

## Description

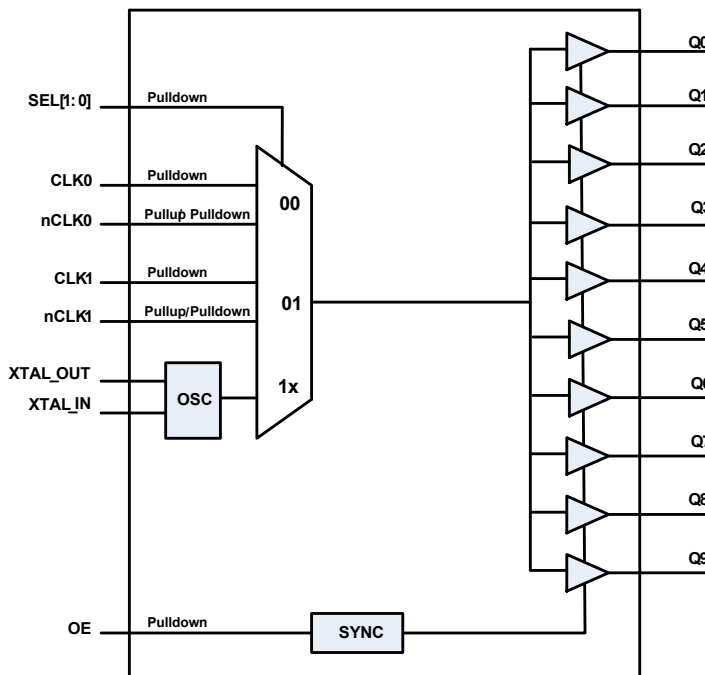
The 8L30110 is a low skew, 1-to-10 LVCMOS / LVTTL Fanout Buffer. The low impedance LVCMOS/LVTTL outputs are designed to drive 50Ω series or parallel terminated transmission lines.

The 8L30110 is characterized at full 3.3V and 2.5V, mixed 3.3V/2.5V, 3.3V/1.8V, 3.3V/1.5V, 2.5V/1.8V and 2.5V/1.5V output operating supply modes. The input clock is selected from two differential clock inputs or a crystal input. The differential input can be wired to accept a single-ended input. The internal oscillator circuit is automatically disabled if the crystal input is not selected.

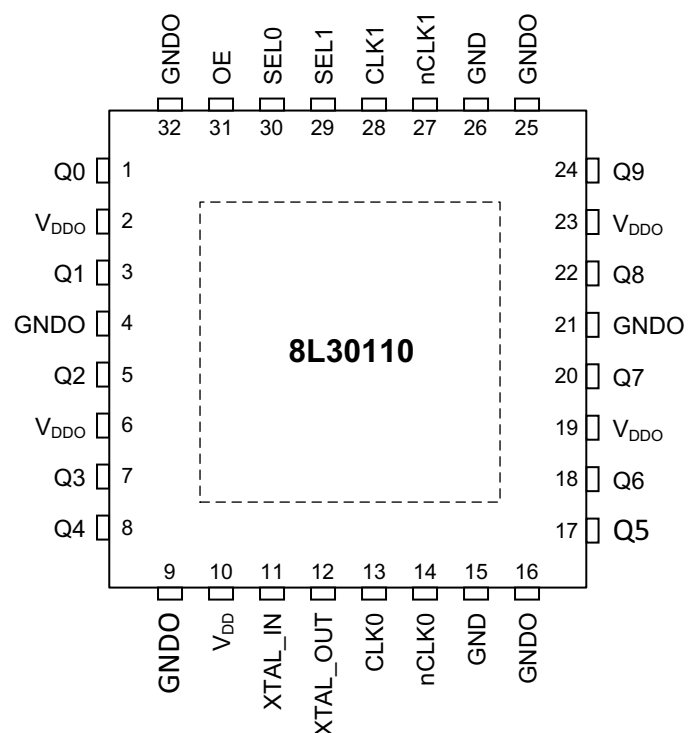
## Features

- Ten LVCMOS / LVTTL outputs up to 200MHz
- Differential input pair can accept the following differential input levels: LVPECL, LVDS, HCSL
- Crystal Oscillator Interface
- Crystal input frequency range: 8MHz to 50MHz
- Output skew: 63ps (typical)
- Additive RMS phase jitter: 22fs (typical)
- Power supply modes:  
Core / Output  
3.3V / 3.3V  
3.3V / 2.5V  
3.3V / 1.8V  
3.3V / 1.5V  
2.5V / 2.5V  
2.5V / 1.8V  
2.5V / 1.5V
- -40°C to 85°C ambient operating temperature
- Lead-free (RoHS 6) packaging
- Supports  $\leq 105^{\circ}\text{C}$  board temperature operations

## Block Diagram



## Pin Assignments



32-pin, 5mm x 5mm VFQFN Package

## Pin Characteristics

**Table 1. Pin Descriptions<sup>[a]</sup>**

Number	Name	Type		Description
1	Q0	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
2	V <sub>DDO</sub>	Power		Output supply.
3	Q1	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
4	GNDO	Power		Power supply output ground.
5	Q2	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
6	V <sub>DDO</sub>	Power		Output supply.
7	Q3	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
8	Q4	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
9	GNDO	Power		Power supply output ground.
10	V <sub>DD</sub>	Power		Power supply.
11	XTAL_IN	Input		Crystal input.
12	XTAL_OUT	Output		Crystal output.
13	CLK0	Input	Pulldown	Non-inverting differential clock.
14	nCLK0	Input	Pullup/ Pulldown	Inverting differential clock. Internal resistor bias to V <sub>DD</sub> /2.
15	GND	Power		Power supply core ground.
16	GNDO	Power		Power supply output ground.
17	Q5	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
18	Q6	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
19	V <sub>DDO</sub>	Power		Output supply.
20	Q7	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
21	GNDO	Power		Power supply output ground.
22	Q8	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
23	V <sub>DDO</sub>	Power		Output supply.
24	Q9	Output		Single-ended clock outputs. LVCMOS/LVTTL interface levels.
25	GNDO	Power		Power supply output ground.
26	GND	Power		Power supply core ground.
27	nCLK1	Input	Pullup/ Pulldown	Inverting differential clock. Internal resistor bias to V <sub>DD</sub> /2.
28	CLK1	Input	Pulldown	Non-inverting differential clock.
29	SEL1	Input	Pulldown	Input clock selection. LVCMOS/LVTTL interface levels. See Table 3A.
30	SEL0	Input	Pulldown	Input clock selection. LVCMOS/LVTTL interface levels. See Table 3A.
31	OE	Input	Pulldown	Output enable. LVCMOS/LVTTL interface levels. See Table 3B.
32	GNDO	Power		Power supply output ground.
ePad	GND_EP	Power		Exposed pad of package. Connect to ground.

[a] Pullup and Pulldown refer to internal input resistors. See Table 2, *Pin Characteristics*, for typical values.

**Table 2. Pin Characteristics<sup>[a]</sup>**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$C_{IN}$	Input Capacitance	SEL[1:0], OE CLK0, nCLK0, CLK1, nCLK1		2		pF
$R_{PULLDOWN}$	Input Pulldown Resistor			51		k $\Omega$
$R_{PULLUP}$	Input Pullup Resistor			51		k $\Omega$
$C_{PD}$	Power Dissipation Capacitance (per output)	$V_{DDO} = 3.465V, f = 100MHz$			9.5	pF
		$V_{DDO} = 2.625V, f = 100MHz$			7.4	pF
		$V_{DDO} = 2V, f = 100MHz$			7.2	pF
		$V_{DDO} = 1.65V, f = 100MHz$			6.9	pF
$R_{OUT}$	Output Impedance	$V_{DDO} = 3.3V$		15		$\Omega$
		$V_{DDO} = 2.5V$		18		$\Omega$
		$V_{DDO} = 1.8V$		25		$\Omega$
		$V_{DDO} = 1.5V$		30		$\Omega$

[a] Measured at ambient temperature (unless otherwise noted.)

## Function Tables

**Table 3A. SELx Function Table**

Control Input	Selected Input Clock
SEL[1:0]	
00 (default)	CLK0, nCLK0
01	CLK1, nCLK1
11 or 10	XTAL

**Table 3B. OE Function Table**

Control Input	Function
OE	Q[0:9]
0 (default)	High-Impedance
1	Enabled

**Table 3C. Input/Output Operation Table<sup>[a]</sup>**

Input State	Output State
CLK0, CLK1 = HIGH nCLK0, nCLK1 = LOW	Logic HIGH
CLK0, CLK1 = LOW nCLK0, nCLK1 = HIGH	Logic LOW
CLK0, nCLK0, CLK1, nCLK1 open	Logic LOW

[a] Device must have switching edge to obtain output states.

## Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of the product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, $V_{DD}/V_{DDO}$	3.6V
Inputs, $V_I$ $CLK_X$ , $nCLK_X$ , $XTAL\_IN$ Other Inputs	3.6V 2V 3.6V
Outputs, $V_O$	3.6V
$T_J$ (Junction Temperature)	125°C
Storage Temperature, $T_{STG}$	-65°C to 150°C

## DC Electrical Characteristics

**Table 4A. Power Supply DC Characteristics,**

$V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$  or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Power Supply Voltage		3.135	3.3	3.465	V
$V_{DDO}$	Output Supply Voltage		3.135	3.3	3.465	V
			2.375	2.5	2.625	V
			1.6	1.8	2	V
			1.35	1.5	1.65	V
$I_{DD}$	Static Supply Current	SEL[1:0] = 00 or 01, Outputs Unloaded		19	22	mA
		SEL[1:0] = 10 or 11, Outputs Unloaded		18	22	mA

**Table 4B. Power Supply DC Characteristics,**

$V_{DD} = 2.5V \pm 5\%$ ,  $V_{DDO} = 2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$  or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Power Supply Voltage		2.375	2.5	2.625	V
$V_{DDO}$	Output Supply Current		2.375	2.5	2.625	V
			1.6	1.8	2	V
			1.35	1.5	1.65	V
$I_{DD}$	Static Supply Current	SEL[1:0] = 00 or 01, Outputs Unloaded		18	22	mA
		SEL[1:0] = 10 or 11, Outputs Unloaded		17	22	mA

**Table 4C. LVCMOS/LVTTL DC Characteristics,**

$V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $V_{DDO} (\leq V_{DD}) = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$  or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage		$V_{DD} = 3.3V \pm 5\%$	2		$V_{DD} + 0.3$	V
			$V_{DD} = 2.5V \pm 5\%$	1.7		$V_{DD} + 0.3$	V
$V_{IL}$	Input Low Voltage		$V_{DD} = 3.3V \pm 5\%$	-0.3		0.8	V
			$V_{DD} = 2.5V \pm 5\%$	-0.3		0.7	V
$I_{IH}$	Input High Current	OE, SEL[1:0]	$V_{DD} = V_{IN} = 3.465V$			150	$\mu\text{A}$
$I_{IL}$	Input Low Current	OE, SEL[1:0]	$V_{DD} = 3.465V$ , $V_{IN} = 0V$	-5			$\mu\text{A}$
$V_{OH}$	Output High Voltage		$V_{DDO} = 3.3V \pm 5\%$ , $I_{OH} = -12\text{mA}$	2.6			V
			$V_{DDO} = 2.5V \pm 5\%$ , $I_{OH} = -8\text{mA}$	1.8			V
			$V_{DDO} = 1.8V \pm 0.2V$ , $I_{OH} = -2\text{mA}$	1.2			V
			$V_{DDO} = 1.5V \pm 0.15V$ , $I_{OH} = -2\text{mA}$	0.95			V
$V_{OL}$	Output Low Voltage		$V_{DDO} = 3.3V \pm 5\%$ , $I_{OL} = 12\text{mA}$			0.5	V
			$V_{DDO} = 2.5V \pm 5\%$ , $I_{OL} = 8\text{mA}$			0.5	V
			$V_{DDO} = 1.8V \pm 0.2V$ , $I_{OL} = 2\text{mA}$			0.4	V
			$V_{DDO} = 1.5V \pm 0.15V$ , $I_{OL} = 2\text{mA}$			0.35	V

**Table 4D. Differential DC Characteristics,**

$V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $V_{DDO} (\leq V_{DD}) = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$  or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
$I_{IH}$	Input High Current	CLK[0:1], nCLK[0:1]	$V_{DD} = V_{IN} = 3.465V$ or $2.625V$			150	$\mu\text{A}$
$I_{IL}$	Input Low Current	CLK[0:1]	$V_{DD} = 3.465V$ or $2.625V$ , $V_{IN} = 0V$	-5			$\mu\text{A}$
		nCLK[0:1]	$V_{DD} = 3.465V$ or $2.625V$ , $V_{IN} = 0V$	-150			$\mu\text{A}$
$V_{PP}$	Peak-to-Peak Input Voltage <sup>[a]</sup>			0.15		1.3	V
$V_{CMR}$	Common Mode Input Voltage [a], [b]			0.5		$V_{DD} - 0.85$	V

[a]  $V_{IL}$  should not be less than -0.3V and  $V_{IH}$  should not be greater than  $V_{DD}$ .

[b] Common mode voltage is defined at the crosspoint.

**Table 5. Crystal Characteristics<sup>[a]</sup>**

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		8		50	MHz
Equivalent Series Resistance (ESR)				50	$\Omega$
Shunt Capacitance				7	pF
Load Capacitance (CL)			12	18	pF

[a] To insure crystal accuracy, the use of external tuning capacitors is required.

## AC Electrical Characteristics

**Table 6. AC Characteristics,**

$V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $V_{DDO} (\leq V_{DD}) = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ <sup>[a][b]</sup> or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
$f_{OUT}$	Output Frequency	Using External Crystal		8		50	MHz
		Using External Clock Source				200	MHz
$\Delta V/\Delta T$	Input Edge Rate	Input Clock from CLK0, nCLK0, CLK1, nCLK1	20% – 80%		2		V/ns
$t_{PD}$	Propagation Delay <sup>[c]</sup>		$V_{DDO} = 3.3V \pm 5\%$	1.5		2.5	ns
			$V_{DDO} = 2.5V \pm 5\%$	1.8		2.7	ns
			$V_{DDO} = 1.8V \pm 0.2V$	1.8		3.8	ns
			$V_{DDO} = 1.5V \pm 0.15V$	2.5		4.6	ns
$t_{sk(o)}$	Output Skew <sup>[d], [e]</sup>		$V_{DDO} = 3.3V \pm 5\%$		40	70	ps
			$V_{DDO} = 2.5V \pm 5\%$		45	85	ps
			$V_{DDO} = 1.8V \pm 0.2V$		55	110	ps
			$V_{DDO} = 1.5V \pm 0.15V$		63	120	ps
$t_{jit}$	Buffer Additive Phase Jitter <sup>[f]</sup> refer to Additive Phase Jitter Section; $f_{OUT} = 156.25\text{MHz}$ , Integration Range: 12kHz - 20MHz	Input Clock from CLK0, nCLK0 or CLK1, nCLK1	$V_{DDO} = 3.3V \pm 5\%$		22		fs
			$V_{DDO} = 2.5V \pm 5\%$		17		fs
			$V_{DDO} = 1.8V \pm 0.2V$		55		fs
			$V_{DDO} = 1.5V \pm 0.15V$		103		fs
NF	Noise Floor <sup>[g]</sup>	Input Clock from CLK0, nCLK0, CLK1, nCLK1 at 125MHz	10kHz Offset		-129		dBc/Hz
			100kHz Offset		-142		dBc/Hz
			1MHz Offset		-158		dBc/Hz
			10MHz Offset		-160		dBc/Hz
			20MHz Offset		-160		dBc/Hz
$t_{jit}(\emptyset)$	RMS Phase Jitter	Input Clock from Crystal 25MHz; 12kHz - 5MHz	$V_{DDO} = 3.3V \pm 5\%$		138		fs
			$V_{DDO} = 2.5V \pm 5\%$		134		fs
			$V_{DDO} = 1.8V \pm 0.2V$		140		fs
			$V_{DDO} = 1.5V \pm 0.15V$		157		fs
NF	Noise Floor	Input Clock from Crystal 25MHz	100Hz Offset		-113		dBc/Hz
			1kHz Offset		-143		dBc/Hz
			10kHz Offset		-157		dBc/Hz
			100kHz Offset		-162		dBc/Hz
			1MHz Offset		-163		dBc/Hz
			5MHz Offset		-163		dBc/Hz
$t_R / t_F$	Output Rise/Fall Time	$V_{DDO} = 3.3V \pm 5\%$	20% to 80%			700	ps
		$V_{DDO} = 2.5V \pm 5\%$	20% to 80%			650	ps
		$V_{DDO} = 1.8V \pm 0.2V$	20% to 80%			605	ps
		$V_{DDO} = 1.5V \pm 0.15V$	20% to 80%			675	ps

**Table 6. AC Characteristics,**

$V_{DD} = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$ ,  $V_{DDO} (\leq V_{DD}) = 3.3V \pm 5\%$  or  $2.5V \pm 5\%$  or  $1.8V \pm 0.2V$  or  $1.5V \pm 0.15V$ ,  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ <sup>[a][b]</sup> or  $T_B = -40^\circ\text{C}$  to  $105^\circ\text{C}$

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
odc	Output Duty Cycle		50% input Duty Cycle at 125MHz	45		55	%
$T_{stab}$	Output Stabilization Time		Measure from $V_{DD\_min}$ to the first rising edge of output See Figure 1	5	18	30	ms
$t_{EN}$	Output Enable Time <sup>[h]</sup>	OE			2 - 3		cycles
$t_{DIS}$	Output Disable Time <sup>[h]</sup>	OE			2 - 3		cycles
MUX <sub>ISOLATION</sub>	MUX Isolation <sup>[h]</sup>		at 125MHz		90		dB

[a] Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

[b] All parameters measured at  $f \leq f_{OUT}$  unless noted otherwise.

[c] Measured from the differential input crossing point to  $V_{DDO}/2$  of the output.

[d] Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at  $V_{DDO}/2$ .

[e] This parameter is defined in accordance with JEDEC Standard 65.

[f] Input source is IDT 8T49NS010.

[g]  $f_{IN} = f_{OUT} = 125\text{MHz}$ ,  $V_{DD} = V_{DDO} = 3.3V$ .

[h] These parameters are guaranteed by characterization. Not tested in production.

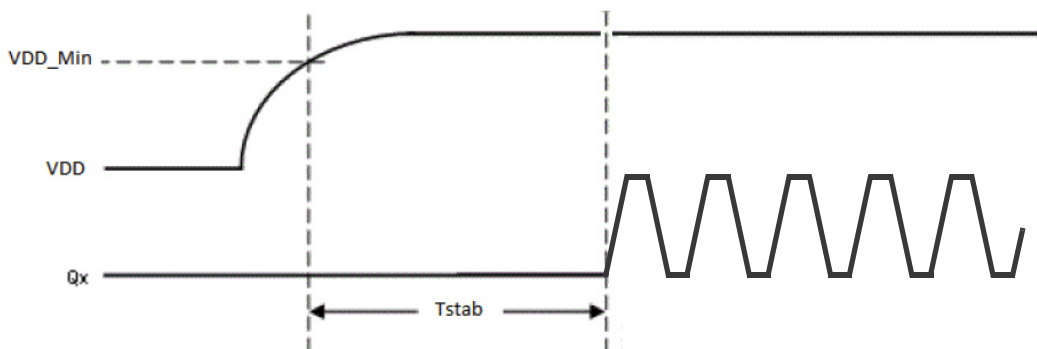
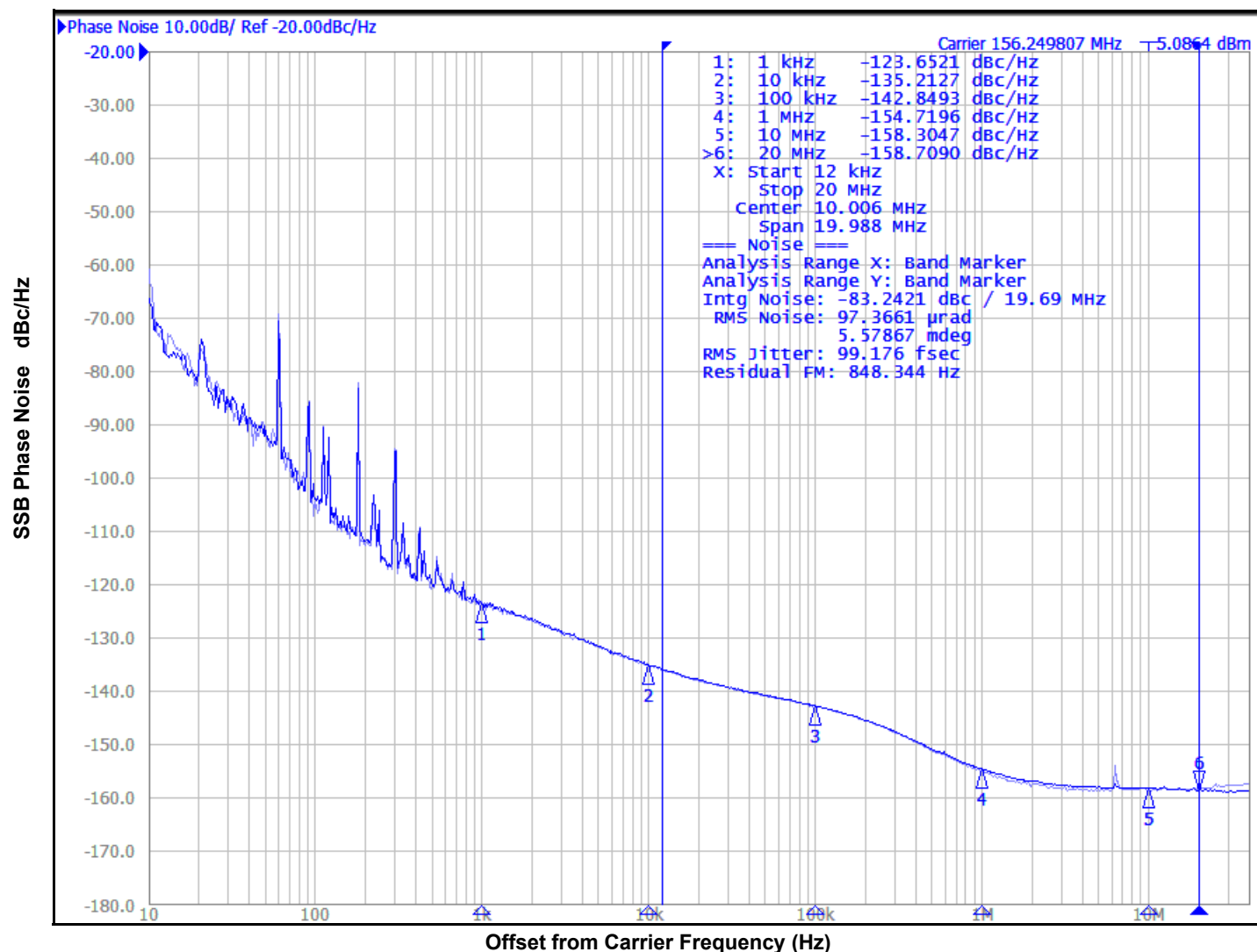


Figure 1. Stabilization Time ( $T_{stab}$ )

## Additive Phase Jitter

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise**. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1 Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1 Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.

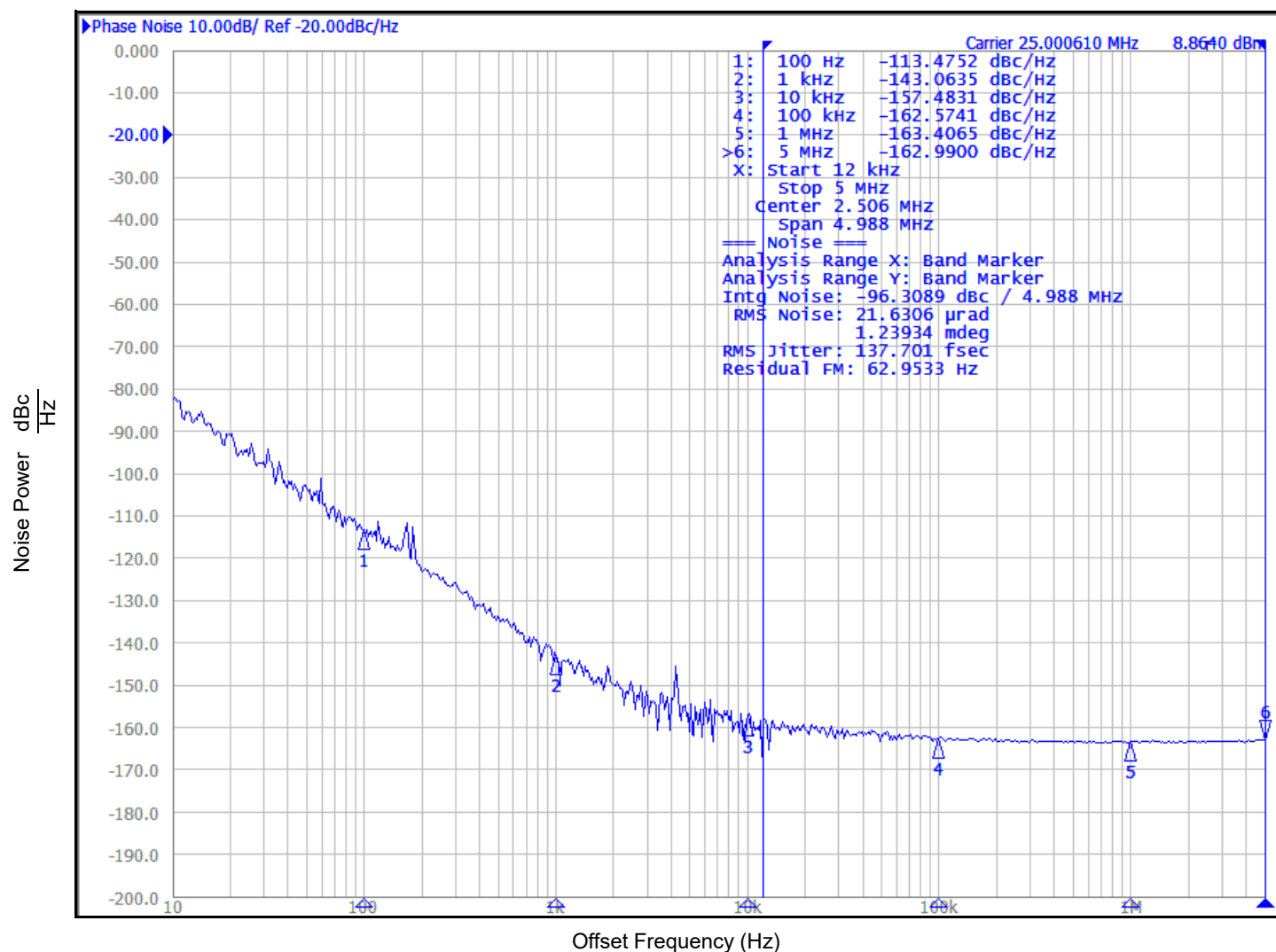


As with most timing specifications, phase noise measurements have issues relating to the limitations of the measurement equipment. The noise floor of the equipment can be higher or lower than the noise floor of the device. Additive phase noise is dependent on both the noise floor of the input source and measurement equipment.

The additive phase jitter for this device was measured using an IDT Clock Driver 8T49NS010 as an input source and Agilent E5052 phase noise analyzer.



## Typical Phase Noise at 25MHz (3.3V) with 25MHz Crystal Input



## Applications Information

### Recommendations for Unused Input and Output Pins

#### Inputs

##### ***CLK/nCLK Inputs***

For applications not requiring the use of the differential input, both CLK and nCLK can be left floating. Though not required, but for additional protection, a 1k $\Omega$  resistor can be tied from CLK to ground.

##### ***Crystal Inputs***

For applications not requiring the use of the crystal oscillator input, both XTAL\_IN and XTAL\_OUT can be left floating. Though not required, but for additional protection, a 1k $\Omega$  resistor can be tied from XTAL\_IN to ground.

#### LVC MOS Control Pins

All control pins have internal pulldowns; additional resistance is not required but can be added for additional protection. A 1k $\Omega$  resistor can be used.

#### Outputs

##### LVC MOS Outputs

All unused LVC MOS outputs can be left floating. IDT recommends that there be no trace attached.

## Wiring the Differential Input to Accept Single-Ended Levels

Figure 2 shows how a differential input can be wired to accept single ended levels. The reference voltage  $V_1 = V_{DD}/2$  is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the  $V_1$  in the center of the input voltage swing. For example, if the input clock is driven from a single-ended 2.5V LVCMOS driver and the DC offset (or swing center) of this signal is 1.25V, the R1 and R2 values should be adjusted to set the  $V_1$  at 1.25V. The values below are for when both the single ended swing and  $V_{DD}$  are at the same voltage. This configuration requires that the sum of the output impedance of the driver ( $R_o$ ) and the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission line impedance. For most 50Ω applications, R3 and R4 can be 100Ω. The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced while maintaining an edge rate faster than 1V/ns. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however  $V_{IL}$  cannot be less than -0.3V and  $V_{IH}$  cannot be more than  $V_{DD} + 0.3V$ . Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

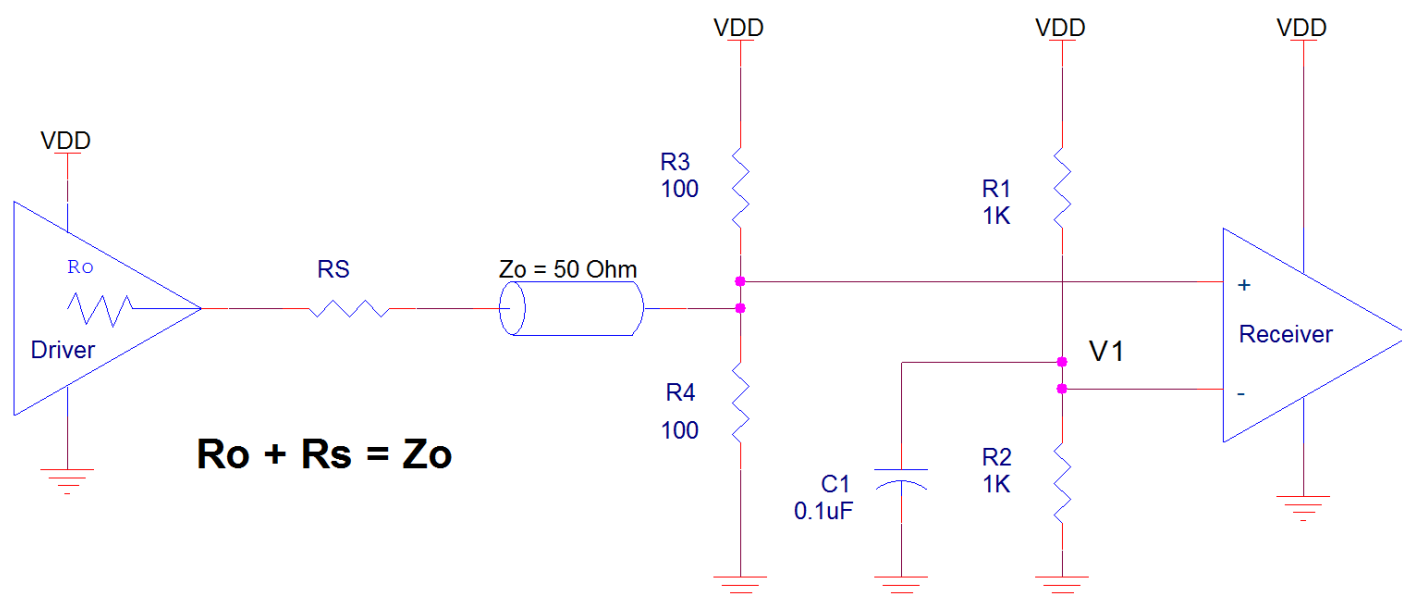


Figure 2. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels

## Crystal Input Interface

The 8L30110 has been characterized with 12pF parallel resonant crystals. The capacitor values, C1 and C2, shown in Figure 3 below were determined using a 12pF parallel resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.

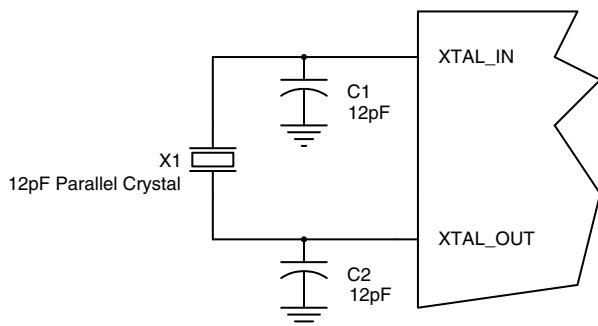
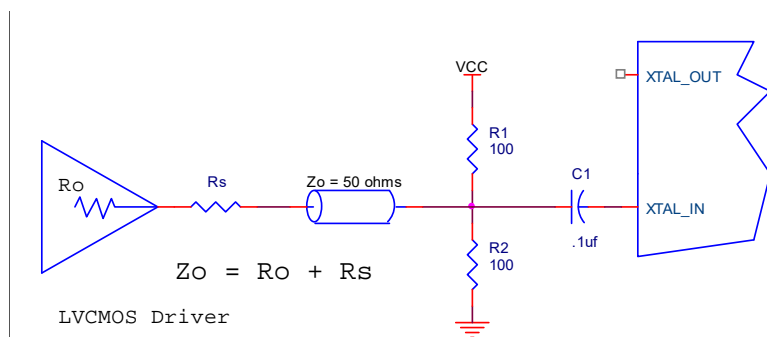


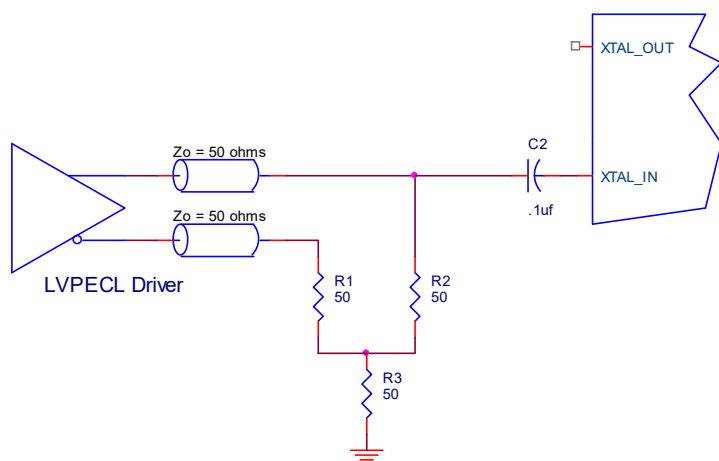
Figure 3. Crystal Input Interface

## Overdriving the XTAL Interface

The XTAL\_IN input can be overdriven by an LVCMOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL\_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/ns. For 3.3V LVCMOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. *Figure 4A* shows an example of the interface diagram for a high speed 3.3V LVCMOS driver. This configuration requires that the sum of the output impedance of the driver ( $R_o$ ) and the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First,  $R1$  and  $R2$  in parallel should equal the transmission line impedance. For most 50Ω applications,  $R1$  and  $R2$  can be 100Ω. This can also be accomplished by removing  $R1$  and changing  $R2$  to 50Ω. The values of the resistors can be increased to reduce the loading for a slower and weaker LVCMOS driver. *Figure 4B* shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL\_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a quartz crystal as the input.



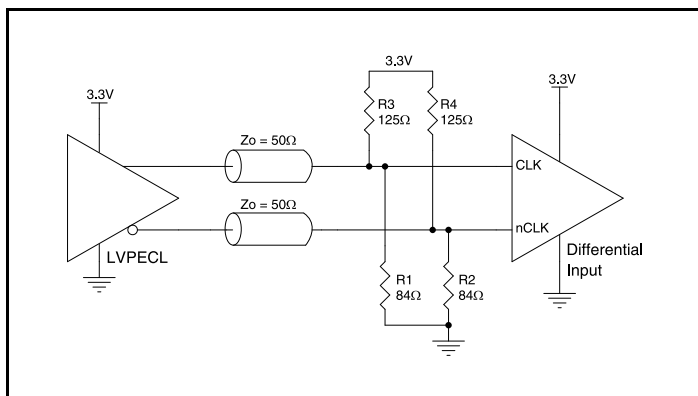
**Figure 4A. General Diagram for LVCMOS Driver to XTAL Input Interface**



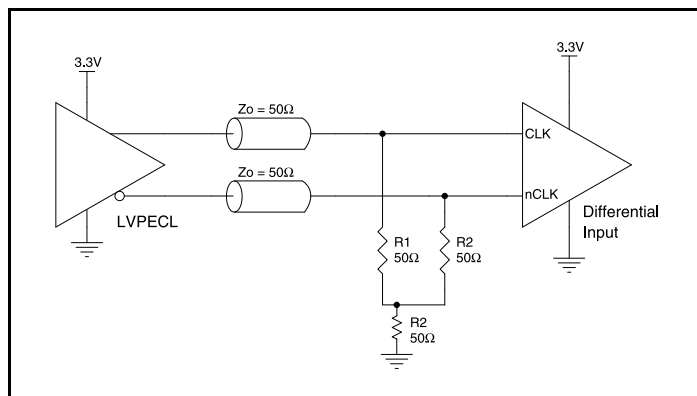
**Figure 4B. General Diagram for LVPECL Driver to XTAL Input Interface**

## 3.3V Differential Clock Input Interface

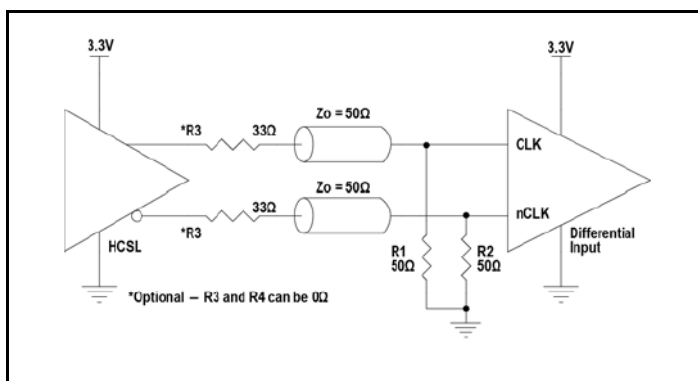
The CLK /nCLK accepts LVDS, LVPECL, HCSL and other differential signals. Both signals must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figures 5A to 5D show interface examples for the CLK /nCLK input with built-in 50 $\Omega$  terminations driven by the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.



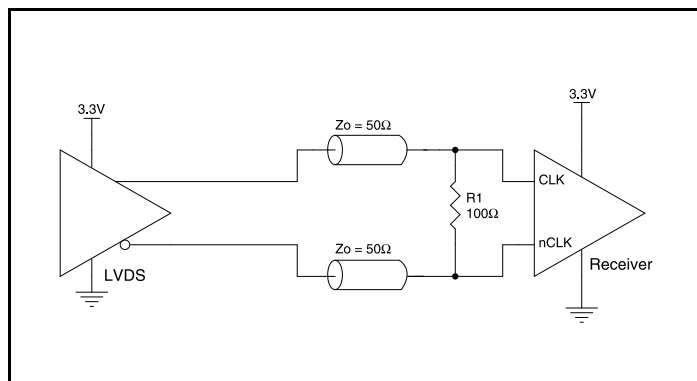
**Figure 5A. CLK/nCLK Input Driven by a 3.3V LVPECL Driver**



**Figure 5B. CLK/nCLK Input Driven by a 3.3V LVPECL Driver**



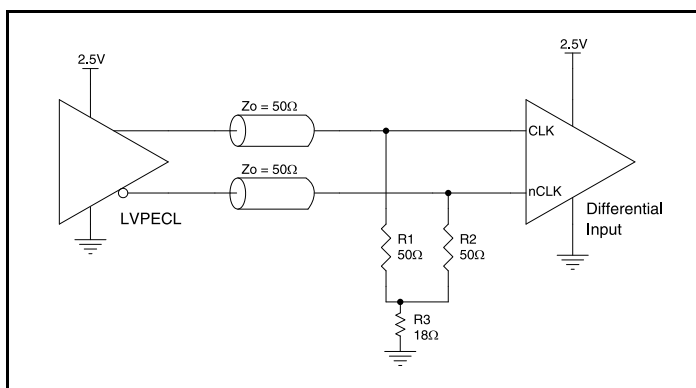
**Figure 5C. CLK/nCLK Input Driven by a 3.3V HCSL Driver**



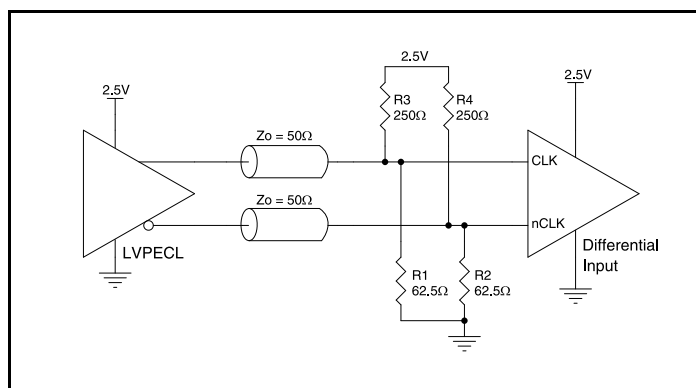
**Figure 5D. CLK/nCLK Input Driven by a 3.3V LVDS Driver**

## 2.5V Differential Clock Input Interface

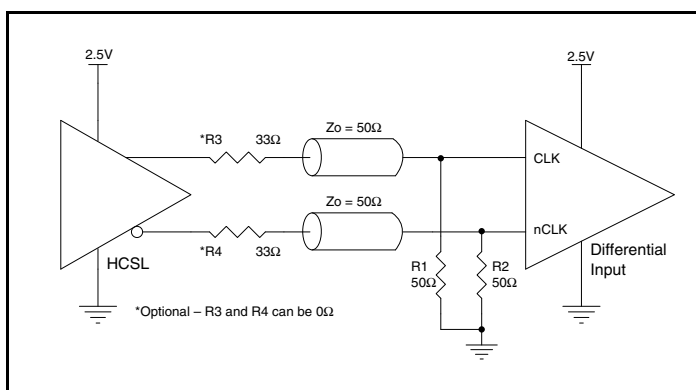
CLKx/nCLKx accepts LVDS, LVPECL, HCSL and other differential signals show interface examples for the CLK /nCLK input with built-in 50 $\Omega$  terminations driven by the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.



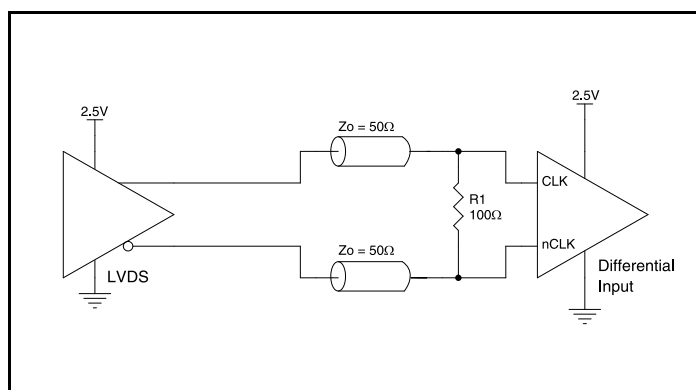
**Figure 6A. CLKx/nCLKx Input Driven by a 2.5V LVPECL Driver**



**Figure 6B. CLKx/nCLKx Input Driven by a 2.5V LVPECL Driver**



**Figure 6C. CLKx/nCLKx Input Driven by a 2.5V HCSL Driver**



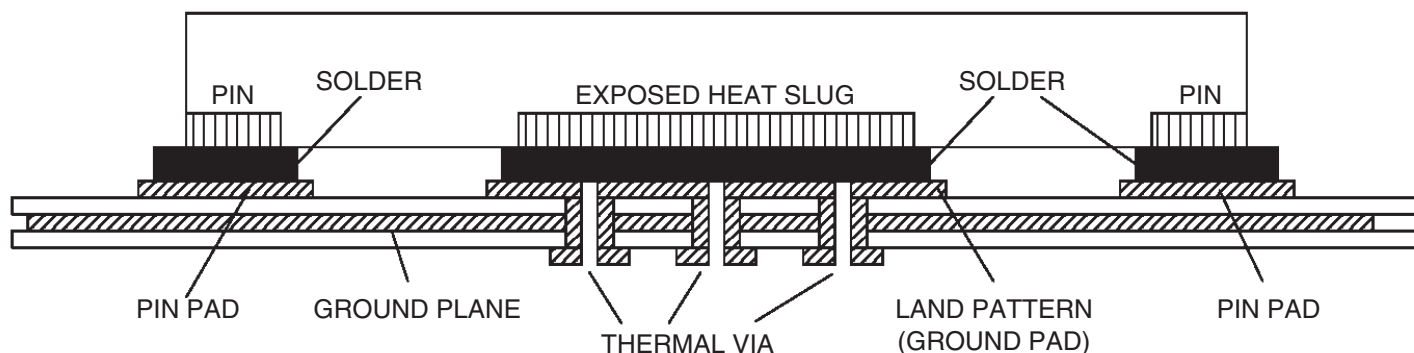
**Figure 6D. CLKx/nCLKx Input Driven by a 2.5V LVDS Driver**

## VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 7*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The

vias act as “heat pipes”. The number of vias (i.e. “heat pipes”) are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor’s Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.



**Figure 7. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)**

## Power Considerations

This section provides information on power dissipation and junction temperature for the 8L30110. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the 8L30110 is the sum of the core power plus the power dissipation in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

#### Total Static Power:

$$\text{Power (core)}_{MAX} = V_{DD\_MAX} * I_{DD} = 3.465V * 22mA = \mathbf{76.23mW}$$

#### Dynamic Power Dissipation at $F_{OUT}$ (200MHz)

$$\text{Total Power (} F_{OUT\_MAX} \text{)} = [(C_{PD} * N) * \text{Frequency} * (V_{DDO})^2] = [(9.5pF * 10) * 200MHz * (3.465V)^2] = \mathbf{228mW}$$

**N = number of outputs**

#### Total Power

$$\begin{aligned} &= \text{Static Power} + \text{Dynamic Power Dissipation} \\ &= 76.23mW + 228mW \\ &= \mathbf{304.23mW} \end{aligned}$$

## 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature,  $T_j$ , to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd_{total} + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd_{total}$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 35.23°C/W per Table 7 below. Therefore,  $T_j$  for an ambient temperature of 85°C with all outputs switching is:

$85^\circ\text{C} + 0.304 * 35.23^\circ\text{C/W} = 95.7^\circ\text{C}$ . This is below the limit of 125°C.

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

**Table 7. Thermal Resistance for 32-VFQFN, Forced Convection**

Thermal Parameters by Velocity <sup>[a]</sup>			
Meters per Second	0 m/s	1 m/s	2 m/s
$\theta_{JA}$ (Junction to Ambient)	35.23°C/W	31.6°C/W	30.0°C/W
$\theta_{JB}$ (Junction to Board)	1.5°C/W		
$\theta_{JC}$ (Junction to Case)	28.4°C/W		

[a] Multi-Layer PCB, JEDEC Standard Test Boards

## Transistor Count

The transistor count for 8L30110 is: 1628

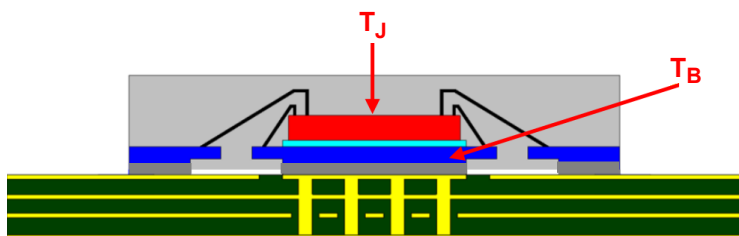


## Case Temperature Considerations

This device supports applications in a natural convection environment which does not have any thermal conductivity through ambient air. The printed circuit board (PCB) is typically in a sealed enclosure without any natural or forced air flow and is kept at or below a specific temperature. The device package design incorporates an exposed pad (ePad) with enhanced thermal parameters which is soldered to the PCB where most of the heat escapes from the bottom exposed pad. For this type of application, it is recommended to use the junction-to-board thermal characterization parameter  $\Psi_{JB}$  (Psi-JB) to calculate the junction temperature ( $T_J$ ) and ensure it does not exceed the maximum allowed junction temperature in the Absolute Maximum Rating table.

The junction-to-board thermal characterization parameter,  $\Psi_{JB}$ , is calculated using the following equation:

- $T_J = T_B + \Psi_{JB} \times P_d$ , where
- $T_J$  = Junction temperature at steady state condition in ( $^{\circ}\text{C}$ ).
- $T_B$  = Board or case temperature (Bottom) at steady state condition in ( $^{\circ}\text{C}$ ).
- $\Psi_{JB}$  = Thermal characterization parameter to report the difference between junction temperature and the temperature of the board measured at the top surface of the board.
- $P_d$  = power dissipation (W) in desired operating configuration.



The ePad provides a low thermal resistance path for heat transfer to the PCB and represents the key pathway to transfer heat away from the IC to the PCB. It's critical that the connection of the exposed pad to the PCB is properly constructed to maintain the desired IC bottom case temperature ( $T_{CB}$ ). A good connection ensures that temperature at the exposed pad ( $T_{CB}$ ) and the board temperature ( $T_B$ ) are relatively the same. An improper connection can lead to increased junction temperature, increased power consumption and decreased electrical performance. In addition, there could be long-term reliability issues and increased failure rate.

Example Calculation for Junction Temperature ( $T_J$ ):  $T_J = T_B + \Psi_{JB} \times P_d$

Package type:	32-VFQFN
Body size:	5 × 5mm
ePad size:	3.3 × 3.3mm
Thermal via:	3 × 3 matrix
$\Psi_{JB}$	1.5C/W
$T_B$	105 $^{\circ}\text{C}$
$P_d$	0.304W

For the variables above, the junction temperature is equal to 105.5 $^{\circ}\text{C}$ . Since this is below the maximum junction temperature of 125 $^{\circ}\text{C}$ , there are no long term reliability concerns. In addition, since the junction temperature at which the device was characterized using forced convection is 95.7 $^{\circ}\text{C}$ , this device can function without the degradation of the specified AC or DC parameters.

## Package Outline Drawings

The package outline drawings are located at the end of this document. The package information is the most current data available.

## Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8L30110NLGI	IDT8L30110NLGI	Lead-Free, 32-Lead VFQFN	Tray	-40°C to 85°C
8L30110NLGI8	IDT8L30110NLGI	Lead-Free, 32-Lead VFQFN	Tape & Reel	-40°C to 85°C

## Revision History

Revision Date	Description of Change
April 11, 2018	Initial release.



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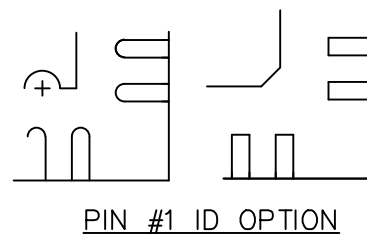
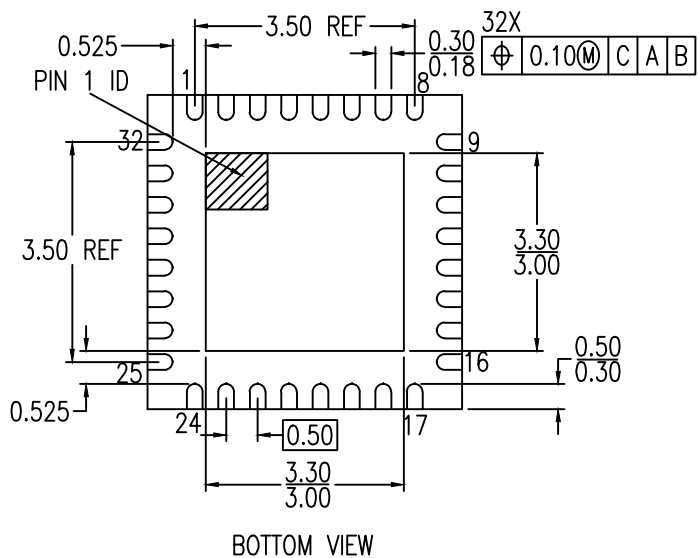
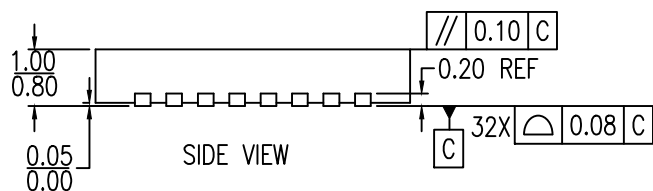
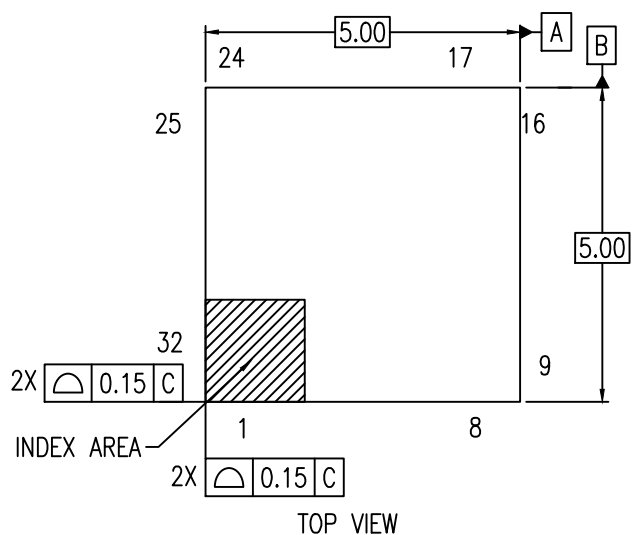
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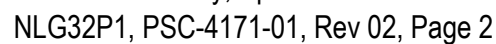
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## NOTE:

1. ALL DIMENSION ARE IN MM. ANGLES IN DEGREES.
2. COPLANARITY APPLIE TO THE EXPOSED PAD AS WELL AS THE TERMINALS.  
COPLANARITY SHALL NOT EXCEED 0.08 MM.
3. WARPAGE SHALL NOT EXCEED 0.10 MM.
4. PIN LOCATION IS UNIDENTIFIED BY EITHER CHAMFER OR NOTCH.





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