

FEATURES

- High saturated output power (P_{SAT}): 36.5 dBm at 30% power added efficiency (PAE)**
- High output third-order intercept (IP3): 44 dBm typical**
- High gain: 28 dB typical**
- High output power for 1 dB compression (P1dB): 36 dBm typical**
- Total supply current: 2200 mA at 7 V**
- 40-lead, 6 mm × 6 mm LFCSP package: 36 mm²**

APPLICATIONS

- Point to point radios**
- Point to multipoint radios**
- Very small aperture terminals (VSATs) and satellite communications (SATCOMs)**
- Military electronic warfare (EW) and electronic counter measures (ECM)**

GENERAL DESCRIPTION

The **HMC1121** is a three-stage, gallium arsenide (GaAs), pseudomorphic high electron mobility transfer (pHEMT), monolithic microwave integrated circuit (MMIC), 4 W power amplifier with an integrated temperature compensated on-chip power detector that operates between 5.5 GHz and 8.5 GHz. The **HMC1121** provides 28 dB of gain, 44 dBm output IP3, and 36.5 dBm of saturated output power at 30% PAE from a 7 V power supply.

FUNCTIONAL BLOCK DIAGRAM

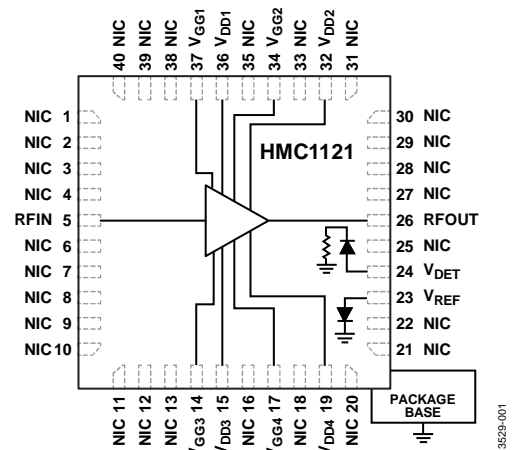


Figure 1.

The **HMC1121** exhibits excellent linearity and it is optimized for high capacity, point to point and point to multipoint radio systems. The amplifier configuration and high gain make it an excellent candidate for last stage signal amplification preceding the antenna.

Ideal for supporting higher volume applications, the **HMC1121** is provided in a 40-lead LFCSP package.

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REVISION HISTORY

10/2017—Rev. 0 to Rev. A

Changes to Figure 32	11
Updated Outline Dimensions	15
Changes to Ordering Guide	15

7/2016—Revision 0: Initial Version

SPECIFICATIONS

ELECTRICAL SPECIFICATIONS

$T_A = 25^\circ\text{C}$, $V_{DD} = V_{DD1} = V_{DD2} = V_{DD3} = V_{DD4} = 7\text{ V}$, $I_{DD} = 2200\text{ mA}$, frequency range = 5.5 GHz to 7.5 GHz.

Table 1.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
FREQUENCY RANGE		5.5		7.5	GHz	
GAIN		24	27		dB	
Gain Variation over Temperature			0.01		dB/°C	
RETURN LOSS						
Input			17		dB	
Output			13		dB	
OUTPUT POWER						
For 1 dB Compression	P1dB	35	36		dBm	36 dBm = 4 W
Saturated	P _{SAT}		36.5		dBm	At 30% PAE
OUTPUT THIRD-ORDER INTERCEPT	IP3		44		dBm	Measurement taken at P _{OUT} /tone = 28 dBm
SUPPLY						
Voltage	V _{DD}	5		7.5	V	Adjust the gate control voltage (V _{GG1} to V _{GG4}) between -2 V to 0 V to achieve an I _{DD} = 2200 mA typical
Total Current	I _{DD}		2200		mA	

$T_A = 25^\circ\text{C}$, $V_{DD} = V_{DD1} = V_{DD2} = V_{DD3} = V_{DD4} = 7\text{ V}$, $I_{DD} = 2200\text{ mA}$, frequency range = 7.5 GHz to 8.5 GHz.

Table 2.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
FREQUENCY RANGE		7.5		8.5	GHz	
GAIN		25	28		dB	
Gain Variation over Temperature			0.009		dB/°C	
RETURN LOSS						
Input			15		dB	
Output			13		dB	
OUTPUT POWER						
For 1 dB Compression	P1dB	35	36		dBm	36 dBm = 4 W
Saturated	P _{SAT}		36.5		dBm	At 30% PAE
OUTPUT THIRD-ORDER INTERCEPT	IP3		43		dBm	Measurement taken at P _{OUT} /tone = 28 dBm
SUPPLY						
Voltage	V _{DD}	5		7.5	V	Adjust the gate control voltage (V _{GG1} to V _{GG4}) between -2 V to 0 V to achieve an I _{DD} = 2200 mA typical
Total Current	I _{DD}		2200		mA	

ABSOLUTE MAXIMUM RATINGS

Table 3.

Parameter	Rating
Drain Voltage Bias	8 V
RF Input Power (RFIN) ¹	24 dBm
Channel Temperature	175°C
Continuous Power Dissipation, P _{DISS} (T _A = 85°C, Derate 227 mW/°C Above 85°C)	20.5 W
Thermal Resistance (R _{TH}) Junction to Ground Paddle	4.4°C/W
Maximum Peak Reflow Temperature (MSL3) ²	260°C
Storage Temperature Range	−65°C to +150°C
Operating Temperature Range	−40°C to +85°C
ESD Sensitivity (Human Body Model)	Class 1A, passed 250 V

¹The maximum input power (P_{IN}) is limited to 24 dBm or to the thermal limits constrained by the maximum power dissipation.

² See the Ordering Guide section.

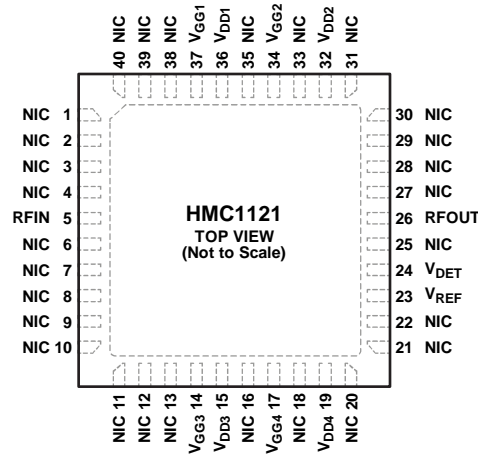
Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES
 1. NIC = NO INTERNAL CONNECTION.
 2. EXPOSED PAD. EXPOSED PAD MUST BE CONNECTED TO THE RF/DC GROUND.

13529-002

Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1 to 4, 6 to 13, 16, 18, 20 to 22, 25, 27 to 31, 33, 35, 38 to 40	NC	No Internal Connection. These pins and exposed ground pad must be connected to RF/dc ground.
5	RFIN	RF Input. This pin is ac-coupled and matched to 50 Ω. See Figure 3 for the RFIN interface schematic.
14, 17, 34, 37	V _{GG3} , V _{GG4} , V _{GG2} , V _{GG1}	Gate Controls for the Amplifier. Adjust V _{GG1} through V _{GG4} to achieve the recommended bias current. External bypass capacitors of 100 pF, 10 nF, and 4.7 μF are required. See Figure 5 for the V _{GG1} to V _{GG4} interface schematic.
15, 19, 32, 36	V _{DD3} , V _{DD4} , V _{DD2} , V _{DD1}	Drain Biases for the Amplifier. External bypass capacitors of 100 pF, 10 nF, and 4.7 μF are required. See Figure 8 for the V _{DD1} to V _{DD4} interface schematic.
23	V _{REF}	Voltage Reference. This pin is the dc bias of the diode biased through the external resistor and is used for the temperature compensation of V _{DET} . See Figure 7 for the V _{REF} interface schematic.
24	V _{DET}	Voltage Detection. This pin is the dc voltage representing the RF output power rectified by the diode that is biased through an external resistor. See Figure 4 for the V _{DET} interface schematic.
26	RFOUT	RF Output. This pin is ac-coupled and matched to 50 Ω. See Figure 6 for the RFOUT interface schematic.
	EPAD	Exposed Pad. The exposed pad must be connected to RF/dc ground.

INTERFACE SCHEMATICS

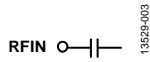


Figure 3. RFIN Interface Schematic

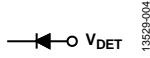


Figure 4. VDET Interface Schematic

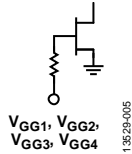


Figure 5. VGG1 to VGG4 Interface Schematic

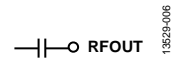


Figure 6. RFOUT Interface Schematic

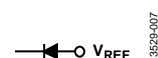


Figure 7. VREF Interface Schematic

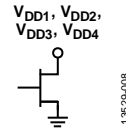


Figure 8. VDD1 to VDD4 Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

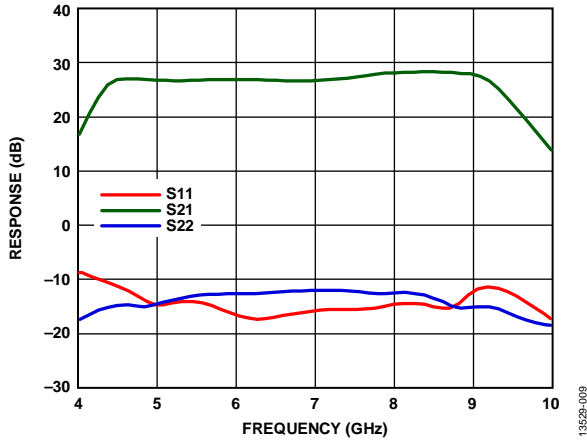


Figure 9. Response (Broadband Gain and Return Loss) vs. Frequency for S21, S11, and S22

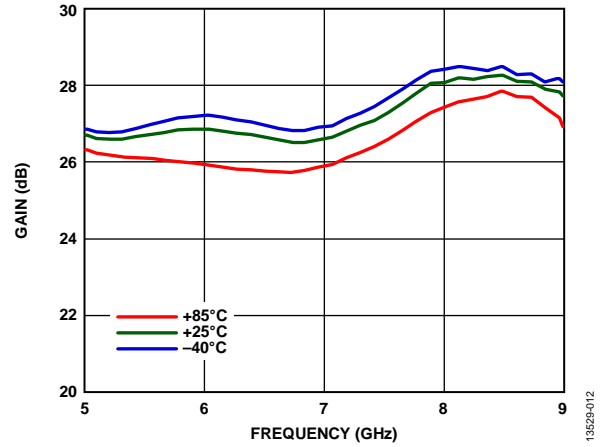


Figure 12. Gain vs. Frequency at Various Temperatures

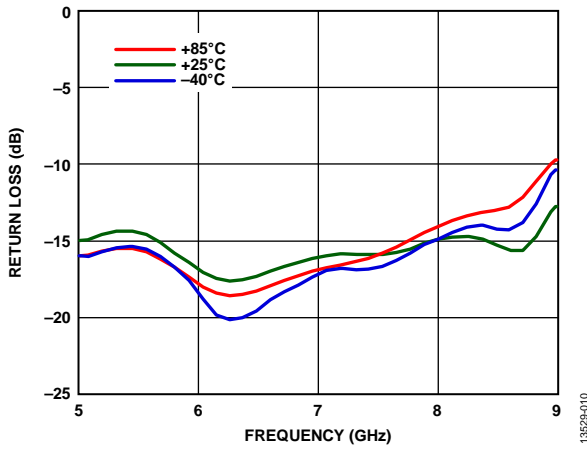


Figure 10. Input Return Loss vs. Frequency at Various Temperatures

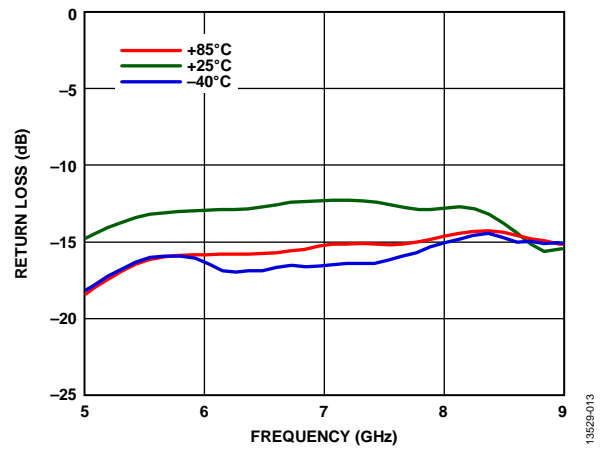


Figure 13. Output Return Loss vs. Frequency at Various Temperatures

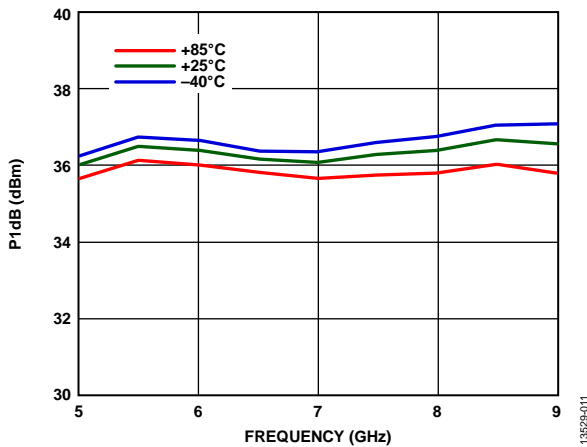


Figure 11. P1dB vs. Frequency at Various Temperatures

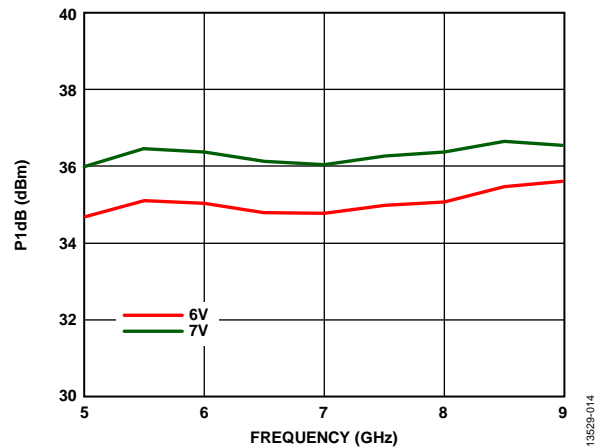


Figure 14. P1dB vs. Frequency at Various Supply Voltages

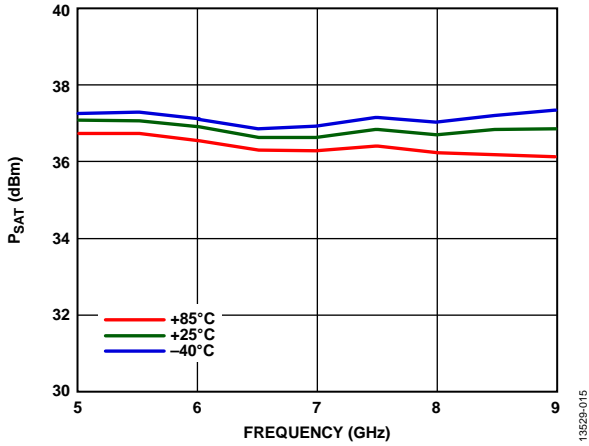


Figure 15. P_{SAT} vs. Frequency at Various Temperatures

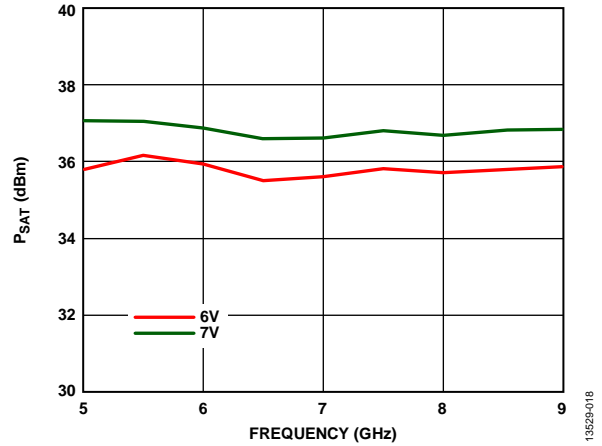


Figure 18. P_{SAT} vs. Frequency at Various Supply Voltages

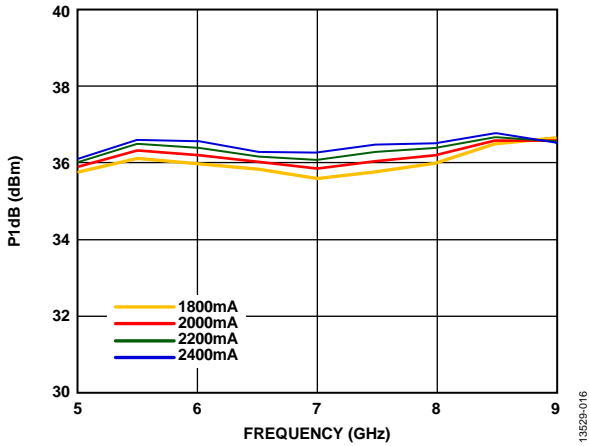


Figure 16. $P1dB$ vs. Frequency at Various Supply Currents (I_{DD})

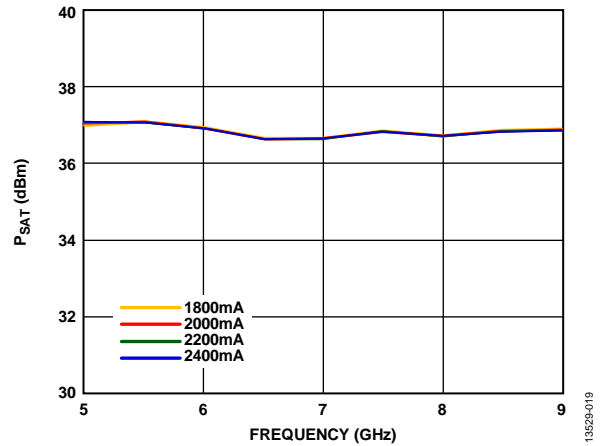


Figure 19. P_{SAT} vs. Frequency at Various Supply Currents (I_{DD})

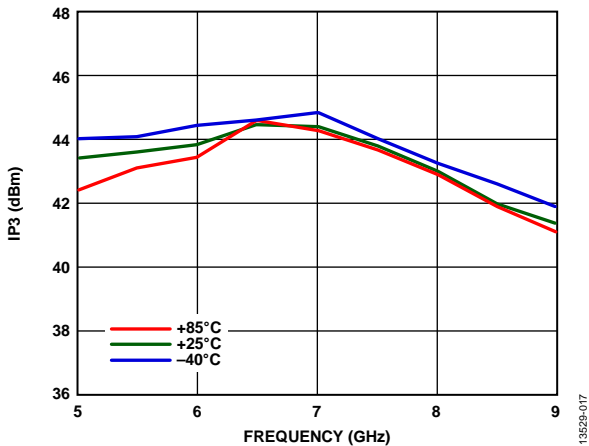


Figure 17. Output $IP3$ vs. Frequency at Various Temperatures, $P_{OUT}/Tone = 28\text{ dBm}$

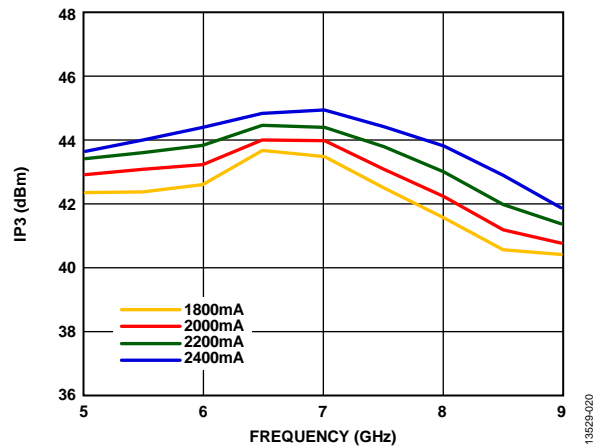


Figure 20. Output $IP3$ vs. Frequency at Various Supply Currents, $P_{OUT}/Tone = 28\text{ dBm}$

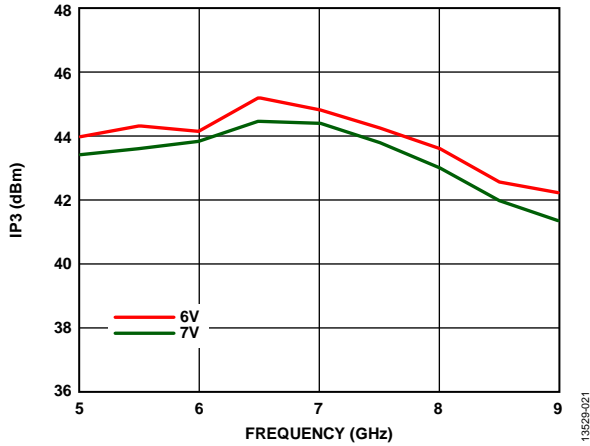


Figure 21. Output IP3 vs. Frequency at Various Supply Voltages, $P_{OUT/Tone} = 28$ dBm

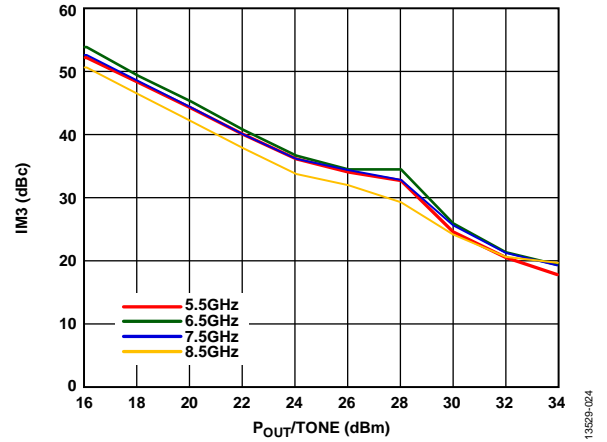


Figure 24. Output IM3 vs. $P_{OUT/Tone}$ at $V_{DD} = 6$ V

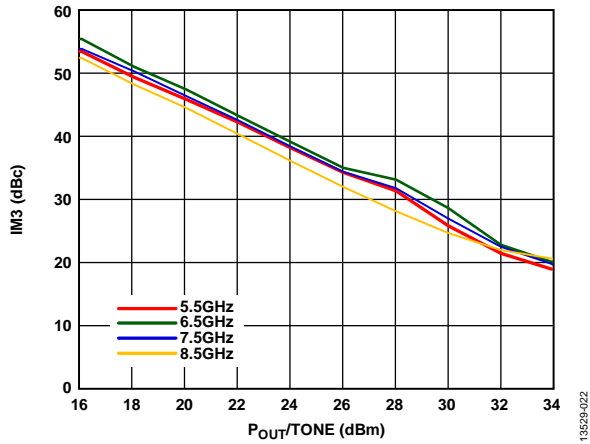


Figure 22. Output Third-Order Intermodulation (IM3) vs. $P_{OUT/Tone}$ at $V_{DD} = 7$ V

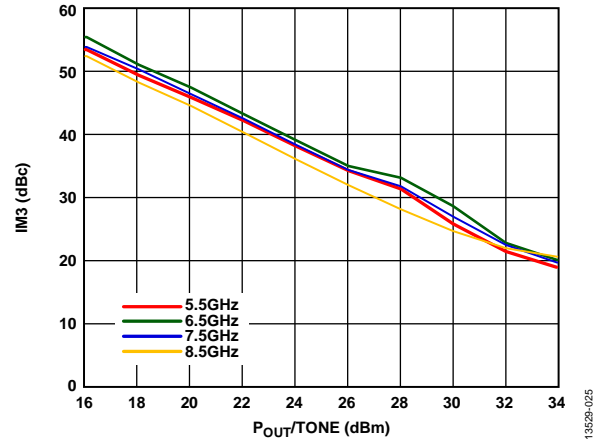


Figure 25. Output IM3 vs. $P_{OUT/Tone}$ at $V_{DD} = 8$ V

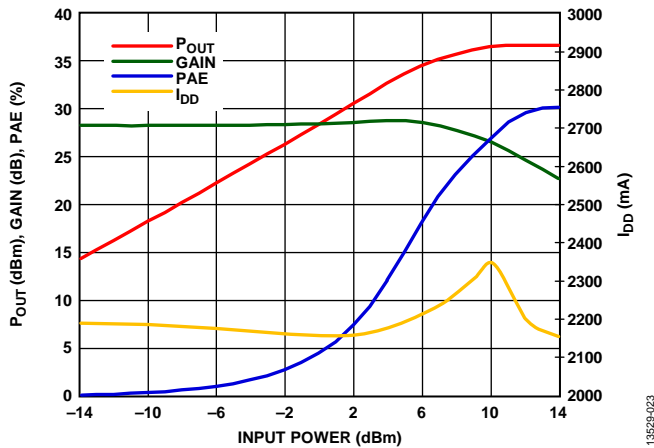


Figure 23. P_{OUT} , Gain, PAE, and I_{DD} vs. Input Power at 7 GHz

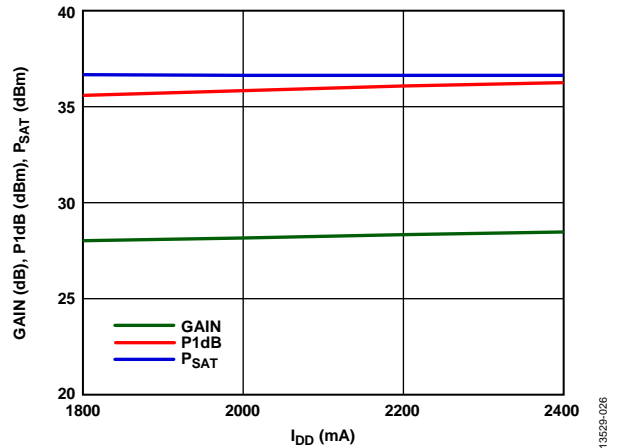


Figure 26. Gain, P1dB, and P_{SAT} vs. Supply Current (I_{DD}) at 7 GHz

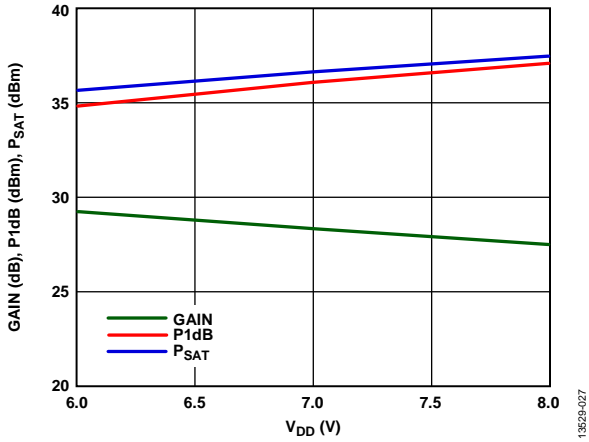


Figure 27. Gain, P1dB, and P_{SAT} vs. Supply Voltage (V_{DD}) at 7 GHz

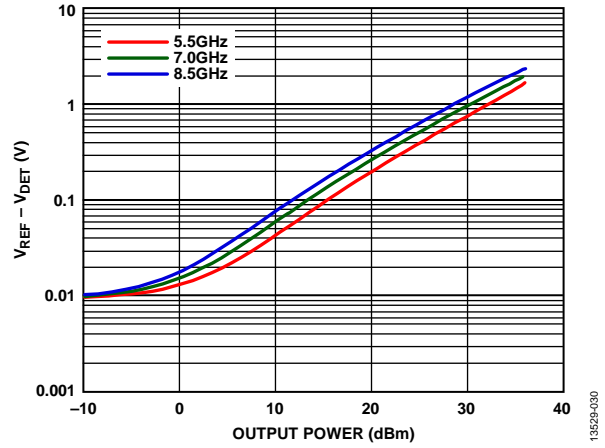


Figure 30. Detector Voltage ($V_{REF} - V_{DET}$) vs. Output Power at Various Frequencies

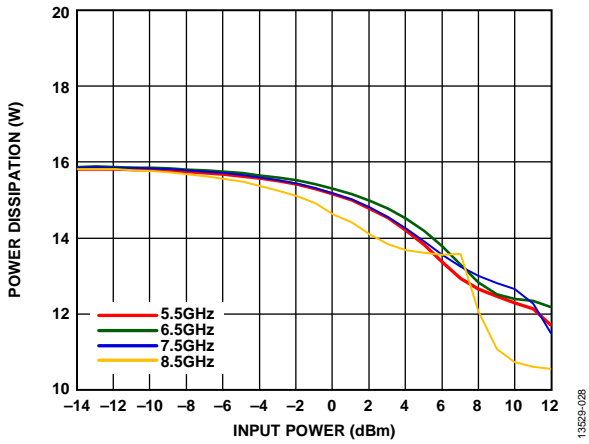


Figure 28. Power Dissipation vs. Input Power at $T_A = 85^\circ\text{C}$

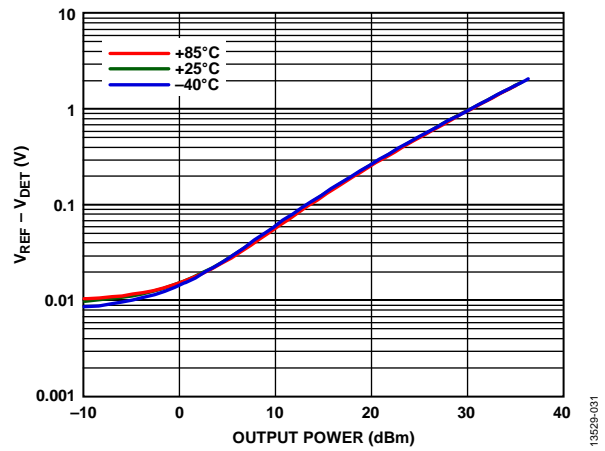


Figure 31. Detector Voltage ($V_{REF} - V_{DET}$) vs. Output Power at Various Temperatures, at 7 GHz

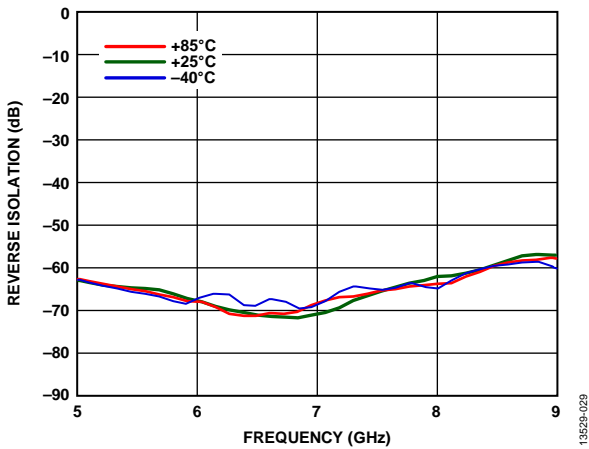


Figure 29. Reverse Isolation vs. Frequency at Various Temperatures

THEORY OF OPERATION

The HMC1121 is a three-stage, GaAs, pHEMT, MMIC, 4 W power amplifier consisting of three gain stages in series. A simplified block diagram is shown in Figure 32. The input signal is divided evenly into two; each of these two paths are amplified through three independent gain stages. The amplified signals are then combined at the output.

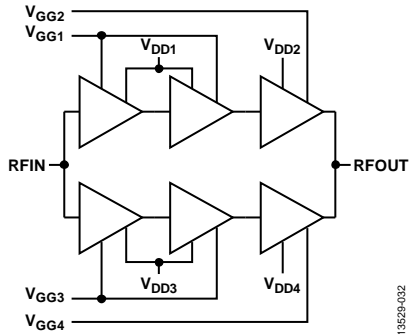


Figure 32. Simplified Block Diagram

The HMC1121 has single-ended input and output ports whose impedances are nominally matched to 50 Ω internally over the 5.5 GHz to 8.5 GHz frequency range. Consequently, it can directly insert into a 50 Ω system without the need for impedance matching circuitry. In addition, multiple HMC1121 amplifiers can be cascaded back to back without the need of external matching circuitry. Similarly, multiple HMC1121 amplifiers can be used with power dividers at the input and power combiners at the output to obtain higher output power levels.

The input and output impedances are sufficiently stable vs. variations in temperature and supply voltage that no impedance matching compensation is required.

It is critical to supply very low inductance ground connections to the ground pins as well as to the backside exposed pad to ensure stable operation.

To ensure the best performance out of the HMC1121, do not exceed the absolute maximum ratings.

APPLICATIONS INFORMATION

Figure 33 shows the basic connections for operating the HMC1121 and also see the Theory of Operation section for additional information. The RF input and RF output are ac-coupled by the internal dc block capacitors.

RECOMMENDED BIAS SEQUENCE

Follow the recommended bias sequencing to avoid damaging the amplifier.

During Power-Up

The recommended bias sequence during power-up is the following:

1. Connect to ground.
2. Set V_{GGx} to -2 V.
3. Set V_{DDx} to 7 V.
4. Increase V_{GGx} to achieve a typical $I_{DD} = 2200$ mA.
5. Apply the RF signal.

During Power-Down

The recommended bias sequence during power-down is the following:

1. Turn the RF signal off.
2. Decrease V_{GGx} to -2 V to achieve a typical $I_{DD} = 0$ mA.

3. Decrease V_{DDx} to 0 V.
4. Increase V_{GGx} to 0 V.

The bias conditions previously listed ($V_{DDx} = 7$ V, $I_{DD} = 2200$ mA), are the recommended operating point to achieve optimum performance. The data used in this datasheet is taken with the recommended bias conditions. When using the HMC1121 with different bias conditions, different performance may result than what is shown in the Typical Performance Characteristics section.

The V_{DET} and V_{REF} pins are the output pins for the internal power detector. The V_{DET} pin is the dc voltage output pin that represents the RF output power rectified by the internal diode, which is biased through an external resistor.

The V_{REF} pin is the dc voltage output pin that represents the reference diode voltage, which is biased through an external resistor. This voltage is then used to compensate for the temperature variation effects on both diodes. A typical circuit is shown in the Typical Application Circuit section that reads out the output voltage and represents the RF output power is shown in Figure 33.

TYPICAL APPLICATION CIRCUIT

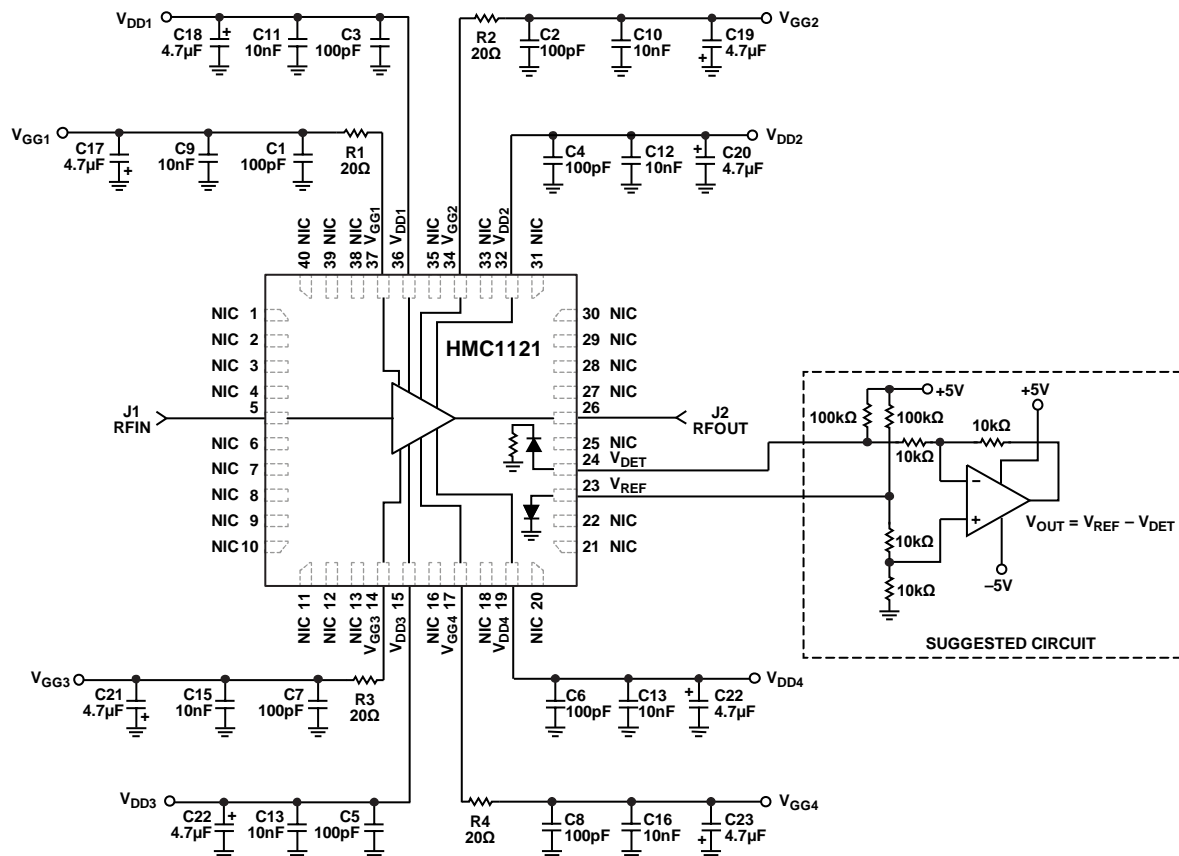


Figure 33. Typical Application Circuit

EVALUATION BOARD

The [HMC1121](#) evaluation printed circuit board (PCB) is a 2-layer board that was fabricated using a Rogers 4350 and best practices for high frequency RF design. The RF input and RF output traces have a 50 Ω characteristic impedance. The PCB is attached to a heat sink by a SN96 solder, which provides a low thermal resistance path. Components are mounted using SN63 solder, allowing rework of the surface-mount components without compromising the circuit board to heat sink attachment.

The evaluation PCB and populated components operate over the -40°C to $+85^{\circ}\text{C}$ ambient temperature range. During operation, attach the evaluation PCB to a temperature controlled plate to control the temperature. For proper bias sequence, see the Applications Information section.

See Figure 35 for the [HMC1121](#) evaluation board schematic. A fully populated and tested evaluation board, which is shown in Figure 34, is available from Analog Devices, Inc., upon request.

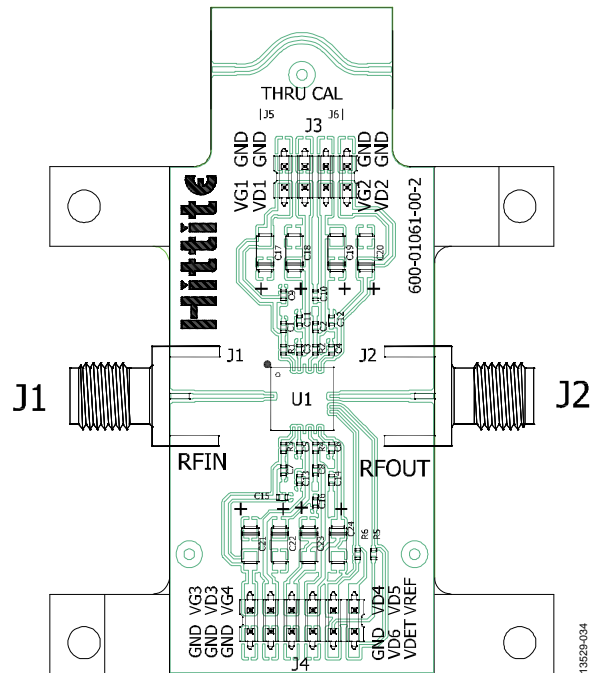


Figure 34. [HMC1121](#) Evaluation Board

BILL OF MATERIALS

Table 5.

Item	Description
J1, J2	PCB mount SMA RF connector, Johnson PN 142-07010851
J3, J4	DC pins
J5, J6	RF connectors for thru line; not populated
C1 to C8	100 pF capacitors, 0402 package
C9 to C16	10 nF capacitors, 0402 package
C17 to C24	4.7 μF capacitor, Case A
R1 to R4	20 Ω resistors, 0402 package
R5, R6	100 k Ω resistors, 0402 package
U1	HMC1121LP6GE
Heat sink	Used for thermal transfer from the HMC1121LP6GE amplifier
PCB	600-01061-00 evaluation board; circuit board material: Rogers 4350

EVALUATION BOARD SCHEMATIC

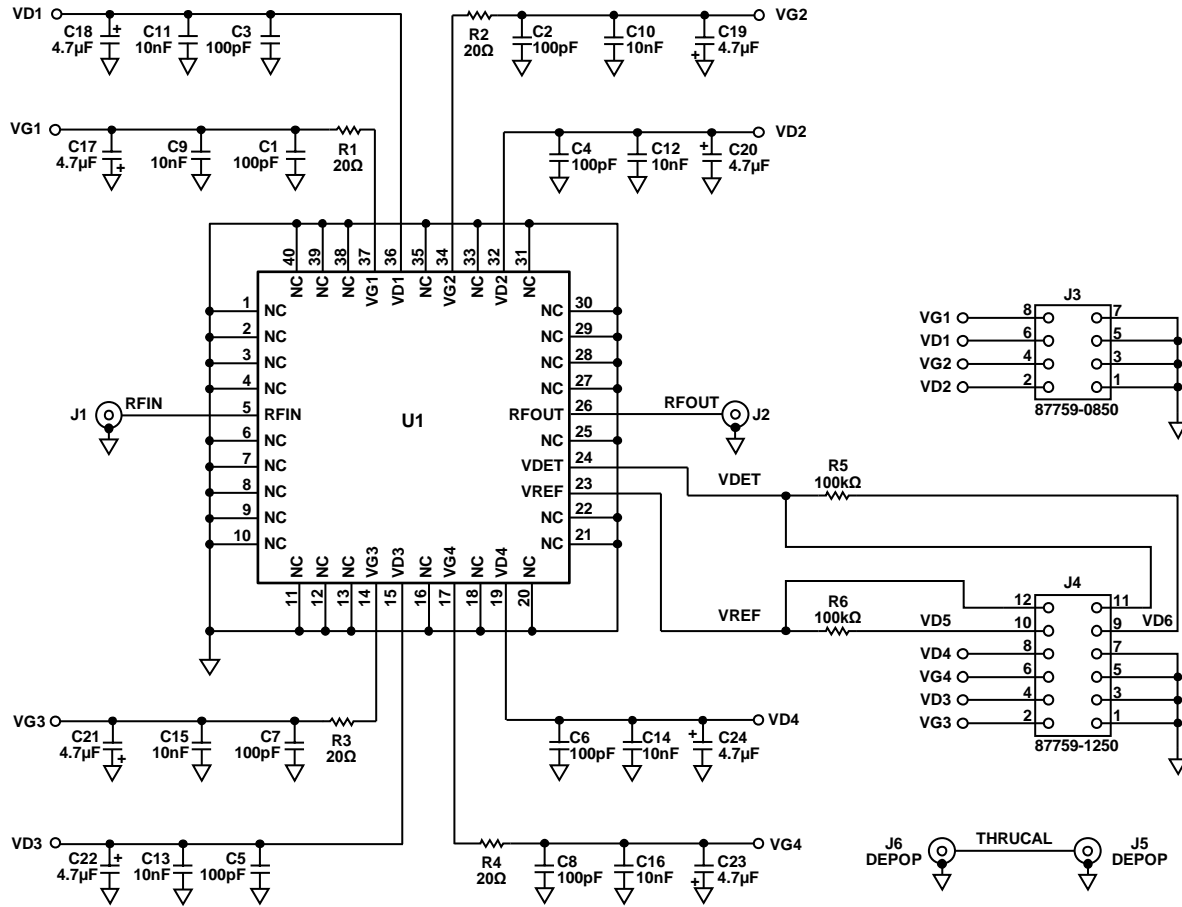
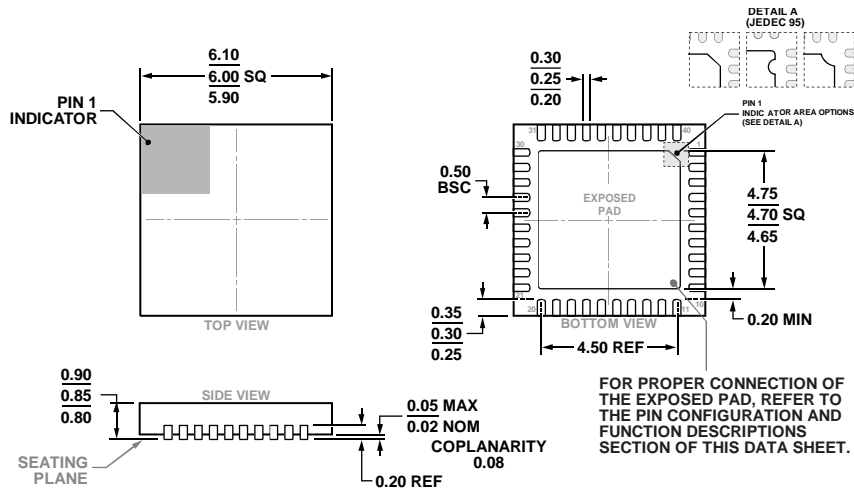


Figure 35. HMC1121 Evaluation Board Schematic

13529-035

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-220-VJJD-5

Figure 36. 40-Lead Lead Frame Chip Scale Package [LFCS]
6 mm × 6 mm Body and 0.85 mm Package Height
(HCP-40-1)
Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Temperature	MSL Rating ²	Package Description ³	Package Option
HMC1121LP6GE	-40°C to +85°C	MSL3	40-Lead Lead Frame Chip Scale Package [LFCS]	HCP-40-1
HMC1121LP6GETR	-40°C to +85°C	MSL3	40-Lead Lead Frame Chip Scale Package [LFCS]	HCP-40-1
EV1HMC1121LP6G			Evaluation Board	

¹ The HMC1121LP6GE and the HMC1121LP6GETR are RoHS-Compliant Parts.

² See the Absolute Maximum Ratings section for additional details.

³ The HMC1121LP6GE and the HMC1121LP6GETR are low stress injection molded plastic and their lead finish is 100% matte Sn.



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)

Факс: 8 (812) 320-02-42

Электронная почта: org@eplast1.ru

Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.