## Data Sheet

## FEATURES

3-wire serial interface
2.7 V to 5.5 V single supply
$2.5 \Omega$ on resistance
$0.75 \Omega$ on-resistance flatness
100 pA leakage currents
Single 8-to-1 multiplexer ADG738
Dual 4-to-1 multiplexer ADG739
Power-on reset
TTL/CMOS-compatible
Qualified for automotive applications

## APPLICATIONS

Data acquisition systems
Communication systems
Relay replacement
Audio and video switching

## GENERAL DESCRIPTION

The ADG738 and ADG739 are CMOS analog matrix switches with a serially-controlled 3-wire interface. The ADG738 is an 8 -channel matrix switch, while the ADG739 is a dual 4-channel matrix switch. On resistance is closely matched between switches and very flat over the full signal range.

The ADG738 and ADG739 utilize a 3-wire serial interface that is compatible with SPI ${ }^{\text {m" }}$, QSPI ${ }^{\text {m"w }}$, MICROWIRE ${ }^{\bullet}$, and some DSP interface standards. The output of the input shift register, DOUT, enables a number of these parts to be daisy-chained. On power-up, the internal input shift register contains all zeros and all switches are in the off state.

Each switch conducts equally well in both directions when on, making these parts suitable for both multiplexing and demultiplexing applications. As each switch is turned on or off by a separate bit, these parts can also be configured as a type of switch array, where any, all, or none of the eight switches may be closed at any time. The input signal range extends to the supply rails.

FUNCTIONAL BLOCK DIAGRAMS


Figure 1.


All channels exhibit break-before-make switching action, preventing momentary shorting when switching channels.
The ADG738 and ADG739 are available in 16-lead TSSOP packages.

## PRODUCT HIGHLIGHTS

1. 3-Wire Serial Interface.
2. Single Supply Operation. The ADG738/ADG739 are fully specified and guaranteed with 3 V and 5 V supply rails.
3. Low On Resistance, $2.5 \Omega$ typical.
4. Any configuration of switches may be on or off at any one time.
5. Guaranteed Break-Before-Make Switching Action.
6. Small 16-lead TSSOP Package.

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## SPECIFICATIONS

$\mathrm{V}_{\mathrm{DD}}=5 \mathrm{~V} \pm 10 \%, \mathrm{GND}=0 \mathrm{~V}$, unless otherwise noted.
Table 1.

| Parameter | $25^{\circ} \mathrm{C}$ | B Version $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $\begin{gathered} \text { W Version } \\ -40^{\circ} \mathrm{C} \text { to }+105^{\circ} \mathrm{C} \end{gathered}$ | Unit | Test Conditions/Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ANALOG SWITCH <br> Analog Signal Range On Resistance ( $\mathrm{R}_{\mathrm{ON}}$ ) <br> On-Resistance Match Between Channels ( $\Delta \mathrm{R}_{\mathrm{ON}}$ ) <br> On-Resistance Flatness ( $\mathrm{R}_{\text {FLAt(ON) }}$ ) | 2.5 <br> 4.5 $0.75$ | $\begin{aligned} & 5 \\ & 0.4 \\ & 0.8 \\ & 1.2 \end{aligned}$ | 0 V to $\mathrm{V}_{\mathrm{DD}}$ <br> 6 <br> 1 <br> 1.5 | V <br> $\Omega$ typ <br> $\Omega$ max <br> $\Omega$ typ <br> $\Omega$ max <br> $\Omega$ typ <br> $\Omega$ max | $\mathrm{V}_{\mathrm{S}}=0 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{DD}}, \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA}$; see Figure 19 $V_{S}=0 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD}}, \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA}$ $\mathrm{V}_{\mathrm{S}}=0 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD}}, \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA}$ |
| LEAKAGE CURRENTS <br> Source Off Leakage $I_{S}$ (Off) <br> Drain Off Leakage $I_{D}$ (Off) <br> Channel On Leakage $\mathrm{I}_{\mathrm{D}}, \mathrm{I}_{\mathrm{S}}(\mathrm{On})$ | $\begin{aligned} & \pm 0.01 \\ & \pm 0.1 \\ & \pm 0.01 \\ & \pm 0.1 \\ & \pm 0.01 \\ & \pm 0.1 \end{aligned}$ | $\pm 0.3$ $\pm 1$ $\pm 1$ | $\begin{aligned} & \pm 0.6 \\ & \pm 1.3 \\ & \pm 1.3 \end{aligned}$ | nA typ <br> nA max <br> nA typ <br> nA max <br> nA typ <br> nA max | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{D}}=4.5 \mathrm{~V} / 1 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}}=1 \mathrm{~V} / 4.5 \mathrm{~V} \text {; see Figure } 20 \\ & \mathrm{~V}_{\mathrm{D}}=4.5 \mathrm{~V} / 1 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}}=1 \mathrm{~V} / 4.5 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{D}}=\mathrm{V}_{\mathrm{S}}=1 \mathrm{~V} / 4.5 \mathrm{~V} \text {, see Figure } 21 \end{aligned}$ |
| DIGITAL INPUTS <br> Input High Voltage, $\mathrm{V}_{\text {INH }}$ <br> Input Low Voltage, $\mathrm{V}_{\mathrm{INL}}$ <br> Input Current, $\mathrm{I}_{\mathrm{INL} \text { or } \mathrm{I}_{\mathrm{NH}}, ~}^{\text {In }}$ <br> $\mathrm{C}_{\mathrm{IN}^{N}}$, Digital Input Capacitance | $\begin{aligned} & 0.005 \\ & 3 \end{aligned}$ | $\pm 0.1$ | $\begin{gathered} 2.4 \\ 0.8 \\ \pm 0.1 \end{gathered}$ | $V$ min <br> $V$ max <br> $\mu \mathrm{A}$ typ <br> $\mu \mathrm{A}$ max <br> pF typ | $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {INL }}$ or $\mathrm{V}_{\text {INH }}$ |
| DIGITAL OUTPUT <br> Output Low Voltage <br> $\mathrm{C}_{\text {out }}$, Digital Output Capacitance | 4 | 0.4 |  | max <br> pF typ | $\mathrm{I}_{\text {SINK }}=6 \mathrm{~mA}$ |
| DYNAMIC CHARACTERISTICS ${ }^{1}$ <br> $\mathrm{t}_{\mathrm{ON}}$ <br> $t_{\text {OFF }}$ <br> Break-Before-Make Time Delay, $\mathrm{t}_{\mathrm{D}}$ <br> Charge Injection <br> Off Isolation <br> Channel-to-Channel Crosstalk <br> -3 dB Bandwidth <br> ADG738 <br> ADG739 <br> $\mathrm{C}_{\mathrm{s}}$ (Off) <br> $C_{D}$ (Off) <br> ADG738 <br> ADG739 <br> $C_{D}, C_{S}$ (On) <br> ADG738 <br> ADG739 | 20 <br> 10 <br> 9 <br> $\pm 3$ <br> $-55$ <br> -75 <br> -55 <br> $-75$ <br> 65 <br> 100 <br> 13 <br> 85 <br> 42 <br> 96 <br> 48 | 32 17 1 | 35 <br> 20 <br> 1 | ns typ <br> ns max <br> ns typ <br> ns max <br> ns typ <br> ns min <br> pC typ <br> dB typ <br> dB typ <br> dB typ <br> dB typ <br> MHz typ <br> MHz typ <br> pF typ <br> pF typ <br> pF typ <br> pF typ <br> pF typ | $\begin{aligned} & R_{L}=300 \Omega, C_{L}=35 \mathrm{pF} \text {, see Figure 22; } \mathrm{V}_{\mathrm{S} 1}=3 \mathrm{~V} \\ & \mathrm{R}_{\mathrm{L}}=300 \Omega, C_{L}=35 \mathrm{pF} \text {, see Figure 22; } \mathrm{V}_{\mathrm{S} 1}=3 \mathrm{~V} \\ & \mathrm{R}_{\mathrm{L}}=300 \Omega, C_{L}=35 \mathrm{pF} \text {; } \\ & \mathrm{V}_{\mathrm{S} 1}=\mathrm{V}_{58}=3 \mathrm{~V} \text {, see Figure } 22 \\ & \mathrm{~V}_{S}=2.5 \mathrm{~V}, R_{S}=0 \Omega, C_{L}=1 \mathrm{nF} \text {; see Figure } 23 \\ & R_{L}=50 \Omega, C_{L}=5 \mathrm{pF}, \mathrm{f}=10 \mathrm{MHz} \\ & R_{L}=50 \Omega, C_{L}=5 \mathrm{pF}, f=1 \mathrm{MHz} \text {; see Figure } 25 \\ & R_{L}=50 \Omega, C_{L}=5 \mathrm{pF}, f=10 \mathrm{MHz} \\ & R_{L}=50 \Omega, C_{L}=5 \mathrm{pF}, f=1 \mathrm{MHz} \text {; see Figure } 24 \\ & R_{L}=50 \Omega, C_{L}=5 \mathrm{pF} \text {, see Figure } 25 \end{aligned}$ |
| POWER REQUIREMENTS $I_{D D}$ | 10 | 20 | 20 | $\mu \mathrm{A}$ typ $\mu \mathrm{A}$ max | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V} \\ & \text { Digital Inputs }=0 \mathrm{~V} \text { or } 5.5 \mathrm{~V} \end{aligned}$ |

[^0]$\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V} \pm 10 \%, \mathrm{GND}=0 \mathrm{~V}$, unless otherwise noted.
Table 2.

| Parameter | $25^{\circ} \mathrm{C}$ | $\begin{gathered} \text { B Version } \\ -40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \end{gathered}$ | $\begin{gathered} \text { W Version } \\ -40^{\circ} \mathrm{C} \text { to }+105^{\circ} \mathrm{C} \end{gathered}$ | Unit | Test Conditions/Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ANALOG SWITCH <br> Analog Signal Range <br> On Resistance ( $\mathrm{RoN}_{\text {on }}$ ) <br> On-Resistance Match Between Channels ( $\Delta \mathrm{R}_{\mathrm{oN}}$ ) <br> On-Resistance Flatness ( $\mathrm{R}_{\text {flationn }}$ ) | 6 <br> 11 $3.5$ | $\begin{aligned} & 12 \\ & 0.4 \\ & 1.2 \end{aligned}$ | 0 V to $\mathrm{V}_{\mathrm{DD}}$ <br> 16 <br> 1.4 | V <br> $\Omega$ typ <br> $\Omega$ max <br> $\Omega$ typ <br> $\Omega$ max <br> $\Omega$ typ | $\begin{aligned} & \mathrm{V}_{\mathrm{S}}=0 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD},} \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA} \text {; see Figure } 19 \\ & \mathrm{~V}_{\mathrm{S}}=0 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD}} \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{S}}=0 \mathrm{~V} \text { to } \mathrm{V}_{\mathrm{DD}} \mathrm{I}_{\mathrm{S}}=10 \mathrm{~mA} \end{aligned}$ |
| LEAKAGE CURRENTS <br> Source Off Leakage $\mathrm{I}_{5}$ (Off) <br> Drain Off Leakage $\mathrm{I}_{\mathrm{D}}$ (Off) <br> Channel On Leakage $\mathrm{I}_{\mathrm{D}}, \mathrm{I}_{\mathrm{S}}(\mathrm{On})$ | $\begin{aligned} & \pm 0.01 \\ & \pm 0.1 \\ & \pm 0.01 \\ & \pm 0.1 \\ & \pm 0.01 \\ & \pm 0.1 \end{aligned}$ | $\begin{aligned} & \pm 0.3 \\ & \pm 1 \\ & \pm 1 \end{aligned}$ | $\begin{aligned} & \pm 0.6 \\ & \pm 1.3 \\ & \pm 1.3 \end{aligned}$ | nA typ nA max nA typ nA max nA typ nA max | $\begin{aligned} & \mathrm{V}_{\mathrm{D}}=3.3 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{S}}=3 \mathrm{~V} / 1 \mathrm{~V}, \mathrm{~V}_{\mathrm{D}}=1 \mathrm{~V} / 3 \mathrm{~V} \text {; see Figure } 20 \\ & \mathrm{~V}_{\mathrm{D}}=3 \mathrm{~V} / 1 \mathrm{~V}, \mathrm{~V}_{\mathrm{D}}=1 \mathrm{~V} / 3 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{D}}=\mathrm{V}_{\mathrm{S}}=3 \mathrm{~V} / 1 \mathrm{~V} \text {, see Figure } 21 \end{aligned}$ |
| DIGITAL INPUTS <br> Input High Voltage, $\mathrm{V}_{\text {INH }}$ <br> Input Low Voltage, $\mathrm{V}_{\mathrm{INL}}$ <br> Input Current, $\mathrm{I}_{\mathbb{N L}}$ or $\mathrm{I}_{\mathrm{INH}}$ <br> $\mathrm{C}_{\text {IN }}$, Digital Input Capacitance | $\begin{aligned} & 0.005 \\ & 3 \end{aligned}$ | $\pm 0.1$ | $\begin{aligned} & 2.0 \\ & 0.4 \\ & \pm 0.1 \end{aligned}$ | $V$ min <br> $V$ max <br> $\mu \mathrm{A}$ typ <br> $\mu \mathrm{A}$ max <br> pF typ | $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {INL }}$ or $\mathrm{V}_{\text {INH }}$ |
| DIGITAL OUTPUT <br> Output Low Voltage <br> $\mathrm{C}_{\text {out, }}$, Digital Output Capacitance | 4 | 0.4 |  | $\begin{aligned} & \max \\ & \mathrm{pF} \text { typ } \end{aligned}$ | $\mathrm{I}_{\text {SINK }}=6 \mathrm{~mA}$ |
| DYNAMIC CHARACTERISTICS ${ }^{1}$ <br> $\mathrm{t}_{\mathrm{ON}}$ <br> $\mathrm{t}_{\text {off }}$ <br> Break-Before-Make Time Delay, $\mathrm{t}_{\mathrm{D}}$ <br> Charge Injection Off Isolation <br> Channel-to-Channel Crosstalk <br> -3 dB Bandwidth <br> ADG738 <br> ADG739 <br> $\mathrm{C}_{\mathrm{S}}$ (Off) <br> $C_{D}$ (Off) <br> ADG738 <br> ADG739 <br> $\mathrm{C}_{\mathrm{D}}, \mathrm{C}_{5}$ (On) <br> ADG738 <br> ADG739 | 40 <br> 14 <br> 12 <br> $\pm 3$ <br> -55 <br> -75 <br> -55 <br> $-75$ <br> 65 <br> 100 <br> 13 <br> 85 <br> 42 <br> 96 <br> 48 | 70 25 | 75 $40$ | ns typ ns max ns typ ns max ns typ ns min pC typ dB typ dB typ dB typ dB typ <br> MHz typ <br> MHz typ <br> pF typ <br> pF typ <br> pF typ <br> pF typ <br> pF typ |  |
| POWER REQUIREMENTS $I_{D D}$ | 10 | 20 | 20 | $\mu A$ typ $\mu \mathrm{A}$ max | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V} \\ & \text { Digital Inputs }=0 \mathrm{~V} \text { or } 3.3 \mathrm{~V} \end{aligned}$ |

[^1]
## Data Sheet

TIMING CHARACTERISTICS
$\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}$ to 5.5 V . All specifications $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$, unless otherwise noted.
Table 3.

|  | Limit at $\mathbf{T}_{\text {MIN }} \mathbf{T}_{\text {MAX }}$ |  |  |
| :--- | :--- | :--- | :--- |
| Parameter ${ }^{1,2}$ | Max | Unit | Test Conditions/Comments |
|  | Min | 30 | MHz |
| $\mathrm{f}_{\text {SCLK }}$ | 33 | SCLK cycle frequency |  |
| $\mathrm{t}_{1}$ | 13 | ns | SCLK cycle time |
| $\mathrm{t}_{2}$ | 13 | ns | SCLK high time |
| $\mathrm{t}_{3}$ | 0 | ns | SCLK low time |
| $\mathrm{t}_{4}$ | 5 | ns | SYNC to SCLK active edge setup time |
| $\mathrm{t}_{5}$ | 4.5 | ns | Data setup time |
| $\mathrm{t}_{6}$ | 0 | ns | Data hold time |
| $\mathrm{t}_{7}$ | 33 | ns | SCLK falling edge to $\overline{\text { SYNC } \text { rising edge }}$ |
| $\mathrm{t}_{8}$ | 20 | ns | Minimum $\overline{\text { SYNC }}$ high time |
| $\mathrm{t}_{9}{ }^{3}$ |  | ns min | SCLK rising edge to DOUT valid |

[^2]
${ }^{1}$ DATA FROM LAST WRITE CYCLE.
Figure 3. 3-Wire Serial Interface Timing Diagram

## ABSOLUTE MAXIMUM RATINGS

$\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ unless otherwise noted.
Table 4.

| Parameter | Rating |
| :---: | :---: |
| VDD to GND | -0.3 V to +7 V |
| Analog, Digital Inputs ${ }^{1}$ | -0.3 V to V DD +0.3 V or 30 mA , Whichever Occurs First |
| Peak Current, S or D | $\begin{aligned} & 100 \mathrm{~mA} \\ & \text { (Pulsed at } 1 \mathrm{~ms}, 10 \% \\ & \text { Duty Cycle Max) } \end{aligned}$ |
| Continuous Current, Each S | 30 mA |
| Continuous Current D |  |
| ADG739 | 80 mA |
| ADG738 | 120 mA |
| Operating Temperature Range |  |
| Industrial (B Version) | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Industrial (W Version) | $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Junction Temperature | $150^{\circ} \mathrm{C}$ |
| TSSOP Package |  |
| $\theta_{\mathrm{JA}}$ Thermal Impedance | $150.4{ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\text {ıc }}$ Thermal Impedance | $27.6^{\circ} \mathrm{C} / \mathrm{W}$ |
| Lead Temperature, Soldering | As per JEDEC J-STD-020 |

[^3]Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ESD CAUTION

|  | ESD (electrostatic discharge) sensitive device. <br> Charged devices and circuit boards can discharge <br> without detection. Although this product features <br> patented or proprietary protection circuitry, damage <br> may occur on devices subjected to high energy ESD. <br> Therefore, proper ESD precautions should be taken to <br> avoid performance degradation or loss of functionality. |
| :--- | :--- |

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS



Figure 4. ADG738 Pin Configuration

Table 5. ADG738 Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | SCLK | Serial Clock Input. Data is clocked into the input shift register on the falling edge of the serial clock input. These devices can accommodate serial input rates of up to 30 MHz . |
| 2 | $\overline{\text { RESET }}$ | Active Low Control Input. This pin clears the input register and turns all switches to the off condition. |
| 3 | DIN | Serial Data Input. Data is clocked into the 8-bit input register on the falling edge of the serial clock input. |
| 4, 5, 6, 7 | S1, S2, S3, S4 | Source. May be an input or output. |
| 8 | D | Drain. May be an input or output. |
| 9, 10, 11, 12 | S8, S7, S6, S5 | Source. May be an input or output. |
| 13 | $\mathrm{V}_{\mathrm{DD}}$ | Power Supply Input. These parts can be operated from a supply of 2.7 V to 5.5 V . |
| 14 | GND | Ground Reference. |
| 15 | DOUT | Data Output. This allows a number a parts to be daisy-chained. Data is clocked out of the input shift register on the rising edge of SCLK. This is an open drain output, which should be pulled to the supply with an external resistor. |
| 16 | $\overline{\text { SYNC }}$ | Active Low Control Input. This is the frame synchronization signal for the input data. When $\overline{\text { SYNC }}$ goes low, it powers on the SCLK and DIN buffers and the input shift register is enabled. Data is transferred on the falling edges of the following clocks. Taking SYNC high updates the switch conditions. |



Figure 5. ADG739 Pin Configuration
Table 6. ADG739 Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | SCLK | Serial Clock Input. Data is clocked into the input shift register on the falling edge of the serial clock input. These devices can accommodate serial input rates of up to 30 MHz . |
| 2 | $\overline{\text { SYNC }}$ | Active Low Control Input. This is the frame synchronization signal for the input data. When $\overline{\text { SYNC }}$ goes low, it powers on the SCLK and DIN buffers and the input shift register is enabled. Data is transferred on the falling edges of the following clocks. Taking $\overline{\text { SYNC }}$ high updates the switch conditions. |
| 3 | DIN | Serial Data Input. Data is clocked into the 8-bit input register on the falling edge of the serial clock input. |
| 4, 5, 6, 7 | S1A, S2A, S3A, S4A | Source. May be an input or output. |
| 8,9 | DA, DB | Drain. May be an input or output. |
| 10, 11, 12, 13 | S4B, S3B, S2B, S1B | Source. May be an input or output. |
| 14 | $V_{\text {D }}$ | Power Supply Input. These parts can be operated from a supply of 2.7 V to 5.5 V . |
| 15 | GND | Ground Reference. |
| 16 | DOUT | Data Output. This allows a number a parts to be daisy-chained. Data is clocked out of the input shift register on the rising edge of SCLK. This is an open drain output, which should be pulled to the supply with an external resistor. |

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 6. On Resistance as a Function of $V_{D}\left(V_{S}\right)$


Figure 7. On Resistance as a Function of $V_{D}\left(V_{s}\right)$ for Different Temperatures, $V_{D D}=5 \mathrm{~V}$


Figure 8. On Resistance as a Function of $V_{D}\left(V_{S}\right)$ for Different Temperatures, $V_{D D}=3 V$


Figure 9. Leakage Currents as a Function of $V_{D}\left(V_{S}\right), V_{D D}=5 \mathrm{~V}$


Figure 10. Leakage Currents as a Function of $V_{D}\left(V_{S}\right), V_{D D}=3 \mathrm{~V}$


Figure 11. Leakage Currents as a Function of Temperature, $V_{D D}=5 \mathrm{~V}$


Figure 12. Leakage Currents as a Function of Temperature, $V_{D D}=3 \mathrm{~V}$


Figure 13. Input Currents vs. Switching Frequency


Figure 14. Charge Injection vs. Source Voltage


Figure 15. $T_{\text {ON }} T_{\text {OFF }}$ Times vs. Temperature


Figure 16. Off Isolation vs. Frequency


Figure 17. Crosstalk vs. Frequency


Figure 18. On Response vs. Frequency

## TEST CIRCUITS



Figure 19. On Resistance


Figure 20. $I_{D}$ (Off), $I_{S}$ (Off)

*SIMILAR CONNECTION FOR ADG739.


Figure 21. $I_{S}, I_{D}(O n)$

再


10758-029

*SIMILAR CONNECTION FOR ADG739.
$\overline{\text { SYNC }}$


Figure 23. Charge Injection


Figure 24. Channel-to-Channel Crosstalk

*SIMILAR CONNECTION FOR ADG739.
S1 IS SWITCHED OFF FOR OFF ISOLATION MEASUREMENTS AND ON FOR BANDWIDTH MEASUREMENTS
OFF ISOLATION $=20$ LOG $_{10}\left(V_{\text {OUT }} / V_{S}\right)$
INSERTION LOSS $=20$ LOG $_{10}\left(\frac{V_{\text {OUT WITH SWITCH }}}{\mathrm{V}_{\text {OUT WITHOUT SWITCH }}}\right)$
Figure 25. Off Isolation and Bandwidth

## ADG738/ADG739

## TERMINOLOGY

$V_{\text {DD }}$
Most positive power supply potential.
$\mathrm{I}_{\mathrm{DD}}$
Positive supply current.
GND
Ground (0 V) reference.
S
Source terminal. May be an input or output.
D
Drain terminal. May be an input or output.
$V_{D}\left(V_{s}\right)$
Analog voltage on Terminal D, Terminal S.
$\mathbf{R}_{\text {ON }}$
Ohmic resistance between D and S.
$\Delta R_{\text {ON }}$
On resistance match between any two channels, that is, $\mathrm{R}_{\mathrm{ON}} \max$
$-R_{\text {ON }} \min$.
$\mathbf{R}_{\text {flat(on) }}$
Flatness is defined as the difference between the maximum and minimum value of on resistance as measured over the specified analog signal range.
$I_{s}$ (Off)
Source leakage current with the switch off.
$I_{D}$ (Off)
Drain leakage current with the switch off.
$\mathrm{I}_{\mathrm{D}}, \mathrm{I}_{\mathrm{S}}(\mathrm{On})$
Channel leakage current with the switch on.
$\mathrm{V}_{\text {INL }}$
Maximum input voltage for Logic 0 .
$\mathrm{V}_{\text {INH }}$
Minimum input voltage for Logic 1.
$\mathrm{I}_{\mathrm{INL}}\left(\mathrm{I}_{\mathrm{INH}}\right)$
Input current of the digital input.
$\mathrm{C}_{\mathrm{S}}$ (Off)
Off switch source capacitance. Measured with reference to ground.
$C_{D}$ (Off)
Off switch drain capacitance. Measured with reference to ground.
$\mathrm{C}_{\mathrm{D}}, \mathrm{C}_{\mathrm{s}}$ (On)
On switch capacitance. Measured with reference to ground.
$\mathrm{C}_{\text {In }}$
Digital input capacitance.
$t_{\text {ON }}$
Delay time between the $50 \%$ and $90 \%$ points of the $\overline{\text { SYNC }}$ rising edge and the switch on condition.
$t_{\text {off }}$
Delay time between the $50 \%$ and $90 \%$ points of the $\overline{\text { SYNC }}$ rising edge and the switch off condition.
$t_{D}$
Off time measured between the $80 \%$ points of both switches when switching from one switch to another.

## Charge Injection

A measure of the glitch impulse transferred from the digital input to the analog output during switching.

## Off Isolation

A measure of unwanted signal coupling through an off switch.

## Crosstalk

A measure of unwanted signal which is coupled through from one channel to another as a result of parasitic capacitance.

## Bandwidth

The frequency at which the output is attenuated by 3 dB .
On Response
The frequency response of the on switch.

## Insertion Loss

The loss due to the on resistance of the switch.

## THEORY OF OPERATION

The ADG738 and ADG739 are serially controlled, 8-channel and dual 4 -channel matrix switches, respectively. While providing the normal multiplexing and demultiplexing functions, these parts also provide the user with more flexibility as to where their signal may be routed. Each bit of the 8 -bit serial word corresponds to one switch of the part. A Logic 1 in the particular bit position turns on the switch, while a Logic 0 turns the switch off. Because each switch is independently controlled by an individual bit, this provides the option of having any, all, or none of the switches on. This feature may be particularly useful in the demultiplexing application where the user may wish to direct one signal from the drain to a number of outputs (sources). Take care, however, in the multiplexing situation where a number of inputs may be shorted together (separated only by the small on resistance of the switch).

When changing the switch conditions, a new 8 -bit word is written to the input shift register. Some of the bits may be the same as the previous write cycle, as the user may not wish to change the state of some switches. To minimize glitches on the output of these switches, the part cleverly compares the state of switches from the previous write cycle. If the switch is already in the on condition, and is required to stay on, there will be minimal glitches on the output of the switch.

## POWER-ON RESET

During device power-up, all switches will be in the off condition and the internal input shift register is filled with zeros and remains so until a valid write takes place.

## SERIAL INTERFACE

The ADG738 and ADG739 have a 3-wire serial interface (SYNC, SCLK, and DIN), which is compatible with SPI, QSPI, MICROWIRE interface standards and most DSPs. Figure 3 shows the timing diagram of a typical write sequence.
Data is written to the 8-bit input shift register via DIN under the control of the $\overline{\text { SYNC }}$ and SCLK signals. Data may be written to the input shift register in more or less than eight bits. In each case, the input shift register retains the last eight bits that were written.
When SYNC goes low, the input shift register is enabled. Data from DIN is clocked into the input shift register on each falling edge of SCLK. Each bit of the 8-bit word corresponds to one of the eight switches. Figure 26 shows the contents of the input shift register. Data appears on the DOUT pin on the rising edge of SCLK suitable for daisy-chaining, delayed, of course, by eight bits. When all eight bits have been written into the shift register, the $\overline{\text { SYNC }}$ line is brought high again. The switches are updated with the new configuration and the input shift register is disabled. With $\overline{\text { SYNC }}$ held high, any further data or noise on the DIN line has no effect on the shift register.


Figure 26. Input Shift Register Contents

## MICROPROCESSOR INTERFACING

Microprocessor interfacing to the ADG738/ADG739 is via a serial bus that uses a standard protocol compatible with microcontrollers and DSP processors. The communications channel is a 3 -wire (minimum) interface consisting of a clock signal, a data signal, and a synchronization signal. The ADG738/ADG739 requires an 8-bit data word with data valid on the falling edge of SCLK.
Data from the previous write cycle is available on the DOUT pin. The following sections illustrate simple 3-wire interfaces with popular microcontrollers and DSPs.

## ADSP-21xx TO ADG738/ADG739

An interface between the ADG738/ADG739 and the ADSP21xx is shown in Figure 27. In the interface example shown, SPORT0 is used to transfer data to the matrix switch. The SPORT control register should be configured as follows: internal clock operation, alternate framing mode; active low framing signal.
Transmission is initiated by writing a word to the Tx register after the SPORT has been enabled. As the data is clocked out of the DSP on the rising edge of SCLK, no glue logic is required to interface the DSP to the matrix switch. The update of each switch condition takes place automatically when TFS is taken high.

*ADDITIONAL PINS OMITTED FOR CLARITY.

Figure 27. ADSP-21xx to ADG738/ADG739 Interface

## 8051 INTERFACE TO ADG738/ADG739

A serial interface between the ADG738/ADG739 and the 8051 is shown in Figure 28. TXD of the 8051 drives SCLK of the ADG738/ADG739, while RXD drives the serial data line, DIN. P3.3 is a bit-programmable pin on the serial port and is used to drive $\overline{\mathrm{SYNC}}$.

The 8051 provides the LSB of its SBUF register as the first bit in the data stream. The user has to ensure that the data in the SBUF register is arranged correctly as the switch expects MSB first.
When data is to be transmitted to the matrix switch, P3.3 is taken low. Data on RXD is clocked out of the microcontroller on the rising edge of TXD and is valid on the falling edge. As a result no glue logic is required between the ADG738/ADG739 and microcontroller interface.

*ADDITIONAL PINS OMITTED FOR CLARITY.
Figure 28.8051 Interface to ADG738/ADG739

## MC68HC11 INTERFACE TO ADG738/ADG739

Figure 29 shows an example of a serial interface between the ADG738/ADG739 and the MC68HC11 microcontroller. SCK of the 68 HC 11 drives the SCLK of the matrix switch, while the MOSI output drives the serial data line, DIN. $\overline{\text { SYNC }}$ is driven from one of the port lines, in this case PC7.


Figure 29. MC68HC11 Interface to ADG738/ADG739
The 68 HC 11 is configured for master mode; $\mathrm{MSTR}=1, \mathrm{CPOL}$ $=0$, and CPHA $=1$. When data is transferred to the part, PC7 is taken low, data is transmitted MSB first. Data appearing on the MOSI output is valid on the falling edge of SCK.
If the user wishes to verify the data previously written to the input shift register, the DOUT line could be connected to MISO of the MC68HC11, and with SYNC low, the input shift register would clock data out on the rising edges of SCLK.

## APPLICATIONS INFORMATION

## EXPAND THE NUMBER OF SELECTABLE SERIAL DEVICES USING AN ADG739

The dual 4-channel ADG739 multiplexer can be used to multiplex a single chip select line to provide chip selects for up to four devices on the SPI bus. Figure 30 illustrates the ADG739 in such a typical configuration. All devices receive the same serial clock and serial data, but only one device receives the $\overline{\text { SYNC }}$ signal at any one time. The ADG739 is a serially controlled device also. One bit programmable pin of the microcontroller is used to enable the ADG739 via $\overline{\text { SYNC2 }}$, while another bit programmable pin is used as the chip select for the other serial devices, $\overline{\text { SYNC1 }}$. Driving $\overline{\text { SYNC2 }}$ low enables changes to be made to the addressed serial devices. By bringing $\overline{\text { SYNC1 }}$ low, the selected serial device hanging from the SPI bus is enabled and data will be clocked into its input shift register on the falling edges of SCLK. The convenient design of the matrix switch allows for different combinations of the four serial devices to be addressed at any one time. If more devices need to be addressed via one chip select line, the ADG738 is an 8 -channel device and would allow further expansion of the chip select scheme. There may be some digital feedthrough from the digital input lines because SCLK and DIN are permanently connected to each device. Using a burst clock minimizes the effects of digital feedthrough on the analog channels.


Figure 30. Addressing Multiple Serial Devices Using an ADG739

## DAISY-CHAINING MULTIPLE ADG738S

A number of ADG738 matrix switches may be daisy-chained simply by using the DOUT pin. DOUT is an open-drain output that should be pulled to the supply with an external resistor. Figure 31 shows a typical implementation. The $\overline{\text { SYNC }}$ pin of all three parts in the example are tied together. When $\overline{\text { SYNC }}$ is brought low, the input shift registers of all parts are enabled, data is written to the parts via DIN, and clocked through the shift registers. When the transfer is complete, $\overline{\text { SYNC }}$ is brought high and all switches are updated simultaneously. Further shift registers may be added in series.


10758-024
Figure 31. Multiple ADG739 Devices in a Daisy-Chained Configuration

## OUTLINE DIMENSIONS



Figure 32. 16-Lead Thin Shrink Small Outline Package [TSSOP]
(RU-16)
Dimensions shown in millimeters

## ORDERING GUIDE

| Model $^{1,2}$ | Temperature Range | Package Description | Package Option |
| :--- | :--- | :--- | :--- |
| ADG738BRU | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG738BRUZ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG738BRUZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG738BRUZ-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG738WBRUZ-REEL | $-40^{\circ} \mathrm{C}$ to $+105^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG739BRU | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG739BRU-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG739BRUZ | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16-Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG739BRUZ-REEL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16-Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |
| ADG739BRUZ-REEL7 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16-Lead Thin Shrink Small Outline Package [TSSOP] | RU-16 |

${ }^{1} Z=$ RoHS Compliant Part.
${ }^{2} \mathrm{~W}=$ Qualified for Automotive Applications.

## AUTOMOTIVE PRODUCTS

The ADG738W model is available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that this automotive model may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

NOTES

## NOTES



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- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
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- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.


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[^0]:    ${ }^{1}$ Guaranteed by design, not subject to production test.

[^1]:    ${ }^{1}$ Guaranteed by design, not subject to production test.

[^2]:    ${ }^{1}$ See Figure 3.
    ${ }^{2}$ All input signals are specified with $\operatorname{tr}=\mathrm{tf}=5 \mathrm{~ns}\left(10 \%\right.$ to $90 \%$ of $\left.\mathrm{V}_{\mathrm{DD}}\right)$ and timed from a voltage level of $\left(\mathrm{V}_{\mathrm{IL}}+\mathrm{V}_{\mathrm{IH}}\right) / 2$.
    ${ }^{3} \mathrm{C}_{\mathrm{L}}=20 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega$.

[^3]:    ${ }^{1}$ Overvoltages at $\mathrm{IN}, \mathrm{S}$, or D are clamped by internal diodes. Limit current to the maximum ratings given.

