

# 16V Pico TLynx<sup>™</sup> 4A: Non-Isolated DC-DC Power Modules 8Vdc -16Vdc input; 0.6Vdc to 8.0Vdc output; 4A Output Current

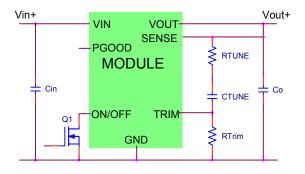




## **RoHS Compliant**

## **Applications**

- Distributed power architectures
- Intermediate bus voltage applications
- Telecommunications equipment
- Servers and storage applications
- Networking equipment
- Industrial equipment



#### **Features**

- Compliant to RoHS EU Directive 2002/95/EC (Z versions)
- Compatible in a Pb-free or SnPb reflow environment (Z versions)
- Wide Input voltage range (8Vdc-16Vdc)
- Output voltage programmable from 0.59Vdc to 8.0Vdc via external resistor
- Tunable Loop<sup>TM</sup> to optimize dynamic output voltage response
- Remote sense
- Power Good signal
- Fixed switching frequency
- Output overcurrent protection (non-latching)
- Overtemperature protection
- Remote On/Off
- Ability to sink and source current
- Cost efficient open frame design
- Small size: 12.2 mm x 12.2 mm x 7.25 mm (0.48 in x 0.48 in x 0.29 in)
- Wide operating temperature range (-40°C to 85°C)
- UL\* Recognized to UL60950-1, CAN/CSA<sup>†</sup> C22.2 No. 60950-1-03, and EN60950-1(VDE<sup>‡</sup> 0805-1) Licensed
- ISO\*\* 9001 and ISO 14001 certified manufacturing facilities

## **Description**

The 16V Pico TLynx<sup>TM</sup> 4A power modules are non-isolated dc-dc converters that can deliver up to 4A of output current. These modules operate over a wide range of input voltage (V<sub>IN</sub> = 8Vdc-16Vdc) and provide a precisely regulated output voltage from 0.59Vdc to 8.0 Vdc, programmable via an external resistor. Features include remote On/Off, adjustable output voltage, over current and over temperature protection. A new feature, the Tunable Loop<sup>™</sup>, allows the user to optimize the dynamic response of the converter to match the load with reduced amount of output capacitance leading to savings on cost and PWB area.

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<sup>\*\*</sup> ISO is a registered trademark of the International Organization of Standards

## **Absolute Maximum Ratings**

Stresses in excess of the absolute maximum ratings can cause permanent damage to the device. These are absolute stress ratings only, functional operation of the device is not implied at these or any other conditions in excess of those given in the operations sections of the data sheet. Exposure to absolute maximum ratings for extended periods can adversely affect the device reliability.

Parameter	Device	Symbol	Min	Max	Unit
Input Voltage	All	V <sub>IN</sub>	-0.3	18	Vdc
Up to 10 seconds					
Operating Ambient Temperature	All	T <sub>A</sub>	-40	85	°C
(see Thermal Considerations section)					
Storage Temperature	All	$T_{stg}$	-55	125	°C

## **Electrical Specifications**

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions.

Parameter	Device	Symbol	Min	Тур	Max	Unit
Operating Input Voltage	All	V <sub>IN</sub>	8.0	_	16.0	Vdc
Maximum Input Current	All	I <sub>IN,max</sub>			6.5	Adc
( $V_{IN}$ =8V to 16V, $I_O$ = $I_{O, max}$ )						
Input No Load Current (V <sub>IN</sub> = 16.0Vdc, I <sub>O</sub> = 0, module enabled)	$V_{O,set}$ = 0.6 Vdc	I <sub>IN,No load</sub>		18		mA
	V <sub>O,set</sub> = 8.0Vdc	I <sub>IN,No load</sub>		96.1		mA
Input Stand-by Current (V <sub>IN</sub> = 16.0Vdc, module disabled)	All	I <sub>IN,stand-by</sub>		1.2		mA
Inrush Transient	All	l <sup>2</sup> t			1	A <sup>2</sup> s
Input Reflected Ripple Current, peak-to-peak (5Hz to 20MHz, 1 $\mu$ H source impedance; V <sub>IN</sub> =0 to 16V, I <sub>O</sub> = I <sub>Omax</sub> ; See Test Configurations)	All			50		mAp-p
Input Ripple Rejection (120Hz)	All			-46		dB

#### CAUTION: This power module is not internally fused. An input line fuse must always be used.

This power module can be used in a wide variety of applications, ranging from simple standalone operation to an integrated part of sophisticated power architecture. To preserve maximum flexibility, internal fusing is not included; however, to achieve maximum safety and system protection, always use an input line fuse. The safety agencies require a fast-acting fuse with a maximum rating of 6 A (see Safety Considerations section). Based on the information provided in this data sheet on inrush energy and maximum dc input current, the same type of fuse with a lower rating can be used. Refer to the fuse manufacturer's data sheet for further information.

# **Electrical Specifications** (continued)

Output Voltage Set-point (with 0.5% tolerance for external resistor used to set output voltage)         All         V <sub>O. set</sub> -1.5         +1.5         % V <sub>O. set</sub> vol. set venteral resistor used to set output voltage, resistive load, and temperature conditions until end of life)           Adjustment Range (selected by an external resistor) (Some output voltages are ynot be possible depending on the input voltage – see Feature Descriptions Section)         All         V <sub>O. set</sub> 2.2.5         —         42.5         % V <sub>O. set</sub> Adjustment Range (selected by an external resistor) (Some output voltages are ynot be possible depending on the input voltage – see Feature Descriptions Section)         All         U         0.59         8.0         Vdc           Remote Sense Range         All         U         0.59         8.0         Vdc           Output Regulation (for V <sub>O. ≥ 2</sub> .5 Vdc)         All         —         —         +0.4         % V <sub>O. set</sub> Line (V <sub>In</sub> =V <sub>N. mon</sub> to V <sub>S. max</sub> )         All         —         —         10 mV         % V <sub>O. set</sub> Load (lo=l <sub>O. mon</sub> to V <sub>O. max</sub> )         All         —         —         10 mV         % V <sub>O. set</sub> Line (V <sub>In</sub> =V <sub>N. mon</sub> to V <sub>O. max</sub> )         All         —         —         10 mV         % V <sub>O. set</sub> Line (V <sub>In</sub> =V <sub>N. mon</sub> to V <sub>O. max</sub> )         All         —         —         10 mV </th <th>Parameter</th> <th>Device</th> <th>Symbol</th> <th>Min</th> <th>Тур</th> <th>Max</th> <th>Unit</th>	Parameter	Device	Symbol	Min	Тур	Max	Unit
load, and temperature conditions until end of life)		All	$V_{O,  set}$	-1.5		+1.5	% V <sub>O, set</sub>
(Some output voltages may not be possible depending on the input voltage – see Feature Descriptions Section)         All         Vo         0.59         8.0         Vdc           Remote Sense Range         All         —         0.5         Vdc           Output Regulation (for V <sub>O</sub> ≥ 2.5Vdc)         —         —         +0.4         % V <sub>O, set</sub> Load (I <sub>O</sub> =I <sub>O, min</sub> to I <sub>O, max</sub> )         All         —         —         +0.4         % V <sub>O, set</sub> Uoutput Regulation (for V <sub>O</sub> ≥ 2.5Vdc)         —         —         10         mV           Line (V <sub>In</sub> =V <sub>MI, min</sub> to I <sub>O, max</sub> )         All         —         —         10         mV           Load (I <sub>O</sub> =I <sub>O, min</sub> to I <sub>O, max</sub> )         All         —         —         10         mV           Temperature (T <sub>ma</sub> =T-I <sub>A</sub> min to T <sub>A, max</sub> )         All         —         —         0.4         % V <sub>O, set</sub> With the Sense Range         All         —         —         0.4         % V <sub>O, set</sub> Output Ripple and Noise on nominal output (V <sub>V=VM, min</sub> and I <sub>O</sub> =I <sub>O, min</sub> to I <sub>O, max</sub> CO = 0.1 µF // 10 µF         —         —         90         100         mV <sub>Pi-Fi</sub> Rest can C-Peak (5Hz to 20MHz bandwidth)         All         C <sub>O, max</sub> 0         —         22         µF		All	V <sub>O, set</sub>	-2.5	_	+2.5	% V <sub>O, set</sub>
Output Regulation (for V <sub>O</sub> ≥ 2.5Vdc)         All         —         +0.4         % V <sub>O.set</sub> Line (V <sub>IN</sub> =V <sub>PIL, min</sub> to V <sub>NL, max</sub> )         All         —         +0.4         % V <sub>O.set</sub> Output Regulation (for V <sub>O</sub> < 2.5Vdc)	(Some output voltages may not be possible depending	All	Vo	0.59		8.0	Vdc
Line (Vij=Vals, min to Vis, max)	Remote Sense Range	All				0.5	Vdc
Load (Io=Io,min to Io,max)	Output Regulation (for V <sub>O</sub> ≥ 2.5Vdc)						
Output Regulation (for Vo < 2.5Vdc)	Line $(V_{IN}=V_{IN, min} \text{ to } V_{IN, max})$	All				+0.4	% V <sub>O, set</sub>
Line (V <sub>N</sub> =V <sub>N,min</sub> to V <sub>N,max</sub> )	Load ( $I_O=I_{O, min}$ to $I_{O, max}$ )	All			_	10mV	% V <sub>O, set</sub>
Load (to=l <sub>O, mix</sub> to l <sub>O, max</sub> )	Output Regulation (for V <sub>o</sub> < 2.5Vdc)						
Temperature (T <sub>ref</sub> =T <sub>A min</sub> to T <sub>A max</sub> ) Remote Sense Range  Output Ripple and Noise on nominal output (V <sub>N</sub> =V <sub>N, nom</sub> and I <sub>O</sub> =I <sub>O, min</sub> to I <sub>O, max</sub> Co = 0.1μF // 10 μF ceramic capacitors) Peak-to-Peak (5Hz to 20MHz bandwidth) All  All  —90 100 mV <sub>pk,pk</sub> RMS (5Hz to 20MHz bandwidth) All  —80 36 38 mV <sub>ms</sub> External Capacitance³ Without the Tunable Loop™ ESR ≥ 1 mΩ  All  Co <sub>, max</sub> 0 — 22 μF  With the Tunable Loop™ ESR ≥ 0.15 mΩ All  Co <sub>, max</sub> 0 — 470 μF ESR ≥ 0.15 mΩ All  Co <sub>, max</sub> 0 — 3000 μF  Output Current (in either sink or source mode) All  I <sub>O, lim</sub> All  I <sub>O, sin</sub> 200 mA <sub>ms</sub> % I <sub>O,max</sub> With short-Circuit Current (V <sub>O</sub> ≤250mV) ( Hiccup Mode ) (Current Imit does not operate in sink mode)  Efficiency  V <sub>O, set</sub> = 0.8Vdc  I <sub>O</sub> =1,2Vdc  I <sub>O</sub> =1,0max  V <sub>O, set</sub> = 1.8Vdc  I <sub>O, set</sub> Solvdc  I <sub>O</sub> Solvec  Sol	Line $(V_{IN}=V_{IN, min} \text{ to } V_{IN, max})$	All				10	mV
Remote Sense Range         All          0.5         V           Output Ripple and Noise on nominal output ( $V_{NN}=V_{Nn,nom}$ and $I_{O=In,min}$ to $I_{O,max}$ Co = 0.1µF // 10 µF ceramic capacitors)         —         90         100         mV <sub>pk-pk</sub> ceramic capacitors)           Peak-to-Peak (5Hz to 20MHz bandwidth)         All         —         90         100         mV <sub>pk-pk</sub> ceramic capacitors)           External Capacitance¹         Without the Tunable Loop™         All         Co, max         0         —         22         µF           With the Tunable Loop™         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Io         0         —         470         µF           Output Current (in either sink or source mode)         All         Io, im         200         w         Vo,max           Output Current Limit Inception (Hiccup Mode)         All         Io, im         200         mA <sub>m</sub> (Vo≤250mV) (Hiccup Mode)         Vo,set = 0.6Vdc         η         74.2	Load (I <sub>O</sub> =I <sub>O, min</sub> to I <sub>O, max</sub> )	All				5	mV
Remote Sense Range         All          0.5         V           Output Ripple and Noise on nominal output ( $V_{NN}=V_{Nn,nom}$ and $I_{O=In,min}$ to $I_{O,max}$ Co = 0.1µF // 10 µF ceramic capacitors)         —         90         100         mV <sub>pk-pk</sub> ceramic capacitors)           Peak-to-Peak (5Hz to 20MHz bandwidth)         All         —         90         100         mV <sub>pk-pk</sub> ceramic capacitors)           External Capacitance¹         Without the Tunable Loop™         All         Co, max         0         —         22         µF           With the Tunable Loop™         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Co, max         0         —         470         µF           ESR ≥ 10 mΩ         All         Io         0         —         470         µF           Output Current (in either sink or source mode)         All         Io, im         200         w         Vo,max           Output Current Limit Inception (Hiccup Mode)         All         Io, im         200         mA <sub>m</sub> (Vo≤250mV) (Hiccup Mode)         Vo,set = 0.6Vdc         η         74.2	Temperature (T <sub>ref</sub> =T <sub>A, min</sub> to T <sub>A, max</sub> )	All				0.4	% V <sub>O. set</sub>
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Remote Sense Range	All				0.5	
ceramic capacitors)         Peak-to-Peak (5Hz to 20MHz bandwidth)         All         —         90         100         mV <sub>pk-pk</sub> mV <sub>pk-pk-pk</sub> mV <sub>pk-pk-pk-pk</sub> mV <sub>pk-pk-pk-pk-pk</sub> mV <sub>pk-pk-pk-pk-pk-pk-pk-pk-pk-pk-pk-pk-pk-p</sub>	Output Ripple and Noise on nominal output						
RMS (5Hz to 20MHz bandwidth)       All       36       38       mV <sub>rms</sub> External Capacitance¹       Without the Tunable Loop™       Image: Common of the common o	ceramic capacitors)	All		_	90	100	$mV_{pk-pk}$
Without the Tunable Loop <sup>TM</sup> ESR ≥ 1 mΩ       All $C_{O,max}$ 0       —       22       μF         With the Tunable Loop <sup>TM</sup> ESR ≥ 0.15 mΩ       All $C_{O,max}$ 0       —       470       μF         ESR ≥ 0.15 mΩ       All $C_{O,max}$ 0       —       470       μF         ESR ≥ 10 mΩ       All $C_{O,max}$ 0       —       3000       μF         Output Current (in either sink or source mode)       All $I_{O,max}$ 0       —       4 Adc         Output Current Limit Inception (Hiccup Mode) (current limit does not operate in sink mode)       All $I_{O,im}$ 200 $\%$ $I_{O,max}$ Output Short-Circuit Current (Vo≤250mV) ( Hiccup Mode )       All $I_{O,sc}$ 200 $mA_{rms}$ Efficiency V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C       V <sub>O,set</sub> = 0.6Vdc       η       74.2       % $I_{O=I_{O,max}, V_O=V_{O,set}}$ V <sub>O,set</sub> = 1.8Vdc       η       83.7       % $I_{O=I_{O,max}, V_O=V_{O,set}}$ V <sub>O,set</sub> = 3.3Vdc       η       90.2       % $I_{O,max}$ (V <sub>O,set</sub> = 5.0Vdc       η       93.7       % $I_{O,set}$ (V <sub>O,set</sub> = 6.5Vdc       η       94.9       % $I_{O,set}$ (V <sub>O,set</sub> =	RMS (5Hz to 20MHz bandwidth)	All			36	38	
Without the Tunable Loop <sup>TM</sup> ESR ≥ 1 mΩ       All $C_{O,max}$ 0       —       22       μF         With the Tunable Loop <sup>TM</sup> ESR ≥ 0.15 mΩ       All $C_{O,max}$ 0       —       470       μF         ESR ≥ 0.15 mΩ       All $C_{O,max}$ 0       —       470       μF         ESR ≥ 10 mΩ       All $C_{O,max}$ 0       —       3000       μF         Output Current (in either sink or source mode)       All $I_{O,max}$ 0       —       4 Adc         Output Current Limit Inception (Hiccup Mode) (current limit does not operate in sink mode)       All $I_{O,im}$ 200 $\%$ $I_{O,max}$ Output Short-Circuit Current (Vo≤250mV) ( Hiccup Mode )       All $I_{O,sc}$ 200 $mA_{rms}$ Efficiency V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C       V <sub>O,set</sub> = 0.6Vdc       η       74.2       % $I_{O=I_{O,max}, V_O=V_{O,set}}$ V <sub>O,set</sub> = 1.8Vdc       η       83.7       % $I_{O=I_{O,max}, V_O=V_{O,set}}$ V <sub>O,set</sub> = 3.3Vdc       η       90.2       % $I_{O,max}$ (V <sub>O,set</sub> = 5.0Vdc       η       93.7       % $I_{O,set}$ (V <sub>O,set</sub> = 6.5Vdc       η       94.9       % $I_{O,set}$ (V <sub>O,set</sub> =	External Capacitance <sup>1</sup>						
With the Tunable Loop TM       All $C_{O,max}$ 0       —       470       μF         ESR ≥ 0.15 mΩ       All $C_{O,max}$ 0       —       3000       μF         Output Current (in either sink or source mode)       All $I_{O,max}$ 0       —       400       μF         Output Current Limit Inception (Hiccup Mode) (current limit does not operate in sink mode)       All $I_{O,lim}$ 200 $W_{I,max}$ Output Short-Circuit Current (Vo≤250mV) ( Hiccup Mode )       All $I_{O,s/c}$ 200 $I_{O,max}$ Efficiency $V_{O,set} = 0.6 Vdc$ $I_{O,s/c}$ 200 $I_{O,max}$ $V_{I,N} = 12 Vdc$ , $I_{A} = 25 °C$ $V_{O,set} = 1.2 Vdc$ $I_{O,s/c}$	Without the Tunable Loop <sup>™</sup>						
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	ESR ≥ 1 mΩ	All	C <sub>O. max</sub>	0	_	22	μF
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	With the Tunable Loop <sup>TM</sup>						
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$		All	C <sub>O. max</sub>	0	_	470	μF
Output Current (in either sink or source mode)         All         I₀         0         4         Adc           Output Current Limit Inception (Hiccup Mode) (current limit does not operate in sink mode)         All         I₀, lim         200         % I₀, max           Output Short-Circuit Current (V₀≤250mV) ( Hiccup Mode )         All         I₀, s/c         200         mA <sub>rms</sub> Efficiency         V₀, set = 0.6Vdc         η         74.2         %           V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C         V₀, set = 1.2Vdc         η         83.7         %           I₀=I₀, max, V₀= V₀, set         V₀, set = 1.8Vdc         η         87.7         %           V₀, set = 2.5Vdc         η         90.2         %           V₀, set = 3.3Vdc         η         91.7         %           V₀, set = 5.0Vdc         η         94.9         %           V₀, set = 6.5Vdc         η         94.9         %           V₀, set = 8.0Vdc         η         96.1         %	ESR ≥ 10 mΩ	All		0	_	3000	μF
Output Current Limit Inception (Hiccup Mode) (current limit does not operate in sink mode)         All         I <sub>O, lim</sub> 200         % I <sub>O,max</sub> Output Short-Circuit Current (V <sub>O</sub> ≤250mV) ( Hiccup Mode )         All         I <sub>O, s/c</sub> 200         mA <sub>ms</sub> Efficiency         V <sub>O,set</sub> = 0.6Vdc η         η         74.2         %           V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C         V <sub>O,set</sub> = 1.2Vdc η         83.7         %           I <sub>O</sub> =I <sub>O, max</sub> , V <sub>O</sub> = V <sub>O,set</sub> V <sub>O,set</sub> = 1.8Vdc η         η         87.7         %           V <sub>O,set</sub> = 2.5Vdc η         90.2         %           V <sub>O,set</sub> = 3.3Vdc η         91.7         %           V <sub>O,set</sub> = 5.0Vdc η         93.7         %           V <sub>O,set</sub> = 6.5Vdc η         94.9         %           V <sub>O,set</sub> = 8.0Vdc η         96.1         %	Output Current (in either sink or source mode)	All	Io	0		4	Adc
Output Short-Circuit Current (Vo≤250mV) ( Hiccup Mode )       All       Io, s/c       200       mA <sub>rms</sub> Efficiency V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C Io=Io, max, Vo=Vo,set       Vo,set = 0.6Vdc Vo,set = 1.2Vdc Vo,set = 1.8Vdc Vo,set = 1.8Vdc Vo,set = 2.5Vdc Vo,set = 2.5Vdc Vo,set = 3.3Vdc Vo,set = 5.0Vdc Vo,set = 5.0Vdc Vo,set = 6.5Vdc Vo,set = 8.0Vdc       η       91.7 94.9 94.9 96.1							24.1
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	(current limit does not operate in sink mode)	All	I <sub>O, lim</sub>		200		% I <sub>o,max</sub>
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Output Short-Circuit Current	All	I <sub>O, s/c</sub>		200		mA <sub>rms</sub>
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	(V <sub>o</sub> ≤250mV) ( Hiccup Mode )						
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Efficiency	V <sub>O.set</sub> = 0.6Vdc	η		74.2		%
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	V <sub>IN</sub> = 12Vdc, T <sub>A</sub> =25°C				83.7		%
$\begin{array}{cccccccccccccccccccccccccccccccccccc$							
$\begin{array}{cccccccccccccccccccccccccccccccccccc$							
$\begin{array}{cccccccccccccccccccccccccccccccccccc$		*					
V <sub>O,set</sub> = 6.5Vdc       η       94.9       %         V <sub>O,set</sub> = 8.0Vdc       η       96.1       %							
V <sub>O,set</sub> = 8.0Vdc η 96.1 %							
Switching Frequency $ $ All $ $ $ $ $ $ $ $ $ $ $ $ $ $ $ $ $ $ 600 $ $ $ $ $ $ $ $ KHz	Switching Frequency	All	f <sub>sw</sub>	_	600		kHz

<sup>&</sup>lt;sup>1</sup>External capacitors may require using the new Tunable Loop<sup>™</sup> feature to ensure that the module is stable as well as getting the best transient response. See the Tunable Loop<sup>™</sup> section for details.

## **General Specifications**

Parameter	Min	Тур	Max	Unit
Calculated MTBF ( $I_0$ =0.8 $I_{O,max},T_A$ =40°C) Telcordia Issue 2 Method 1 Case 3		14,353,850		Hours
Weight	_	1.92 (0.0677)	_	g (oz.)

# **Feature Specifications**

Unless otherwise indicated, specifications apply over all operating input voltage, resistive load, and temperature conditions. See Feature Descriptions for additional information.

Parameter	Device	Symbol	Min	Тур	Max	Unit
On/Off Signal Interface						
$(V_{IN}=V_{IN, min} \text{ to } V_{IN, max}; \text{ open collector or equivalent,}$						
Signal referenced to GND)						
Device is with suffix "4" – Positive Logic (See Ordering Information)						
Logic High (Module ON)						
Input High Current	All	Iн			10	μΑ
Input High Voltage	All	VIH	3.5		$V_{IN,max}$	V
Logic Low (Module OFF)						
Input Low Current	All	lıL	_		1	mA
Input Low Voltage	All	VIL	-0.3		0.8	V
Device Code with no suffix – Negative Logic (See Ordering Information)						
(On/OFF pin is open collector/drain logic input with						
external pull-up resistor; signal referenced to GND)						
Logic High (Module OFF)						
Input High Current	All	Iн	_	_	1	mA
Input High Voltage	All	VIH	3.5	_	$V_{\text{IN, max}}$	Vdc
Logic Low (Module ON)						
Input low Current	All	Iı∟	_	_	10	μΑ
Input Low Voltage	All	VIL	-0.2	_	0.3	Vdc
Turn-On Delay and Rise Times						
( $V_{IN}=V_{IN, nom}$ , $I_O=I_{O, max}$ , $V_O$ to within ±1% of steady state)						
Case 1: On/Off input is enabled and then input power is applied (delay from instant at which $V_{IN}$ = $V_{IN,  min}$ until $V_0$ = 10% of $V_0$ , set)	All	Tdelay	_	2	_	msec
Case 2: Input power is applied for at least one second and then the On/Off input is enabled (delay from instant at which Von/Off is enabled until Vo = 10% of Vo, set)	All	Tdelay	_	2	_	msec
Output voltage Rise time (time for Vo to rise from 10% of Vo, set to 90% of Vo, set)	All	Trise	_	4	_	msec
Output voltage overshoot (T <sub>A</sub> = 25°C					3	% V <sub>O, set</sub>
$V_{IN}$ = $V_{IN, min}$ to $V_{IN, max}$ , $I_{O}$ = $I_{O, min}$ to $I_{O, max}$ )						
With or without maximum external capacitance						
Over Temperature Protection	All	$T_{ref}$		140		°C
(See Thermal Considerations section)						

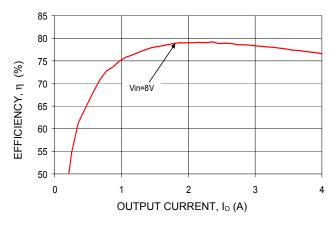
# **Feature Specifications (cont.)**

Parameter	Device	Symbol	Min	Тур	Max	Units
Input Undervoltage Lockout						
Turn-on Threshold	All				4.0	Vdc
Turn-off Threshold	All		3.6			Vdc
Hysteresis	All		0.4			Vdc
PGOOD (Power Good)						
Signal Interface Open Drain, V <sub>supply</sub> ≤ 5VDC						
Output Voltage Limit for PGOOD	All		90%		110%	$V_{\text{O, set}}$
Pulldown resistance of PGOOD pin	All			7	50	Ω

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 0.6Vo and at 25°C.

UUIPUI VULIAGE

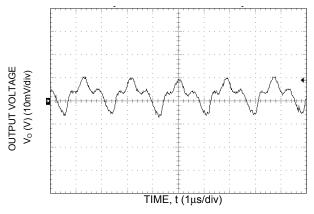
OUTPUT CURRENT,



4.5 OL LABRENT TEMPERATURE, TA °C

Figure 1. Converter Efficiency versus Output Current.

Figure 2. Derating Output Current versus Ambient Temperature and Airflow.



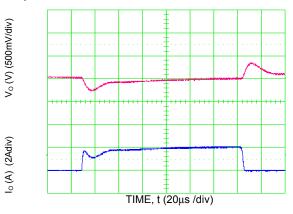
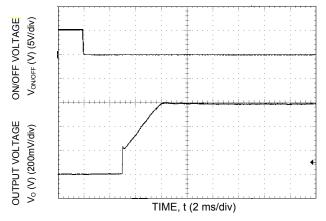


Figure 3. Typical output ripple and noise ( $V_{IN}$  = 8V,  $I_{o}$  =  $I_{o,max}$ ).

Figure 4. Transient Response to Dynamic Load Change from 0% to 50% to 0% .



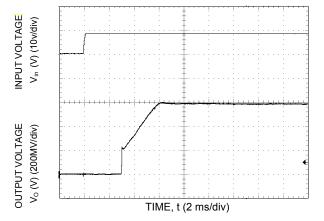
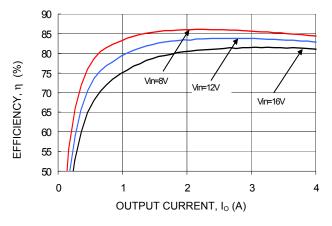


Figure 5. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ,  $V_{IN} = 8V$ )

Figure 6. Typical Start-up Using Input Voltage ( $V_{IN} = 8V$ ,  $I_0 = I_{0,max}$ ).

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 1.2Vo and at 25°C.



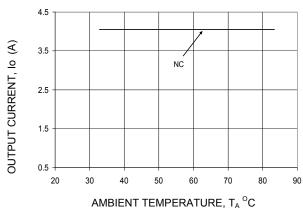
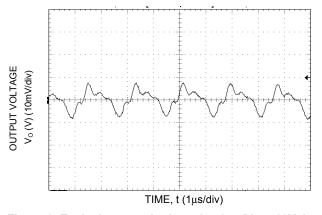


Figure 7. Converter Efficiency versus Output Current.

Figure 8. Derating Output Current versus Ambient Temperature and Airflow.



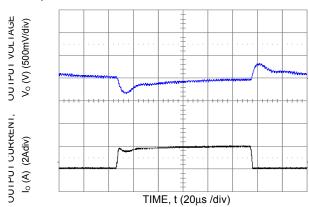
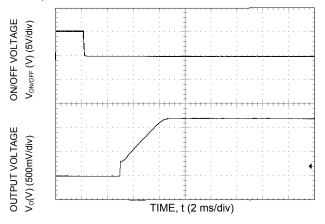


Figure 9. Typical output ripple and noise ( $V_{IN}$  = 12V,  $I_0$  =  $I_{0,max}$ ).

Figure 10. Transient Response to Dynamic Load Change from 0% to 50% to 0%.



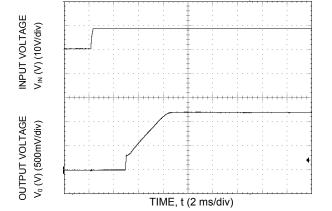
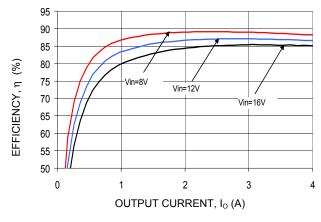


Figure 11. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ,  $V_{IN} = 8V$ ).

Figure 12. Typical Start-up Using Input Voltage ( $V_{IN} = 8V$ ,  $I_0 = I_{0,max}$ ).

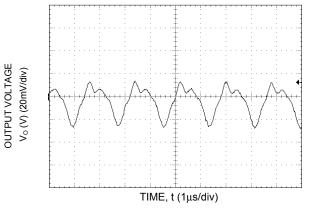
The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 1.8Vo and at 25°C.



AMBIENT TEMPERATURE, T<sub>A</sub> °C

Figure 13. Converter Efficiency versus Output Current.

Figure 14. Derating Output Current versus Ambient Temperature and Airflow.



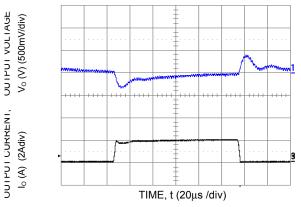
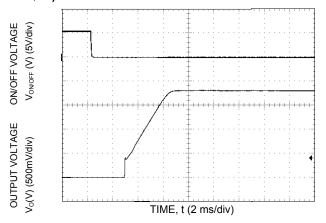


Figure 15. Typical output ripple and noise ( $V_{IN} = 12V$ ,  $I_{O} = I_{O,max}$ ).

Figure 16. Transient Response to Dynamic Load Change from 0% to 50% to 0%.



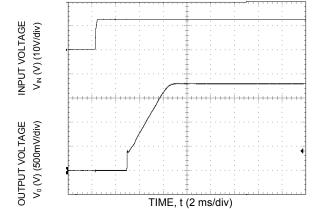


Figure 17. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ).

Figure 18. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 2.5Vo and at 25°C.

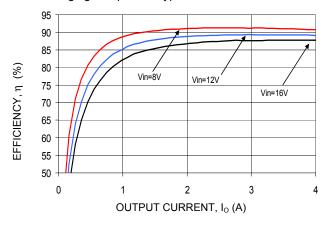
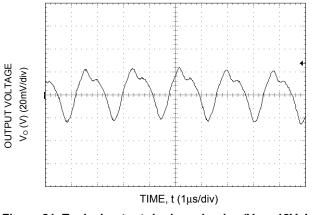


Figure 19. Converter Efficiency versus Output Current.

Figure 20. Derating Output Current versus Ambient Temperature and Airflow.



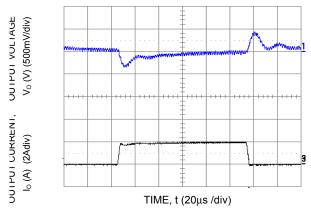
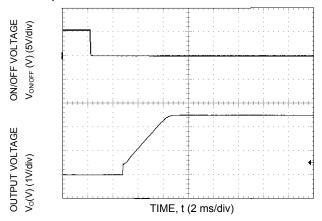


Figure 21. Typical output ripple and noise ( $V_{IN}$  = 12V,  $I_{o}$  =  $I_{o,max}$ ).

Figure 22. Transient Response to Dynamic Load Change from 0% to 50% to 0%.



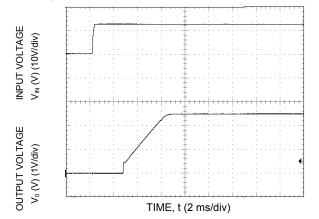


Figure 23. Typical Start-up Using On/Off Voltage (Io = Io.max).

Figure 24. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 3.3Vo and at 25°C.

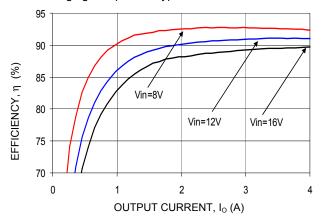
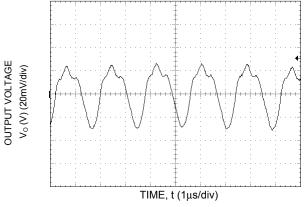


Figure 25. Converter Efficiency versus Output Current.

Figure 26. Derating Output Current versus Ambient Temperature and Airflow.



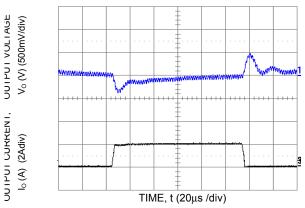
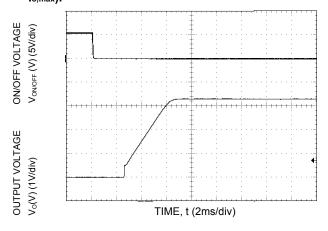


Figure 27. Typical output ripple and noise ( $V_{IN}$  = 12V,  $I_{O}$  =  $I_{O,max}$ ).

Figure 28. Transient Response to Dynamic Load Change from 0% 50% to 0%.



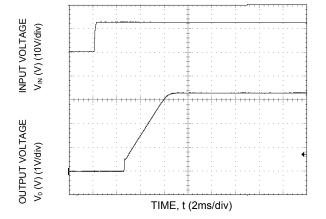


Figure 29. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ).

Figure 30. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 5.0 Vo and at 25°C.

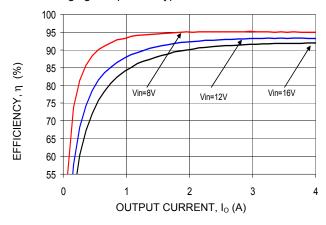
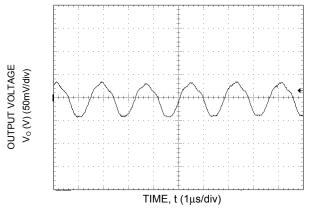


Figure 31. Converter Efficiency versus Output Current.

Figure 31. Derating Output Current versus Ambient Temperature and Airflow.



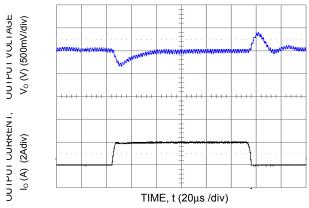
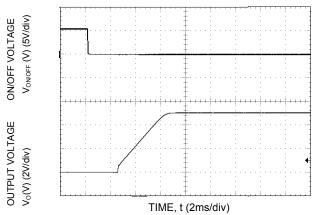


Figure 33. Typical output ripple and noise ( $V_{IN} = 12V$ ,  $I_{O} = I_{O,max}$ ).

Figure 34. Transient Response to Dynamic Load Change from 0% 50% to 0%.



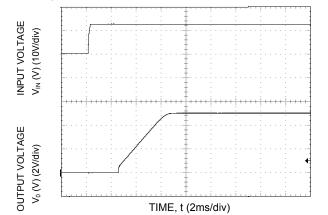
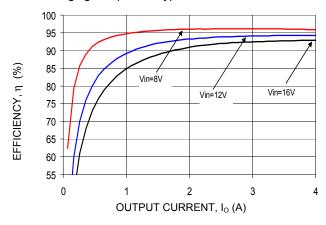


Figure 35. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ).

Figure 36. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

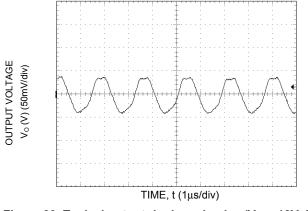
The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 6.5 Vo and at 25°C.



4.5 OUTPUT CURRENT, Io (A) 3.5 NC 2.5 1.5 0.5 60 70 20 30 40 50 80 90 AMBIENT TEMPERATURE, TA OC

Figure 37. Converter Efficiency versus Output Current.

Figure 38. Derating Output Current versus Ambient Temperature and Airflow.



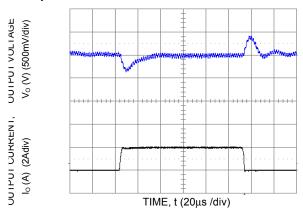
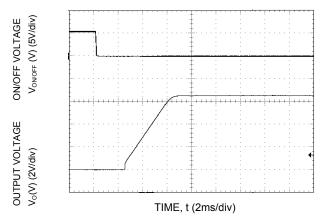


Figure 39. Typical output ripple and noise ( $V_{IN}$  = 12V,  $I_{o}$  =  $I_{o,max}$ ).

Figure 40. Transient Response to Dynamic Load Change from 0% 50% to 0%.



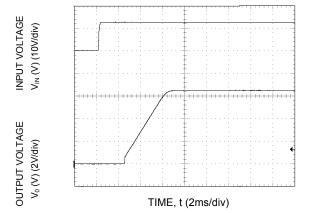


Figure 41. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ).

Figure 42. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

The following figures provide typical characteristics for the 16V Pico TLynx<sup>TM</sup> 4A at 8.0Vo and at 25°C.

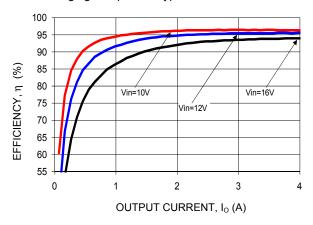
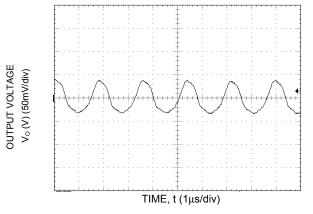


Figure 43. Converter Efficiency versus Output Current.

Figure 44. Derating Output Current versus Ambient Temperature and Airflow.



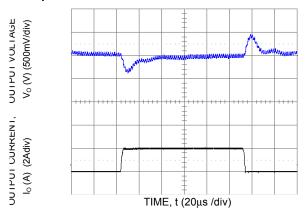
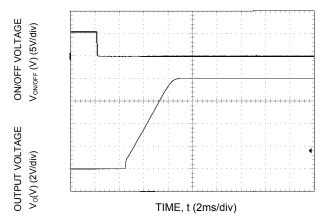


Figure 45. Typical output ripple and noise ( $V_{IN}$  = 12V,  $I_{O}$  =  $I_{O,max}$ ).

Figure 46. Transient Response to Dynamic Load Change from 0% 50% to 0%.



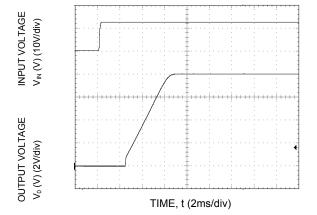
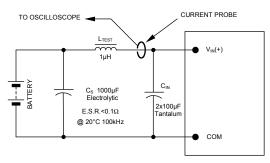


Figure 47. Typical Start-up Using On/Off Voltage ( $I_0 = I_{0,max}$ ).

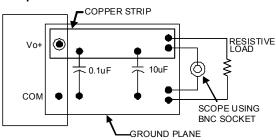
Figure 48. Typical Start-up Using Input Voltage ( $V_{IN} = 12V$ ,  $I_0 = I_{0,max}$ ).

## **Test Configurations**



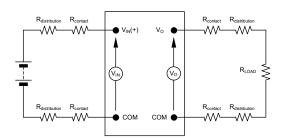
NOTE: Measure input reflected ripple current with a simulated source inductance ( $L_{TST}$ ) of 1µH. Capacitor C<sub>S</sub> offsets possible battery impedance. Measure current as shown above.

# Figure 49. Input Reflected Ripple Current Test Setup.



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

#### Figure 50. Output Ripple and Noise Test Setup



NOTE: All voltage measurements to be taken at the module terminals, as shown above. If sockets are used then Kelvin connections are required at the module terminals to avoid measurement errors due to socket contact resistance.

# Figure 51. Output Voltage and Efficiency Test Setup.

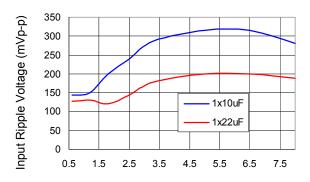
Efficiency 
$$\eta = \frac{V_0. I_0}{V_{IN}. I_{IN}} \times 100 \%$$

## **Design Considerations**

#### **Input Filtering**

The 16V Pico TLynx<sup>TM</sup> 4A module should be connected to a low ac-impedance source. A highly inductive source can affect the stability of the module. An input capacitance must be placed directly adjacent to the input pin of the module, to minimize input ripple voltage and ensure module stability.

To minimize input voltage ripple, ceramic capacitors are recommended at the input of the module. Figure 52 shows the input ripple voltage for various output voltages at 4A of load current with 1x10  $\mu F$  or 1x22  $\mu F$  ceramic capacitors and an input of 12V.



Output Voltage (Vdc)

Figure 52. Input ripple voltage for various output voltages with 1x10  $\mu$ F or 1x22  $\mu$ F ceramic capacitors at the input (4A load). Input voltage is 12V.

## **Output Filtering**

The 16V Pico TLynx<sup>TM</sup> 4A modules are designed for low output ripple voltage and will meet the maximum output ripple specification with 0.1  $\mu$ F ceramic and 10  $\mu$ F ceramic capacitors at the output of the module. However, additional output filtering may be required by the system designer for a number of reasons. First, there may be a need to further reduce the output ripple and noise of the module. Second, the dynamic response characteristics may need to be customized to a particular load step change.

To reduce the output ripple and improve the dynamic response to a step load change, additional capacitance at the output can be used. Low ESR polymer and ceramic capacitors are recommended to improve the dynamic response of the module. Figure 53 provides output ripple information for different external capacitance values at various Vo and for a load current of 4A. For stable operation of the module, limit the capacitance to less than the maximum output

capacitance as specified in the electrical specification table. Optimal performance of the module can be achieved by using the Tunable Loop<sup>TM</sup> feature described later in this data sheet.

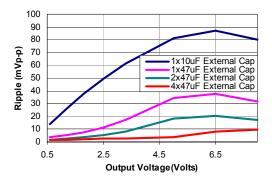


Figure 53. Output ripple voltage for various output voltages with external 1x10  $\mu$ F, 1x47  $\mu$ F, 2x47  $\mu$ F or 4x47  $\mu$ F ceramic capacitors at the output (4A load). Input voltage is 12V.

## **Safety Considerations**

For safety agency approval the power module must be installed in compliance with the spacing and separation requirements of the end-use safety agency standards, i.e., UL 60950-1, CSA C22.2 No. 60950-1-03, and VDE 0850:2001-12 (EN60950-1) Licensed.

For the converter output to be considered meeting the requirements of safety extra-low voltage (SELV), the input must meet SELV requirements. The power module has extra-low voltage (ELV) outputs when all inputs are ELV.

The input to these units is to be provided with a fastacting fuse with a maximum rating of 6A in the positive input lead.

## **Feature Descriptions**

#### **Remote Enable**

The 16V Pico TLynx<sup>TM</sup> 4A power modules feature an On/Off pin for remote On/Off operation. Two On/Off logic options are available. In the Positive Logic On/Off option, (device code suffix "4" – see Ordering Information), the module turns ON during a logic High on the On/Off pin and turns OFF during a logic Low. With the Negative Logic On/Off option, (no device code suffix, see Ordering Information), the module turns OFF during logic High and ON during logic Low. The On/Off signal is always referenced to ground. For either On/Off logic option, leaving the On/Off pin disconnected will turn the module ON when input voltage is present.

For positive logic modules, the circuit configuration for using the On/Off pin is shown in Figure 54.

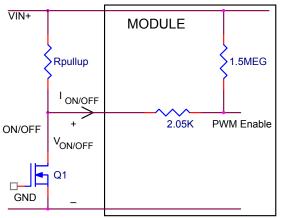


Figure 54. Circuit configuration for using positive On/Off logic.

For negative logic On/Off modules, the circuit configuration is shown in Fig. 55.

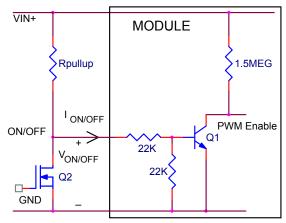


Figure 55. Circuit configuration for using negative On/Off logic.

#### **Overcurrent Protection**

To provide protection in a fault (output overload) condition, the unit is equipped with internal current-limiting circuitry and can endure current limiting continuously. At the point of current-limit inception, the unit enters hiccup mode. The unit operates normally once the output current is brought back into its specified range.

#### **Overtemperature Protection**

To provide protection in a fault condition, the unit is equipped with a thermal shutdown circuit. The unit will shutdown if the overtemperature threshold of  $140^{\circ}$ C is exceeded at the thermal reference point  $T_{ref}$ . The thermal shutdown is not intended as a guarantee that the unit will survive temperatures beyond its rating. Once the unit goes into thermal shutdown it will then wait to cool before attempting to restart.

#### **Input Undervoltage Lockout**

At input voltages below the input undervoltage lockout limit, the module operation is disabled. The module will begin to operate at an input voltage above the undervoltage lockout turn-on threshold.

#### **Output Voltage Programming**

The output voltage of the 16V Pico TLynx<sup>TM</sup> 4A modules can be programmed to any voltage from 0.59dc to 8.0Vdc by connecting a resistor between the Trim and GND pins of the module. Certain restrictions apply on the output voltage set point depending on the input voltage. These are shown in the Output Voltage vs. Input Voltage Set Point Area plot in Fig. 56.

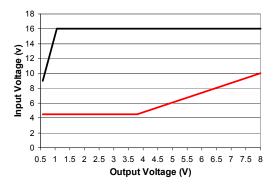


Figure 56. Output Voltage vs. Input Voltage Set Point Area plot showing limits where the output voltage can be set for different input voltages.

Without an external resistor between Trim and GND pins, the output of the module will be 0.59Vdc. To calculate the value of the trim resistor, *Rtrim* for a desired output voltage, use the following equation:

$$Rtrim = \left[ \frac{5.91}{(Vo - 0.591)} \right] k\Omega$$

Rtrim is the external resistor in  $k\Omega$ 

Vo is the desired output voltage.

Table 1 provides Rtrim values required for some common output voltages.

Table 1

V <sub>O, set</sub> (V)	Rtrim (KΩ)
0.6	656.7
1.0	14.45
1.2	9.704
1.5	6.502
1.8	4.888
2.5	3.096
3.3	2.182
5.0	1.340
6.5	1.000
8.0	0.798

By using a  $\pm 0.5\%$  tolerance trim resistor with a TC of  $\pm 100$ ppm, a set point tolerance of  $\pm 1.5\%$  can be achieved as specified in the electrical specification.

#### **Remote Sense**

The 16V Pico TLynx<sup>TM</sup> 4A power modules have a Remote Sense feature to minimize the effects of distribution losses by regulating the voltage at the SENSE pin. The voltage between the SENSE pin and VOUT pin must not exceed 0.5V. Note that the output voltage of the module cannot exceed the specified maximum value. This includes the voltage drop between the SENSE and Vout pins. When the Remote Sense feature is not being used, connect the SENSE pin to the VOUT pin.

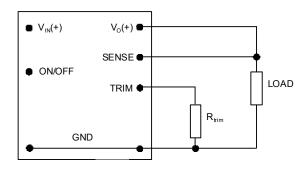


Figure 57. Circuit configuration for programming output voltage using an external resistor.

#### **Voltage Margining**

Output voltage margining can be implemented in the 16V Pico TLynx<sup>TM</sup> 4A modules by connecting a resistor, R<sub>margin-up</sub>, from the Trim pin to the ground pin for margining-up the output voltage and by connecting a resistor, R<sub>margin-down</sub>, from the Trim pin to output pin for margining-down. Figure 10 shows the circuit configuration for output voltage margining. The POL Programming Tool, available at www.lineagepower.com under the Design Tools section, also calculates the values of R<sub>margin-up</sub> and R<sub>margin-down</sub> for a specific output voltage and % margin. Please consult your local Lineage Power technical representative for additional details.

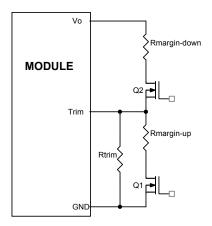


Figure 58. Circuit Configuration for margining Output voltage.

#### Monotonic Start-up and Shutdown

The 16V Pico TLynx<sup>TM</sup> 4A modules have monotonic start-up and shutdown behavior for any combination of rated input voltage, output current and operating temperature range.

#### Startup into Pre-biased Output

The 16V Pico TLynx<sup>TM</sup> 4A modules can start into a prebiased output as long as the prebias voltage is 0.5V less than the set output voltage. Note that prebias operation is not supported when output voltage sequencing is used.

#### **Power Good**

The 16V Pico TLynx<sup>TM</sup> 4A modules provide a Power Good (PGOOD) signal that is implemented with an open-drain output to indicate that the output voltage is within the regulation limits of the power module. The PGOOD signal will be de-asserted to a low state if any condition such as overtemperature, overcurrent or loss of regulation occurs that would result in the output voltage going ±10% outside the setpoint value. The PGOOD terminal should be connected through a

pullup resistor (suggested value 100K $\Omega$ ) to a source of 5VDC or lower.

## Tunable Loop<sup>™</sup>

The 16V Pico TLynx<sup>TM</sup> 4A modules have a new feature that optimizes transient response of the module called Tunable Loop<sup>TM</sup>.

External capacitors are usually added to the output of the module for two reasons: to reduce output ripple and noise (see Fig. 53) and to reduce output voltage deviations from the steady-state value in the presence of dynamic load current changes. Adding external capacitance however affects the voltage control loop of the module, typically causing the loop to slow down with sluggish response. Larger values of external capacitance could also cause the module to become unstable.

The Tunable Loop<sup>TM</sup> allows the user to externally adjust the voltage control loop to match the filter network connected to the output of the module. The Tunable Loop<sup>TM</sup> is implemented by connecting a series R-C between the SENSE and TRIM pins of the module, as shown in Fig. 59. This R-C allows the user to externally adjust the voltage loop feedback compensation of the module.

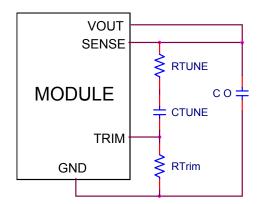


Figure. 59. Circuit diagram showing connection of  $R_{\text{TUME}}$  and  $C_{\text{TUNE}}$  to tune the control loop of the module.

Recommended values of  $R_{TUNE}$  and  $C_{TUNE}$  for different output capacitor combinations are given in Tables 2 and 3. Table 2 shows the recommended values of  $R_{TUNE}$  and  $C_{TUNE}$  for different values of ceramic output capacitors up to  $470\mu F$  that might be needed for an application to meet output ripple and noise requirements. Selecting  $R_{TUNE}$  and  $C_{TUNE}$  according to Table 2 will ensure stable operation of the module.

In applications with tight output voltage limits in the presence of dynamic current loading, additional output capacitance will be required. Table 3 lists recommended values of  $R_{\text{TUNE}}$  and  $C_{\text{TUNE}}$  in order to meet 2% output voltage deviation limits for some

common output voltages in the presence of a 2A to 4A step change (50% of full load), with an input voltage of 16V.

Please contact your Lineage Power technical representative to obtain more details of this feature as well as for guidelines on how to select the right value of external R-C to tune the module for best transient performance and stable operation for other output capacitance values or input voltages other than 16V.

Table 2. General recommended values of of  $R_{\text{TUNE}}$  and  $C_{\text{TUNE}}$  for Vin=12V and various external ceramic capacitor combinations.

#### Vo=6.5

Co	1x47μF	2x47μF	4x47μF	6x47μF	10x47μF
R <sub>TUNE</sub>	150	150	100	75	75
C <sub>TUNE</sub>	2700pF	5600pF	10nF	12nF	18nF

#### Vo=1.8

Со	1x47μF	2x47μF	4x47μF	6x47μF	10x47μF
R <sub>TUNE</sub>	100	75	75	75	75
C <sub>TUNE</sub>	6800pF	12nF	18nF	18nF	18nF

Table 3. Recommended values of  $R_{TUNE}$  and  $C_{TUNE}$  to obtain transient deviation of  $\leq$ 2% of Vout for a 2A step load with Vin=16V.

Vo	8V	6.5V	5V	3.3V	2.5V	1.8V
Со	1x47μF	2x47μF	2x47μF	2x47μF	3x47μF	4x47μF
R <sub>TUNE</sub>	150	150	100	100	75	75
C <sub>TUNE</sub>	2700pF	5600pF	6900pF	8200pF	12nF	18nF
ΔV	160mV	87mV	69mV	61mV	43mV	33mV

Vo	1.2V	0.6V
		1x47μF
	1 x	+
Co	330μF	2
	Polymer	x330μF
		Polymer
R <sub>TUNE</sub>	75	75
C <sub>TUNE</sub>	56nF	100nF
ΔV	21mV	11.4mV

#### **Thermal Considerations**

Power modules operate in a variety of thermal environments; however, sufficient cooling should always be provided to help ensure reliable operation.

Considerations include ambient temperature, airflow, module power dissipation, and the need for increased reliability. A reduction in the operating temperature of the module will result in an increase in reliability. The thermal data presented here is based on physical measurements taken in a wind tunnel. The test setup is shown in Figure 60. The preferred airflow direction for the module is in Figure 61.

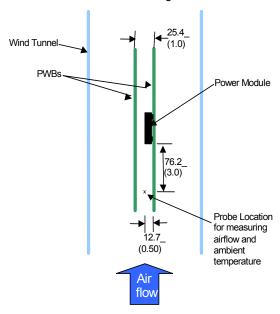


Figure 60. Thermal Test Setup.

The thermal reference points,  $T_{ref}$  used in the specifications are also shown in Figure 60. For reliable operation the temperatures at these points should not exceed  $120^{\circ}$ C. The output power of the module should not exceed the rated power of the module (Vo,set x lo,max).

Please refer to the Application Note "Thermal Characterization Process For Open-Frame Board-Mounted Power Modules" for a detailed discussion of thermal aspects including maximum device temperatures.

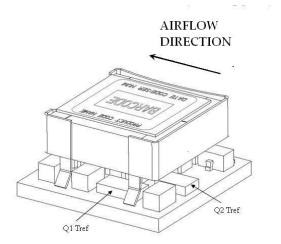


Figure 61. Preferred airflow direction and location of hot-spot of the module (Tref).

#### Shock and Vibration

The ruggedized (-D version) of the modules are designed to withstand elevated levels of shock and vibration to be able to operate in harsh environments. The ruggedized modules have been successfully tested to the following conditions:

#### Non operating random vibration:

Random vibration tests conducted at 25C, 10 to 2000Hz, for 30 minutes each level, starting from 30Grms (Z axis) and up to 50Grms (Z axis). The units were then subjected to two more tests of 50Grms at 30 minutes each for a total of 90 minutes.

#### Operating shock to 40G per Mil Std. 810F, Method 516.4 Procedure I:

The modules were tested in opposing directions along each of three orthogonal axes, with waveform and amplitude of the shock impulse characteristics as follows:

All shocks were half sine pulses, 11 milliseconds (ms) in duration in all 3 axes.

Units were tested to the Functional Shock Test of MIL-STD-810, Method 516.4, Procedure I - Figure 516.4-4. A shock magnitude of 40G was utilized. The operational units were subjected to three shocks in each direction along three axes for a total of eighteen shocks.

#### Operating vibration per Mil Std 810F, Method 514.5 Procedure I:

The ruggedized (-D version) modules are designed and tested to vibration levels as outlined in MIL-STD-810F, Method 514.5, and Procedure 1, using the Power Spectral Density (PSD) profiles as shown in Table 4 and Table 5 for all axes. Full compliance with performance specifications was required during the performance test. No damage was allowed to the module and full compliance to performance specifications was required when the endurance environment was removed. The module was tested per MIL-STD-810, Method 514.5, Procedure I, for functional (performance) and endurance random vibration using the performance and endurance levels shown in Table 4 and Table 5 for all axes. The performance test has been split, with one half accomplished before the endurance test and one half after the endurance test (in each axis). The duration of the performance test was at least 16 minutes total per axis and at least 120 minutes total per axis for the endurance test. The endurance test period was 2 hours minimum per axis.

**Table 4: Performance Vibration Qualification - All Axes** 

Frequency (Hz)	PSD Level (G2/Hz)	Frequency (Hz)	PSD Level (G2/Hz)	Frequency (Hz)	PSD Level (G2/Hz)
10	1.14E-03	170	2.54E-03	690	1.03E-03
30	5.96E-03	230	3.70E-03	800	7.29E-03
40	9.53E-04	290	7.99E-04	890	1.00E-03
50	2.08E-03	340	1.12E-02	1070	2.67E-03
90	2.08E-03	370	1.12E-02	1240	1.08E-03
110	7.05E-04	430	8.84E-04	1550	2.54E-03
130	5.00E-03	490	1.54E-03	1780	2.88E-03
140	8.20E-04	560	5.62E-04	2000	5.62E-04

Table 5: Endurance Vibration Qualification - All Axes

Frequency (Hz)	PSD Level (G2/Hz)	Frequency (Hz)	PSD Level (G2/Hz)	Frequency (Hz)	PSD Level (G2/Hz)
10	0.00803	170	0.01795	690	0.00727
30	0.04216	230	0.02616	800	0.05155
40	0.00674	290	0.00565	890	0.00709
50	0.01468	340	0.07901	1070	0.01887
90	0.01468	370	0.07901	1240	0.00764
110	0.00498	430	0.00625	1550	0.01795
130	0.03536	490	0.01086	1780	0.02035
140	0.0058	560	0.00398	2000	0.00398

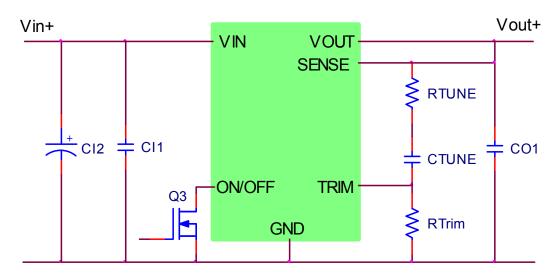
## **Example Application Circuit**

#### Requirements:

Vin: 16V Vout: 1.8V

lout: 3A max., worst case load transient is from 2A to 3A  $\Delta$ Vout: 1.5% of Vout (27mV) for worst case load transient

Vin, ripple 1.5% of Vin (240mV, p-p)



CI1 10µF/25V ceramic capacitor (e.g. Murata GRM31CR61E106KA12)

CI2 100µF/25V bulk electrolytic

CO1 4 x 47μF/10V ceramic capacitor (e.g. Murata GRM32ER61A476KE20)

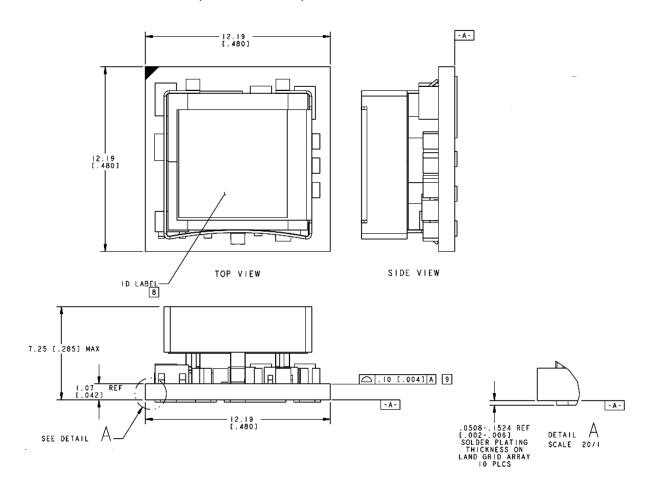
CTune 18nF/50V ceramic capacitor (can be 1206, 0805 or 0603 size)
RTune 75 ohms SMT resistor (can be 1206, 0805 or 0603 size)

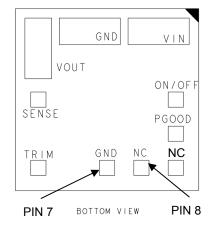
RTrim 4.87k $\Omega$  SMT resistor (can be 1206, 0805 or 0603 size, recommended tolerance of 0.1%)

## **Mechanical Outline**

Dimensions are in millimeters and (inches).

Tolerances: x.x mm  $\pm$  0.5 mm (x.xx in.  $\pm$  0.02 in.) [unless otherwise indicated] x.xx mm  $\pm$  0.25 mm (x.xxx in  $\pm$  0.010 in.)



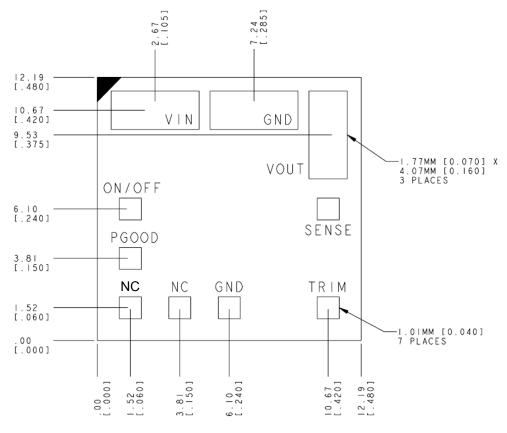


PIN	FUNCTION			
1	ON/OFF			
2	VIN			
3	GND			
4	VOUT			
5	SENSE			
6	TRIM			
7	GND			
8	NC			
9	NC			
10	PGOOD			

## **Recommended Pad Layout**

Dimensions are in millimeters and (inches)

Tolerances: x.x mm  $\pm$  0.5 mm (x.xx in.  $\pm$  0.02 in.) [unless otherwise indicated] x.xx mm  $\pm$  0.25 mm (x.xxx in  $\pm$  0.010 in.)



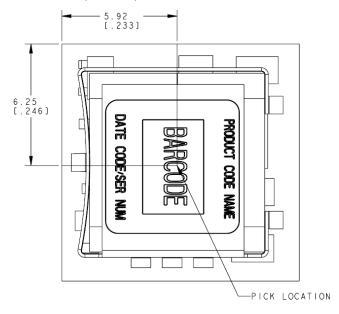
RECOMMENDED FOOTPRINT - THROUGH THE BOARD-

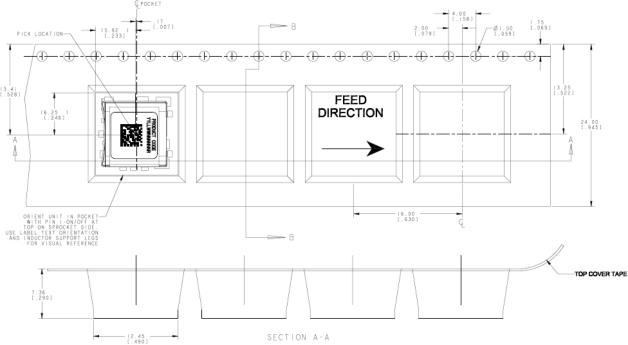
FUNCTION				
ON/OFF				
VIN				
GND				
VOUT				
SENSE				
TRIM				
GND				
NC				
NC				
PGOOD				

## **Packaging Details**

The 16V Pico TLynx $^{\text{TM}}$  4A modules are supplied in tape & reel as standard. Modules are shipped in quantities of 400 modules per reel.

All Dimensions are in millimeters and (in inches).





Reel Dimensions:

Outside Dimensions: 330.2 mm (13.00)
Inside Dimensions: 177.8 mm (7.00")
Tape Width: 24.00 mm (0.945")

## **Surface Mount Information**

#### Pick and Place

The 16V Pico TLynx<sup>TM</sup> 4A modules use an open frame construction and are designed for a fully automated assembly process. The modules are fitted with a label designed to provide a large surface area for pick and place operations. The label meets all the requirements for surface mount processing, as well as safety standards, and is able to withstand reflow temperatures of up to 300°C. The label also carries product information such as product code, serial number and the location of manufacture.

#### **Nozzle Recommendations**

The module weight has been kept to a minimum by using open frame construction. Variables such as nozzle size, tip style, vacuum pressure and placement speed should be considered to process. optimize this The minimum recommended inside nozzle diameter for reliable operation is 3mm. The maximum nozzle outer diameter, which will safely fit within the allowable component spacing, is 7 mm.

#### **Bottom Side / First Side Assembly**

This module is not recommended for assembly on the bottom side of a customer board. If such an assembly is attempted, components may fall off the module during the second reflow process. If assembly on the bottom side is planned, please contact Lineage Power for special manufacturing process instructions.

Only ruggedized (-D version) modules with additional epoxy will work with a customer's first side assembly. For other versions, first side assembly should be avoided

#### **Lead Free Soldering**

The 16V Pico TLynx<sup>TM</sup> 4A modules are lead-free (Pb-free) and RoHS compliant and fully compatible in a Pb-free soldering process. Failure to observe the instructions below may result in the failure of or cause damage to the modules and can adversely affect long-term reliability.

#### **Pb-free Reflow Profile**

Power Systems will comply with J-STD-020 Rev. C (Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices) for both Pb-free solder profiles and MSL classification procedures. This standard provides a recommended forced-air-convection reflow profile based on the volume and thickness of the package (table 4-2). The suggested Pb-free Sn/Ag/Cu solder paste is (SAC). The recommended linear reflow profile

Sn/Ag/Cu solder is shown in Fig. 62. Soldering outside of the recommended profile requires testing to verify results and performance. For questions regarding Land grid array(LGA) soldering, solder volume; please contact Lineage Power for special manufacturing process instructions

#### **MSL** Rating

The 16V Pico TLynx<sup>TM</sup> 4A modules have a MSL rating of 2.

## Storage and Handling

The recommended storage environment and handling procedures for moisture-sensitive surface mount packages is detailed in J-STD-033 Rev. A (Handling, Packing, Shipping and Use of Moisture/Reflow Sensitive Surface Mount Devices). Moisture barrier bags (MBB) with desiccant are required for MSL ratings of 2 or greater. These sealed packages should not be broken until time of use. Once the original package is broken, the floor life of the product at conditions of ≤ 30°C and 60% relative humidity varies according to the MSL rating (see J-STD-033A). The shelf life for dry packed SMT packages will be a minimum of 12 months from the bag seal date, when stored at the following conditions: < 40° C, < 90% relative humidity.

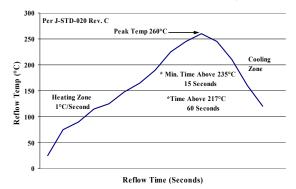


Figure 62. Recommended linear reflow profile using Sn/Ag/Cu solder.

#### **Post Solder Cleaning and Drying** Considerations

Post solder cleaning is usually the final circuit-board assembly process prior to electrical board testing. The result of inadequate cleaning and drying can affect both the reliability of a power module and the testability of the finished circuit-board assembly. For quidance on appropriate soldering, cleaning and drying procedures, refer to Board Mounted Power Modules: Soldering and Cleaning Application Note (AN04-001).

**LINEAGE POWER** 25

Document No: DS08-006 ver. 1.08 PDF name: APXK004A0X.pdf

## **Ordering Information**

Please contact your Lineage Power Sales Representative for pricing, availability and optional features.

#### **Table 6. Device Codes**

Device Code Input Voltage Range		Output Voltage	Output Current	On/Off Logic	Sequencing	Comcodes
APXK004A0X-SRZ	8.0 – 16Vdc	0.59 – 8.0Vdc	4A	Negative	No	CC109146651
APXK004A0X4-SRZ	8.0 – 16Vdc	0.59 - 8.0Vdc	4A	Positive	No	CC109146643
APXK004A0X-SRDZ	8.0 – 16Vdc	0.59 - 8.0Vdc	4A	Negative	No	CC109158804

#### **Table 7. Coding Scheme**

TLynx family	Sequencing feature.	Input voltage range	Output current	=	On/Off logic	Options		ROHS Compliance
AP	Х	K	004A0	X	4	-SR	-D	Z
	X = w/o Seq.	K = 8.0 - 16V	4.0A	X = programmable output	4 = positive No entry = negative	S = Surface Mount R = Tape&Reel	D = 105C operating ambient, 40G operating shock as per MIL Std 810F	Z = ROHS6



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LINEAGE POWER 26

Document No: DS08-006 ver. 1.08 PDF name: APXK004A0X.pdf



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- Подбор аналогов;
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- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



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