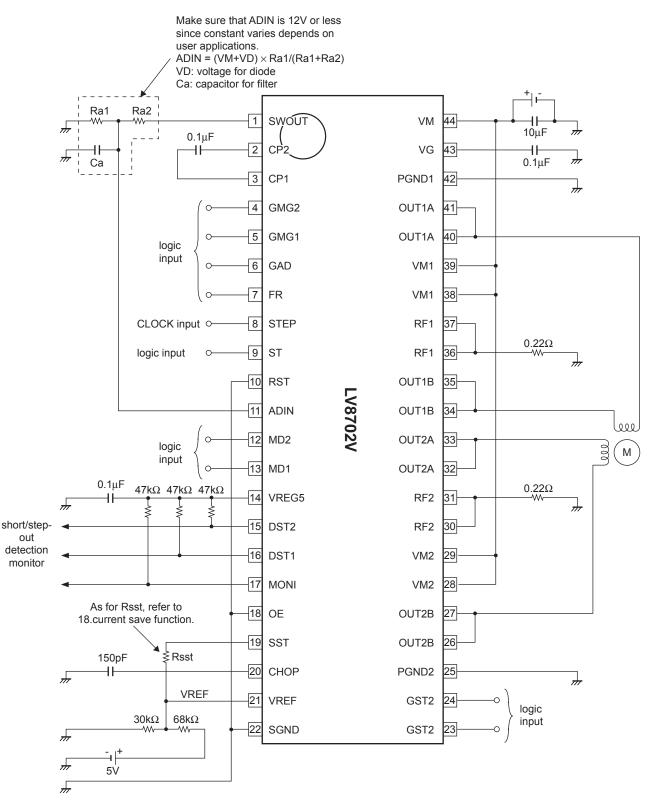
The thermal shutdown circuit doesn't guarantee protection of the set and the destruction prevention of IC,

because it works at the temperature that is higher than rating (Tjmax=150°C) of the junction temperature.

 $TSD = 180^{\circ}C (typ)$ 

 $\Delta TSD = 40^{\circ}C (typ)$ 

#### Example of application circuit



Calculation for each constant setting according to the above circuit diagram is as follows.

- 1) Constant current (100%) setting
  - $$\label{eq:VREF} \begin{split} VREF &= 5V \times 30 k \Omega / (68 k \Omega + 30 k \Omega) \approx 1.53 V \\ When \ VREF &= 1.53 V : \\ IOUT &= VREF / 5 / 0.22 \Omega \approx 1.39 A \end{split}$$

2) Chopping frequency setting Fchop = Ichop/(Cchop×Vtchop×2) =10 $\mu$ A/(150pF×0.5V×2)  $\approx$  66.7kHz

#### **ORDERING INFORMATION**

Device	Package	Shipping (Qty / Packing)
LV8702V-TLM-H	SSOP44J (275mil) (Pb-Free / Halogen-Free)	2000 / Tape & Reel
LV8702V-MPB-H	SSOP44J (275mil) (Pb-Free / Halogen-Free)	30 / Fan-Fold

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- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



#### Как с нами связаться

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