

Thermocouple EMF to Temperature Converter, ±1.5°C Maximum Accuracy

Features

- Thermocouple Electromotive Force (EMF) to °C Converter:
	- Integrated Cold-Junction Compensation
- Supported Types (designated by NIST ITS-90): - Type K, J, T, N, S, E, B and R
- Sensor Accuracy for Thermocouple Hot-Junction
- MCP9600 ±0.5°C/±1.5°C (Typ./Max.)
- MCP96L00 $\pm 2.0^{\circ}$ C/ $\pm 4.0^{\circ}$ C (Typ./Max.)
- Measurement Resolution:
	- Hot and Cold-Junctions: +0.0625°C (typical)
- Four Programmable Temperature Alert Outputs:
- Monitor Hot or Cold-Junction Temperatures
- Detect Rising or Falling Temperatures
- Up to 255°C of Programmable Hysteresis
- Programmable Digital Filter for Temperature
- Low Power:
	- Shutdown Mode
	- Burst Mode: 1 to 128 Temperature Samples
- 2-Wire Interface: 1^2C Compatible, 100 kHz:
	- Supports Eight Devices per 1^2C Bus
- Operating Voltage Range: 2.7V to 5.5V
- Operating Current: 300 µA (typical)
- Shutdown Current: 2 µA (typical)
- Package: 20-lead MQFN

Typical Applications

- Petrochemical Thermal Management
- Hand-Held Measurement Equipment
- Industrial Equipment Thermal Management
- Ovens
- Industrial Engine Thermal Monitor
- **Temperature Detection Racks**

Description

Microchip Technology Inc.'s MCP9600/L00 converts thermocouple EMF to degree Celsius with integrated cold-junction compensation. This MCP9600 corrects the thermocouple nonlinear error characteristics of eight thermocouple types and outputs ±0.5°C/±1.5°C (Typ./Max.) accurate temperature data for the selected thermocouple Hot-Junction temperature. And the MCP96L00 outputs ±2.0°C/±4.0°C (Typ./Max.) accurate temperature data. The correction coefficients are derived from the National Institute of Standards and Technology (NIST) ITS-90 Thermocouple Database.

The MCP9600/L00 digital temperature sensor comes with user-programmable registers which provide design flexibility for various temperature sensing applications. The registers allow user-selectable settings, such as Low-Power modes for battery-powered applications, adjustable digital filter for fast transient temperatures and four individually programmable temperature alert outputs which can be used to detect multiple temperature zones.

The temperature alert limits have multiple user-programmable configurations, such as alert polarity as either an active-low or active-high push-pull output, and output function as a Comparator mode (useful for thermostat-type operation) or Interrupt mode for microprocessor-based systems. In addition, the alerts can detect either a rising or a falling temperature with up to +255°C hysteresis.

This sensor uses an industry standard 2-wire, 1^2C compatible serial interface and supports up to eight devices per bus by setting the device address using the ADDR pin.

Package Type

MCP9600/L00 Registers

MCP9600 Evaluation Board (ADM00665)

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings†

† Notice: Stresses above those listed under "Maximum ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 2.7V$ to 5.5V, GND = Ground, $T_A = -40^{\circ}C$ to +125°C (where: $T_A = T_C$, defined as Device Ambient Temperature).

Note 1 The T_C and T_A summation is implemented in milli-volt (mV) domain. The result, T_H (mV), is converted to Degree Celsius using the NIST ITS-90 Conversion database.

2 The T_A _{ACY} temperature accuracy specification is defined as the device accuracy to the NIST ITS-90 Thermocouple EMF to Degree Celsius Conversion Database. T_A is also defined as the temperature difference between the hot and cold-junctions or temperatures from the NIST ITS-90 database with $T_c = 0^\circ C$.

3 The device measures temperature below the specified range, however, the sensitivity to changes in temperature reduces exponentially. Type R and S measure down to -50 $^{\circ}$ C, or -0.226 mV_{EMF} and -0.235 mV_{EMF}, respectively. Type B measures down to 500°C or 1.242 mV_{EMF} (see [Figures 2-7,](#page-8-0) 2-8, 2-10, 2-11, 2-14 and 2-17).

4 Exceeding the V_{IN} $_{\text{CM}}$ input range may cause leakage current through the ESD protection diodes at the thermocouple input pins. This parameter is characterized but not production tested.

DC CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 2.7V$ to 5.5V, GND = Ground, $T_A = -40^{\circ}C$ to +125°C (where: $T_A = T_C$, defined as Device Ambient Temperature).

Note 1 The T_C and T_A summation is implemented in milli-volt (mV) domain. The result, T_H (mV), is converted to Degree Celsius using the NIST ITS-90 Conversion database.

2 The T_A _{ACY} temperature accuracy specification is defined as the device accuracy to the NIST ITS-90 Thermocouple EMF to Degree Celsius Conversion Database. T_{Δ} is also defined as the temperature difference between the hot and cold-junctions or temperatures from the NIST ITS-90 database with $T_{\rm C}$ = 0°C.

3 The device measures temperature below the specified range, however, the sensitivity to changes in temperature reduces exponentially. Type R and S measure down to -50°C, or -0.226 mV $_{EMF}$ and -0.235 mV $_{EMF}$, respectively. Type B measures down to 500°C or 1.242 mV $_{EMF}$ (see Figures 2-7, 2-8, 2-10, 2-11, 2-14 and 2-17).

4 Exceeding the V_{IN} $_{\text{CM}}$ input range may cause leakage current through the ESD protection diodes at the thermocouple input pins. This parameter is characterized but not production tested.

DC CHARACTERISTICS (CONTINUED)

Note 1 The T_C and T_A summation is implemented in milli-volt (mV) domain. The result, T_H (mV), is converted to Degree Celsius using the NIST ITS-90 Conversion database.

2 The T_A _{ACY} temperature accuracy specification is defined as the device accuracy to the NIST ITS-90 Thermocouple EMF to Degree Celsius Conversion Database. T_A is also defined as the temperature difference between the hot and cold-junctions or temperatures from the NIST ITS-90 database with $T_{C} = 0^{\circ}C$.

3 The device measures temperature below the specified range, however, the sensitivity to changes in temperature reduces exponentially. Type R and S measure down to -50°C, or -0.226 mV $_{EMF}$ and -0.235 mV $_{EMF}$, respectively. Type B measures down to 500°C or 1.242 mV_{EMF} (see Figures 2-7, 2-8, 2-10, 2-11, 2-14 and 2-17).

4 Exceeding the V_{IN} $_{\text{CM}}$ input range may cause leakage current through the ESD protection diodes at the thermocouple input pins. This parameter is characterized but not production tested.

INPUT/OUTPUT PIN DC CHARACTERISTICS

Note 1 The ADDR pin can be tied to V_{DD} or V_{SS}. For additional slave addresses, a resistive divider network can be used to set voltage levels that are rationed to $\rm V_{DD}$. The device supports up to 8 levels (see **Section 6.3.1 "I²C Addressing"** for recommended resistor values).

2 $V_{\text{ADDR_TYP}}$ = Address * $V_{\text{DD}}/8$ + $V_{\text{DD}}/16$, V_{ADDR_L} = V_{ADDR_TYP} – V_{DD}/32 and V_{ADDR_H} = V_{ADDR_TYP} + V_{DD}/32 (where: Address = 1, 2, 3, 4, 5, 6).

TEMPERATURE CHARACTERISTICS

Note 1 Operation in this range must not cause T_J to exceed the Maximum Junction Temperature (+150°C).

SENSOR SERIAL INTERFACE TIMING SPECIFICATIONS

Electrical Specifications: Unless otherwise indicated, GND = Ground, $T_A = -40^{\circ}C$ to $+125^{\circ}C$, $V_{DD} = 2.7V$ to 5.5V and C_1 = 80 pF (**[Note 1](#page-6-2)**).

Note 1 All values referred to $V_{IL\,MAX}$ and $V_{IH\,MIN}$ levels.

2 This device can be used in a Standard mode I^2C bus system, but the requirement, $t_{SU:DATA} \ge 250$ ns, must be met.

3 Characterized, but not production tested.

4 Master controllers without features to detect clock stretching by Slave devices, should reduce f_{SCL} for proper I²C communication for Read commands. See [Figure 2-29](#page-11-0) for a typical t_{STRETCH} performance.

FIGURE 1-1: Timing Diagram.

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

FIGURE 2-2: Typical Temperature Accuracy from NIST ITS-90 Database, Type J.

FIGURE 2-3: Typical Temperature Accuracy from NIST ITS-90 Database, Type N.

FIGURE 2-5: Temperature Sensitivity with 18-Bit Resolution, Type J.

FIGURE 2-6: Temperature Sensitivity with 18-Bit Resolution, Type N.

FIGURE 2-7: Typical Temperature Accuracy from NIST ITS-90 Database, Type S.

FIGURE 2-8: Typical Temperature Accuracy from NIST ITS-90 Database, Type R.

FIGURE 2-9: Typical Temperature Accuracy from NIST ITS-90 Database, Type E.

FIGURE 2-10: Temperature Sensitivity with 18-Bit Resolution, Type S.

FIGURE 2-11: Temperature Sensitivity with 18-Bit Resolution, Type R.

FIGURE 2-12: Temperature Sensitivity with 18-Bit Resolution, Type E.

FIGURE 2-13: Typical Temperature Accuracy from NIST ITS-90 Database, Type T.

FIGURE 2-14: Typical Temperature Accuracy from NIST ITS-90 Database, Type B.

FIGURE 2-15: Input Offset Error Voltage (VIN+, VIN-).

FIGURE 2-16: Temperature Sensitivity with 18-Bit Resolution, Type T.

FIGURE 2-17: Temperature Sensitivity with 18-Bit Resolution, Type B.

FIGURE 2-18: Full-Scale Gain Error.

FIGURE 2-20: Cold-Junction Sensor Temperature Accuracy.

FIGURE 2-21: SDA and Alert Outputs, V_{OL} Across V_{DD}.

FIGURE 2-23: Cold-Junction Sensor Temperature Accuracy Distribution.

VDD.

Note: Unless otherwise indicated, V_{DD} = 2.7V to 5.5V, GND = Ground, SDA/SCL pulled-up to V_{DD} and T_A = -40°C to +125°C.

FIGURE 2-25: I

 $\overline{I^2C}$ *Inactive, I_{DD} Across V_{DD}.*

FIGURE 2-26: I 1²C Active, I_{DD} Across V_{DD}.

Across V_{DD}.

FIGURE 2-27: Shutdown Current, I_{SHDN}

FIGURE 2-28: SDA, SCL and ADDR Input Pins Leakage Current, I_{LEAK} Across V_{DD}.

FIGURE 2-29: I 2C Interface Clock Stretch Duration, t_{STRETCH} Across V_{DD}.

FIGURE 2-30: Temperature Calculation Duration, t_{CALC} Change Across V_{DD}.

3.0 PIN DESCRIPTION

The descriptions of the pins are listed in [Table 3-1.](#page-12-0)

TABLE 3-1: PIN FUNCTION TABLE

3.1 Ground Pin (GND)

The GND pin is the system ground pin. Pins 1, 3, 5, 13 and 17 are system ground pins and they are at the same potential. However, pins 6, 7, 9, 10 and 18 must be connected to ground for normal operation.

3.2 Thermocouple Input (V_{IN+}, V_{IN-})

The thermocouple wires are directly connected to these inputs. The positive node is connected to the V_{IN+} pin, while the negative node connects to the V_{IN-} node. The thermocouple voltage is converted to degree Celsius.

3.3 Power Pin (V_{DD})

 V_{DD} is the power pin. The operating voltage range, as specified in the DC Characteristics table, is applied on this pin.

3.4 Push-Pull Alert Outputs (Alert 1, 2, 3, 4)

The Alert pins are user- programmable push-pull outputs which can be used to detect rising or falling temperatures. The device outputs signal when the ambient temperature exceeds the user-programmed temperature alert limit.

3.5 I2C Slave Address Pin (ADDR)

This pin is used to set the I^2C slave address. This pin can be tied to V_{DD} , GND, or a ratio of V_{DD} can be selected to set up to eight address levels using a resistive voltage divider network.

3.6 Serial Clock Line (SCL)

The SCL is a clock input pin. All communication and timing is relative to the signal on this pin. The clock is generated by the host or master controller on the bus (see **[Section 4.0 "Serial Communication"](#page-14-0)**).

3.7 Serial Data Line (SDA)

SDA is a bidirectional input/output pin used to serially transmit data to/from the host controller. This pin requires a pull-up resistor (see **[Section 4.0 "Serial](#page-14-0) [Communication"](#page-14-0)**).

NOTES:

4.0 SERIAL COMMUNICATION

4.1 2-Wire Standard Mode I2C Protocol-Compatible Interface

The MCP9600/L00 Serial Clock Input (SCL) and the bidirectional Serial Data Line (SDA) form a 2-wire bidirectional data communication line (refer to the **[Input/Output Pin DC Characteristics](#page-5-3)** table and **[Sensor Serial Interface Timing Specifications](#page-6-4)** table).

The following bus protocol has been defined:

TABLE 4-1: MCP9600/L00 SERIAL BUS PROTOCOL DESCRIPTIONS

4.1.1 DATA TRANSFER

Data transfers are initiated by a Start condition (START), followed by a 7-bit device address and a read/write bit. An Acknowledge (ACK) from the slave confirms the reception of each byte. Each access must be terminated by a Stop condition (STOP).

Repeated communication is initiated after $t_{\text{B-FREE}}$.

This device supports the Receive Protocol. The register can be specified using the pointer for the initial read. Each repeated read or receive begins with a Start condition and address byte. The MCP9600/L00 retains the previously selected register. Therefore, it outputs data from the previously-specified register (repeated pointer specification is not necessary).

4.1.2 MASTER/SLAVE

The bus is controlled by a master device (typically a microcontroller) that controls the bus access, and generates the Start and Stop conditions. The MCP9600/L00 is a slave device and does not control other devices in the bus. Both master and slave devices can operate as either transmitter or receiver. However, the master device determines which mode is activated.

4.1.3 START/STOP CONDITION

A high-to-low transition of the SDA line (while SCL is high) is the Start condition. All data transfers must be preceded by a Start condition from the master. A low-to-high transition of the SDA line (while SCL is high) signifies a Stop condition.

If a Start or Stop condition is introduced during data transmission, the MCP9600/L00 releases the bus. All data transfers are ended by a Stop condition from the master.

4.1.4 ADDRESS BYTE

Following the Start condition, the host must transmit an 8-bit address byte to the MCP9600/L00. The address for the MCP9600/L00 temperature sensor is '11,0,0,A2,A1,A0' in binary, where the A2, A1 and A0 bits are set externally by connecting the corresponding VADDR voltage levels on the ADDR pin (see the **["Input/Output Pin DC Characteristics"](#page-5-3)** section). The 7-bit address transmitted in the serial bit stream must match the selected address for the MCP9600/L00 to respond with an ACK. Bit 8 in the address byte is a read/write bit. Setting this bit to '1' commands a read operation, while '0' commands a write operation (see [Figure 4-1](#page-14-1)).

FIGURE 4-1: Device Addressing.

4.1.5 DATA VALID

After the Start condition, each bit of data in transmission needs to be settled for a time specified by t_{SU-DATA} before SCL toggles from low-to-high (see the **["Sensor Serial Interface Timing Specifications"](#page-6-4)** section).

4.1.6 ACKNOWLEDGE (ACK/NAK)

Each receiving device, when addressed, is expected to generate an ACK bit after the reception of each byte. The master device must generate an extra clock pulse for ACK to be recognized.

The Acknowledging device pulls down the SDA line for t_{SU-DATA} before the low-to-high transition of SCL from the master. SDA also needs to remain pulled down for t_{HD-DAT} after a high-to-low transition of SCL.

During read, the master must signal an End-of-Data (EOD) to the slave by not generating an ACK bit (NAK) once the last bit has been clocked out of the slave. In this case, the slave will leave the data line released to enable the master to generate the Stop condition.

4.1.7 CLOCK STRETCHING

During the I^2C read operation, this device will hold the I^2C clock line low for t_{STRECH} after the falling edge of the ACK signal. In order to prevent bus contention, the master controller must release or hold the SCL line low during this period.

In addition, the master controller must provide eight consecutive clock cycles after generating the ACK bit from a read command. This allows the device to push out data from the SDA Output Shift registers. Missing clock cycles could result in bus contention. At the end of the data transmission, the master controller must provide the NAK bit, followed by a Stop bit to terminate communication.

FIGURE 4-2: Clock Stretching.

4.1.8 SEQUENTIAL READ

During a sequential read, the device transmits data from the proceeding register, starting from the previously set Register Pointer. The MCP9600/L00 maintains an Internal Address Pointer, which is incremented at the completion of each read data transmission, followed by an ACK from the master. A Stop bit terminates the sequential read.

FIGURE 4-3: Timing Diagram to Set a Register Pointer and Read a Two-Byte Data.

FIGURE 4-4: Timing Diagram to Set a Register Pointer, Write One Byte, and Read the Data.

FIGURE 4-5: Timing Diagram to Set a Register Pointer, Write Two Bytes, and Read the Data.

FIGURE 4-6: Timing Diagram to Sequential Read All Registers Starting from T_H Register.

5.0 FUNCTIONAL DESCRIPTION

The MCP9600/L00 temperature sensor consists of an 18-bit Delta-Sigma Analog-to-Digital Converter (ADC), which is used to measure the thermocouple voltage or EMF, a digital temperature sensor used to measure cold-junction or ambient temperature and a processor core which is used to compute the EMF to degree Celsius conversion using coefficients derived from the NIST ITS-90 coefficients. [Figure 5-1](#page-20-0) shows a block diagram of how these functions are structured in the device.

The MCP9600/L00 device has several registers that are user-accessible. These registers include the Thermocouple Temperature (cold-junction compensated), Hot-Junction Temperature, Cold-Junction Temperature, Raw ADC Data, user-programmable Alert Limit registers, and STATUS and Configuration registers.

The Temperature and the Raw ADC Data registers are read-only registers, used to access the thermocouple and the ambient temperature data. In addition, the four Alert Temperature registers are individually controlled, and can be used to detect a rising and/or a falling temperature change. If the ambient temperature drifts beyond the user-specified limits, the MCP9600/L00 device outputs an alert flag at the corresponding pin (refer to **[Section 5.3.3 "Alert Configuration Regis](#page-33-0)[ters"](#page-33-0)**). The alert limits can also be used to detect critical temperature events.

The MCP9600/L00 also provides STATUS and Configuration registers, which allow users to detect device statuses. The Configuration registers provide various features, such as adjustable temperature measurement resolution and Shutdown modes. The thermocouple types can also be selected using the Configuration registers.

The registers are accessed by sending a Register Pointer to the MCP9600/L00 using the serial interface. This is an 8-bit write-only pointer. [Register 5-1](#page-21-0) describes the pointer definitions.

REGISTER 5-1: REGISTER POINTER

bit 7-4 **Unimplemented:** Read as '0'

bit 3-0 **P<3:0>:** Pointer bits

TABLE 5-1: SUMMARY OF REGISTERS AND BIT ASSIGNMENTS

5.1 Thermocouple Temperature Sensor Registers

This device integrates three Temperature registers that are used to read the cold and hot-junction temperatures, and the sum of the two junctions to output the absolute thermocouple temperature. In addition, the Raw ADC Data register, which is used to derive the thermocouple temperature, is available. The following sections describe each register in detail.

5.1.1 THERMOCOUPLE TEMPERATURE $REGISTER (T_H)$

This register contains the cold-junction compensated and error-corrected thermocouple temperature in degree Celsius. The temperature data from this register is the absolute Thermocouple Hot-Junction temperature, T_H , to the specified accuracy (see **[Section 1.0 "Electrical](#page-2-4) [Characteristics"](#page-2-4)**. T_H is the sum of the values in the T_{Λ} and $T_{\rm C}$ registers, as shown in [Figure 5-2.](#page-23-1)

EQUATION 5-1: TEMPERATURE CONVERSION

Temperature $\geq 0^{\circ}C$

 $T_H = (UpperByte \times 16 + LowerByte/16)$

Temperature $< 0^{\circ}$ C

 $T_H = (UpperByte \times 16 + LowerByte/16) - 4096$

The temperature bits are in two's complement format; therefore, positive temperature data and negative temperature data are computed differently. [Equation 5-1](#page-23-0) shows how to convert the binary data to temperature in degree Celsius.

REGISTER 5-2: THERMOCOUPLE TEMPERATURE REGISTER (READ-ONLY)

bit 14-0 **TH:** Data in Two's Complement Format bits This register contains the error corrected and cold-junction compensated thermocouple temperature.

5.1.2 THERMOCOUPLE JUNCTIONS DELTA TEMPERATURE REGISTER (T∆)

This register contains the error corrected Thermocouple Hot-Junction temperature without the Cold-Junction compensation. The error correction methodology uses several coefficients to convert the digitized Thermocouple EMF voltage to degree Celsius. Each Thermocouple type has a unique set of coefficients as specified by NIST, and these coefficients are available in the configuration register for user selection as shown in [Figure 5-3.](#page-24-1)

EQUATION 5-2: TEMPERATURE CONVERSION

Temperature $\geq 0^{\circ}C$

T∆ = (UpperByte x 16 + LowerByte / 16)

Temperature $< 0°C$

T∆ = (UpperByte x 16 + LowerByte / 16) - 4096

The temperature bits are in two's complement format, therefore, positive temperature data and negative temperature data are computed differently, as shown in [Equation 5-2](#page-24-0).

FIGURE 5-3: Thermocouple Hot-Junction Register (T∆) Block Diagram.

REGISTER 5-3: HOT-JUNCTION TEMPERATURE REGISTER (READ-ONLY)

This register contains Thermocouple Hot-Junction temperature data.

5.1.3 COLD-JUNCTION/AMBIENT TEMPERATURE REGISTER (T_C)

The MCP9600/L00 integrates an ambient temperature sensor which can be used to measure the thermocouple cold-junction temperature. For accurate measurement, the device will have to be placed at close proximity to the thermocouple cold-junction to detect the junction ambient temperature. This is a 16-bit double-buffered, read-only register. The temperature resolution is user-selectable to 0.0625°C/LSb or 0.25°C/LSb resolutions and setting the resolution determines the temperature update rate, as shown in [Table 5-2.](#page-25-1)

EQUATION 5-3: TEMPERATURE CONVERSION

Temperature $\geq 0^{\circ}C$

 $T_C = (UpperByte \times 16 + LowerByte/16)$

Temperature < 0 °C

TC = (UpperByte x 16 + LowerByte/16) – 4096

The temperature bits are in two's complement format; therefore, positive temperature data and negative temperature data are computed differently, as shown in [Equation 5-3.](#page-25-0)

TABLE 5-2: RESOLUTION vs. CONVERSION TIME

Note 1: 's' is Sign and 'x' is unknown bit.

FIGURE 5-4: Thermocouple Cold-Junction Register (TC) Block Diagram.

REGISTER 5-4: COLD-JUNCTION TEMPERATURE REGISTER

bit 11-0 **T_C:** Data in Two's Complement Format bits This register contains the thermocouple cold-junction temperature or the device ambient temperature data. Bits 1 and 0 may remain clear ('0') depending on the status of the Resolution register.

5.1.4 ANALOG-TO-DIGITAL CONVERTER (ADC)

The MCP9600/L00 uses an 18-bit Delta-Sigma Analog-to-Digital Converter to digitize the Thermocouple EMF voltage and the data is available in the ADC register. The ADC measurement resolution is selectable, which enables the user to choose faster conversion times with reduced resolution. This feature is useful to detect fast transient temperatures.

Resolution/ Sensitivity (typical)	Conversion Time (typical)	Raw ADC Register Bit Format (Note 1:)			
18 bit/2 µV	320 ms	SSSS		SSSX XXXX XXXX	
					XXXX XXXX
16 bit/8 μ V	80 ms	SSSS		SSSX XXXX XXXX	
				xxxx xx00	
14 bit/32 µV	20 _{ms}	SSSS		SSSX XXXX XXXX	
				xxxx 0000	
12 bit/128 µV	5 _{ms}	SSSS		SSSX XXXX XXXX	
				xx00 0000	

TABLE 5-3: ADC RESOLUTION([27\)](#page-26-1)

Note 1: 's' is the Sign bit and 'x' is the ADC data bit. **2:** See **[Section 6.2.2 "Conversion Time vs.](#page-37-0) [Self-Heat"](#page-37-0)**.

REGISTER 5-5: ADC SAMPLE: 24-BIT REGISTER

FIGURE 5-5: Delta-Sigma Analog-to-Digital Converter, ADC Core Block Diagram.

bit 23-0 **ADC Data<23:0>:** Raw ADC Data bits Includes sign bits.

5.2 Sensor STATUS and Configuration Registers

This device provides various temperature and measurement Status bits which can be monitored regularly by the master controller. In addition, this device integrates various user-programmable features which can be useful to develop complex thermal management applications. The following sections describe each feature in detail.

REGISTER 5-6: STATUS REGISTER

5.2.1 STATUS REGISTER

The STATUS register contains several flag bits that indicate statuses, such as temperature alert, the ADC input range status for the selected thermocouple type and the Temperature register update status for both single conversion or Burst mode conversions.

5.2.2 THERMOCOUPLE SENSOR CONFIGURATION REGISTER

The MCP9600/L00 Sensor Configuration register is used to select the thermocouple sensor types and to select the digital filter options. This device supports eight thermocouple types. Each type has a unique set of error correction coefficients that are derived from the NIST Thermocouple EMF Voltage Conversion database.

In addition, this device integrates a first order. recursive Infinite Impulse Response (IIR) filter, also known as Exponential Moving Average (EMA). The filter uses the current new temperature sample and the previous filter output to calculate the next filter output. It also adds more weight to the current temperature data, allowing a faster filter response to the immediate change in temperature. This feature can be used to filter out fast thermal transients or thermal instability at the thermocouple hot-junction temperature. Writing this register resets the filter.

The filter equation is shown in [Equation 5-4](#page-28-1) and the Filter Coefficient n is user-selectable, from Level 0 to 7. A coefficient of 0 disables the filter function and a 7 coefficient provides a maximum digital filter. [Figure 5-6](#page-28-2) shows the filter response to a step function, which can be used to extrapolate the filter performance to various temperature changes.

EQUATION 5-4: DIGITAL FILTER

$$
Y = k \times X + (1 - k) \times Y_{-1}
$$

$$
k = 2 \times (2^n + 1)
$$

Where:

- Y = New filtered temperature in T_Λ
- $X =$ Current, unfiltered hot-junction temperatures
- Y_{-1} = Previous filtered temperature
	- n = User-selectable filter coefficient

REGISTER 5-7: SENSOR CONFIGURATION REGISTER

bit 7 **Unimplemented:** Read as '0'

5.2.3 DEVICE CONFIGURATION REGISTER

The device Configuration register allows the user to configure various functions, such as sensor measurement resolutions and Power modes. The Resolution register is used to select the sensor resolution for the desired temperature conversion time. When resolutions are changed, the change takes effect when the next measurement cycle begins.

This device integrates two Low-Power Operating modes: Shutdown mode and Burst mode, which can be selected using bit 0 and bit 1. When the Shutdown mode is executed, all power consuming activities are disabled and the operating current remains at I_{SHDN}. During the Shutdown mode, all registers are accessible; however, I^2C activity on the bus increases the current.

The Burst mode enables users to execute a given number of temperature samples (defined by bits<4-2>) before entering Shutdown mode. Each temperature sample is compared to the user-settable alert temperature limits, and if the alert conditions are true, then the device asserts the corresponding alert output. In addition, if the filter option is enabled, then the filter engine is applied to each temperature sample. The alert thresholds are also compared to the filtered temperature data. This feature is useful for battery power applications, where temperature is sampled upon request from the master controller.

FIGURE 5-7: Burst Mode Operation.

REGISTER 5-8: DEVICE CONFIGURATION REGISTER

5.3 Temperature Alert Registers

This device provides four Temperature Alert registers that are individually configured, which allow users to monitor multiple temperature zones with a single device. The following sections describe each alert feature in detail.

5.3.1 ALERT LIMIT REGISTERS

This device integrates four individually controlled Temperature Alert Limit registers. Each alert limit is individually set to detect a rising or falling temperature, or either the Thermocouple Temperature (T_H) register or the Cold-Junction (T_C) register. The corresponding alert limit outputs can also be enabled for temperature status indicators. All alert functions are configured using the Alert Limit Configuration registers [\(Register 5-11](#page-33-1)) and the alert output hysteresis function is set using the Alert Hysteresis registers ([Register 5-10\)](#page-32-0).

TABLE 5-4: ALERT LIMIT REGISTERS

REGISTER 5-9: ALERT LIMITS 1, 2, 3 AND 4 REGISTERS

FIGURE 5-9: Alert Limits Boundary Conditions and Output Characteristics when Set to Detect TH.

5.3.2 ALERT HYSTERESIS REGISTER

This device integrates four individually controlled temperature Alert Hysteresis registers for each alert output, with a range of 0°C to +255°C.

The alert hysteresis directions are set using bit 3 of the corresponding Alert Configuration registers ([Register 5-10\)](#page-32-0) to detect rising or falling temperatures. For rising temperatures, the hysteresis range is below the alert limit where, as for falling temperatures, the hysteresis range is above the alert limit, as shown in [Figure 5-10.](#page-32-1)

TABLE 5-5: ALERT HYSTERESIS REGISTERS

REGISTER 5-10: T_{HYSTx}: ALERT 1, 2, 3 AND 4 HYSTERESIS REGISTERS

bit 7-0 **Alert Hysteresis:** Alert Hysteresis Range 0x00 to 0xFF bits (which represents +1°C to +255°C)

 $-n =$ Value at POR $1' = B$ it is set $D' = B$ it is cleared $X = B$ it is unknown

FIGURE 5-10: Graphical Description of Alert Output Hysteresis Direction.

5.3.3 ALERT CONFIGURATION **REGISTERS**

This device integrates four individually controlled temperature alert outputs. Each output is configured for the corresponding alert output using the Alert Output Configuration registers.

The Configuration registers are used to enable each output, select the Alert Function mode as Comparator or Interrupt mode, active-high or active-low output, detect rising or falling temperatures and detect T_H or $T_{\rm C}$ Temperature registers.

The Comparator mode is useful for thermostat-type applications, such as on/off switches for fan controllers, buzzer or LED indicators. The alert output asserts and deasserts when the temperature exceeds the user-specified limit, and the user-specified hysteresis limit. The Interrupt mode is useful for interrupt driven microcontroller-based systems. The alert outputs are asserted each time the temperature exceeds the user-specified alert limit and hysteresis limits.

The microcontroller will have Acknowledged the interrupt signal from the corresponding alert output by clearing the interrupt using bit 7 of the corresponding Configuration register.

The Rise/Fall bit (bit 3) and the Monitor T_H/T_C bit (bit 4) can be used to detect and maintain the thermocouple temperature or the cold-junction temperature to the desired temperature window.

REGISTER 5-11: ALERT 1, 2, 3 AND 4 CONFIGURATION REGISTER

5.3.4 DEVICE ID AND REVISION ID REGISTER

The Device ID (Identification) and Revision ID register is a 16-bit read-only register, which can be used to identify this device among other devices on the $I²C$ bus. The upper 8 bits indicate the Device ID of 0x40, while the lower 8 bits indicate the device revision. The device revision byte is divided into nibbles, where the upper nibble indicates the major revision and the lower nibble indicates minor revisions for each major revision. The initial release is indicated by a major revision of '1' and a minor revision of '0' or 0x4010 (Refer to the Silicon Errata, DS80000741, for changes and revision IDs).

REGISTER 5-12: MCP9600/L00 DEVICE ID AND REVISION ID REGISTER

bit 15-8 **Device ID:** Device ID bits (0x40)

bit 7-0 **Revision:** Major/Minor Revision ID bits (0x10) for the initial Release, or Revision 1.0 (Refer to the Silicon Errata, DS80000741, for change date codes and revision IDs).

NOTES:

6.0 APPLICATION INFORMATION

6.1 Layout Considerations

The MCP9600/L00 does not require any additional components to digitize thermocouples. However, it is recommended that a decoupling capacitor of 0.1 µF to 1 μ F be used between the V_{DD} and GND pins. A high-frequency ceramic capacitor is recommended. It is necessary for the capacitor to be located as close as possible to the V_{DD} and ground pins of the device in order to provide effective noise protection.

In addition, good PCB layout is key for better thermal conduction from the PCB temperature to the sensor die. The PCB provides thermal conduction from the die to the thermocouple cold-junction; therefore, the component placement positioning and the copper layout techniques are key for optimum cold-junction compensation. The recommended implementation for optimum temperature sensitivity is to extend a copper ground pad around the device pins, as shown in [Figure 6-1.](#page-36-0)

FIGURE 6-1: Recommended PCB Layout.

6.1.1 COLD-JUNCTION COMPENSATION

Copper provides better thermal conductivity than PCB FR4 to the ambient temperature. It also provides better thermal conduction than the 5 mm x 5 mm MQFN plastic package, which houses the temperature sensor die. Therefore, when connecting the thermocouple wire to the PCB, it is recommended to place the ground copper between the thermocouple connector footprint, where dissimilar conductive material is attached to the PCB and the MCP9600/L00 exposed pad. This allows temperature to stabilize to the local ambient temperature (between the thermocouple connector junction and the PCB copper) and the copper trace conducts the temperature to the package exposed pad where the temperature sensor die is placed. The placement of the sensor exposed pad to the thermocouple connector junction greatly determines the temperature sensor's sensitivity to the local junction temperature changes. [Figure 6-2](#page-36-1) demonstrates the recommended techniques.

FIGURE 6-2: Recommended Component Placement.

6.2 Thermal Considerations

The potential for self-heating errors exist if the MCP9600/L00 SDA, SCL and alert outputs are heavily loaded (high current) with pull-up resistors and circuits, such as high-current LEDs or buzzer loads. The temperature rise due to self-heat increases the ambient temperature sensor output, resulting in an increased temperature offset error compared to the thermocouple cold-junction ambient temperature.

6.2.1 SELF-HEAT DURING OPERATION

During normal operation, the typical self-heating error is negligible due to the relatively small current consumption of the MCP9600/L00. However, this device integrates a processor to compute the equations necessary to convert the thermocouple EMF voltage to degrees Celsius. The processor also maintains the I^2C bus. During I^2C communication, the device operating current increases to $I_{DD} = 1.5$ mA (typical), ²C Active specification. If the bus is continually polled for data at frequent intervals, then the processor power dissipates heat to the temperature sensor and the effect of self-heat can be detected. Therefore, the recommended implementation is to maintain polling to no more than three times per temperature conversion period of 320 ms or use the Burst mode feature to manage self-heat (refer to **[Section 6.2.3 "Using Burst](#page-37-1) [Mode to Manage Self-Heat"](#page-37-1)**). [Equation 6-1](#page-37-2) can also be used to determine the effect of self-heat.

EQUATION 6-1: EFFECT OF SELF-HEATING

At room temperature (T_A = +25°C) with I_{DD} = 2.5 mA (maximum) and V_{DD} = 3.3V, the self-heating due to power dissipation, T_A , is 0.32°C for the MQFN package.

6.2.2 CONVERSION TIME vs. SELF-HEAT

Once the ADC completes digitization, the processor initiates the data computation routine for t_{CALC} , which also increases I_{DD} . During the 18-bit ADC conversion time (3 SPS, Samples per Second), the increased current lasts for approximately 5% of the one-second period. The effect of self-heat for the total power consumed per second, including the 5% t_{CALC} period, is negligible. However, as the ADC resolution is reduced from 18-bit to 16-bit, the power consuming t_{CALC} period increases to 20% per second. This change in resolution adds approximately 0.04°C (typical) temperature error due to self-heat. [Table 6-1](#page-37-3) provides an estimate for self-heat for all resolutions using [Equation 6-1.](#page-37-2)

In order to reduce the effects of self-heat for lower resolution settings, the Burst mode feature is recommended to manage the effects of self-heat.

TABLE 6-1: ADC RESOLUTION vs. SELF-HEAT

Resolution	SPS (typ.)	t _{CALC} Duration per Second	T_A	
18 bit		5%	0.0096° C	
16 bit	15	20%	0.0384 °C	
14 bit	60	80%	0.1536° C	
12 bit	240	100%	0.1920° C	

Note: $V_{DD} = 3.3V$ and $I_{DD} = 1.5$ mA (typical).

6.2.3 USING BURST MODE TO MANAGE SELF-HEAT

The Burst mode feature is useful to manage power dissipation while maintaining the device sensitivity to changes in temperature (see **[Section 5.2.3 "Device](#page-29-1) [Configuration Register"](#page-29-1)**). While the device is in Low-Power or Shutdown mode, the master controller executes Burst mode to sample temperature. The number of temperature samples and the measurement resolution settings are selected while executing the command. While in Burst mode, if the temperature data exceeds the alert limits, the device asserts the corresponding alert output. The alert outputs are used so the master controller does not need to continually poll the latest temperature data and potentially increase the temperature error.

In addition, with some applications monitoring several hundred degrees of temperature changes, 18-bit resolution may not be necessary. In this case, a fewer number of burst samples reducing the resolution enables the user to monitor fast transient temperatures at the burst intervals. The 12-bit ADC resolution provides approximately 3°C resolution (for Type K) and a new sample of temperature data is computed at approximately 20 ms intervals. Therefore, the number of Burst mode Samples per Second can be selected to manage the effects of self-heat using these estimates.

The temperature conversion status during Burst mode can also be momentarily polled (using bit 7 of [Register 5-6\)](#page-27-0) to detect whether the on-going sample bursts are completed. The master controller may terminate an on-going burst by executing a shutdown command or resetting the Burst mode by sending another burst command.

6.2.4 ALERT OUTPUTS

The alert outputs are intended to drive high-impedance loads. Typically, the outputs are connected to a microcontroller input pin. However, if the outputs are used to drive indicators, such as LEDs or buzzers, then a buffer circuit is recommended in order to minimize the effects of self-heat due to the applied load (see [Figure 6-3](#page-37-4)).

FIGURE 6-3: Alert Output Buffer for LED Indicator.

6.3 Device Features

 $6.3.1$ I²C ADDRESSING

The MCP9600/L00 supports up to eight devices on the I²C bus. Applications, such as large thermal management racks with several thermocouple sensor interfaces, are able to monitor various temperature zones with minimal pin count microcontrollers. This reduces the total solution cost, while providing a highly accurate thermal management solution using the MCP9600/L00.

FIGURE 6-4: *2C Address Selection Implementation.*

6.3.2 INPUT IMPEDANCE

The MCP9600/L00 uses a switched capacitor amplifier input stage to gain the input signal to a maximum resolution of 2 µV/LSb at an 18-bit ADC setting. An internal input capacitor is used for charge storage. The differential input impedance, $Z_{\text{IN-DF}}$, is dominated by the sampling capacitor and the switched capacitor amplifier sampling frequency. During a sampling period, the charging and discharging of the sampling capacitor creates dynamic input currents at the input pins. Adding a 10-100 nF capacitor between the inputs can improve stability.

Since the sampling capacitor is only switching to the input pins during a conversion process, the input impedance is only valid during conversion periods. During Low-Power or Shutdown mode, the input amplifier stage is disabled; therefore, the input impedance is Z_{IN-CM} , which is due to the leakage current from ESD protection diodes, as shown in [Figure 6-5](#page-38-1).

6.3.3 OPEN AND SHORT DETECTION **CIRCUITS**

External circuits can be added to detect the thermocouple status as open (physically disconnected) or as short (thermocouple wire in contact with the system ground or V_{DD}). If a passive circuit is added to the input stage, then the circuit loading effect to the MCP9600/L00 ADC inputs must be considered. System calibration is also required to ensure proper accuracy. In addition, external loads can degrade the device performance, such as input offset, gain and Integral Nonlinearity (INL) errors. If a low-impedance active circuit is added, then both offset and gain errors must be calibrated.

6.3.3.1 Open-Circuit Detection Technique

For open-circuit detection, the Input Range Flag bit, bit 4 of the STATUS register ([Register 5-6\)](#page-27-0), can be used to detect open-circuit conditions. This would require a few external resistors, as shown in [Figure 6-6.](#page-39-1) The passive circuit does not affect the MCP9600/L00 accuracy (the recommended value for R_B is set to 1 M Ω). When the thermocouple is connected, the input Common-mode voltage is 0.5 $*$ V_{DD}. When the thermocouple is disconnected, the voltage at the V_{IN+} input is 0.66 $*V_{DD}$ and the voltage at the V_{IN-} input is pulled down to V_{SS} . This change forces the Input Range Flag bit to be set. The master controller can momentarily poll the Status bit to detect the open-circuit condition.

6.3.4 ALIASING AND ANTI-ALIASING FILTER

Aliasing occurs when the input signal contains time varying signals with frequency greater than half the sample rate. In the aliasing conditions, the ADC can output unexpected codes. The ADC integrates a first order sync filter; however, an external anti-aliasing filter can provide an added filter for high noise applications. This can be done with a simple RC low-pass filter at the inputs, as shown in [Figure 6-7](#page-39-2). Open-circuit detection resistors can also be added, as shown in [Figure 6-8.](#page-39-3)

FIGURE 6-7: Adding a Low-Pass Filter.

FIGURE 6-8: Adding Open-Circuit Detection Resistors with an Input Low-Pass Filter.

6.3.5 ESD PROTECTION USING FERRITE BEADS

Ferrite beads are highly recommended to protect the MCP9600/L00 and other circuits from ESD discharge through the thermocouple wire. The beads suppress fast transient signals, such as ESD, and can be added in-line to the ADC inputs. In addition, protection diodes are also recommended, as shown in [Figure 6-9.](#page-39-0)

7.0 PACKAGING INFORMATION

7.1 Package Marking Information

PIN 1² | VVVVVVVV | PIN 1 20-Lead MQFN (5x5x1.0 mm) Example

© 2015-2018 Microchip Technology Inc. **DS20005426D-page 41**

20-Lead More Thin Plastic Quad Flat, No Lead Package (NU) - 5x5x1.0 mm Body [MQFN] - (Also called VQFN)

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging **Note:**

Microchip Technology Drawing C04-186A Sheet 1 of 2

20-Lead More Thin Plastic Quad Flat, No Lead Package (NU) - 5x5x1.0 mm Body [MQFN] - (Also called VQFN)

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging **Note:**

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

- 2. Package is saw singulated
- 3. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-186A Sheet 2 of 2

20-Lead More Thin Plastic Quad Flat, No Lead Package (NU) - 5x5x1.0 mm Body [MQFN] - (Also called VQFN)

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging **Note:**

RECOMMENDED LAND PATTERN

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

2. For best soldering results, thermal vias, if used, should be filled or tented to avoid solder loss during reflow process

Microchip Technology Drawing C04-286B

APPENDIX A: REVISION HISTORY

Revision D (August 2018)

The following is the list of modifications:

1. Added the MCP96L00 device and related information throughout the document.

Revision C (September 2017)

The following is the list of modifications:

- 1. Updated **[Figure 4-3](#page-16-1)**, **[Equation 5-1](#page-23-2)**, **[Equation 5-2](#page-24-2)** and **[Equation 5-3](#page-25-3)**.
- 2. Updated **[Section 6.3.3.1 "Open-Circuit](#page-39-4) [Detection Technique"](#page-39-4)**.
- 3. Various typographical edits.

Revision B (June 2016)

The following is the list of modifications:

- 1. Corrected the pin description error for pins 19 and 20 on page 1.
- 2. Added the MCP9600 Evaluation Board picture on page 2.
- 3. Added **[Section 6.3.3.1 "Open-Circuit](#page-39-4) [Detection Technique"](#page-39-4)** and updated **[Section 6.3.4 "Aliasing and Anti-Aliasing](#page-39-5) [Filter"](#page-39-5)** and **[Section 6.3.5 "ESD Protection](#page-39-6) [Using Ferrite Beads"](#page-39-6)**.
- 4. Updated the [Product Identification System](#page-46-0) section.

Revision A (August 2015)

• Original Release of this Document.

NOTES:

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

NOTES:

Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION, INCLUDING BUT NOT LIMITED TO ITS CONDITION, QUALITY, PERFORMANCE, MERCHANTABILITY OR FITNESS FOR PURPOSE**.** Microchip disclaims all liability arising from this information and its use. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.

QUALITY MANAGEMENT SYSTEM CERTIFIED BY DNV $=$ **ISO/TS 16949** $=$

Trademarks

The Microchip name and logo, the Microchip logo, AnyRate, AVR, AVR logo, AVR Freaks, BitCloud, chipKIT, chipKIT logo, CryptoMemory, CryptoRF, dsPIC, FlashFlex, flexPWR, Heldo, JukeBlox, KeeLoq, Kleer, LANCheck, LINK MD, maXStylus, maXTouch, MediaLB, megaAVR, MOST, MOST logo, MPLAB, OptoLyzer, PIC, picoPower, PICSTART, PIC32 logo, Prochip Designer, QTouch, SAM-BA, SpyNIC, SST, SST Logo, SuperFlash, tinyAVR, UNI/O, and XMEGA are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

ClockWorks, The Embedded Control Solutions Company, EtherSynch, Hyper Speed Control, HyperLight Load, IntelliMOS, mTouch, Precision Edge, and Quiet-Wire are registered trademarks of Microchip Technology Incorporated in the U.S.A. Adjacent Key Suppression, AKS, Analog-for-the-Digital Age, Any Capacitor, AnyIn, AnyOut, BodyCom, CodeGuard, CryptoAuthentication, CryptoAutomotive, CryptoCompanion, CryptoController, dsPICDEM, dsPICDEM.net, Dynamic Average Matching, DAM, ECAN, EtherGREEN, In-Circuit Serial Programming, ICSP, INICnet, Inter-Chip Connectivity, JitterBlocker, KleerNet, KleerNet logo, memBrain, Mindi, MiWi, motorBench, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, PowerSmart, PureSilicon, QMatrix, REAL ICE, Ripple Blocker, SAM-ICE, Serial Quad I/O, SMART-I.S., SQI, SuperSwitcher, SuperSwitcher II, Total Endurance, TSHARC, USBCheck, VariSense, ViewSpan, WiperLock, Wireless DNA, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

Silicon Storage Technology is a registered trademark of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2018, Microchip Technology Incorporated, All Rights Reserved. ISBN: 978-1-5224-3434-4

2015-2018 Microchip Technology Inc. DS20005426D-page 49

Worldwide Sales and Service

AMERICAS

Corporate Office 2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200 Fax: 480-792-7277 Technical Support: [http://www.microchip.com/](http://support.microchip.com) support

Web Address: www.microchip.com

Atlanta Duluth, GA Tel: 678-957-9614 Fax: 678-957-1455

Austin, TX Tel: 512-257-3370

Boston Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL Tel: 630-285-0071 Fax: 630-285-0075

Dallas Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit Novi, MI Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Indianapolis Noblesville, IN Tel: 317-773-8323 Fax: 317-773-5453 Tel: 317-536-2380

Los Angeles Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608 Tel: 951-273-7800

Raleigh, NC Tel: 919-844-7510

New York, NY Tel: 631-435-6000 **San Jose, CA**

Tel: 408-735-9110 Tel: 408-436-4270

Canada - Toronto Tel: 905-695-1980 Fax: 905-695-2078

ASIA/PACIFIC

Australia - Sydney Tel: 61-2-9868-6733

China - Beijing Tel: 86-10-8569-7000 **China - Chengdu**

Tel: 86-28-8665-5511 **China - Chongqing** Tel: 86-23-8980-9588

China - Dongguan Tel: 86-769-8702-9880

China - Guangzhou Tel: 86-20-8755-8029

China - Hangzhou Tel: 86-571-8792-8115

China - Hong Kong SAR Tel: 852-2943-5100

China - Nanjing Tel: 86-25-8473-2460

China - Qingdao Tel: 86-532-8502-7355

China - Shanghai Tel: 86-21-3326-8000

China - Shenyang Tel: 86-24-2334-2829

China - Shenzhen Tel: 86-755-8864-2200

China - Suzhou Tel: 86-186-6233-1526 **China - Wuhan**

Tel: 86-27-5980-5300

China - Xian Tel: 86-29-8833-7252

China - Xiamen Tel: 86-592-2388138 **China - Zhuhai**

Tel: 86-756-3210040

ASIA/PACIFIC

India - Bangalore Tel: 91-80-3090-4444

India - New Delhi Tel: 91-11-4160-8631 **India - Pune**

Tel: 91-20-4121-0141 **Japan - Osaka**

Tel: 81-6-6152-7160 **Japan - Tokyo**

Tel: 81-3-6880- 3770 **Korea - Daegu**

Tel: 82-53-744-4301 **Korea - Seoul**

Tel: 82-2-554-7200 **Malaysia - Kuala Lumpur**

Tel: 60-3-7651-7906

Malaysia - Penang Tel: 60-4-227-8870

Philippines - Manila Tel: 63-2-634-9065

Singapore Tel: 65-6334-8870

Taiwan - Hsin Chu Tel: 886-3-577-8366

Taiwan - Kaohsiung Tel: 886-7-213-7830

Taiwan - Taipei Tel: 886-2-2508-8600

Thailand - Bangkok Tel: 66-2-694-1351

Vietnam - Ho Chi Minh Tel: 84-28-5448-2100

Tel: 49-8931-9700 **Germany - Haan**

Tel: 49-2129-3766400 **Germany - Heilbronn**

EUROPE Austria - Wels Tel: 43-7242-2244-39 Fax: 43-7242-2244-393 **Denmark - Copenhagen** Tel: 45-4450-2828 Fax: 45-4485-2829 **Finland - Espoo** Tel: 358-9-4520-820 **France - Paris** Tel: 33-1-69-53-63-20 Fax: 33-1-69-30-90-79 **Germany - Garching**

Tel: 49-7131-67-3636 **Germany - Karlsruhe** Tel: 49-721-625370

Germany - Munich Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Germany - Rosenheim Tel: 49-8031-354-560

Israel - Ra'anana Tel: 972-9-744-7705

Italy - Milan Tel: 39-0331-742611 Fax: 39-0331-466781

Italy - Padova Tel: 39-049-7625286

Netherlands - Drunen Tel: 31-416-690399 Fax: 31-416-690340

Norway - Trondheim Tel: 47-7288-4388

Poland - Warsaw Tel: 48-22-3325737

Romania - Bucharest Tel: 40-21-407-87-50

Spain - Madrid Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

Sweden - Gothenberg Tel: 46-31-704-60-40

Sweden - Stockholm Tel: 46-8-5090-4654

UK - Wokingham Tel: 44-118-921-5800 Fax: 44-118-921-5820

Mouser Electronics

Authorized Distributor

Click to View Pricing, Inventory, Delivery & Lifecycle Information:

[Microchip](https://www.mouser.com/microchip):

 [MCP9600-I/MX](https://www.mouser.com/access/?pn=MCP9600-I/MX) [MCP9600T-I/MX](https://www.mouser.com/access/?pn=MCP9600T-I/MX) [MCP9600-E/MX](https://www.mouser.com/access/?pn=MCP9600-E/MX) [MCP9600T-E/MX](https://www.mouser.com/access/?pn=MCP9600T-E/MX) [MCP96L00T-E/MX](https://www.mouser.com/access/?pn=MCP96L00T-E/MX) [MCP96L00-E/MX](https://www.mouser.com/access/?pn=MCP96L00-E/MX) [MCP96RL00-E/MX](https://www.mouser.com/access/?pn=MCP96RL00-E/MX) [MCP96RL00T-E/MX](https://www.mouser.com/access/?pn=MCP96RL00T-E/MX)

Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits,General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.

Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный) **Факс:** 8 (812) 320-02-42 **Электронная почта:** org@eplast1.ru **Адрес:** 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.