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#### ISO7841, ISO7841F

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# ISO7841 ISO7841F High-Performance, 8000 V<sub>PK</sub> Reinforced Quad-Channel Digital Isolator

Technical

Documents

#### 1 Features

- Signaling Rate: Up to 100 Mbps
- Wide Supply Range: 2.25 V to 5.5 V
- 2.25 V to 5.5 V Level Translation
- Wide Temperature Range: -55°C to 125°C
- Low Power Consumption, Typical 1.7 mA per Channel at 1 Mbps
- Low Propagation Delay: 11 ns Typical (5 V Supplies)
- Industry leading CMTI: ±100 kV/µs
- Robust Electromagnetic Compatibility (EMC)
- System-Level ESD, EFT, and Surge Immunity
- Low Emissions
- Isolation Barrier Life: >25 Years
- Wide Body SOIC-16 Package
- Safety and Regulatory Approvals:
  - 8000 V<sub>PK</sub> V<sub>IOTM</sub> and 2121 V<sub>PK</sub> V<sub>IORM</sub> Reinforced Isolation per DIN V VDE V 0884-10 (VDE V 0884-10):2006-12
  - 5.7 kV<sub>RMS</sub> Isolation for 1 Minute per UL 1577
  - CSA Component Acceptance Notice 5A, IEC 60950-1, IEC 60601-1 and IEC 61010-1 End **Equipment Standards**
  - CQC Certification per GB4943.1-2011

# 2 Applications

- Industrial Automation
- Motor Control
- Power Supplies
- Solar Inverters
- Medical Equipment
- Hybrid Electric Vehicles

# 3 Description

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The ISO7841 is a high-performance, quad-channel digital isolator with 8000  $V_{\text{PK}}$  isolation voltage. This reinforced isolation certifications device has according to VDE, CSA, and CQC. The isolator provides high electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVCMOS digital I/Os. Each isolation channel has a logic input and output buffer separated by silicon dioxide (SiO<sub>2</sub>) insulation barrier. This device comes with enable pins which can be used to put the respective outputs in high impedance for multi-master drivina applications and to reduce power consumption. ISO7841 has three forward and one reverse-direction channels. If the input power or signal loss, default output is 'high' for the ISO7841 device and 'low' for the ISO7841F device. See Device *Functional Modes* for further details. Used in conjunction with isolated power supplies, this device prevents noise currents on a data bus or other circuits from entering the local ground and interfering with or damaging sensitive circuitry. Through innovative chip design and layout techniques, electromagnetic compatibility of ISO7841 has been significantly enhanced to ease system-level ESD, EFT, Surge and Emissions compliance. ISO7841 is available in 16-pin SOIC wide-body (DW) package.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
ISO7841 / ISO7841F	SOIC, DW (16)	10.30 mm × 7.50 mm		

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### V<sub>cco</sub> Vcci Isolation Capacitor OUTx INx ENx

GNDO

**Simplified Schematic** 

- GNDI (1) V<sub>CCI</sub> and GNDI are supply and ground connections respectively for the input channels.
- (2) V<sub>CCO</sub> and GNDO are supply and ground connections respectively for the output channels.



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# 4 Revision History

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•	Added device ISO7481F to the datasheet 1
•	Changed the Description to include: " default output is 'high' for the ISO7841 device and 'low' for the ISO7841F device 1
•	Changed the Supply current TYP and MAX values in the <i>Electrical Characteristics, 5 V</i>
•	Changed the Supply current TYP and MAX values in the <i>Electrical Characteristics, 3.3 V</i>
•	Changed From: t <sub>PLH</sub> and t <sub>PHL</sub> at 5.5V To: t <sub>PLH</sub> and t <sub>PHL</sub> at 5.0 V 12
•	Changed Figure 9 14
•	Added the Device I/O Schematics section

#### Changes from Revision A (January 2015) to Revision B

Submit Documentation Feedback

Changes from Revision B (January 2015) to Revision C

•	Changed the document title From: "Quad-Channel Digital Isolator" To: "Quad-Channel 3/1 Digital Isolator" Added Features 2.25 V to 5.5 V Level Translation	
•	Changed <i>Features</i> From: Wide Body SOIC-16 Package To: Wide Body and Extra-Wide Body SOIC-16 Package Options	
•	Changed the Safety and Regulatory Approvals list of <i>Features</i>	
•	Changed the Simplified Schematic and added Notes 1 and 2	1
•	Added the Power Rating table	
•	Changed Figure 7	13
•	Changed Figure 8	13
•	Changed From: V <sub>CC1</sub> To: V <sub>CC1</sub> in Figure 9	14
•	Changed Figure 10	14
•	Changed title From: IEC Insulation and Safety-Related Specifications for DW-16 Package To: <i>Package Insulation</i> and Safety-Related Specifications	16
•	Changed the table in Package Insulation and Safety-Related Specifications	
•	Changed the Test Condition of CTI of the table in Package Insulation and Safety-Related Specifications	



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•	Changed the MIN value of CTI From" > 600 V To: 600 V	. 16
•	Changed title From: DIN V VDE 0884-10 (VDE V 0884-10) and UL 1577 Insulation Characteristics To: Insulation Characteristics	. 17
•	Changed the table in Insulation Characteristics	. 17
•	Changed title From: Regulatory Information (All Certifications Planned)Regulatory Information (All Certifications Planned) To: Regulatory Information	. 18
•	Changed the table in Regulatory Information	. 18
•	Changed the table in Safety Limiting Values, added I <sub>S</sub> DW-16 package options	. 19
•	Deleted INPUT-SIDE and OUTPUT-SIDE from columns 1 and 2 of Table 1	. 20
•	Changed the Application Information section	. 22
•	Changed the Typical Application section	. 22

	Page	
•	Updated device from preview to production data	1



# 5 Pin Configuration and Functions



#### **Pin Functions**

PIN //		1/0	DESCRIPTION	
		1/0	DESCRIPTION	
EN1	7	I	Output enable 1. Output pins on side 1 are enabled when EN1 is high or open and in high- impedance state when EN1 is low.	
EN2	10	I	Output enable 2. Output pins on side 2 are enabled when EN2 is high or open and in high- impedance state when EN2 is low.	
GND1	2		Cround connection for V	
GNDT	8	_	Ground connection for V <sub>CC1</sub>	
GND2	9		Cround connection for V	
GND2	15	_	Ground connection for V <sub>CC2</sub>	
INA	3	I	Input, channel A	
INB	4	I	Input, channel B	
INC	5	I	Input, channel C	
IND	11	I	Input, channel D	
OUTA	14	0	Output, channel A	
OUTB	13	0	Output, channel B	
OUTC	12	0	Output, channel C	
OUTD	6	0	Output, channel D	
V <sub>CC1</sub>	1	_	Power supply, V <sub>CC1</sub>	
V <sub>CC2</sub>	16	_	Power supply, V <sub>CC2</sub>	

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#### Specifications 6

# 6.1 Absolute Maximum Ratings

See (1)

			MIN	MAX	UNIT
V <sub>CC1</sub> , V <sub>CC2</sub>	Supply voltag	ge <sup>(2)</sup>	-0.5	6	V
		INx			
	Voltage	OUTx	-0.5	$V_{CCX} + 0.5^{(3)}$	V
		ENx			
I <sub>O</sub>	Output curre	nt	-15	15	mA
Surge immunity				12.8	kV
TJ	Maximum jur	nction temperature		150	°C
T <sub>stg</sub>	Storage temp	perature	-65	150	°C

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings (1) only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak (2)

voltage values. (3)Maximum voltage must not exceed 6 V

#### 6.2 ESD Ratings

			VALUE	UNIT
		Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	±6000	
V <sub>(ESD)</sub>	) Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101, all $pins^{(2)}$	±1500	V

JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. (1)

JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. (2)

#### 6.3 Recommended Operating Conditions

			MIN	NOM MAX	UNIT
V <sub>CC1</sub> , V <sub>CC2</sub>	Supply voltage		2.25	5.5	5 V
		$V_{CCO}^{(1)} = 5 V$	-4		
I <sub>OH</sub>	High-level output current	$V_{\rm CCO}^{(1)} = 3.3 \ V$	-2		mA
		$V_{\rm CCO}^{(1)} = 2.5 \ V$	-1		
		$V_{\rm CCO}^{(1)} = 5  \rm V$		2	ł
I <sub>OL</sub>	Low-level output current	$V_{\rm CCO}^{(1)} = 3.3 \ V$		2	2 mA
		$V_{\rm CCO}^{(1)} = 2.5 \ V$			
VIH	High-level input voltage		$0.7 \times V_{CCI}$ <sup>(1)</sup>	V <sub>CCI</sub> <sup>(1</sup>	) V
V <sub>IL</sub>	Low-level input voltage		0	0.3 × V <sub>CCI</sub> <sup>(1</sup>	) V
DR	Signaling rate		0	100	) Mbps
TJ	Junction temperature <sup>(2)</sup>		-55	150	O°C
T <sub>A</sub>	Ambient temperature		-55	25 125	5 °C

(1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ . (2) To maintain the recommended operating conditions for T<sub>J</sub>, see *Thermal Information*.

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### 6.4 Thermal Information

	THERMAL METRIC <sup>(1)</sup>	DW (SOIC)	UNIT
		16 Pins	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	78.9	
R <sub>0JC(top)</sub>	Junction-to-case(top) thermal resistance	41.6	
$R_{ extsf{ heta}JB}$	Junction-to-board thermal resistance	43.6	°C/W
ΨJT	Junction-to-top characterization parameter	15.5	°C/vv
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	43.1	
$R_{\theta JC(bottom)}$	Junction-to-case(bottom) thermal resistance	N/A	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

### 6.5 Power Rating

			VALUE	UNIT
$P_D$	Maximum power dissipation by ISO7841	$V_{CC1} = V_{CC2} = 5.5 \text{ V}, \text{ T}_{1} = 150^{\circ}\text{C},$	200	
P <sub>D1</sub>	Maximum power dissipation by side-1 of ISO7841	$C_L = 15 \text{ pF}$ , input a 50 MHz 50% duty cycle	50	mW
P <sub>D2</sub>	Maximum power dissipation by side-2 of ISO7841	square wave	150	

# 6.6 Electrical Characteristics, 5 V

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	$I_{OH} = -4$ mA; see Figure 7	$V_{CCO}^{(1)} - 0.4$	$V_{CCO}^{(1)} - 0.2$		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 4 mA; see Figure 7		0.2	0.4	V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis		$0.1 \times V_{CCO}^{(1)}$			V
I <sub>IH</sub>	High-level input current	$V_{IH} = V_{CCI}^{(1)}$ at INx or ENx			10	μA
IIL	Low-level input current	$V_{IL} = 0 V$ at INx or ENx	-10			μA
CMTI	Common-mode transient immunity	$V_I = V_{CCI}^{(1)}$ or 0 V; see Figure 10	70	100		kV/µs



# Electrical Characteristics, 5 V (continued)

$V_{CC1} = V_{CC2} = 5 \text{ V} \pm 10\%$ (over recommended operating conditions unless otherwise noted	$V_{CC1}$ :	$= V_{CC2} = 1$	5 V ±10%	(over	recommended	operating	conditions	unless	otherwise	noted)
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	PARAMETER	TEST	CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>CC1</sub>		Disable; EN1 = EN2 = 0 V	$\begin{array}{l} DC \ signal: \ V_I = 0 \ V \\ (Devices \ with \ suffix \ F) \ , \\ V_I = \ V_{CCI} \ (Devices \\ without \ suffix \ F) \end{array}$		1.1	1.8	
I <sub>CC2</sub>		Disable; EN1 = EN2 = 0 V	$\begin{array}{l} DC \ signal: \ V_l = 0 \ V \\ (Devices \ with \ suffix \ F) \ , \\ V_l = \ V_{CCl} \ (Devices \\ without \ suffix \ F) \end{array}$		0.8	1.3	
I <sub>CC1</sub>		Disable; EN1 = EN2 = 0 V	$\begin{array}{l} DC \ signal: V_I = V_{CCI} \\ (Devices \ with \ suffix \ F) \ , \\ V_I = 0 \ V(Devices \ without \\ suffix \ F) \end{array}$		4.5	6.6	
I <sub>CC2</sub>		Disable; EN1 = EN2 = 0 V	$\begin{array}{l} DC \ signal: V_I = V_{CCI} \\ (Devices \ with \ suffix \ F) \ , \\ V_I = 0 \ V(Devices \ without \ suffix \ F) \end{array}$		2	2.9	
I <sub>CC1</sub>	- Supply current	DC signal	$\begin{array}{l} DC \ signal: \ V_I = 0 \ V \\ (Devices \ with \ suffix \ F) \ , \\ V_I = \ V_{CCI} \ (Devices \\ without \ suffix \ F) \end{array}$		1.5	2.4	mA
I <sub>CC2</sub>	— Supply current	DC signal	$\begin{array}{l} DC \ signal: \ V_I = 0 \ V \\ (Devices \ with \ suffix \ F) \ , \\ V_I = \ V_{CCI} \ (Devices \\ without \ suffix \ F) \end{array}$		2.1	3.1	
I <sub>CC1</sub>		DC signal	$\begin{array}{l} DC \ signal: V_I = V_{CCI} \\ (Devices \ with \ suffix \ F) \ , \\ V_I = 0 \ V(Devices \ without \ suffix \ F) \end{array}$		5	7.3	
I <sub>CC2</sub>		DC signal	$\begin{array}{l} DC \ signal: V_I = V_{CCI} \\ (Devices \ with \ suffix \ F) \ , \\ V_I = 0 \ V(Devices \ without \ suffix \ F) \end{array}$		3.4	4.9	
I <sub>CC1</sub>		1 Mbps			3.3	4.9	
I <sub>CC2</sub>		1 Mbps	AC signal: All channels		2.9	4.2	
I <sub>CC1</sub>		10 Mbps	switching with square		3.9	5.5	
I <sub>CC2</sub>		10 Mbps	wave clock input; C <sub>L</sub> = 15 pF		4.4	5.7	
I <sub>CC1</sub>		100 Mbps			9.2	11.6	
I <sub>CC2</sub>		100 Mbps			19	21.9	

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# 6.7 Electrical Characteristics, 3.3 V

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST	CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>ОН</sub>	High-level output voltage	$I_{OH} = -2 \text{ mA}; \text{ see F}$	Figure 7	$V_{CCO}^{(1)} - 0.4$	$V_{CCO}^{(1)} - 0.2$		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 2 mA; see Fig	gure 7		0.2	0.4	V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis			$0.1 \times V_{CCO}^{(1)}$			V
I <sub>IH</sub>	High-level input current	$V_{IH} = V_{CCI}^{(1)}$ at INx	or ENx			10	μA
I <sub>IL</sub>	Low-level input current	$V_{IL} = 0 V at INx or$	ENx	-10			μΑ
CMTI	Common-mode transient immunity	$V_{I} = V_{CCI}^{(1)} \text{ or } 0 V;$	see Figure 10	70	100		kV/µs
I <sub>CC1</sub>		Disable; EN1 = EN2 = 0 V	DC signal: $V_I = 0 V$ (Devices with suffix F), $V_I$ = $V_{CCI}$ (Devices without suffix F)		1.1	1.8	
I <sub>CC2</sub>	_	Disable; EN1 = EN2 = 0 V	DC signal: $V_1 = 0 V$ (Devices with suffix F), $V_1$ = $V_{CC1}$ (Devices without suffix F)		0.8	1.3	
I <sub>CC1</sub>	_	Disable; EN1 = EN2 = 0 V	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		4.5	6.6	
I <sub>CC2</sub>	_	Disable; EN1 = EN2 = 0 V	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		1.9	2.9	
I <sub>CC1</sub>	Currely surrout	DC signal	DC signal: $V_1 = 0 V$ (Devices with suffix F), $V_1$ = $V_{CC1}$ (Devices without suffix F)		1.5	2.4	
I <sub>CC2</sub>	<ul> <li>Supply current</li> </ul>	DC signal	$      DC signal: V_I = 0 V \\ (Devices with suffix F) , V_I \\ = V_{CCI} (Devices without \\ suffix F) $		2.1	3.1	mA
I <sub>CC1</sub>		DC signal	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		5	7.3	
I <sub>CC2</sub>	_	DC signal	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		3.3	4.9	
I <sub>CC1</sub>		1 Mbps			3.3	4.9	
I <sub>CC2</sub>		1 Mbps	AC signal: All shannels		2.8	4.1	
I <sub>CC1</sub>		10 Mbps	<ul> <li>AC signal: All channels switching with square</li> </ul>		3.7	5.3	
I <sub>CC2</sub>		10 Mbps	wave clock input;		3.9	5.2	
I <sub>CC1</sub>		100 Mbps	C <sub>L</sub> = 15 pF		7.4	9.3	
I <sub>CC2</sub>		100 Mbps			14.5	16.9	

(1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .

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# 6.8 Electrical Characteristics, 2.5 V

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST	CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>ОН</sub>	High-level output voltage	I <sub>OH</sub> = -1 mA; see F	igure 7	$V_{CCO}^{(1)} - 0.4$	V <sub>CCO</sub> <sup>(1)</sup> – 0.2		V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 1 mA; see Fig	jure 7		0.2	0.4	V
V <sub>I(HYS)</sub>	Input threshold voltage hysteresis			$0.1 \times V_{CCO}^{(1)}$			V
I <sub>IH</sub>	High-level input current	$V_{IH} = V_{CCI}^{(1)}$ at INx	or ENx			10	μA
IIL	Low-level input current	$V_{IL} = 0 V \text{ at INx or}$	ENx	-10			μA
CMTI	Common-mode transient immunity	$V_{I} = V_{CCI}^{(1)} \text{ or } 0 V;$	see Figure 10	70	100		kV/µs
I <sub>CC1</sub>		Disable; EN1 = EN2 = 0 V	DC signal: $V_I = 0 V$ (Devices with suffix F) , $V_I$ = $V_{CCI}$ (Devices without suffix F)		1.1	1.7	
I <sub>CC2</sub>		Disable; EN1 = EN2 = 0 V	DC signal: $V_I = 0 V$ (Devices with suffix F) , $V_I$ = $V_{CCI}$ (Devices without suffix F)		0.8	1.2	*
I <sub>CC1</sub>		Disable; EN1 = EN2 = 0 V	DC signal: $V_I = V_{CCI}$ (Devices with suffix F), $V_I$ = 0 V(Devices without suffix F)		4.5	6.5	*
I <sub>CC2</sub>		Disable; EN1 = EN2 = 0 V	DC signal: $V_I = V_{CCI}$ (Devices with suffix F), $V_I$ = 0 V(Devices without suffix F)		1.9	2.8	*
I <sub>CC1</sub>	Currely surrent	DC signal	DC signal: $V_1 = 0 V$ (Devices with suffix F), $V_1$ = $V_{CC1}$ (Devices without suffix F)		1.6	2.3	0
I <sub>CC2</sub>	<ul> <li>Supply current</li> </ul>	DC signal	DC signal: $V_1 = 0 V$ (Devices with suffix F), $V_1$ = $V_{CC1}$ (Devices without suffix F)		2.2	3.1	mA
I <sub>CC1</sub>	-	DC signal	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		5.1	7.2	+
I <sub>CC2</sub>	-	DC signal	DC signal: $V_1 = V_{CC1}$ (Devices with suffix F), $V_1$ = 0 V(Devices without suffix F)		3.5	4.8	+
I <sub>CC1</sub>		1 Mbps			3.4	4.8	]
I <sub>CC2</sub>		1 Mbps	AC signal: All channels		2.9	4	ļ
CC1		10 Mbps	switching with square		3.7	5.1	
I <sub>CC2</sub>		10 Mbps	wave clock input;		3.7	4.8	
I <sub>CC1</sub>		100 Mbps	C <sub>L</sub> = 15 pF		6.8	8.1	ļ
CC2		100 Mbps			12	14.2	

(1)  $V_{CCI}$  = Input-side  $V_{CC}$ ;  $V_{CCO}$  = Output-side  $V_{CC}$ .

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# 6.9 Switching Characteristics, 5 V

 $V_{CC1} = V_{CC2} = 5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		6	11	16	
PWD <sup>(1)</sup>	Pulse width distortion  t <sub>PHL</sub> - t <sub>PLH</sub>	See Figure 7		0.55	4.1	
t <sub>sk(o)</sub> <sup>(2)</sup>	Channel-to-channel output skew time	Same-direction channels			2.5	
t <sub>sk(pp)</sub> <sup>(3)</sup>	Part-to-part skew time				4.5	
t <sub>r</sub>	Output signal rise time			1.7	3.9	ns
t <sub>f</sub>	Output signal fall time	See Figure 7		1.9	3.9	
t <sub>PHZ</sub>	Disable propagation delay, high-to-high impedance output			12	20	
t <sub>PLZ</sub>	Disable propagation delay, low-to-high impedance output			12	20	
	Enable propagation delay, high impedance-to-high output for ISO7841			10	20	ns
t <sub>PZH</sub>	Enable propagation delay, high impedance-to-high output for ISO7841F	See Figure 8		2	2.5	μs
	Enable propagation delay, high impedance-to-low output for ISO7841			2	2.5	μs
t <sub>PZL</sub>	Enable propagation delay, high impedance-to-low output for ISO7841F			10	20	ns
t <sub>fs</sub>	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 9		0.2	9	μs
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		0.90		ns

(1) Also known as pulse skew.

(2) t<sub>sk(0)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

(3) t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

# 6.10 Switching Characteristics, 3.3 V

 $V_{CC1} = V_{CC2} = 3.3 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time	0 Firmer 7	6	10.8	16	
PWD <sup>(1)</sup>	Pulse width distortion $ t_{PHL} - t_{PLH} $	See Figure 7		0.7	4.2	
t <sub>sk(o)</sub> <sup>(2)</sup>	Channel-to-channel output skew time	Same-direction channels			2.2	
t <sub>sk(pp)</sub> <sup>(3)</sup>	Part-to-part skew time				4.5	
t <sub>r</sub>	Output signal rise time	See Figure 7		0.8	3	ns
t <sub>f</sub>	Output signal fall time	See Figure 7		0.8	3	
t <sub>PHZ</sub>	Disable propagation delay, high-to-high impedance output			17	32	
t <sub>PLZ</sub>	Disable propagation delay, low-to-high impedance output			17	32	
	Enable propagation delay, high impedance-to-high output for ISO7841			17	32	ns
t <sub>PZH</sub>	Enable propagation delay, high impedance-to-high output for ISO7841F	- See Figure 8		2	2.5	μs
	Enable propagation delay, high impedance-to-low output for ISO7841			2	2.5	μs
t <sub>PZL</sub>	Enable propagation delay, high impedance-to-low output for ISO7841F			17	32	ns
t <sub>fs</sub>	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 9		0.2	9	μs
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		0.91		ns

(1) Also known as Pulse Skew.

(2) t<sub>sk(0)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

(3) t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.



# 6.11 Switching Characteristics, 2.5 V

 $V_{CC1} = V_{CC2} = 2.5 \text{ V} \pm 10\%$  (over recommended operating conditions unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>PLH</sub> , t <sub>PHL</sub>	Propagation delay time		7.5	11.7	17.5	
PWD <sup>(1)</sup>	Pulse width distortion  t <sub>PHL</sub> - t <sub>PLH</sub>	See Figure 7		0.66	4.2	
t <sub>sk(o)</sub> <sup>(2)</sup>	Channel-to-channel output skew time	Same-direction Channels			2.2	
t <sub>sk(pp)</sub> <sup>(3)</sup>	Part-to-part skew time				4.5	
t <sub>r</sub>	Output signal rise time	Out Finner 7		1	3.5	ns
t <sub>f</sub>	Output signal fall time	See Figure 7		1.2	3.5	
t <sub>PHZ</sub>	Disable propagation delay, high-to-high impedance output			22	45	
t <sub>PLZ</sub>	Disable propagation delay, low-to-high impedance output			22	45	
	Enable propagation delay, high impedance-to-high output for ISO7841			18	45	ns
t <sub>PZH</sub>	Enable propagation delay, high impedance-to-high output for ISO7841F	─ See Figure 8		2	2.5	μs
	Enable propagation delay, high impedance-to-low output for ISO7841	_		2	2.5	μs
t <sub>PZL</sub>	Enable propagation delay, high impedance-to-low output for ISO7841F			18	45	ns
t <sub>fs</sub>	Default output delay time from input power loss	Measured from the time $V_{CC}$ goes below 1.7 V. See Figure 9		0.2	9	μs
t <sub>ie</sub>	Time interval error	2 <sup>16</sup> – 1 PRBS data at 100 Mbps		0.91		ns

(1) Also known as pulse skew.

(2) t<sub>sk(0)</sub> is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

(3) t<sub>sk(pp)</sub> is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

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# 6.12 Typical Characteristics





# 7 Parameter Measurement Information



- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  50 kHz, 50% duty cycle, t<sub>r</sub>  $\leq$  3 ns, t<sub>f</sub>  $\leq$  3ns, Z<sub>O</sub> = 50  $\Omega$ . At the input, 50  $\Omega$  resistor is required to terminate Input Generator signal. It is not needed in actual application.
- B.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within  $\pm 20\%$ .

#### Figure 7. Switching Characteristics Test Circuit and Voltage Waveforms



- A. The input pulse is supplied by a generator having the following characteristics: PRR  $\leq$  10 kHz, 50% duty cycle,  $t_r \leq$  3 ns,  $t_f \leq$  3 ns,  $Z_O =$  50  $\Omega$ .
- B.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within ±20%.

#### Figure 8. Enable/Disable Propagation Delay Time Test Circuit and Waveform

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#### **Parameter Measurement Information (continued)**



A.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within ±20%.

#### Figure 9. Default Output Delay Time Test Circuit and Voltage Waveforms



A.  $C_L = 15 \text{ pF}$  and includes instrumentation and fixture capacitance within ±20%.

#### Figure 10. Common-Mode Transient Immunity Test Circuit



# 8 Detailed Description

# 8.1 Overview

ISO7841 employs an ON-OFF Keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. If the EN pin is low then the output goes to high impedance. ISO7841 also incorporates advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions due the high frequency carrier and IO buffer switching. The conceptual block diagram of a digital capacitive isolator, Figure 11, shows a functional block diagram of a typical channel.

# 8.2 Functional Block Diagram



Figure 11. Conceptual Block Diagram of a Digital Capacitive Isolator

Also a conceptual detail of how the ON/OFF Keying scheme works is shown in Figure 12.



Figure 12. On-Off Keying (OOK) Based Modulation Scheme

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## 8.3 Feature Description

PRODUCT	CHANNEL DIRECTION	RATED ISOLATION	MAX DATA RATE	DEFAULT OUTPUT
ISO7841	3 Forward,	5700 V <sub>RMS</sub> / 8000 V <sub>PK</sub> <sup>(1)</sup>	100 Mbps	High
1307641	1 Reverse	5700 V <sub>RMS</sub> / 6000 V <sub>PK</sub> ( /		High
10070445	3 Forward,	EZOD V (2000 V <sup>(1)</sup>	100 Mbro	Low
ISO7841F	1 Reverse	5700 V <sub>RMS</sub> / 8000 V <sub>PK</sub> <sup>(1)</sup>	100 Mbps	Low

(1) See *Regulatory Information* for detailed isolation ratings.

#### 8.3.1 High Voltage Feature Description

#### 8.3.1.1 Package Insulation and Safety-Related Specifications

over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
L(I01)	Minimum air gap (clearance)	Shortest terminal-to-terminal distance through air	8			mm	
L(I02) <sup>(1)</sup>	Minimum external tracking (creepage)	Shortest terminal-to-terminal distance across the package surface	8			mm	
СТІ	Tracking resistance (comparative tracking index)	DIN EN 60112 (VDE 0303-11); IEC 6011	2; UL 746A	600			V
Р	Isolation resistance, input to	$V_{IO} = 500 \text{ V}, \text{ T}_{A} = 25^{\circ}\text{C}$					Ω
R <sub>IO</sub>	output <sup>(2)</sup>	$V_{IO} = 500 \text{ V}, 100^{\circ}\text{C} \le T_A \le \text{max}$					Ω
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(2)</sup>	$V_{IO} = 0.4 \text{ x} \sin (2\pi ft), f = 1 \text{ MHz}$		2		pF	
CI	Input capacitance <sup>(3)</sup>	$V_{I} = V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ x sin } (2\pi \text{ft}), \text{ f} = 1 \text{ MHz}, V_{CC}/2 + 0.4 \text{ m}$		2		pF	

(1) Per JEDEC package dimensions.

(2) All pins on each side of the barrier tied together creating a two-terminal device.

(3) Measured from input pin to ground.

#### NOTE

Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance.

Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.

#### 8.3.1.2 Insulation Characteristics

	PARAMETER <sup>(1)</sup>	TEST CONDITIONS	SPECIFICATION	UNIT
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	21	μm
V	Moving un inclution working voltage	Time dependent dialectric breakdown (TDDD) Test	1500	V <sub>RMS</sub>
V <sub>IOWM</sub>	Maximum isolation working voltage	Time dependent dielectric breakdown (TDDB) Test	2121	V <sub>DC</sub>
DIN V VI	DE V 0884-10 (VDE V 0884-10):2006-12			
V <sub>IOTM</sub>	Maximum transient isolation voltage	V <sub>TEST</sub> = V <sub>IOTM</sub> t = 60 sec (qualification) t= 1 sec (100% production)	8000	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage	Test method per IEC 60065, 1.2/50 $\mu$ s waveform, V <sub>TEST</sub> = 1.6 x V <sub>IOSM</sub> = 12800 V <sub>PK</sub> (qualification)	8000	V <sub>PK</sub>
VIORM	Maximum repetitive peak isolation voltage		2121	V <sub>PK</sub>
		Method a, After Input/Output safety test subgroup 2/3, $V_{PR} = V_{IORM} \times 1.2$ , t = 10 s, Partial discharge < 5 pC	2545	
V <sub>PR</sub>	Input-to-output test voltage	Method a, After environmental tests subgroup 1, $V_{PR} = V_{IORM} \times 1.6$ , t = 10 s, Partial Discharge < 5 pC	3394	V <sub>PK</sub>
		Method b1, $V_{PR} = V_{IORM} \times 1.875$ , t = 1 s (100% Production test) Partial discharge < 5 pC	3977	
R <sub>S</sub>	Isolation resistance	$V_{IO} = 500 \text{ V at } T_{S}$	>10 <sup>9</sup>	Ω
	Pollution degree		2	
UL 1577		· ·		•
V <sub>ISO</sub>	Withstanding isolation voltage		5700	V <sub>RMS</sub>

(1) Climatic Classification 55/125/21

# 8.3.1.3 IEC 60664-1 Ratings Table

PARAMETER	TEST CONDITIONS	SPECIFICATION
Basic isolation group	Material group	I
Installation elegation	Rated mains voltage ≤ 600 V <sub>RMS</sub>	I–IV
Installation classification	Rated mains voltage ≤ 1000 V <sub>RMS</sub>	I–III

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#### 8.3.1.4 Regulatory Information

DW package certifications are complete.

VDE	CSA	UL	CQC
Certified according to DIN V VDE V 0884-10 (VDE V 0884- 10):2006-12 and DIN EN 60950- 1 (VDE 0805 Teil 1):2011-01	Approved under CSA Component Acceptance Notice 5A, IEC 60950-1, IEC 61010-1, and IEC 60601-1	Certified according to UL 1577 Component Recognition Program	Certified according to GB 4943.1- 2011
Reinforced insulation Maximum transient isolation voltage, 8000 V <sub>PK</sub> ; Maximum repetitive peak isolation voltage, 2121 V <sub>PK</sub> ; Maximum surge isolation voltage, 8000 V <sub>PK</sub>	Reinforced insulation per CSA 61010-1-12 and IEC 61010-1 3rd Ed., 300 V <sub>RMS</sub> max working voltage; Reinforced insulation per CSA 60950-1-07+A1+A2 and IEC 60950-1 2nd Ed., 800 V <sub>RMS</sub> max working voltage (pollution degree 2, material group I) ; 2 MOPP (Means of Patient Protection) per CSA 60601- 1:14 and IEC 60601-1 Ed. 3.1, 250 V <sub>RMS</sub> (354 V <sub>PK</sub> ) max working voltage	Single protection, 5700 V <sub>RMS</sub> <sup>(1)</sup>	Reinforced Insulation, Altitude ≤ 5000 m, Tropical Climate, 250 V <sub>RMS</sub> maximum working voltage
Certificate number: 40040142	Master contract number: 220991	File number: E181974	Certificate number: CQC15001121716

(1) Production tested  $\ge$  6840 V<sub>RMS</sub> for 1 second in accordance with UL 1577.



#### 8.3.1.5 Safety Limiting Values

Safety limiting intends to prevent potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier potentially leading to secondary system failures.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		$R_{\theta JA} = 78.9^{\circ}C/W, V_I = 5.5 V, T_J = 150^{\circ}C, T_A = 25^{\circ}C$			288	
I <sub>S</sub>	Is Safety input, output, or supply current for DW-16 Package					
	ouncil for DW To Fackage	$R_{\theta JA} = 78.9^{\circ}C/W, V_{I} = 2.75 V, T_{J} = 150^{\circ}C, T_{A} = 25^{\circ}C$			576	
Ps	Safety input, output, or total power	$R_{\theta JA} = 78.9^{\circ}C/W, T_{J} = 150^{\circ}C, T_{A} = 25^{\circ}C$			1584	mW
Τ <sub>S</sub>	Maximum case temperature				150	°C

The maximum safety temperature is the maximum junction temperature specified for the device. The power dissipation and junction-to-air thermal impedance of the device installed in the application hardware determines the junction temperature. The assumed junction-to-air thermal resistance in the *Thermal Information* is that of a device installed on a High-K test board for Leaded Surface Mount Packages. The power is the recommended maximum input voltage times the current. The junction temperature is then the ambient temperature plus the power times the junction-to-air thermal resistance





# 8.4 Device Functional Modes

ISO7841 functional modes are shown in Table 1.

V <sub>CCI</sub>	V <sub>cco</sub>	INPUT (INx) <sup>(2)</sup>	OUTPUT ENABLE (ENx)	OUTPUT (OUTx)	COMMENTS				
		н	H or open	н	Normal Operation:				
		L	H or open	L	A channel output assumes the logic state of its input.				
PU	PU	Open	H or open	Default	Default mode: When INx is open, the corresponding channel output goes to its default logic state. Default= High for ISO7841 and Low for ISO7841F.				
х	PU	х	L	Z	A low value of Output Enable causes the outputs to be high- impedance				
PD	PU	x	H or open	Default	Default mode: When $V_{CCI}$ is unpowered, a channel output assumes the logic state based on the selected default option. Default= High for ISO7841 and Low for ISO7841F. When $V_{CCI}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input. When $V_{CCI}$ transitions from powered-up to unpowered, channel output assumes the selected default state.				
х	PD	х	х	Undetermined	When $V_{CCO}$ is unpowered, a channel output is undetermined <sup>(3)</sup> . When $V_{CCO}$ transitions from unpowered to powered-up, a channel output assumes the logic state of its input				

### Table 1. Function Table<sup>(1)</sup>

V<sub>CCI</sub> = Input-side V<sub>CC</sub>; V<sub>CCO</sub> = Output-side V<sub>CC</sub>; PU = Powered up (V<sub>CC</sub> ≥ 2.25 V); PD = Powered down (V<sub>CC</sub> ≤ 1.7 V); X = Irrelevant; H = High level; L = Low level; Z = High Impedance
 A strongly driven input signal can weakly power the floating V<sub>CC</sub> via an internal protection diode and cause undetermined output.
 The outputs are in undetermined state when 1.7 V < V<sub>CCI</sub>, V<sub>CCO</sub> < 2.25 V.</li>



## 8.4.1 Device I/O Schematics



Figure 15. Device I/O Schematics

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# 9 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

#### 9.1 Application Information

The ISO7841 is a high-performance, quad-channel digital isolator with 5.7 kV<sub>RMS</sub> isolation voltage. The device comes with enable pins on each side which can be used to put the respective outputs in high impedance for multi master driving applications and reduce power consumption. ISO7841 uses single-ended CMOS-logic switching technology. Its supply voltage range is from 2.25 V to 5.5 V for both supplies, V<sub>CC1</sub> and V<sub>CC2</sub>. When designing with digital isolators, it is important to keep in mind that due to the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is,  $\mu$ C or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

# 9.2 Typical Application

The Isolated SPI Interface is shown in Figure 16.



Figure 16. Isolated SPI Interface for an Analog Input Module With 16 Input



### **Typical Application (continued)**

## 9.2.1 Design Requirements

For ISO7841, use the parameters shown in Table 2.

#### **Table 2. Design Parameters**

PARAMETER	VALUE
Supply voltage	2.25 to 5.5 V
Decoupling capacitor between V <sub>CC1</sub> and GND1	0.1 μF
Decoupling capacitor from V <sub>CC2</sub> and GND2	0.1 μF

#### 9.2.2 Detailed Design Procedure

Unlike optocouplers, which need external components to improve performance, provide bias, or limit current, ISO7841 only needs two external bypass capacitors to operate.



Figure 17. Typical ISO7841 Circuit Hook-up

#### 9.2.2.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 22. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO7841 incorporate many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection for input and output signal pins and inter-chip bond pads.
- Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by ensuring purely differential internal operation.

## 9.2.3 Application Curve

Typical eye diagram of ISO7841 indicate low jitter and wide open eye at the maximum data rate of 100 Mbps.



Figure 18. Eye Diagram at 100 Mbps PRBS, 5 V and 25°C

# **10** Power Supply Recommendations

To ensure reliable operation at all data rates and supply voltages, a  $0.1-\mu$ F bypass capacitor is recommended at input and output supply pins (V<sub>CC1</sub> and V<sub>CC2</sub>). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments' SN6501. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501 data sheet (SLLSEA0).



# 11 Layout

#### 11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Figure 19). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/inch<sup>2</sup>.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power / ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, see application note SLLA284, Digital Isolator Design Guide.

#### 11.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 epoxy-glass as PCB material. FR-4 (flame retardant 4) meets the requirements of Underwriters Laboratories UL94-V0, and is preferred over cheaper alternatives due to its lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and its self-extinguishing flammability-characteristics.

### 11.2 Layout Example



Figure 19. Layout Example Schematic



# **12 Device and Documentation Support**

#### **12.1** Documentation Support

#### 12.1.1 Related Documentation

See the Isolation Glossary (SLLA353)

#### 12.2 Related Links

The table below lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 3. Related Links

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY			
ISO7841	Click here	Click here	Click here	Click here	Click here			
ISO7841F	Click here	Click here	Click here	Click here	Click here			

#### 12.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E<sup>™</sup> Online Community *TI's Engineer-to-Engineer (E2E) Community.* Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support TI's Design Support** Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 12.4 Trademarks

E2E is a trademark of Texas Instruments. All other trademarks are the property of their respective owners.

#### 12.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

# 12.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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# PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
ISO7841DW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7841	Samples
ISO7841DWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7841	Samples
ISO7841DWW	PREVIEW	SOIC	DWW	16	45	Green (RoHS & no Sb/Br)	Call TI	Level-2-260C-1 YEAR		XISO7841	
ISO7841FDW	ACTIVE	SOIC	DW	16	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7841F	Samples
ISO7841FDWR	ACTIVE	SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-55 to 125	ISO7841F	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



# PACKAGE OPTION ADDENDUM

20-Aug-2015

<sup>(6)</sup> Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE MATERIALS INFORMATION

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# TAPE AND REEL INFORMATION





# QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All	dimensions are nominal												
	Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
	ISO7841DWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
	ISO7841FDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

TEXAS INSTRUMENTS

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# PACKAGE MATERIALS INFORMATION

14-Jul-2015



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO7841DWR	SOIC	DW	16	2000	367.0	367.0	38.0
ISO7841FDWR	SOIC	DW	16	2000	367.0	367.0	38.0

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