DATA SHEET

For a complete data sheet, please also download:

- The IC06 74HC/HCT/HCU/HCMOS Logic Family Specifications
- The IC06 74HC/HCT/HCU/HCMOS Logic Package Information
- The IC06 74HC/HCT/HCU/HCMOS Logic Package Outlines

74HC/HCT191
Presettable synchronous 4-bit binary up/down counter

Product specification
File under Integrated Circuits, IC06

December 1990

Philips Semiconductors

PHILIPS

Presettable synchronous 4-bit binary up/down counter

74HC/HCT191

FEATURES

- Synchronous reversible counting
- Asynchronous parallel load
- Count enable control for synchronous expansion
- Single up/down control input
- Output capability: standard
- I_{CC} category: MSI

GENERAL DESCRIPTION

The 74HC/HCT191 are high-speed Si-gate CMOS devices and are pin compatible with low power Schottky TTL (LSTTL). They are specified in compliance with JEDEC standard no. 7A.

The 74HC/HCT191 are asynchronously presettable 4-bit binary up/down counters. They contain four master/slave flip-flops with internal gating and steering logic to provide asynchronous preset and synchronous count-up and count-down operation.

Asynchronous parallel load capability permits the counter to be preset to any desired number. Information present on the parallel data inputs (D_0 to D_3) is loaded into the counter and appears on the outputs when the parallel load ($\overline{PL}$) input is LOW. As indicated in the function table, this operation overrides the counting function.

Counting is inhibited by a HIGH level on the count enable ($\overline{CE}$) input. When $\overline{CE}$ is LOW internal state changes are initiated synchronously by the LOW-to-HIGH transition of the clock input. The up/down ($\overline{U}$/D) input signal determines the direction of counting as indicated in the function table. The $\overline{CE}$ input may go LOW when the clock is in either state, however, the LOW-to-HIGH $\overline{CE}$ transition must occur only when the clock is HIGH. Also, the $\overline{U}$/D input should be changed only when either $\overline{CE}$ or CP is HIGH.

Overflow/underflow indications are provided by two types of outputs, the terminal count (TC) and ripple clock ($\overline{RC}$). The TC output is normally LOW and goes HIGH when a circuit reaches zero in the count-down mode or reaches “15” in the count-up-mode. The TC output will remain HIGH until a state change occurs, either by counting or presetting, or until $\overline{U}$/D is changed. Do not use the TC output as a clock signal because it is subject to decoding spikes. The TC signal is used internally to enable the $\overline{RC}$ output. When TC is HIGH and $\overline{CE}$ is LOW, the $\overline{RC}$ output follows the clock pulse (CP). This feature simplifies the design of multistage counters as shown in Figs 5 and 6.

In Fig.5, each $\overline{RC}$ output is used as the clock input to the next higher stage. It is only necessary to inhibit the first stage to prevent counting in all stages, since a HIGH on $\overline{CE}$ inhibits the $\overline{RC}$ output pulse as indicated in the function table. The timing skew between state changes in the first and last stages is represented by the cumulative delay of the clock as it ripples through the preceding stages. This can be a disadvantage of this configuration in some applications.

Fig.6 shows a method of causing state changes to occur simultaneously in all stages. The $\overline{RC}$ outputs propagate the carry/borrow signals in ripple fashion and all clock inputs are driven in parallel. In this configuration the duration of the clock LOW state must be long enough to allow the negative-going edge of the carry/borrow signal to ripple through to the last stage before the clock goes HIGH. Since the $\overline{RC}$ output of any package goes HIGH shortly after its CP input goes HIGH there is no such restriction on the HIGH-state duration of the clock.

In Fig.7, the configuration shown avoids ripple delays and their associated restrictions. Combining the TC signals from all the preceding stages forms the $\overline{CE}$ input for a given stage. An enable must be included in each carry gate in order to inhibit counting. The TC output of a given stage it not affected by its own $\overline{CE}$ signal therefore the simple inhibit scheme of Figs 5 and 6 does not apply.

QUICK REFERENCE DATA

GND = 0 V; T_{amb} = 25 °C; $t_r = t_f$ = 6 ns

SYMBOL	PARAMETER	CONDITIONS	TYPICAL		UNIT
			HC	HCT	
t_{PHL}/ t_{PLH}	propagation delay CP to Q_n	C_L = 15 pF; V_{CC} = 5 V	22	22	ns
f_{max}	maximum clock frequency		36	36	MHz
C_I	input capacitance		3.5	3.5	pF
C_{PD}	power dissipation capacitance per package	notes 1 and 2	31	33	pF

Notes

1. C_{PD} is used to determine the dynamic power dissipation (P_D in µW):
   $P_D = C_{PD} \times V_{CC}^2 \times f_i + \sum (C_L \times V_{CC}^2 \times f_o)$ where:
   f_i = input frequency in MHz
   f_o = output frequency in MHz
   $\sum (C_L \times V_{CC}^2 \times f_o)$ = sum of outputs
   C_L = output load capacitance in pF
   V_{CC} = supply voltage in V
2. For HC the condition is V_I = GND to V_{CC}
   For HCT the condition is V_I = GND to V_{CC} –1.5 V

ORDERING INFORMATION

See *"74HC/HCT/HCU/HCMOS Logic Package Information"*.

PIN DESCRIPTION

PIN NO.	SYMBOL	NAME AND FUNCTION
3, 2, 6, 7	Q_0 to Q_3	flip-flop outputs
4	$\overline{CE}$	count enable input (active LOW)
5	$\overline{U}$/D	up/down input
8	GND	ground (0 V)
11	$\overline{PL}$	parallel load input (active LOW)
12	TC	terminal count output
13	$\overline{RC}$	ripple clock output (active LOW)
14	CP	clock input (LOW-to-HIGH, edge triggered)
15, 1, 10, 9	D_0 to D_3	data inputs
16	V_{CC}	positive supply voltage

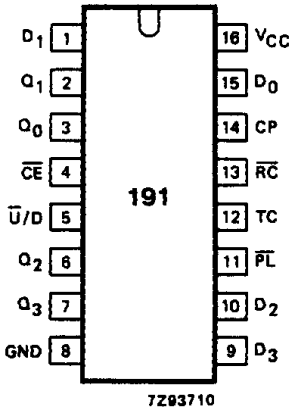


Fig.1 Pin configuration.

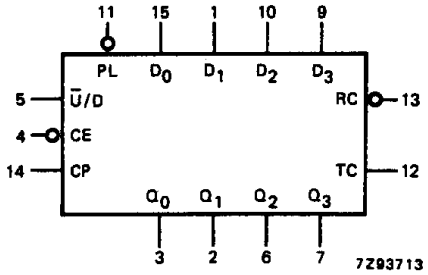


Fig.2 Logic symbol.

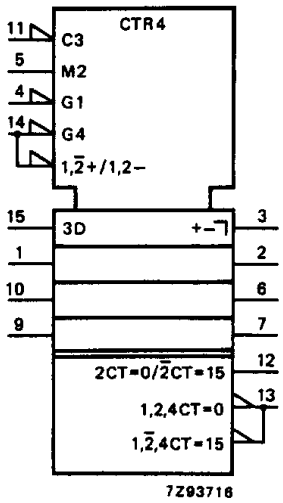


Fig.3 IEC logic symbol.

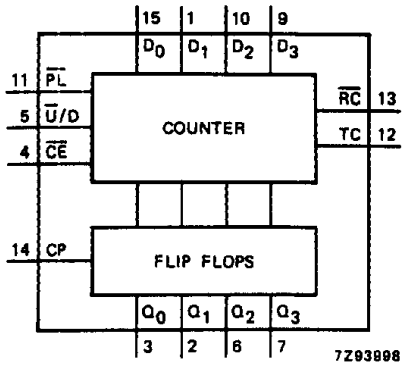


Fig.4 Functional diagram.

FUNCTION TABLE

OPERATING MODE	INPUTS					OUTPUTS
	$\overline{PL}$	$\overline{U}$/D	$\overline{CE}$	CP	D_n	Q_n
parallel load	L L	X X	X X	X X	L H	L H
count up	H	L	I	↑	X	count up
count down	H	H	I	↑	X	count down
hold (do nothing)	H	X	H	X	X	no change

TC AND RC FUNCTION TABLE

INPUTS			TERMINAL COUNT STATE				OUTPUTS	
$\overline{U}$/D	$\overline{CE}$	CP	Q_0	Q_1	Q_2	Q_3	TC	$\overline{RC}$
H	H	X	H	H	H	H	L	H
L	H	X	H	H	H	H	H	H
L	L	⊓⊔	H	H	H	H	⊓	⊓⊔
L	H	X	L	L	L	L	L	H
H	H	X	L	L	L	L	H	H
H	L	⊓⊔	L	L	L	L	⊓	⊓⊔

Notes

1. H = HIGH voltage level
   L = LOW voltage level
   I = LOW voltage level one set-up time prior to the LOW-to-HIGH CP transition
   X = don't care
   ↑ = LOW-to-HIGH CP transition

   ⊓⊔ = one LOW level pulse

   ⊓ = TC goes LOW on a LOW-to-HIGH CP transition

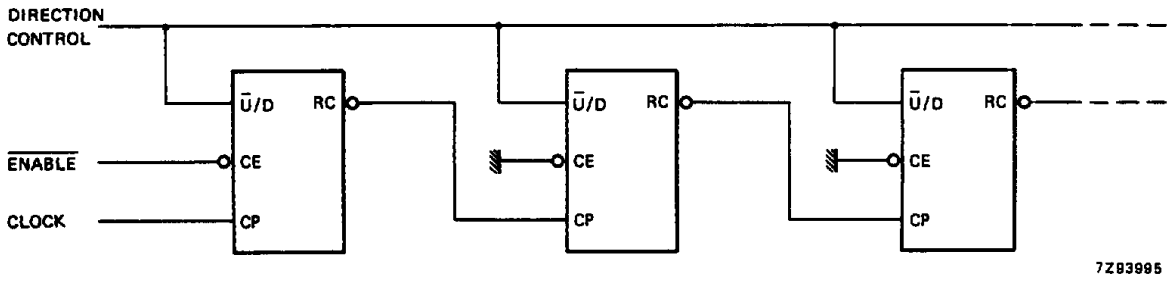


Fig.5 N-stage ripple counter using ripple clock.

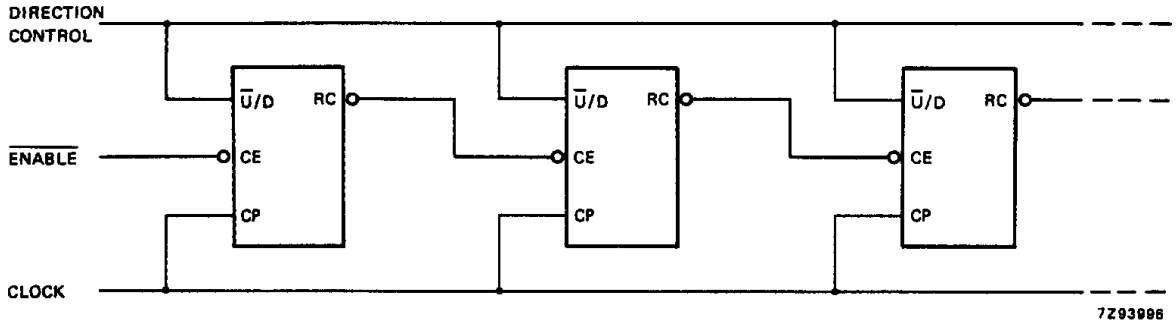


Fig.6 Synchronous n-stage counter using ripple carry/borrow.

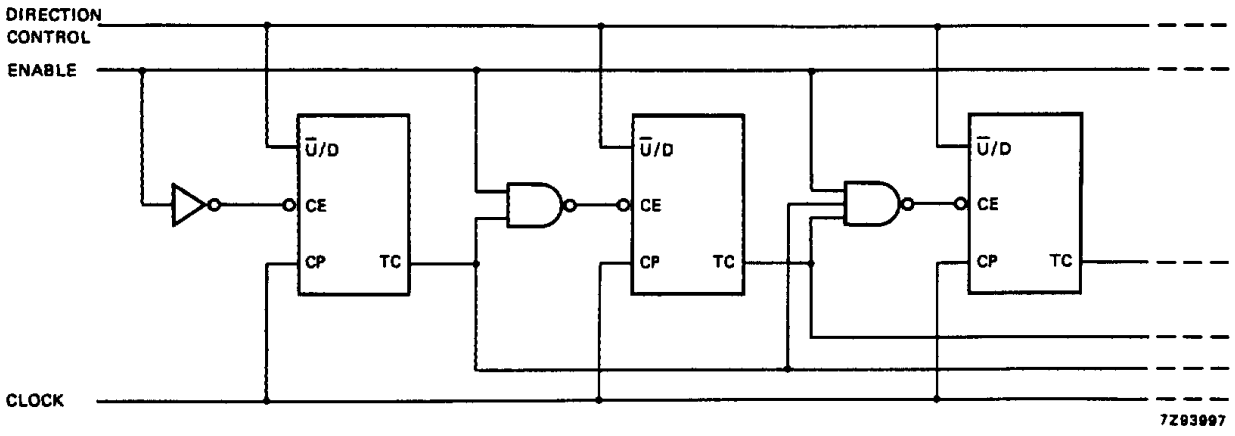
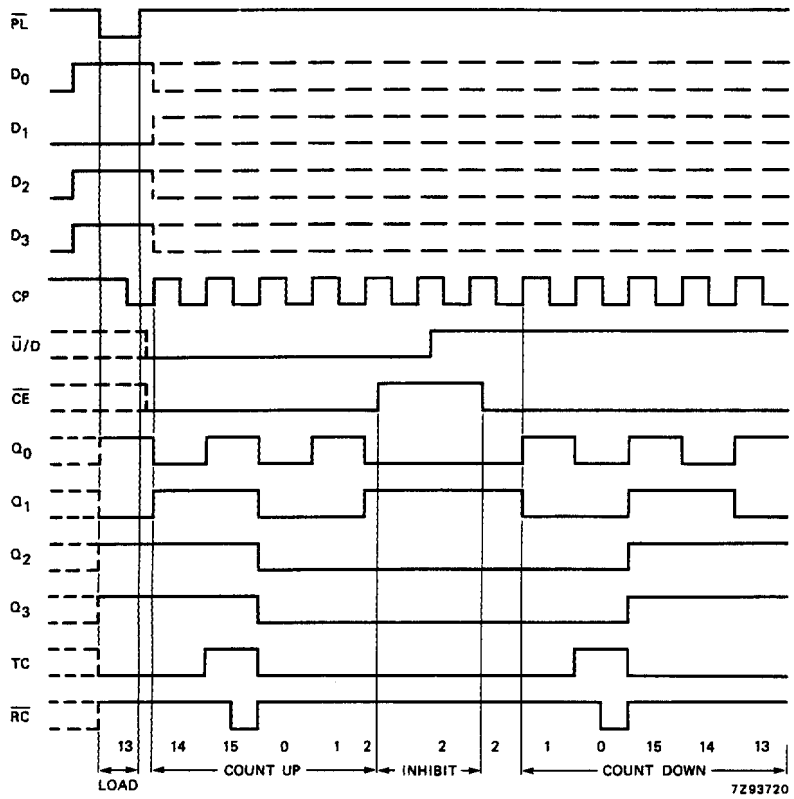


Fig.7 Synchronous n-stage counter with parallel gated carry/borrow.

Sequence

Load (preset) to binary thirteen;
count up to fourteen, fifteen, zero, one and two;
inhibit;
count down to one, zero, fifteen, fourteen and thirteen.

Fig.8 Typical load, count and inhibit sequence.



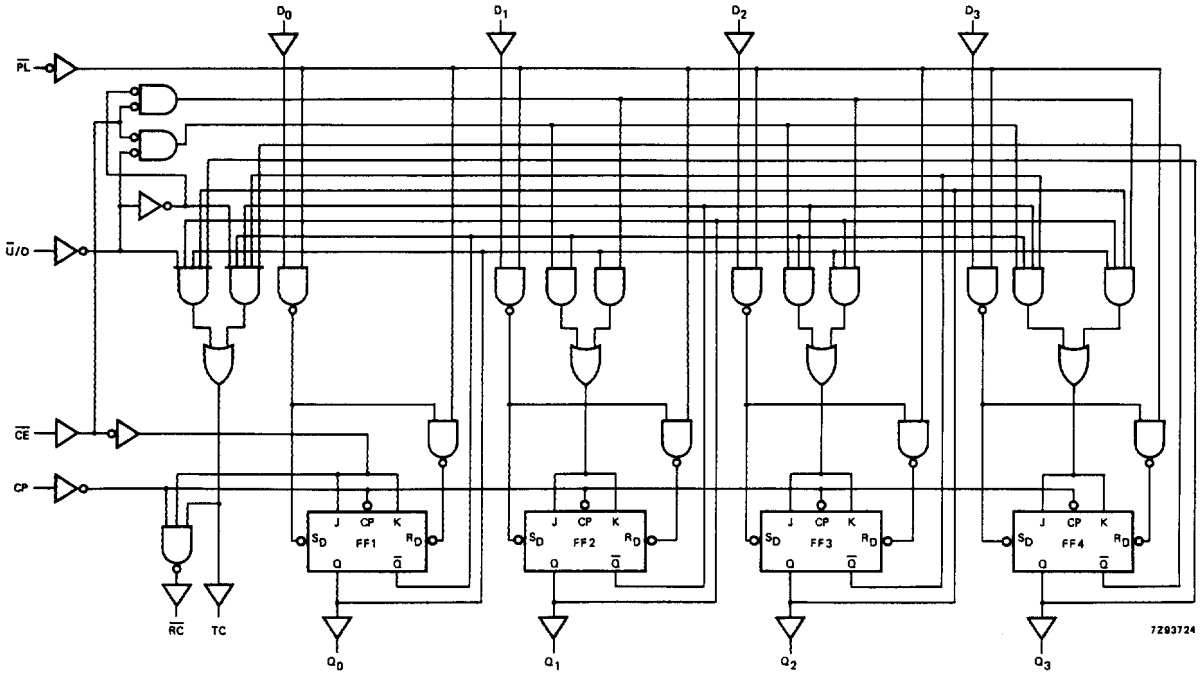


Fig.9 Logic diagram.

DC CHARACTERISTICS FOR 74HC

For the DC characteristics see *"74HC/HCT/HCU/HCMOS Logic Family Specifications"*.

Output capability: standard
I_{CC} category: MSI

AC CHARACTERISTICS FOR 74HC

GND = 0 V; $t_r = t_f$ = 6 ns; C_L = 50 pF

SYMBOL	PARAMETER	T_{amb} (°C)							UNIT	TEST CONDITIONS	
		74HC								V_{CC} (V)	WAVEFORMS
		+25			−40 to +85		−40 to +125				
		min.	typ.	max.	min.	max.	min.	max.			
t_{PHL}/ t_{PLH}	propagation delay CP to Q_n		72 26 21	220 44 37		275 55 47		330 66 56	ns	2.0 4.5 6.0	Fig.10
t_{PHL}/ t_{PLH}	propagation delay CP to TC		83 30 24	255 51 43		320 64 54		395 77 65	ns	2.0 4.5 6.0	Fig.10
t_{PHL}/ t_{PLH}	propagation delay CP to $\overline{RC}$		47 17 14	150 30 26		190 38 33		225 45 38	ns	2.0 4.5 6.0	Fig.11
t_{PHL}/ t_{PLH}	propagation delay $\overline{CE}$ to $\overline{RC}$		33 12 10	130 26 22		165 33 28		195 39 33	ns	2.0 4.5 6.0	Fig.11
t_{PHL}/ t_{PLH}	propagation delay D_n to Q_n		61 22 18	220 44 37		275 55 47		330 66 56	ns	2.0 4.5 6.0	Fig.12
t_{PHL}/ t_{PLH}	propagation delay $\overline{PL}$ to Q_n		61 22 18	220 44 37		275 55 47		330 66 56	ns	2.0 4.5 6.0	Fig.13
t_{PHL}/ t_{PLH}	propagation delay $\overline{U}$/D to TC		44 16 13	190 38 32		240 48 41		285 57 48	ns	2.0 4.5 6.0	Fig.14
t_{PHL}/ t_{PLH}	propagation delay $\overline{U}$/D to $\overline{RC}$		50 18 14	210 42 36		265 53 45		315 63 54	ns	2.0 4.5 6.0	Fig.14
t_{THL}/ t_{TLH}	output transition time		19 7 6	75 15 13		95 19 16		110 22 19	ns	2.0 4.5 6.0	Fig.15
t_W	clock pulse width HIGH or LOW	125 25 21	28 10 8		155 31 26		195 39 33		ns	2.0 4.5 6.0	Fig.10
t_W	parallel load pulse width LOW	100 20 17	22 8 6		125 25 21		150 30 26		ns	2.0 4.5 6.0	Fig.15

SYMBOL	PARAMETER	T_{amb} (°C)							UNIT	TEST CONDITIONS	
		74HC								V_{CC} (V)	WAVEFORMS
		+25			−40 to +85		−40 to +125				
		min.	typ.	max.	min.	max.	min.	max.			
t_{rem}	removal time $\overline{PL}$ to CP	35 7 6	8 3 2		45 9 8		55 11 9		ns	2.0 4.5 6.0	Fig.15
t_{su}	set-up time $\overline{U}$/D to CP	205 41 35	50 18 14		255 51 43		310 62 53		ns	2.0 4.5 6.0	Fig.17
t_{su}	set-up time D_n to $\overline{PL}$	100 20 17	19 7 6		125 25 21		150 30 26		ns	2.0 4.5 6.0	Fig.16
t_{su}	set-up time $\overline{CE}$ to CP	140 28 24	44 16 13		175 35 30		210 42 36		ns	2.0 4.5 6.0	Fig.17
t_h	hold time $\overline{U}$/D to CP	0 0 0	−39 −14 −11		0 0 0		0 0 0		ns	2.0 4.5 6.0	Fig.17
t_h	hold time D_n to $\overline{PL}$	0 0 0	−11 −4 −3		0 0 0		0 0 0		ns	2.0 4.5 6.0	Fig.16
t_h	hold time $\overline{CE}$ to CP	0 0 0	−28 −10 −8		0 0 0		0 0 0		ns	2.0 4.5 6.0	Fig.17
f_{max}	maximum clock pulse frequency	4.0 20 24	11 33 39		3.2 16 19		2.6 13 15		MHz	2.0 4.5 6.0	Fig.10

DC CHARACTERISTICS FOR 74HCT

For the DC characteristics see *"74HC/HCT/HCU/HCMOS Logic Family Specifications"*.

Output capability: standard
I_{CC} category: MSI

Note to HCT types

The value of additional quiescent supply current (ΔI_{CC}) for a unit load of 1 is given in the family specifications. To determine ΔI_{CC} per input, multiply this value by the unit load coefficient shown in the table below.

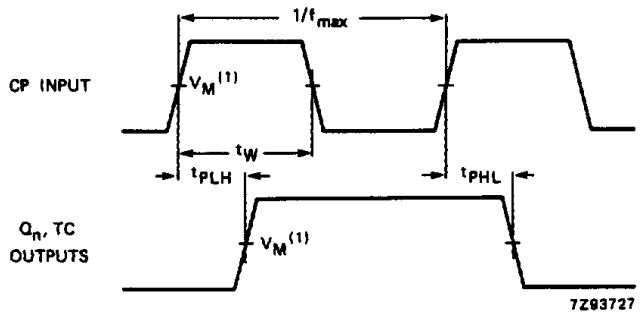
INPUT	UNIT LOAD COEFFICIENT
D_n	0.5
$\overline{CP}$	0.65
$\overline{U}/D$	1.15
$\overline{CE}$, $\overline{PL}$	1.5

AC CHARACTERISTICS FOR 74HCT

GND = 0 V; $t_r = t_f$ = 6 ns; C_L = 50 pF

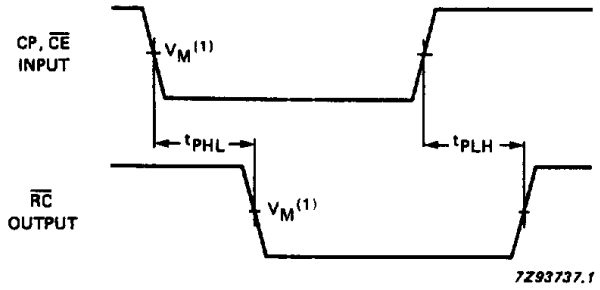
SYMBOL	PARAMETER	T_{amb} (°C)							UNIT	TEST CONDITIONS	
		74HC								V_{CC} (V)	WAVEFORMS
		+25			−40 to +85		−40 to +125				
		min.	typ.	max.	min.	max.	min.	max.			
t_{PHL}/ t_{PLH}	propagation delay CP to Q_n		26	48		60		72	ns	4.5	Fig.10
t_{PHL}/ t_{PLH}	propagation delay CP to TC		32	51		64		77	ns	4.5	Fig.10
t_{PHL}/ t_{PLH}	propagation delay CP to $\overline{RC}$		19	35		44		53	ns	4.5	Fig.11
t_{PHL}/ t_{PLH}	propagation delay $\overline{CE}$ to $\overline{RC}$		19	33		41		50	ns	4.5	Fig.11
t_{PHL}/ t_{PLH}	propagation delay D_n to Q_n		22	44		55		66	ns	4.5	Fig.12
t_{PHL}/ t_{PLH}	propagation delay $\overline{PL}$ to Q_n		27	46		58		69	ns	4.5	Fig.13
t_{PHL}/ t_{PLH}	propagation delay $\overline{U}$/D to TC		23	45		56		68	ns	4.5	Fig.14
t_{PHL}/ t_{PLH}	propagation delay $\overline{U}$/D to $\overline{RC}$		24	45		56		68	ns	4.5	Fig.14
t_{THL}/ t_{TLH}	output transition time		7	15		19		22	ns	4.5	Fig.15
t_W	clock pulse width HIGH or LOW	16	9		20		24		ns	4.5	Fig.10
t_W	parallel load pulse width LOW	22	11		28		33		ns	4.5	Fig.15
t_{rem}	removal time $\overline{PL}$ to CP	7	1		9		11		ns	4.5	Fig.15
t_{su}	set-up time $\overline{U}$/D to CP	41	20		51		62		ns	4.5	Fig.17
t_{su}	set-up time D_n to $\overline{PL}$	20	9		25		30		ns	4.5	Fig.16
t_{su}	set-up time $\overline{CE}$ to CP	30	18		38		45		ns	4.5	Fig.17
t_h	hold time $\overline{U}$/D to CP	0	−18		0		0		ns	4.5	Fig.17
t_h	hold time D_n to $\overline{PL}$	0	−5		0		0		ns	4.5	Fig.16
t_h	hold time $\overline{CE}$ to CP	0	−10		0		0		ns	4.5	Fig.17
f_{max}	maximum clock pulse frequency	20	33		16		13		MHz	4.5	Fig.10

AC WAVEFORMS



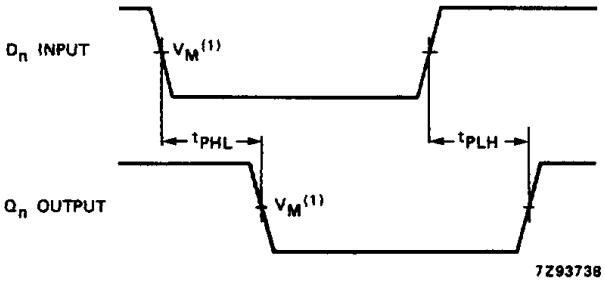
(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.10 Waveforms showing the clock (CP) to output (Q_n) propagation delays, the clock pulse width and the maximum clock pulse frequency.



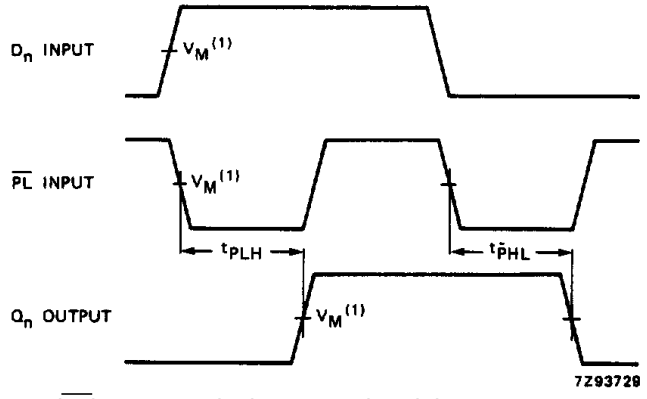
(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.11 Waveforms showing the clock and count enable inputs (CP, $\overline{CE}$) to ripple clock output ($\overline{RC}$) propagation delays.



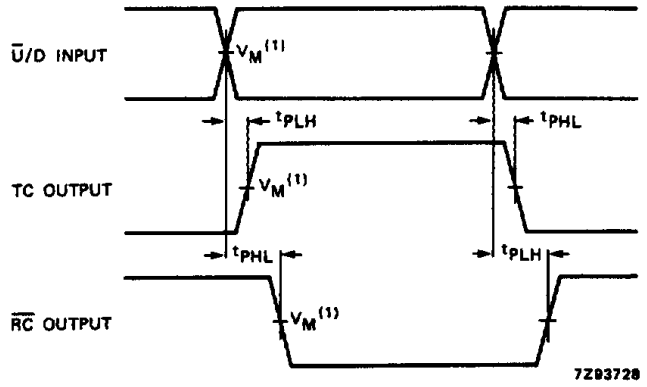
(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.12 Waveforms showing the input (D_n) to output (Q_n) propagation delays.



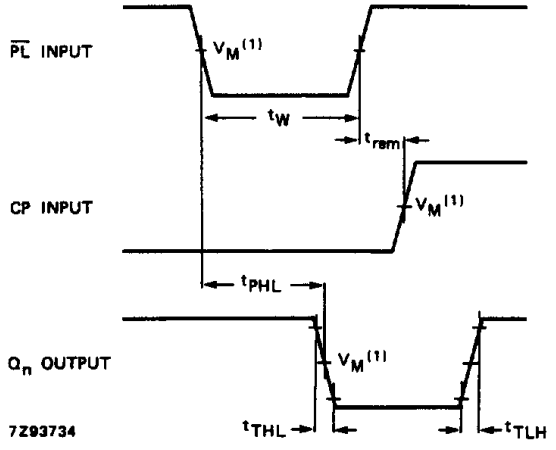
(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.13 Waveforms showing the input ($\overline{PL}$) to output (Q_n) propagation delays.



(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.14 Waveforms showing the up/down count input ($\overline{U}$/D) to terminal count and ripple clock output (TC, $\overline{RC}$) propagation delays.



(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

Fig.15 Waveforms showing the parallel load input ($\overline{PL}$) pulse width, removal time to clock (CP) and the output (Q_n) transition times.

The shaded areas indicate when the input is permitted to change for predictable output performance.

(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

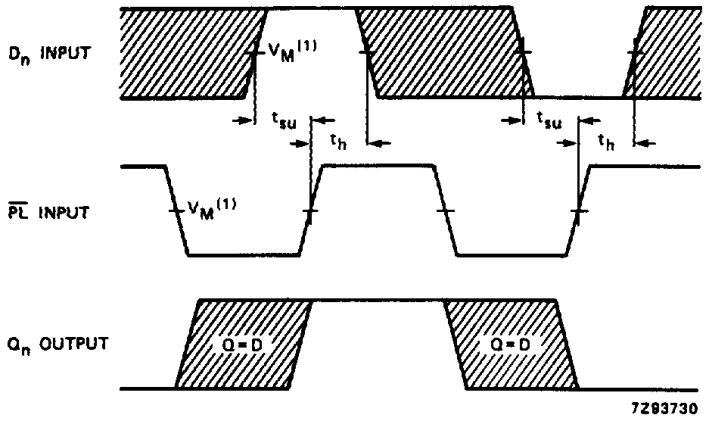


Fig.16 Waveforms showing the set-up and hold times from the parallel load input ($\overline{PL}$) to the data input (D_n).

The shaded areas indicate when the input is permitted to change for predictable output performance.

(1) HC : V_M = 50%; V_I = GND to V_{CC}.
HCT : V_M = 1.3 V; V_I = GND to 3 V.

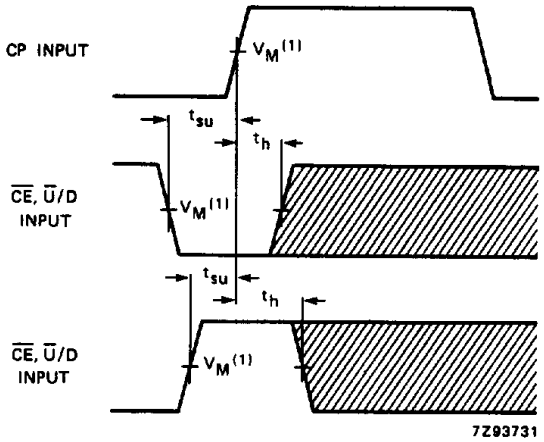


Fig.17 Waveforms showing the set-up and hold times from the count enable and up/down inputs ($\overline{CE}$, $\overline{U}$/D) to the clock (CP).

PACKAGE OUTLINES

See *"74HC/HCT/HCU/HCMOS Logic Package Outlines"*.

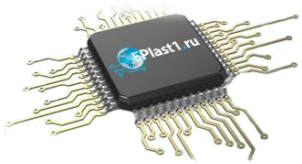
Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits,General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)
Факс: 8 (812) 320-02-42
Электронная почта: org@eplast1.ru
Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.