## Single Channel, 128-/256-Position, $I^{2}$ C/SPI, Nonvolatile Digital Potentiometer

## Data Sheet

## FEATURES

## $10 \mathrm{k} \Omega$ and $100 \mathrm{k} \Omega$ resistance options

Resistor tolerance: 8\% maximum
Wiper current: $\pm 6 \mathrm{~mA}$
Low temperature coefficient: $35 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$
Wide bandwidth: 3 MHz
Fast start-up time $<75 \mu \mathrm{~s}$
Linear gain setting mode
Single- and dual-supply operation
Independent logic supply: 1.8 V to 5.5 V
Wide operating temperature: $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
$3 \mathrm{~mm} \times 3 \mathrm{~mm}$ LFCSP package
4 kV ESD protection

## APPLICATIONS

Portable electronics level adjustment
LCD panel brightness and contrast controls
Programmable filters, delays, and time constants
Programmable power supplies

## GENERAL DESCRIPTION

The AD5121/AD5141 potentiometers provide a nonvolatile solution for 128-/256-position adjustment applications, offering guaranteed low resistor tolerance errors of $\pm 8 \%$ and up to $\pm 6 \mathrm{~mA}$ current density in the $\mathrm{A}, \mathrm{B}$, and W pins.

The low resistor tolerance and low nominal temperature coefficient simplify open-loop applications as well as applications requiring tolerance matching.

The linear gain setting mode allows independent programming of the resistance between the digital potentiometer terminals, through $R_{A W}$ and $R_{w B}$ string resistors, allowing very accurate resistor matching.

The high bandwidth and low total harmonic distortion (THD) ensure optimal performance for ac signals, making it suitable for filter design.

The low wiper resistance of only $40 \Omega$ at the ends of the resistor array allows for pin-to-pin connection.

The wiper values can be set through an SPI-/I ${ }^{2} \mathrm{C}$-compatible digital interface that is also used to read back the wiper register and EEPROM contents.

The AD5121/AD5141 is available in a compact, 16 -lead, $3 \mathrm{~mm} \times$ 3 mm LFCSP. The parts are guaranteed to operate over the extended industrial temperature range of $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$.

## AD5121/AD5141

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## AD5121/AD5141

## SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS—AD5121

$\mathrm{V}_{\mathrm{DD}}=2.3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{DD}}=2.25 \mathrm{~V}$ to $2.75 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=-2.25 \mathrm{~V}$ to $-2.75 \mathrm{~V} ; \mathrm{V}_{\text {LOGIC }}=1.8 \mathrm{~V}$ to $5.5 \mathrm{~V},-40^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{A}}<+125^{\circ} \mathrm{C}$, unless otherwise noted.

Table 2.



AD5121/AD5141

| Parameter | Symbol | Test Conditions/Comments | Min | Typ ${ }^{1}$ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DYNAMIC CHARACTERISTICS ${ }^{9}$ |  |  |  |  |  |  |
| Bandwidth | BW | -3 dB |  |  |  |  |
|  |  | $\mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ |  | 3 |  | MHz |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 0.43 |  | MHz |
| Total Harmonic Distortion | THD | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{SS}}= \pm 2.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{A}}=1 \mathrm{Vrms}, \\ & \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz} \end{aligned}$ |  |  |  |  |
|  |  | $\mathrm{R}_{A B}=10 \mathrm{k} \Omega$ |  | -80 |  | dB |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | -90 |  | dB |
| Resistor Noise Density | $\mathrm{e}_{\text {N_wb }}$ | $\begin{aligned} & \text { Code }=\text { half scale, } T_{A}=25^{\circ} \mathrm{C}, \\ & \mathrm{f}=10 \mathrm{kHz} \end{aligned}$ |  |  |  |  |
|  |  | $\mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ |  | 7 |  | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 20 |  | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
| Vw Settling Time | $\mathrm{ts}_{5}$ | $\mathrm{V}_{\mathrm{A}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}$, from zero scale to full scale, $\pm 0.5$ LSB error band |  |  |  |  |
|  |  | $\mathrm{R}_{\text {AB }}=10 \mathrm{k} \Omega$ |  | 2 |  | $\mu \mathrm{s}$ |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 12 |  |  |
| Endurance ${ }^{10}$ |  | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 1 |  | Mcycles |
|  |  |  | 100 |  |  | kcycles |
| Data Retention ${ }^{11}$ |  |  |  | 50 |  | Years |

${ }^{1}$ Typical values represent average readings at $25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{SS}}=0 \mathrm{~V}$, and $\mathrm{V}_{\text {LOGIC }}=5 \mathrm{~V}$.
${ }^{2}$ Resistor integral nonlinearity ( $\mathrm{R}-\mathrm{INL}$ ) error is the deviation from an ideal value measured between the maximum resistance and the minimum resistance wiper positions. R-DNL measures the relative step change from ideal between successive tap positions. The maximum wiper current is limited to $\left(0.7 \times V_{D D}\right) / R_{A B}$.
${ }^{3}$ Guaranteed by design and characterization, not subject to production test.
${ }^{4}$ INL and DNL are measured at $V_{\text {WB }}$ with the RDAC configured as a potentiometer divider similar to a voltage output $D A C . V_{A}=V_{D D}$ and $V_{B}=0 \mathrm{~V}$. DNL specification limits of $\pm 1$ LSB maximum are guaranteed monotonic operating conditions.
${ }^{5}$ Resistor Terminal A, Resistor Terminal B, and Resistor Terminal W have no limitations on polarity with respect to each other. Dual-supply operation enables ground referenced bipolar signal adjustment.
${ }^{6}$ Different from operating current; supply current for EEPROM program lasts approximately 30 ms .
${ }^{7}$ Different from operating current; supply current for EEPROM read lasts approximately $20 \mu \mathrm{~s}$.
${ }^{8} \mathrm{P}_{\text {DISS }}$ is calculated from ( $\left.\mathrm{I}_{\text {DD }} \times \mathrm{V}_{\text {DD }}\right)+\left(\mathrm{I}_{\text {LOGIC }} \times \mathrm{V}_{\text {LOGIC }}\right)$.
${ }^{9}$ All dynamic characteristics use $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{S S}= \pm 2.5 \mathrm{~V}$, and $\mathrm{V}_{\text {LOGIC }}=2.5 \mathrm{~V}$.
${ }^{10}$ Endurance is qualified to 100,000 cycles per JEDEC Standard 22 , Method A117 and measured at $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$.
${ }^{11}$ Retention lifetime equivalent at junction temperature $\left(\mathrm{T}_{\mathrm{J}}\right)=125^{\circ} \mathrm{C}$ per JEDEC Standard 22 , Method A117. Retention lifetime, based on an activation energy of 1 eV , derates with junction temperature in the Flash/EE memory.

## AD5121/AD5141

## ELECTRICAL CHARACTERISTICS—AD5141

$\mathrm{V}_{\mathrm{DD}}=2.3 \mathrm{~V}$ to 5.5 V , $\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{DD}}=2.25 \mathrm{~V}$ to 2.75 V, $\mathrm{V}_{\mathrm{SS}}=-2.25 \mathrm{~V}$ to -2.75 V ; $\mathrm{V}_{\text {Logic }}=1.8 \mathrm{~V}$ to $5.5 \mathrm{~V},-40^{\circ} \mathrm{C}<\mathrm{T}_{\mathrm{A}}<+125^{\circ} \mathrm{C}$, unless otherwise noted.

Table 3.

| Parameter | Symbol | Test Conditions/Comments | Min | Typ ${ }^{1}$ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DC CHARACTERISTICS—RHEOSTAT MODE (ALL RDACs) |  |  |  |  |  |  |
| Resolution | N R-INL |  | 8 |  |  | Bits |
| Resistor Integral Nonlinearity ${ }^{2}$ |  | $\mathrm{R}_{\text {AB }}=10 \mathrm{k} \Omega$ |  |  |  |  |
|  |  | $V_{D D} \geq 2.7 \mathrm{~V}$ | -2 | $\pm 0.2$ | +2 | $\begin{aligned} & \text { LSB } \\ & \text { LSB } \end{aligned}$ |
|  |  | $\mathrm{V}_{\mathrm{DD}}<2.7 \mathrm{~V}$ | -5 | $\pm 1.5$ | +5 |  |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  |  |  |  |
|  |  | $V_{D D} \geq 2.7 \mathrm{~V}$ | -1 | $\pm 0.1$ | +1 | $\begin{aligned} & \text { LSB } \\ & \text { LSB } \end{aligned}$ |
|  |  | $\mathrm{V}_{\mathrm{DD}}<2.7 \mathrm{~V}$ | $\begin{aligned} & -2 \\ & -0.5 \end{aligned}$ | $\pm 0.5$ | +2 |  |
| Resistor Differential Nonlinearity ${ }^{2}$ | R-DNL |  |  | $\pm 0.2$ | +0.5 | $\begin{aligned} & \text { LSB } \\ & \text { LSB } \end{aligned}$ |
| Nominal Resistor Tolerance | $\Delta \mathrm{R}_{\mathrm{AB}} / \mathrm{R}_{\mathrm{AB}}$$\left(\Delta \mathrm{R}_{A B} / \mathrm{R}_{A B}\right) / \Delta \mathrm{T} \times 10^{6}$ |  | -8 | $\begin{aligned} & \pm 1 \\ & 35 \end{aligned}$ | +8 |  |
| Resistance Temperature Coefficient ${ }^{3}$ |  | Code = full scale |  |  |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| Wiper Resistance ${ }^{3}$ | Rw | Code $=$ zero scale |  |  |  |  |
|  |  | $\mathrm{R}_{\text {AB }}=10 \mathrm{k} \Omega$ |  | 55 | 125 | $\Omega$ |
|  |  | $\mathrm{R}_{\mathrm{AB}}=100 \mathrm{k} \Omega$ |  | 130 | 400 | $\Omega$ |
| Bottom Scale or Top Scale | $\mathrm{R}_{\text {BS }}$ or $\mathrm{R}_{\text {TS }}$ |  |  |  |  |  |
|  |  | $\mathrm{R}_{\text {AB }}=10 \mathrm{k} \Omega$ |  | 40 | 80 | $\Omega$ |
|  |  | $\mathrm{R}_{\mathrm{AB}}=100 \mathrm{k} \Omega$ |  | 60 | 230 |  |
| DC CHARACTERISTICS—POTENTIOMETER DIVIDER MODE (ALL RDACs) |  |  |  |  |  |  |
| Integral Nonlinearity ${ }^{4}$ | INL |  |  |  |  |  |
|  |  | $\begin{aligned} & \mathrm{R}_{A B}=10 \mathrm{k} \Omega \\ & \mathrm{R}_{A B}=100 \mathrm{k} \Omega \end{aligned}$ | -1 | $\pm 0.2$ | +1 | LSB |
|  |  |  | -0.5 | $\pm 0.1$ | +0.5 | LSB |
| Differential Nonlinearity ${ }^{4}$ | DNL |  | -0.5 | $\pm 0.2$ | +0.5 | LSB |
| Full-Scale Error | $V_{\text {WFSE }}$ |  |  |  |  |  |
|  |  | $\mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ | -2.5-1 | -0.1 |  | $\begin{aligned} & \text { LSB } \\ & \text { LSB } \end{aligned}$ |
|  |  | $\mathrm{R}_{\mathrm{AB}}=100 \mathrm{k} \Omega$ |  | $\pm 0.2$ | +1 |  |
| Zero-Scale Error | $\mathrm{V}_{\text {wZSE }}$ |  |  |  |  |  |
|  |  | $\mathrm{R}_{\text {AB }}=10 \mathrm{k} \Omega$ |  | 1.2 | 3 | LSB |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 0.5 | 1 |  |
| Voltage Divider Temperature Coefficient ${ }^{3}$ | $\left(\Delta \mathrm{V}_{\mathrm{w}} / \mathrm{V}_{\mathrm{w}}\right) / \Delta \mathrm{T} \times 10^{6}$ | Code $=$ half scale |  | $\pm 5$ |  | ppm/ ${ }^{\circ} \mathrm{C}$ |



| Parameter | Symbol | Test Conditions/Comments | Min | Typ ${ }^{1}$ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DYNAMIC CHARACTERISTICS ${ }^{9}$ |  |  |  |  |  |  |
| Bandwidth | BW | -3 dB |  |  |  |  |
|  |  | $\mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ |  | 3 |  | MHz |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 0.43 |  | MHz |
| Total Harmonic Distortion | THD | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{SS}}= \pm 2.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{A}}=1 \mathrm{Vrms}, \\ & \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}, \mathrm{f}=1 \mathrm{kHz} \end{aligned}$ |  |  |  |  |
|  |  | $\mathrm{R}_{A B}=10 \mathrm{k} \Omega$ |  | -80 |  | dB |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | -90 |  | dB |
| Resistor Noise Density | $\mathrm{e}_{\text {__wb }}$ | $\begin{aligned} & \text { Code }=\text { half scale, } \mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \\ & \mathrm{f}=10 \mathrm{kHz} \end{aligned}$ |  |  |  |  |
|  |  | $\mathrm{R}_{\mathrm{AB}}=10 \mathrm{k} \Omega$ |  | 7 |  | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 20 |  | $\mathrm{nV} / \sqrt{ } \mathrm{Hz}$ |
| Vw Settling Time | $\mathrm{ts}_{5}$ | $\mathrm{V}_{\mathrm{A}}=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{B}}=0 \mathrm{~V}$, from zero scale to full scale, $\pm 0.5$ LSB error band |  |  |  |  |
|  |  | $\mathrm{R}_{A B}=10 \mathrm{k} \Omega$ |  | 2 |  | $\mu \mathrm{s}$ |
|  |  | $\mathrm{R}_{\text {AB }}=100 \mathrm{k} \Omega$ |  | 12 |  |  |
| Endurance ${ }^{10}$ |  | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 1 |  | Mcycles |
|  |  |  | 100 |  |  | kcycles |
| Data Retention ${ }^{11}$ |  |  |  | 50 |  | Years |

${ }^{1}$ Typical values represent average readings at $25^{\circ} \mathrm{C}, \mathrm{V}_{D D}=5 \mathrm{~V}, \mathrm{~V}_{S S}=0 \mathrm{~V}$, and $\mathrm{V}_{\text {LOGIC }}=5 \mathrm{~V}$.
${ }^{2}$ Resistor integral nonlinearity error ( $\mathrm{R}-\mathrm{INL}$ ) is the deviation from an ideal value measured between the maximum resistance and the minimum resistance wiper positions. R-DNL measures the relative step change from ideal between successive tap positions. The maximum wiper current is limited to $\left(0.7 \times V_{D D}\right) / R_{A B}$.
${ }^{3}$ Guaranteed by design and characterization, not subject to production test.
${ }^{4}$ INL and DNL are measured at $V_{\text {WB }}$ with the RDAC configured as a potentiometer divider similar to a voltage output DAC. $V_{A}=V_{D D}$ and $V_{B}=0 V$. DNL specification limits of $\pm 1$ LSB maximum are guaranteed monotonic operating conditions.
${ }^{5}$ Resistor Terminal A, Resistor Terminal B, and Resistor Terminal W have no limitations on polarity with respect to each other. Dual-supply operation enables ground referenced bipolar signal adjustment.
${ }^{6}$ Different from operating current; supply current for EEPROM program lasts approximately 30 ms .
${ }^{7}$ Different from operating current; supply current for EEPROM read lasts approximately $20 \mu \mathrm{~s}$.
${ }^{8} \mathrm{P}_{\text {DISS }}$ is calculated from ( $\left.\mathrm{I}_{\text {DD }} \times \mathrm{V}_{\text {DD }}\right)+\left(\mathrm{I}_{\text {LOGIC }} \times \mathrm{V}_{\text {LOGIC }}\right)$.
${ }^{9}$ All dynamic characteristics use $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\mathrm{SS}}= \pm 2.5 \mathrm{~V}$, and $\mathrm{V}_{\text {LOGIC }}=2.5 \mathrm{~V}$.
${ }^{10}$ Endurance is qualified to 100,000 cycles per JEDEC Standard 22 , Method A117 and measured at $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$.
${ }^{11}$ Retention lifetime equivalent at junction temperature $\left(\mathrm{T}_{\mathrm{J}}\right)=125^{\circ} \mathrm{C}$ per JEDEC Standard 22, Method A117. Retention lifetime, based on an activation energy of 1 eV, derates with junction temperature in the Flash/EE memory.

## INTERFACE TIMING SPECIFICATIONS

$\mathrm{V}_{\text {LOGIC }}=1.8 \mathrm{~V}$ to 5.5 V ; all specifications $\mathrm{T}_{\text {MIN }}$ to $\mathrm{T}_{\text {MAX }}$, unless otherwise noted.
Table 4. SPI Interface

| Parameter ${ }^{1}$ | Test Conditions/Comments | Min | Typ | Max | Unit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{1}$ | $\mathrm{V}_{\text {LoGIC }}>1.8 \mathrm{~V}$ | 20 |  |  | ns | SCLK cycle time |
|  | $V_{\text {LoGic }}=1.8 \mathrm{~V}$ | 30 |  |  | ns |  |
| $\mathrm{t}_{2}$ | $\mathrm{V}_{\text {LoGII }}>1.8 \mathrm{~V}$ | 10 |  |  | ns | SCLK high time |
|  | $\mathrm{V}_{\text {LoGII }}=1.8 \mathrm{~V}$ | 15 |  |  | ns |  |
| $\mathrm{t}_{3}$ | $\mathrm{V}_{\text {LoGII }}>1.8 \mathrm{~V}$ | 10 |  |  | ns | SCLK low time |
|  | $\mathrm{V}_{\text {LoGII }}=1.8 \mathrm{~V}$ | 15 |  |  | ns |  |
| $\mathrm{t}_{4}$ |  | 10 |  |  | ns | $\overline{\text { SYNC-to-SCLK falling edge setup time }}$ |
| $\mathrm{t}_{5}$ |  | 5 |  |  | ns | Data setup time |
| $\mathrm{t}_{6}$ |  | 5 |  |  | ns | Data hold time |
| $\mathrm{t}_{7}$ |  | 10 |  |  | ns | $\overline{\text { SYNC }}$ rising edge to next SCLK fall ignored |
| $\mathrm{t}_{8}{ }^{2}$ |  | 20 |  |  | ns | Minimum $\overline{\text { SYNC }}$ high time |
| $\mathrm{t}_{9}{ }^{3}$ |  |  | 50 |  | ns | SCLK rising edge to SDO valid |
| $\mathrm{t}_{10}$ |  |  |  | 500 | ns | $\overline{\text { SYNC }}$ rising edge to SDO pin disable |

${ }^{1}$ All input signals are specified with $\mathrm{t}_{\mathrm{r}}=\mathrm{t}_{\mathrm{f}}=1 \mathrm{~ns} / \mathrm{V}\left(10 \%\right.$ to $90 \%$ of $\left.\mathrm{V}_{\mathrm{DD}}\right)$ and timed from a voltage level of $\left(\mathrm{V}_{\mathrm{IL}}+\mathrm{V}_{\mathbb{H}}\right) / 2$.
${ }^{2}$ Refer to $\mathrm{t}_{\text {EEPROM_PROGRAM }}$ and $\mathrm{t}_{\text {EEPROM_READBACK }}$ for memory commands operations (see Table 6).
${ }^{3}$ RPULL_UP $=2.2 \mathrm{k} \Omega$ to $\mathrm{V}_{\text {DD }}$ with a capacitance load of 168 pF .

Table 5. $\mathrm{I}^{2} \mathrm{C}$ Interface

| Parameter ${ }^{1}$ | Test Conditions/Comments | Min | Typ | Max | Unit | Description |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{fscL}^{2}$ | Standard mode |  |  | 100 | kHz | Serial clock frequency |
|  | Fast mode |  |  | 400 | kHz |  |
| $\mathrm{t}_{1}$ | Standard mode | 4.0 |  |  | $\mu \mathrm{s}$ | SCL high time, $\mathrm{t}_{\text {HIGH }}$ |
|  | Fast mode | 0.6 |  |  | $\mu \mathrm{s}$ |  |
| $\mathrm{t}_{2}$ | Standard mode | 4.7 |  |  | $\mu \mathrm{s}$ | SCL low time, tıow |
|  | Fast mode | 1.3 |  |  | $\mu \mathrm{s}$ |  |
| $t_{3}$ | Standard mode | 250 |  |  | ns | Data setup time, tsu; DAT |
|  | Fast mode | 100 |  |  | ns |  |
| $\mathrm{t}_{4}$ | Standard mode | 0 |  | 3.45 | $\mu \mathrm{s}$ | Data hold time, $\mathrm{thD;}^{\text {dat }}$ |
|  | Fast mode | 0 |  | 0.9 | $\mu \mathrm{s}$ |  |
| $\mathrm{t}_{5}$ | Standard mode | 4.7 |  |  | $\mu \mathrm{s}$ | Setup time for a repeated start condition, $\mathrm{tsu}_{\text {; STA }}$ |
|  | Fast mode | 0.6 |  |  | $\mu \mathrm{s}$ |  |
| $t_{6}$ | Standard mode | 4 |  |  | $\mu \mathrm{s}$ | Hold time (repeated) for a start condition, $\mathrm{thD}_{\text {; STA }}$ |
|  | Fast mode | 0.6 |  |  | $\mu \mathrm{s}$ |  |
| $\mathrm{t}_{7}$ | Standard mode | 4.7 |  |  | $\mu \mathrm{s}$ | Bus free time between a stop and a start condition, $\mathrm{t}_{\text {BuF }}$ |
|  | Fast mode | 1.3 |  |  | $\mu \mathrm{s}$ |  |
| $\mathrm{t}_{8}$ | Standard mode | 4 |  |  | $\mu \mathrm{s}$ | Setup time for a stop condition, tsu; sto |
|  | Fast mode | 0.6 |  |  | $\mu \mathrm{s}$ |  |
| $\mathrm{t}_{9}$ | Standard mode |  |  | 1000 | ns | Rise time of SDA signal, trda |
|  | Fast mode | $20+0.1 \mathrm{CL}$ |  | 300 | ns |  |
| $\mathrm{t}_{10}$ | Standard mode |  |  | 300 | ns | Fall time of SDA signal, trdA |
|  | Fast mode | $20+0.1 C_{L}$ |  | 300 | ns |  |
| $\mathrm{t}_{11}$ | Standard mode |  |  | 1000 | ns | Rise time of SCL signal, $\mathrm{t}_{\text {RCL }}$ |
|  | Fast mode | $20+0.1 \mathrm{CL}$ |  | 300 | ns |  |
| $\mathrm{t}_{11 \mathrm{~A}}$ | Standard mode |  |  | 1000 | ns | Rise time of SCL signal after a repeated start condition and after an acknowledge bit, trCL1 (not shown in Figure 3) |
|  | Fast mode | $20+0.1 \mathrm{CL}$ |  | 300 | ns |  |


| Parameter $^{1}$ | Test Conditions/Comments | Min | Typ | Max | Unit | Description |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{12}$ | Standard mode |  | 300 | ns | Fall time of SCL signal, $\mathrm{t} \subset \mathrm{CL}$ |  |
|  | Fast mode | $20+0.1 \mathrm{C}_{\mathrm{L}}$ |  | 300 | ns |  |
| $\mathrm{t}_{\text {SP }}{ }^{3}$ | Fast mode | 0 | 50 | ns | Pulse width of suppressed spike (not shown in Figure 3) |  |

${ }^{1}$ Maximum bus capacitance is limited to 400 pF .
${ }^{2}$ The SDA and SCL timing is measured with the input filters enabled. Switching off the input filters improves the transfer rate; however, it has a negative effect on the EMC behavior of the part.
${ }^{3}$ Input filtering on the SCL and SDA inputs suppresses noise spikes that are less than 50 ns for fast mode.

Table 6. Control Pins

| Parameter | Min | Typ | Max | Unit | Description |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{t}_{1}$ | 1 |  |  | $\mu \mathrm{~s}$ | End command to $\overline{\text { LRDAC falling edge }}$ |
| $\mathrm{t}_{2}$ | 50 |  |  | ns | Minimum $\overline{\text { LRDAC low time }}$ |
| $\mathrm{t}_{3}$ | 0.1 |  | 10 | $\mu \mathrm{~s}$ | $\overline{\text { RESET low time }}$ |
| $\mathrm{t}_{\text {EEPROM_PROGRAM }}{ }^{1}$ |  | 15 | 50 | ms | Memory program time (not shown in Figure 6) |
| $\mathrm{t}_{\text {EEPROM_READBACK }}$ |  | 7 | 30 | $\mu \mathrm{~s}$ | Memory readback time (not shown in Figure 6) |
| tpowER_UP $^{2}$ |  |  | 75 | $\mu \mathrm{~s}$ | Power-on EEPROM restore time (not shown in Figure 6) |
| $\mathrm{t}_{\text {RESET }}$ |  | 30 |  | $\mu \mathrm{~s}$ | Reset EEPROM restore time (not shown in Figure 6) |

${ }^{1}$ EEPROM program time depends on the temperature and EEPROM write cycles. Higher timing is expected at lower temperatures and higher write cycles.
${ }^{2}$ Maximum time after $\mathrm{V}_{\mathrm{DD}}-\mathrm{V}_{S S}$ is equal to 2.3 V .

## SHIFT REGISTER AND TIMING DIAGRAMS



Figure 2. Input Shift Register Contents


Figure 3. $I^{2}$ C Serial Interface Timing Diagram (Typical Write Sequence)


Figure 4. SPI Serial Interface Timing Diagram, $C P O L=0, C P H A=1$


Figure 5. SPI Serial Interface Timing Diagram, $C P O L=1, C P H A=0$


Figure 6. Control Pins Timing Diagram

## ABSOLUTE MAXIMUM RATINGS

$\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.
Table 7.

| Parameter | Rating |
| :---: | :---: |
| $V_{\text {DD }}$ to GND | -0.3 V to +7.0 V |
| $V_{\text {ss }}$ to GND | +0.3 V to -7.0 V |
| $V_{\text {DD }}$ to $V_{S S}$ | 7 V |
| V Logic to GND | $\begin{aligned} & -0.3 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \text { or } \\ & +7.0 \mathrm{~V} \text { (whichever is less) } \end{aligned}$ |
| $\mathrm{V}_{\mathrm{A}}, \mathrm{V}_{\mathrm{W}}, \mathrm{V}_{\mathrm{B}}$ to GND | $\begin{aligned} & \mathrm{V}_{S S}-0.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \text { or } \\ & +7.0 \mathrm{~V} \text { (whichever is less) } \end{aligned}$ |
| $I_{\text {A }}, I_{\text {w }}, I_{\text {b }}$ |  |
| Pulsed ${ }^{1}$ |  |
| Frequency > 10 kHz |  |
| RAw $=10 \mathrm{k} \Omega$ | $\pm 6 \mathrm{~mA} / \mathrm{d}^{2}$ |
| $\mathrm{R}_{\mathrm{Aw}}=100 \mathrm{k} \Omega$ | $\pm 1.5 \mathrm{~mA} / \mathrm{d}^{2}$ |
| Frequency $\leq 10 \mathrm{kHz}$ |  |
| $\mathrm{R}_{\text {Aw }}=10 \mathrm{k} \Omega$ | $\pm 6 \mathrm{~mA} / \sqrt{ } \mathrm{d}^{2}$ |
| $\mathrm{R}_{\text {Aw }}=100 \mathrm{k} \Omega$ | $\pm 1.5 \mathrm{~mA} / \sqrt{ } \mathrm{d}^{2}$ |
| Digital Inputs | $\begin{aligned} & -0.3 \mathrm{~V} \text { to } \mathrm{V} \text { LoGic }+0.3 \mathrm{~V} \text { or } \\ & +7 \mathrm{~V} \text { (whichever is less) } \end{aligned}$ |
| Operating Temperature Range, $\mathrm{T}_{\mathrm{A}}{ }^{3}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Maximum Junction Temperature, TJ, Maximum | $150^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Reflow Soldering |  |
| Peak Temperature | $260^{\circ} \mathrm{C}$ |
| Time at Peak Temperature | 20 sec to 40 sec |
| Package Power Dissipation |  |

${ }^{1}$ Maximum terminal current is bounded by the maximum current handling of the switches, maximum power dissipation of the package, and maximum applied voltage across any two of the $A, B$, and $W$ terminals at a given resistance.
${ }^{2} d$ = pulse duty factor.
${ }^{3}$ Includes programming of EEPROM memory.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## THERMAL RESISTANCE

$\theta_{\mathrm{JA}}$ is defined by the JEDEC JESD51 standard, and the value is dependent on the test board and test environment.

Table 8. Thermal Resistance

| Package Type | $\boldsymbol{\theta}_{\mathrm{JA}}$ | $\boldsymbol{\theta}_{\mathbf{\prime c}}$ | Unit |
| :--- | :--- | :--- | :--- |
| 16-Lead LFCSP | $89.5^{1}$ | 3 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

${ }^{1}$ JEDEC 2S2P test board, still air ( $0 \mathrm{~m} / \mathrm{sec}$ airflow).

## ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Figure 7. Pin Configuration

Table 9. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | GND | Ground Pin, Logic Ground Reference. |
| 2 | A | Terminal A of RDAC. $\mathrm{V}_{S S} \leq \mathrm{V}_{\mathrm{A}} \leq \mathrm{V}_{\mathrm{DD}}$. |
| 3 | W | Wiper terminal of RDAC. $\mathrm{V}_{S S} \leq \mathrm{V}_{\mathrm{W}} \leq \mathrm{V}_{\mathrm{DD}}$. |
| 4 | B | Terminal B of RDAC. $\mathrm{V}_{S S} \leq \mathrm{V}_{\mathrm{B}} \leq \mathrm{V}_{\mathrm{DD}}$. |
| 5 | $\mathrm{V}_{\text {ss }}$ | Negative Power Supply. Decouple this pin with $0.1 \mu \mathrm{~F}$ ceramic capacitors and $10 \mu \mathrm{~F}$ capacitors. |
| 6 | $\overline{\text { SYNC/ADDRO }}$ | Programmable Address (ADDRO) for Multiple Package Decoding, DIS $=1$. |
|  |  | Synchronization Data Input, Active Low. When $\overline{\text { SYNC }}$ returns high, data is loaded into the RDAC register, DIS $=0$. |
| 7 | $\overline{\text { RESET }}$ | Hardware Reset Pin. Refresh the RDAC registers from EEPROM. $\overline{\text { RESET }}$ is activated at logic low. If this pin is not used, tie $\overline{\text { RESET }}$ to V Logic. |
| 8 | DIS | Digital Interface Select (SPI// ${ }^{2} \mathrm{C}$ Select). SPI when DIS $=0$ (GND), $I^{2} \mathrm{C}$ when DIS $=1$ (V Logic). This pin cannot be left floating. |
| 9 | $V_{\text {D }}$ | Positive Power Supply. Decouple this pin with $0.1 \mu \mathrm{~F}$ ceramic capacitors and $10 \mu \mathrm{~F}$ capacitors. |
| 10 | $V_{\text {Logic }}$ | Logic Power Supply; 1.8 V to VDD. Decouple this pin with $0.1 \mu \mathrm{~F}$ ceramic capacitors and $10 \mu \mathrm{~F}$ capacitors. |
| 11 | SCLK/SCL | SPI Serial Clock Line (SCLK). Data is clocked in at logic low transition. $1^{2} \mathrm{C}$ Serial Clock Line (SCL). Data is clocked in at logic low transition. |
| 12 | SDI/SDA | Serial Data Input/Output (SDA), When DIS $=1$. Serial Data Input (SDI), When DIS $=0$. |
| 13 | $\overline{W P}$ | Optional Write Protect. This pin prevents any changes to the present RDAC and EEPROM contents, except when reloading the content of the EEPROM into the RDAC register. $\overline{W P}$ is activated at logic low. If this pin is not used, tie $\overline{W P}$ to $V_{\text {Logic }}$ |
| 14 | SDO/ADDR1 | Programmable Address (ADDR1) for Multiple Package Decoding, When DIS $=1$. <br> Serial Data Output (SDO). This is an open-drain output pin, and it needs an external pull-up resistor when DIS $=0$. |
| 15 | INDEP | Linear Gain Setting Mode at Power-Up. Each string resistor is loaded from its associate memory location. If INDEP is enabled, it cannot be disabled by the software. |
| 16 | $\overline{\text { LRDAC }}$ | Load RDAC. Transfers the contents of the input register to the RDAC register. This allows asynchronous RDAC update. $\overline{\text { LRDAC }}$ is activated low. If this pin is not used, tie $\overline{\text { LRDAC }}$ to $\mathrm{V}_{\text {Logic. }}$ |
|  | EPAD | Internally Connect the Exposed Pad to $\mathrm{V}_{\text {ss }}$. |

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 8. R-INL vs. Code (AD5141)


Figure 9. R-INL vs. Code (AD5121)


Figure 10. INL vs. Code (AD5141)


Figure 11. R-DNL vs. Code (AD5141)


Figure 12. R-DNL vs. Code (AD5121)


Figure 13. DNL vs. Code (AD5141)


Figure 14. INL vs. Code (AD5121)


Figure 15. Potentiometer Mode Temperature Coefficient $\left(\left(\Delta V_{w} / V_{w}\right) / \Delta T \times 10^{6}\right)$ vs. Code


Figure 16. Supply Current vs. Temperature


Figure 17. DNL vs. Code (AD5121)


Figure 18. Rheostat Mode Temperature Coefficient $\left(\left(\Delta R_{w B} / R_{w B}\right) / \Delta T \times 10^{6}\right)$ vs. Code


Figure 19. I LOGIC Current vs. Digital Input Voltage


Figure 20. $10 \mathrm{k} \Omega$ Gain vs. Frequency vs. Code


Figure 21. Total Harmonic Distortion Plus Noise $(T H D+N)$ vs. Frequency


Figure 22. Normalized Phase Flatness vs. Frequency, $R_{A B}=10 \mathrm{k} \Omega$


Figure 23. $100 \mathrm{k} \Omega$ Gain vs. Frequency vs. Code


Figure 24. Total Harmonic Distortion Plus Noise $(T H D+N)$ vs. Amplitude


Figure 25. Normalized Phase Flatness vs. Frequency, $R_{A B}=100 \mathrm{k} \Omega$


Figure 26. Incremental Wiper On Resistance vs. VDD


Figure 27. Maximum Bandwidth vs. Code vs. Net Capacitance


Figure 28. Maximum Transition Glitch


Figure 29. Resistor Lifetime Drift


Figure 30. Power Supply Rejection Ratio (PSRR) vs. Frequency


Figure 31. Digital Feedthrough


Figure 32. Shutdown Isolation vs. Frequency


Figure 33. Theoretical Maximum Current vs. Code

## TEST CIRCUITS

Figure 34 to Figure 38 define the test conditions used in the Specifications section.


Figure 34. Resistor Integral Nonlinearity Error (Rheostat Operation; R-INL, R-DNL)


Figure 35. Potentiometer Divider Nonlinearity Error (INL, DNL)


Figure 37. Power Supply Sensitivity and Power Supply Rejection Ratio (PSS and PSRR)


Figure 38. Incremental on Resistance


Figure 36. Wiper Resistance

## THEORY OF OPERATION

The AD5121/AD5141 digital programmable potentiometers are designed to operate as true variable resistors for analog signals within the terminal voltage range of $\mathrm{V}_{\mathrm{SS}}<\mathrm{V}_{\mathrm{TERM}}<\mathrm{V}_{\mathrm{DD}}$. The resistor wiper position is determined by the RDAC register contents. The RDAC register acts as a scratchpad register that allows unlimited changes of resistance settings. A secondary register (the input register) can be used to preload the RDAC register data.
The RDAC register can be programmed with any position setting using the $\mathrm{I}^{2} \mathrm{C}$ or SPI interface (depending on the model). When a desirable wiper position is found, this value can be stored in the EEPROM memory. Thereafter, the wiper position is always restored to that position for subsequent power-ups. The storing of EEPROM data takes approximately 18 ms ; during this time, the device is locked and does not acknowledge any new command, preventing any changes from taking place.

## RDAC REGISTER AND EEPROM

The RDAC register directly controls the position of the digital potentiometer wiper. For example, when the RDAC register is loaded with 0x80 (AD5141, 256 taps), the wiper is connected to half scale of the variable resistor. The RDAC register is a standard logic register; there is no restriction on the number of changes allowed.
It is possible to both write to and read from the RDAC register using the digital interface (see Table 16).

The contents of the RDAC register can be stored to the EEPROM using Command 9 (see Table 16). Thereafter, the RDAC register always sets at that position for any future on-off-on power supply sequence. It is possible to read back data saved into the EEPROM with Command 3 (see Table 16).
Alternatively, the EEPROM can be written to independently using Command 1 (see Table 16).

## INPUT SHIFT REGISTER

For the AD5121/AD5141, the input shift register is 16 bits wide, as shown in Figure 2. The 16-bit word consists of four control bits, followed by four address bits and by eight data bits

If the AD5121 RDAC or EEPROM registers are read from or written to the lowest data bit (Bit 0 ) is ignored.
Data is loaded MSB first (Bit 15). The four control bits determine the function of the software command, as listed in Table 11 and Table 16.

## SERIAL DATA DIGITAL INTERFACE SELECTION, DIS

The AD5121/AD5141 LFSCP provides the flexibility of a selectable interface. When the digital interface select (DIS) pin is tied low, the SPI mode is engaged. When the DIS pin is tied high, the $\mathrm{I}^{2} \mathrm{C}$ mode is engaged.

## SPI SERIAL DATA INTERFACE

The AD5121/AD5141 contain a 4 -wire, SPI-compatible digital interface (SDI, $\overline{S Y N C}$, SDO, and SCLK). The write sequence begins by bringing the $\overline{\text { SYNC }}$ line low. The $\overline{\text { SYNC }}$ pin must be held low until the complete data-word is loaded from the SDI pin. Data is loaded in at the SCLK falling edge transition, as shown in Figure 4. When SYNC returns high, the serial data-word is decoded according to the instructions in Table 16.
The AD5121/AD5141 do not require a continuous SCLK when $\overline{S Y N C}$ is high. To minimize power consumption in the digital input buffers when the part is enabled, operate all serial interface pins close to the $\mathrm{V}_{\text {Logic }}$ supply rails.

## $\overline{\text { SYNC Interruption }}$

In a standalone write sequence for the AD5121/AD5141, the $\overline{\text { SYNC }}$ line is kept low for 16 falling edges of SCLK, and the instruction is decoded when $\overline{S Y N C}$ is pulled high. However, if the $\overline{\text { SYNC }}$ line is kept low for less than 16 falling edges of SCLK, the input shift register content is ignored, and the write sequence is considered invalid.

## SDO Pin

The serial data output pin (SDO) serves two purposes: to read back the contents of the control, EEPROM, RDAC, and input registers using Command 3 (see Table 11 and Table 16), and to connect the AD5121/AD5141 to daisy-chain mode.
The SDO pin contains an internal open-drain output that needs an external pull-up resistor. The SDO pin is enabled when $\overline{\text { SYNC }}$ is pulled low, and the data is clocked out of SDO on the rising edge of SCLK.

## Daisy-Chain Connection

Daisy chaining minimizes the number of port pins required from the controlling IC. As shown in Figure 39, the SDO pin of one package must be tied to the SDI pin of the next package. The clock period may need to be increased because the propagation delay of the line between subsequent devices. When two AD5121/ AD5141 devices are daisy chained, 32 bits of data are required. The first 16 bits assigned to U 2 , and the second 16 bits assigned to U1, as shown in Figure 40. Keep the SYNC pin low until all 32 bits are clocked into their respective serial registers. The $\overline{\text { SYNC }}$ pin is then pulled high to complete the operation. A typical connection is shown in Figure 39.

To prevent data from mislocking (for example, due to noise) the part includes an internal counter, if the clock falling edges count is not a multiple of 8 , the part ignores the command. A valid clock count is 16,24 , or 32 . The counter resets when $\overline{\text { SYNC }}$ returns high.


Figure 39. Daisy-Chain Configuration


Figure 40. Daisy-Chain Diagram

## $I^{2}$ C SERIAL DATA INTERFACE

The AD5141 has 2-wire, $\mathrm{I}^{2} \mathrm{C}$-compatible serial interface. These devices can be connected to an $\mathrm{I}^{2} \mathrm{C}$ bus as a slave device, under the control of a master device. See Figure 3 for a timing diagram of a typical write sequence.

The AD5141 supports standard ( 100 kHz ) and fast ( 400 kHz ) data transfer modes. Support is not provided for 10-bit addressing and general call addressing.
The 2-wire serial bus protocol operates as follows:

1. The master initiates a data transfer by establishing a start condition, which is when a high-to-low transition on the SDA line occurs while SCL is high. The following byte is the address byte, which consists of the 7-bit slave address and an $\mathrm{R} / \overline{\mathrm{W}}$ bit. The slave device corresponding to the transmitted address responds by pulling SDA low during the ninth clock pulse (this is called the acknowledge bit). At this stage, all other devices on the bus remain idle while the selected device waits for data to be written to, or read from, its shift register.
If the $\mathrm{R} / \overline{\mathrm{W}}$ bit is set high, the master reads from the slave device. However, if the $\mathrm{R} / \overline{\mathrm{W}}$ bit is set low, the master writes to the slave device.
2. Data is transmitted over the serial bus in sequences of nine clock pulses (eight data bits followed by an acknowledge bit). The transitions on the SDA line must occur during the low period of SCL and remain stable during the high period of SCL.
3. When all data bits have been read from or written to, a stop condition is established. In write mode, the master pulls the SDA line high during the tenth clock pulse to establish a stop condition. In read mode, the master issues a no acknowledge for the ninth clock pulse (that is, the SDA line remains high). The master then brings the SDA line low before the tenth clock pulse, and then high again during the tenth clock pulse to establish a stop condition.

## $1^{2}$ C ADDRESS

The AD5141 has two different pin address options available, as shown in Table 10.

Table 10. 24-Lead LFCSP Device Address Selection

| ADDRO Pin | ADDR1 Pin | 7-Bit ${ }^{1} \mathrm{C}$ Device Address |
| :---: | :---: | :---: |
| $V_{\text {LOGIC }}$ | V LOGIC | 0100000 |
| No connect ${ }^{1}$ | V Logic | 0100010 |
| GND | $V_{\text {LoGic }}$ | 0100011 |
| $V_{\text {Logic }}$ | No connect ${ }^{1}$ | 0101000 |
| No connect ${ }^{1}$ | No connect ${ }^{1}$ | 0101010 |
| GND | No connect ${ }^{1}$ | 0101011 |
| $V_{\text {LoGic }}$ | GND | 0101100 |
| No connect ${ }^{1}$ | GND | 0101110 |
| GND | GND | 0101111 |

${ }^{1}$ Not available in bipolar mode $\left(\mathrm{V}_{\mathrm{SS}}<0 \mathrm{~V}\right)$ or in low voltage mode $\left(\mathrm{V}_{\text {LOGIC }}=1.8 \mathrm{~V}\right)$.

Table 11. Simple Command Operation Truth Table

| Command Number | ControlBits[DB15:DB12] |  |  |  | Address Bits[DB11:DB8] ${ }^{1}$ |  |  |  | Data Bits[DB7:DB0] ${ }^{1}$ |  |  |  |  |  |  |  | Operation |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | C3 | C2 | C1 | C0 | A3 | A2 | A1 | A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |  |
| 0 | 0 | 0 | 0 | 0 | X | X | X | X | X | X | X | X | X | X | X | X | NOP: do nothing |  |  |
| 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Write contents of serial register data to RDAC |  |  |
| 2 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Write contents of serial register data to input register |  |  |
| 3 | 0 | 0 | 1 | 1 | X | 0 | 0 | 0 | X | X | X | X | X | X | D1 | D0 | Read back contents |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | D1 | D0 | Data |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{array}{\|l\|} \hline 0 \\ 1 \end{array}$ | $\begin{aligned} & 1 \\ & 1 \end{aligned}$ | EEPROM RDAC |
| 9 | 0 | 1 | 1 | 1 | X | X | 0 | 0 | X | X | X | X | X | X | X | 1 | Copy RDAC register to EEPROM |  |  |
| 10 | 0 | 1 | 1 | 1 | X | X | 0 | 0 | X | X | X | X | X | X | X | 0 | Copy EEPROM into RDAC |  |  |
| 14 | 1 | 0 | 1 | 1 | X | X | X | X | X | X | X | X | X | X | X | X | Software reset |  |  |
| 15 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | X | X | X | X | X | X | X | D0 | Software shutdown |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | D0 | Condition |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{array}{\|l\|} \hline 0 \\ 1 \end{array}$ | Normal mode Shutdown mode |  |

[^0]
## ADVANCED CONTROL MODES

The AD5121/AD5141 digital potentiometers include a set of user programming features to address the wide number of applications for these universal adjustment devices (see Table 16 and Table 18).

Key programming features include the following:

- Input register
- Linear gain setting mode
- Low wiper resistance feature
- Linear increment and decrement instructions
- $\pm 6 \mathrm{~dB}$ increment and decrement instructions
- Burst mode ( $\mathrm{I}^{2} \mathrm{C}$ only)
- Reset
- Shutdown mode


## Input Register

The AD5121/AD5141 include one input register per RDAC register. This register allows preloading of the value for the associated RDAC register.
This feature allows a synchronous and asynchronous update of one or all the RDAC registers at the same time.
These registers can be written to using Command 2 and read back from using Command 3 (see Table 16).
The transfer from the input register to the RDAC register is done asynchronously by the $\overline{\text { LRDAC }}$ pin or synchronously by Command 8 (see Table 16).
If new data is loaded in an RDAC register, this RDAC register automatically overwrites the associated input register.

## Linear Gain Setting Mode

The patented architecture of the AD5121/AD5141 allows the independent control of each string resistor, $\mathrm{R}_{A W}$ and $\mathrm{R}_{\text {WB }}$. To enable this feature, use Command 16 (see Table 16) to set Bit D2 of the control register (see Table 18).
This mode of operation can control the potentiometer as two independent rheostats connected at a single point, W terminal, as opposed to potentiometer mode where each resistor is complementary, $\mathrm{R}_{A W}=\mathrm{R}_{A B}-\mathrm{R}_{\mathrm{wb}}$.
This feature enables a second input and an RDAC register per channel, as shown in Table 16; however, the actual RDAC contents remain unchanged. The same operations are valid for potentiometer and linear gain setting modes.
If the INDEP pin is pulled high, the device powers up in linear gain setting mode and loads the values stored in the associated memory locations for each channel (see Table 17). The INDEP pin and D2 bit are connected internally to a logic OR gate, if any or both are 1, the parts cannot operate in potentiometer mode.

## Low Wiper Resistance Feature

The AD5121/AD5141 include two commands to reduce the wiper resistance between the terminals when they achieve full scale or zero scale. These extra positions are called bottom scale, BS, and top scale, TS. The resistance between Terminal A and Terminal W at top scale is specified as $\mathrm{R}_{\text {Ts. }}$ Similarly, the bottom scale resistance between Terminal B and Terminal W is specified as $\mathrm{R}_{\mathrm{BS}}$.
The contents of the RDAC registers are unchanged by entering in these positions. There are two ways to exit from top scale and bottom scale: by using Command 12 or Command 13 (see Table 16); or by loading new data in an RDAC register, which includes increment/decrement operations and a shutdown command.

Table 12 and Table 13 show the truth tables for the top scale position and the bottom scale position, respectively, when linear gain setting mode is enabled.

Table 12. Top Scale Truth Table

| Linear Gain Setting Mode |  | Potentiometer Mode |  |
| :--- | :--- | :--- | :--- |
| $\mathrm{R}_{A W}$ | $\mathrm{R}_{\text {WB }}$ | $\mathrm{R}_{A W}$ | $\mathrm{R}_{\mathrm{WB}}$ |
| $\mathrm{R}_{A B}$ | $\mathrm{R}_{A B}$ | $\mathrm{R}_{T S}$ | $\mathrm{R}_{A B}$ |

Table 13. Bottom Scale Truth Table

| Linear Gain Setting Mode |  | Potentiometer Mode |  |
| :---: | :---: | :---: | :---: |
| Raw | Rws | Raw | Rwi |
| RTS | RBS | RAB | RBS |

## Linear Increment and Decrement Instructions

The increment and decrement commands (Command 4 and Command 5 in Table 16) are useful for linear step adjustment applications. These commands simplify microcontroller software coding by allowing the controller to send an increment or decrement command to the device. The adjustment can be individual or in a ganged potentiometer arrangement, where all wiper positions are changed at the same time.
For an increment command, executing Command 4 automatically moves the wiper to the next resistance segment position. This command can be executed in a single channel or multiple channels.

## $\pm 6$ dB Increment and Decrement Instructions

Two programming instructions produce logarithmic taper increment or decrement of the wiper position control by an individual potentiometer or by a ganged potentiometer arrangement where all RDAC register positions are changed simultaneously. The +6 dB increment is activated by Command 6 , and the -6 dB decrement is activated by Command 7 (see Table 16). For example, starting with the zero-scale position and executing Command 6 ten times moves the wiper in 6 dB steps to the fullscale position. When the wiper position is near the maximum setting, the last 6 dB increment instruction causes the wiper to go to the full-scale position (see Table 14).

Incrementing the wiper position by +6 dB essentially doubles the RDAC register value, whereas decrementing the wiper position by -6 dB halves the register content. Internally, the AD5121/AD5141 use shift registers to shift the bits left and right to achieve a $\pm 6 \mathrm{~dB}$ increment or decrement. These functions are useful for various audio/video level adjustments, especially for white LED brightness settings in which human visual responses are more sensitive to large adjustments than to small adjustments.

Table 14. Detailed Left Shift and Right Shift Functions for the $\pm 6 \mathrm{~dB}$ Step Increment and Decrement

| Left Shift (+6 dB/Step) | Right Shift (-6 dB/Step) |
| :--- | :--- |
| 00000000 | 11111111 |
| 00000001 | 01111111 |
| 00000010 | 00111111 |
| 00000100 | 00011111 |
| 00001000 | 00001111 |
| 00010000 | 00000111 |
| 00100000 | 00000011 |
| 01000000 | 00000001 |
| 10000000 | 00000000 |
| 1111111 | 00000000 |

## Burst Mode ( $1^{2}$ C Only)

By enabling the burst mode, multiple data bytes can be sent to the part consecutively. After the command byte, the part interprets the consecutive bytes as data bytes for the first command.
A new command can be sent by generating a repeat start or by a stop and start condition.
The burst mode is activated by setting Bit D3 of the control register (see Table 18), and if a reset or power-down is performed, it automatically resets.

## Reset

The AD5121/AD5141 can be reset through software by executing Command 14 (see Table 16) or through hardware on the low pulse of the $\overline{\text { RESET }}$ pin. The reset command loads the RDAC register with the contents of the EEPROM and takes approximately $30 \mu \mathrm{~s}$. The EEPROM is preloaded to midscale at the factory, and initial power-up is, accordingly, at midscale.
Tie $\overline{\text { RESET }}$ to $V_{D D}$ if the $\overline{\text { RESET }}$ pin is not used.

## Shutdown Mode

The AD5121/AD5141 can be placed in shutdown mode by executing the software shutdown command, Command 15 (see Table 16); and by setting the LSB (D0) to 1 . This feature places the RDAC in a special state. The contents of the RDAC register are unchanged by entering shutdown mode. However, all commands listed in Table 16 are supported while in shutdown mode. Execute Command 15 (see Table 16) and set the LSB (D0) to 0 to exit shutdown mode.

Table 15. Truth Table for Shutdown Mode

| A2 | Linear Gain Setting Mode |  | Potentiometer Mode |  |
| :--- | :--- | :--- | :--- | :--- |
|  | AW | WB | AW | WB |
| 0 | $N / A^{1}$ | Open | Open | $\mathrm{R}_{B S}$ |
| 1 | Open | $\mathrm{N} / \mathrm{A}^{1}$ | $\mathrm{~N} / \mathrm{A}^{1}$ | $\mathrm{~N} / \mathrm{A}^{1}$ |

${ }^{1} \mathrm{~N} / \mathrm{A}=$ not applicable.

## EEPROM OR RDAC REGISTER PROTECTION

The EEPROM and RDAC registers can be protected by disabling any update to these registers. This can be done by using software or by using hardware. If these registers are protected by software, set Bit D0 and/or Bit D1 (see Table 18), which protects the RDAC and EEPROM registers independently.
If the registers are protected by hardware, pull the $\overline{\mathrm{WP}}$ pin low. If the $\overline{\mathrm{WP}}$ pin is pulled low when the part is executing a command, the protection is not enabled until the command is completed.
When RDAC is protected, the only operation allowed is to copy the EEPROM into the RDAC register.

## LOAD RDAC INPUT REGISTER (LRDAC)

$\overline{\text { LRDAC }}$ software or hardware transfers data from the input register to the RDAC register (and therefore updates the wiper position). By default, the input register has the same value as the RDAC register; therefore, only the input register that has been updated using Command 2 is updated.
Software $\overline{\mathrm{LRDAC}}$, Command 8 , allows updating of a single RDAC register or all of the channels at once (see Table 16). This is a synchronous update.
The hardware $\overline{\text { LRDAC }}$ is completely asynchronous and copies the content of all the input registers into the associated RDAC registers. If a command is executed, to avoid data corruption, any transition in the $\overline{\text { LRDAC }}$ pin is ignored by the part.

## INDEP PIN

If the INDEP pin is pulled high at power-up, the part operates in linear gain setting mode, loading each string resistor, RAW and $\mathrm{R}_{\mathrm{WB}}$, with the value stored into the EEPROM (see Table 17). If the pin is pulled low, the part powers up in potentiometer mode.
The INDEP pin and the D2 bit are connected internally to a logic OR gate, if any or both are 1 , the part cannot operate in potentiometer mode (see Table 18).

Table 16. Advance Commands Operation Truth Table

| Command Number | ControlBits[DB15:DB12] |  |  |  | AddressBits[DB11:DB8] ${ }^{1}$ |  |  |  | Data Bits[DB7:DB0] ${ }^{1}$ |  |  |  |  |  |  |  | Operation |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | C3 | C2 | C1 | C0 | A3 | A2 | A1 | A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |  |
| 0 | 0 | 0 | 0 | 0 | X | X | X | X | X | X | X | X | X | X | X | X | NOP: do nothing |  |  |
| 1 | 0 | 0 | 0 | 1 | 0 | A2 | 0 | A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Write contents of serial register data to RDAC |  |  |
| 2 | 0 | 0 | 1 | 0 | 0 | A2 | 0 | A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Write contents of serial register data to input register |  |  |
| 3 | 0 | 0 | 1 | 1 | X | A2 | A1 | A0 | X | X | X | X | X | X | D1 | D0 | Read back contents |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | D1 | D0 | Data |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  | $\begin{aligned} & \hline 0 \\ & 0 \\ & 1 \\ & 1 \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline 0 \\ & 1 \\ & 0 \\ & 1 \end{aligned}$ | Input register <br> EEPROM <br> Control register RDAC |
| 4 | 0 | 1 | 0 | 0 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | 1 | Linear RDAC increment |  |  |
| 5 | 0 | 1 | 0 | 0 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | 0 | Linear RDAC decrement |  |  |
| 6 | 0 | 1 | 0 | 1 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | 1 | +6 dB RDAC increment |  |  |
| 7 | 0 | 1 | 0 | 1 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | 0 | -6 dB RDAC decrement |  |  |
| 8 | 0 | 1 | 1 | 0 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | X | Copy input register to RDAC (software LRDAC) |  |  |
| 9 | 0 | 1 | 1 | 1 | 0 | A2 | 0 | A0 | X | X | X | X | X | X | X | 1 | Copy RDAC register to EEPROM |  |  |
| 10 | 0 | 1 | 1 | 1 | 0 | A2 | 0 | A0 | X | X | X | X | X | X | X | 0 | Copy EEPROM into RDAC |  |  |
| 11 | 1 | 0 | 0 | 0 | 0 | A2 | 0 | A0 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Write contents of serial register data to EEPROM |  |  |
| 12 | 1 | 0 | 0 | 1 | A3 | A2 | 0 | A0 | 1 | X | X | X | X | X | X | D0 | Top scale D0 $=0$; normal mode D0 $=1$; shutdown mode |  |  |
| 13 | 1 | 0 | 0 | 1 | A3 | A2 | 0 | A0 | 0 | X | X | X | X | X | X | D0 | Bottom scale D0 $=1$; enter D0 = 0; exit |  |  |
| 14 | 1 | 0 | 1 | 1 | X | X | X | X | X | X | X | X | X | X | X | X | Software reset |  |  |
| 15 | 1 | 1 | 0 | 0 | A3 | A2 | 0 | A0 | X | X | X | X | X | X | X | D0 | Software shutdown D0 $=0$; normal mode D0 $=1$; device placed in shutdown mode |  |  |
| 16 | 1 | 1 | 0 | 1 | X | X | X | X | X | X | X | X | D3 | D2 | D1 | D0 | Copy serial register data to control register |  |  |

${ }^{1} \mathrm{X}=$ don't care.

Table 17. Address Bits

| A3 | A2 | A1 | A0 | Potentiometer Mode |  | Linear Gain Setting Mode |  | Stored RDAC Memory |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Input Register | RDAC Register | Input Register | RDAC Register |  |
| 1 | X ${ }^{1}$ | X ${ }^{1}$ | $\mathrm{X}^{1}$ | All channels | All channels | All channels | All channels | Not applicable |
| 0 | 0 | 0 | 0 | RDAC | RDAC | Rwb | Rwb | RDAC/Rwв |
| 0 | 1 | 0 | 0 | Not applicable | Not applicable | Raw | Raw | Not applicable |
| 0 | 0 | 0 | 1 | Not applicable | Not applicable | Not applicable | Not applicable | Raw |
| 0 | 0 | 1 | 0 | Not applicable | Not applicable | Not applicable | Not applicable | MSB tolerance |
| 0 | 0 | 1 | 1 | Not applicable | Not applicable | Not applicable | Not applicable | LSB tolerance |

[^1]Table 18. Control Register Bit Descriptions

| Bit Name | Description |
| :--- | :--- |
| D0 | RDAC register write protect |
|  | $0=$ wiper position frozen to value in EEPROM memory |
|  | $1=$ allows update of wiper position through digital interface (default) |
| D1 | EEPROM program enable |
|  | $0=$ EEPROM program disabled |
|  | $1=$ enables device for EEPROM program (default) |
| D2 | Linear setting mode/potentiometer mode |
|  | $0=$ potentiometer mode (default) |
|  | $1=$ linear gain setting mode |
| D3 | Burst mode ( $I^{2}$ C only) <br>  <br>  <br>  <br>  <br>  <br>  <br> $=$ disabled (default) |
|  |  |

## RDAC ARCHITECTURE

To achieve optimum performance, Analog Devices, Inc., has patented the RDAC segmentation architecture for all the digital potentiometers. In particular, the AD5121/AD5141 employ a three-stage segmentation approach, as shown in Figure 41. The AD5121/AD5141 wiper switch is designed with the transmission gate CMOS topology and with the gate voltage derived from $V_{\text {DD }}$ and $V_{s s}$.


Figure 41. AD5121/AD5141 Simplified RDAC Circuit

## Top Scale/Bottom Scale Architecture

In addition, the AD5121/AD5141 include new positions to reduce the resistance between terminals. These positions are called bottom scale and top scale. At bottom scale, the typical wiper resistance decreases from $130 \Omega$ to $60 \Omega\left(\mathrm{R}_{A B}=100 \mathrm{k} \Omega\right)$. At top scale, the resistance between Terminal A and Terminal W is decreased by 1 LSB , and the total resistance is reduced to $60 \Omega$ $\left(\mathrm{R}_{A B}=100 \mathrm{k} \Omega\right)$.

## PROGRAMMING THE VARIABLE RESISTOR

## Rheostat Operation— $\pm 8 \%$ Resistor Tolerance

The AD5121/AD5141 operate in rheostat mode when only two terminals are used as a variable resistor. The unused terminal can be floating, or it can be tied to Terminal W, as shown in Figure 42.


The nominal resistance between Terminal A and Terminal B, $R_{A B}$, is $10 \mathrm{k} \Omega$ or $100 \mathrm{k} \Omega$, and has 128/256 tap points accessed by the wiper terminal. The 7-bit/8-bit data in the RDAC latch is decoded to select one of the 128/256 possible wiper settings. The general equations for determining the digitally programmed output resistance between Terminal W and Terminal B are
AD5121:

$$
\begin{equation*}
R_{W B}(D)=\frac{D}{128} \times R_{A B}+R_{W} \quad \text { From } 0 \times 00 \text { to } 0 \mathrm{x} 7 \mathrm{~F} \tag{1}
\end{equation*}
$$

AD5141:

$$
\begin{equation*}
R_{W B}(D)=\frac{D}{256} \times R_{A B}+R_{W} \quad \text { From 0x00 to 0xFF } \tag{2}
\end{equation*}
$$

where:
$D$ is the decimal equivalent of the binary code in the 7-bit/8-bit RDAC register.
$R_{A B}$ is the end-to-end resistance.
$R_{W}$ is the wiper resistance.
In potentiometer mode, similar to the mechanical potentiometer, the resistance of the RDAC between Terminal W and Terminal A also produces a digitally controlled complementary resistance, Rwa. Rwa also gives a maximum of $8 \%$ absolute resistance error. R ${ }_{w A}$ starts at the maximum resistance value and decreases as the data loaded into the latch increases. The general equations for this operation are

AD5121:

$$
\begin{equation*}
R_{A W}(D)=\frac{128-D}{128} \times R_{A B}+R_{W} \tag{3}
\end{equation*}
$$

From 0x00 to 0x7F
AD5141:

$$
\begin{equation*}
R_{A W}(D)=\frac{256-D}{256} \times R_{A B}+R_{W} \quad \text { From } 0 \times 00 \text { to } 0 \mathrm{xFF} \tag{4}
\end{equation*}
$$

where:
$D$ is the decimal equivalent of the binary code in the 7-bit/8-bit RDAC register.
$R_{A B}$ is the end-to-end resistance.
$R_{W}$ is the wiper resistance.

If the part is configured in linear gain setting mode, the resistance between Terminal W and Terminal A is directly proportional to the code loaded in the associate RDAC register. The general equations for this operation are
AD5121:

$$
\begin{equation*}
R_{A W}(D)=\frac{D}{128} \times R_{A B}+R_{W} \quad \text { From } 0 \mathrm{x} 00 \text { to } 0 \mathrm{x} 7 \mathrm{~F} \tag{5}
\end{equation*}
$$

AD5141:

$$
\begin{equation*}
R_{A W}(D)=\frac{D}{256} \times R_{A B}+R_{W} \quad \text { From } 0 x 00 \text { to } 0 \mathrm{xFF} \tag{6}
\end{equation*}
$$

where:
$D$ is the decimal equivalent of the binary code in the 7-bit/8-bit RDAC register.
$R_{A B}$ is the end-to-end resistance.
$R_{W}$ is the wiper resistance.
In the bottom scale condition or top scale condition, a finite total wiper resistance of $40 \Omega$ is present. Regardless of which setting the part is operating in, limit the current between Terminal A to Terminal B, Terminal W to Terminal A, and Terminal $W$ to Terminal $B$, to the maximum continuous current of $\pm 6 \mathrm{~mA}$ or to the pulse current specified in Table 7. Otherwise, degradation or possible destruction of the internal switch contact can occur.

## Calculate the Actual End-to-End Resistance

The resistance tolerance is stored in the internal memory during factory testing. Therefore, the actual end-to-end resistance can be calculated (which is valuable for calibration, tolerance matching, and precision applications).
The resistance tolerance (in percentage) is stored in fixed point format, using a 16 -bit sign magnitude binary. The sign bit ( $0=$ negative and $1=$ positive) and the integer part are located in Address 0x02, as shown in Table 19. Address 0x03 contains the fractional part, as shown in Table 19.

That is, if the data readback from Address $0 \times 02$ is 00000010 , and the data readback from Address $0 \times 03$ is 10110000, the end-to-end resistance can be calculated as follows.

For Memory Map Address $0 \times 02, \mathrm{DB}[7]=0=$ negative, and $\mathrm{DB}[6: 0]=0000010=2$.

For Memory Map Address $0 \times 03, \mathrm{DB}[7: 0]=10110000=176 \times 2^{-8}=$ 0.6875 , and therefore, tolerance $=-2.6875 \%$, and $R_{A B}=9.731 \mathrm{k} \Omega$.

## PROGRAMMING THE POTENTIOMETER DIVIDER Voltage Output Operation

The digital potentiometer easily generates a voltage divider at wiper-to- B and wiper-to- A that is proportional to the input voltage at A to B , as shown in Figure 43.


Connecting Terminal A to 5 V and Terminal B to ground produces an output voltage at the Wiper W to Terminal B ranging from 0 V to 5 V . The general equation defining the output voltage at $\mathrm{V}_{\mathrm{w}}$ with respect to ground for any valid input voltage applied to Terminal A and Terminal B is

$$
\begin{equation*}
V_{W}(D)=\frac{R_{W B}(D)}{R_{A B}} \times V_{A}+\frac{R_{A W}(D)}{R_{A B}} \times V_{B} \tag{7}
\end{equation*}
$$

where:
$R_{W B}(D)$ can be obtained from Equation 1 and Equation 2. $R_{A W}(D)$ can be obtained from Equation 3 and Equation 4.
Operation of the digital potentiometer in the divider mode results in a more accurate operation over temperature. Unlike the rheostat mode, the output voltage is dependent mainly on the ratio of the internal resistors, $\mathrm{R}_{\mathrm{Aw}}$ and $\mathrm{R}_{\mathrm{wb}}$, and not the absolute values. Therefore, the temperature drift reduces to $5 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$.

Table 19. End-to-End Resistance Tolerance Bytes

| Memory Map Address | Data Byte |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | DB7 | DB6 | DB5 | DB4 | DB3 | DB2 | DB1 | DB0 |
|  | Sign | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |
| $0 \times 03$ | $2^{-1}$ | $2^{-2}$ | $2^{-3}$ | $2^{-4}$ | $2^{-5}$ | $2^{-6}$ | $2^{-7}$ | $2^{-8}$ |

## TERMINAL VOLTAGE OPERATING RANGE

The AD5121/AD5141 are designed with internal ESD diodes for protection. These diodes also set the voltage boundary of the terminal operating voltages. Positive signals present on Terminal A, Terminal B, or Terminal W that exceed VDD clamped by the forward-biased diode. There is no polarity constraint between $V_{A}, V_{W}$, and $V_{B}$, but they cannot be higher than $V_{D D}$ or lower than $V_{\text {Ss. }}$.


Figure 44. Maximum Terminal Voltages Set by $V_{D D}$ and $V_{S S}$

## POWER-UP SEQUENCE

Because there are diodes to limit the voltage compliance at Terminal A, Terminal B, and Terminal W (see Figure 44), it is important to power up $V_{D D}$ first before applying any voltage to Terminal A, Terminal B, and Terminal W. Otherwise, the diode is forward-biased such that $\mathrm{V}_{\mathrm{DD}}$ is powered unintentionally. The ideal power-up sequence is $\mathrm{V}_{\text {SS }}, \mathrm{V}_{\text {DD }}, \mathrm{V}_{\text {LoGIC }}$, digital inputs, and $V_{A}, V_{B}$, and $V_{w}$. The order of powering $V_{A}, V_{B}, V_{W}$, and digital inputs is not important as long as they are powered after $\mathrm{V}_{\mathrm{ss}}$, $V_{D D}$, and $V_{\text {LoGic. }}$. Regardless of the power-up sequence and the ramp rates of the power supplies, once $V_{\text {Logic }}$ is powered, the power-on preset activates, which restores EEPROM values to the RDAC registers.

## LAYOUT AND POWER SUPPLY BIASING

It is always a good practice to use a compact, minimum lead length layout design. Ensure that the leads to the input are as direct as possible with a minimum conductor length. Ground paths should have low resistance and low inductance. It is also good practice to bypass the power supplies with quality capacitors. Apply low equivalent series resistance (ESR) $1 \mu \mathrm{~F}$ to $10 \mu \mathrm{~F}$ tantalum or electrolytic capacitors at the supplies to minimize any transient disturbance and to filter low frequency ripple. Figure 45 illustrates the basic supply bypassing configuration for the AD5121/AD5141.


Figure 45. Power Supply Bypassing

## AD5121/AD5141

## OUTLINE DIMENSIONS



Figure 46. 16-Lead Lead Frame Chip Scale Package [LFCSP_WQ]
$3 \mathrm{~mm} \times 3 \mathrm{~mm}$ Body, Very Very Thin Quad
(CP-16-22)
Dimensions shown in millimeters

ORDERING GUIDE

| Model ${ }^{1,2}$ | $\mathrm{R}_{\text {AB }}(\mathrm{k} \Omega$ ) | Resolution | Interface | Temperature Range | Package Description | Package Option | Branding |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AD5121BCPZ10-RL7 | 10 | 128 | SPI/ $/{ }^{2} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16-Lead LFCSP_WQ | CP-16-22 | DHE |
| AD5121BCPZ100-RL7 | 100 | 128 | SPI/ $/ 1^{2} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16-Lead LFCSP_WQ | CP-16-22 | DHF |
| AD5141BCPZ10-RL7 | 10 | 256 | SPI/ $/{ }^{2} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16-Lead LFCSP_WQ | CP-16-22 | DHC |
| AD5141BCPZ100-RL7 <br> EVAL-AD5141DBZ | 100 | 256 | SPI/ $/{ }^{2} \mathrm{C}$ | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 16-Lead LFCSP_WQ <br> Evaluation Board | CP-16-22 | DHD |

[^2]$\square$
Data Sheet
NOTES

## NOTES



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

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[^0]:    ${ }^{1} \mathrm{X}=$ don't care.

[^1]:    ${ }^{1} \mathrm{X}=$ don't care.

[^2]:    ${ }^{1} \mathrm{Z}=$ RoHS Compliant Part
    ${ }^{2}$ The evaluation board is shipped with the $10 \mathrm{k} \Omega \mathrm{R}_{A B}$ resistor option; however, the board is compatible with both of the available resistor value options.

