Dual 5 A High Speed Low-Side MOSFET Drivers with Enable

NCP81071 is a high speed dual low-side MOSFETs driver. It is capable of providing large peak currents into capacitive loads. This driver can deliver 5 A peak current at the Miller plateau region to help reduce the Miller effect during MOSFETs switching transition. This driver also provides enable functions to give users better control capability in different applications. ENA and ENB are implemented on pin 1 and pin 8 which were previously unused in the industry standard pin-out. They are internally pulled up to driver's input voltage for active high logic and can be left open for standard operations. This part is available in MSOP8–EP package, SOIC8 package and WDFN8 3 mm x 3 mm package.

Features

- High Current Drive Capability ±5 A
- TTL/CMOS Compatible Inputs Independent of Supply Voltage
- Industry Standard Pin-out
- Enable Functions for Each Driver
- 8 ns Typical Rise and 8 ns Typical Fall Times with 1.8 nF Load
- Typical Propagation Delay Times of 20 ns with Input Falling and 20 ns with Input Rising
- Input Voltage from 4.5 V to 20 V
- Dual Outputs can be Paralleled for Higher Drive Current
- These Devices are Pb–Free, Halogen Free/BFR Free and are RoHS Compliant

Applications

- Server Power
- Telecommunication, Datacenter Power
- Synchronous Rectifier
- Switch Mode Power Supply
- DC/DC Converter
- Power Factor Correction
- Motor Drive
- Renewable Energy, Solar Inverter



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(Note: Microdot may be in either location)



ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 2 of this data sheet.

ORDERING INFORMATION

Part Number	Package Type	Tape and Reel	Output Configuration	Temperature Range (°C)
NCP81071ADR2G	SOIC-8	2500	dual inverting	
NCP81071BDR2G	SOIC-8	2500	dual non inverting	
NCP81071CDR2G	SOIC-8	2500	One inverting one non inverting	
NCP81071AZR2G	MSOP8 EP	3000	dual inverting	
NCP81071BZR2G	MSOP8 EP	3000	dual non inverting	-40 to +140
NCP81071CZR2G	MSOP8 EP	3000	One inverting one non inverting	
NCP81071AMNTXG	WDFN8 3x3	3000	dual inverting	
NCP81071BMNTXG	WDFN8 3x3	3000	dual non inverting]
NCP81071CMNTXG	WDFN8 3x3	3000	One inverting one non inverting	



Figure 1. NCP81071 Dual Non-inverting Block Diagram

Table 1. PIN DESCRIPTION

Pin No.	Symbol	Description
1	ENA	Enable input for the driver channel A with logic compatible threshold and hysteresis. This pin is used to enable and disable the driver output. It is internally pulled up to VDD with a 200 k Ω resistor for active high operation. The output of the pin when the device is disabled will be always low.
2	INA	Input of driver channel A which has logic compatible threshold and hysteresis. If not used, this pin should be connected to either VDD or GND. It should not be left unconnected.
3	GND	Common ground. This ground should be connected very closely to the source of the power MOSFET.
4	INB	Input of driver channel B which has logic compatible threshold and hysteresis. If not used, this pin should be connected to either VDD or GND. It should not be left unconnected.
5	OUTB	Output of driver channel B. The driver is able to provide 5 A drive current to the gate of the power MOSFET.
6	VDD	Supply voltage. Use this pin to connect the input power for the driver device.
7	OUTA	Output of driver channel A. The driver is able to provide 5 A drive current to the gate of the power MOSFET.
8	ENB	Enable input for the driver channel B with logic compatible threshold and hysteresis. This pin is used to enable and disable the driver output. It is internally pulled up to VDD with a 200 k Ω resistor for active high operation. The output of the pin when the device is disabled will be always low.

TYPICAL APPLICATION CIRCUIT



Table 2. ABSOLUTE MAXIMUM RATINGS

		Va	Value	
		Min	Max	Unit
Supply Voltage	VDD	-0.3	24	V
Output Current (DC)	lout_dc		0.3	А
Output Current (Pulse < 0.5 μ s)	lout_pulse		6	А
Input Voltage	INA, INB	-6	VDD+0.3	V
Enable Voltage	ENA, ENB	-0.3	VDD+0.3	
Output Voltage	OUTA, OUTB	-0.3	VDD+0.3	V
Output Voltage (Pulse < 0.5 μs)	OUTA, OUTB	-1.0	VDD+1.0	V
Junction Operation Temperature	TJ	-40	150	°C
Storage Temperature	T _{stg}	-65	160	
Electrostatic Discharge	Human body model, HBM	4	000	V
	Charge device model, CDM	1	000	1
OUTA OUTB Latch-up Protection		Ę	500	mA

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

Table 3. RECOMMENDED OPERATING CONDITIONS

Parameter	Rating	Unit
VDD supply Voltage	4.5 to 20	V
INA, INB input voltage	–5 to VDD	V
ENA, ENB input voltage	0 to VDD	V
Junction Temperature Range	-40 to +140	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

Table 4. THERMAL INFORMATION

Package	θ _{JA} (°C/W)	θ _{JC} (°C/W)
SOIC-8	115	50
MSOP-8 EP	39	4.7
WDFN8 3x3	39	4.7

Table 5. INPUT/OUTPUT TABLE

				NCP81071A		NCP8 ²	1071B	NCP8	1071C
ENA	ENB	INA	INB	OUTA	OUTB	OUTA	OUTB	OUTA	OUTB
Н	Н	L	L	Н	Н	L	L	Н	L
Н	Н	L	Н	Н	L	L	Н	Н	Н
Н	Н	Н	L	L	н	Н	L	L	L
Н	Н	Н	Н	L	L	Н	Н	L	Н
L	L	Any	Any	L	L	L	L	L	L
Any	Any	x (Note 1)	x (Note 1)	L	L	L	L	L	L
x (Note 1)	x (Note 1)	L	L	Н	Н	L	L	Н	L
x (Note 1)	x (Note 1)	L	Н	Н	L	L	Н	Н	Н
x (Note 1)	x (Note 1)	Н	L	L	н	Н	L	L	L
x (Note 1)	x (Note 1)	Н	Н	L	L	Н	Н	L	Н

1. Floating condition, internal resistive pull up or pull down configures output condition

PRODUCT MATRIX







NCP81071A

NCP81071B

NCP81071C

Table 6. ELECTRICAL CHARACTERISTICS

(Typical values: V_{DD} =12 V, 1 μ F from V_{DD} to GND, $T_A = T_J = -40^{\circ}$ C to 140°C, typical at $T_{AMB} = 25^{\circ}$ C, unless otherwise specified)

Parameter	Parameter Symbol Test Conditions		Min	Тур	Max	Units
SUPPLY VOLTAGE		-		-	-	
VDD Under Voltage Lockout (rising)	V _{CCR}	VDD rising	3.5	4.0	4.5	V
VDD Under Voltage Lockout (hysteresis)	V _{CCH}			400		mV
Operating Current (no switching)	I _{DD}	INA = 0, INB = 5 V, ENA = ENB = 0 INA = 5 V, INB = 0, ENA = ENB = 0 INA = 0, INB = 5 V, ENA = ENB = 5 V INA = 5 V, INB = 0, ENA = ENB = 5 V		1.4	3	mA
VDD Under Voltage Lockout to Output Delay (Note 2)		VDD rising		10		μS
INPUTS		·				
High Threshold	V _{thH}	Input rising from logic low	1.8	2.0	2.2	V
Low Threshold	V _{thL}	Input falling from logic high	0.8	1.0	1.2	V
INA, INB Pull–Up Resistance		OUTA = OUTB = Inverter Configuration		200		kΩ
INA, INB Pull-Down Resistance		OUTA = OUTB = Buffer Configuration		200		kΩ
OUTPUTS		-		-	-	
Output Resistance High	R _{OH}	IOUT = -10 mA		0.8	2	Ω
Output Resistance Low	R _{OL}	IOUT = +10 mA		0.8	2	Ω
Peak Source Current (Note 3)	I _{Source}	OUTA/OUTB = GND 200 ns Pulse		5		A
Miller Plateau Source Current (Note 3)	I _{Source}	OUTA/OUTB = 5.0 V 200 ns Pulse		4.5		A
Peak Sink Current (Note 3)	I _{Sink}	OUTA/OUTB = VDD 200 ns Pulse		5		A
Miller Plateau Sink Current (Note 3)	I _{Sink}	OUTA/OUTB = 5.0 V 200 ns Pulse		3.5		A
ENABLE						
High-Level Input Voltage	V _{IN_H}	Low to High Transition	1.8	2.0	2.2	V
Low-Level Input Voltage	V _{IN_L}	High to Low Transition	0.8	1.0	1.2	V
ENA, ENB pull-up resistance				200		kΩ
Propagation Delay Time (EN to OUT) (Note 4)	t _{d3}	C _{Load} = 1.8 nF		20	35	ns
Propagation Delay Time (EN to OUT) (Note 4)	t _{d4}	C _{Load} = 1.8 nF		20	35	ns
SWITCHING CHARACTERISTICS		•				
Propagation Delay Time Low to High, IN Rising (IN to OUT) (Note 4)	t _{d1}	C _{Load} = 1.8 nF		20	35	ns
Propagation Delay Time High to Low, IN Falling (IN to OUT) (Note 4)	t _{d2}	C _{Load} = 1.8 nF		20	35	ns
Rise Time (Note 4)	t _r	C _{Load} = 1.8 nF 8		15	ns	
Fall Time (Note 4)	t _f	C _{Load} = 1.8 nF		8	15	ns
Delay Matching between 2 Channels (Note 5)	t _m	INA = INB, OUTA and OUTB at 50% Transition Point		1	4	ns

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions. 2. Guaranteed by design.

3. Not production tested, guaranteed by design and statistical analysis.

See timing diagrams in Figure 2, Figure 3, Figure 4 and Figure 5.
 Guaranteed by characterization.



Figure 2. Enable Function for Non-inverting Input Driver Operation



Figure 4. Non-inverting Input Driver Operation



Figure 3. Enable Function for Inverting Input Driver Operation



Figure 5. Inverting Input Driver Operation







Figure 23. Output Behavior vs. Supply Voltage NCP81071A (Inverting) 10 nF between Output and GND, INA = GND, ENA = VDD

Figure 22. Output Behavior vs. Supply Voltage NCP81071A (Inverting) 10 nF between Output and GND, INA = GND, ENA = VDD



Figure 24. Output Behavior vs. Supply Voltage NCP81071A (Inverting) 10 nF between Output and GND, INA = VDD, ENA = VDD



Figure 26. Output Behavior vs. Supply Voltage NCP81071B (Non–Inverting) 10 nF between Output and GND, INA = VDD, ENA = VDD



Figure 28. Output Behavior vs. Supply Voltage NCP81071B (Non–Inverting) 10 nF between Output and GND, INA = GND, ENA = VDD



Figure 25. Output Behavior vs. Supply Voltage NCP81071A (Inverting) 10 nF between Output and GND, INA = VDD, ENA = VDD



Figure 27. Output Behavior vs. Supply Voltage NCP81071B (Non–Inverting) 10 nF between Output and GND, INA = VDD, ENA = VDD



Figure 29. Output Behavior vs. Supply Voltage NCP81071B (Non–Inverting) 10 nF between Output and GND, INA = GND, ENA = VDD

LAYOUT GUIDELINES

The switching performance of NCP81071 highly depends on the design of PCB board. The following layout design guidelines are recommended when designing boards using these high speed drivers.

Place the driver as close as possible to the driven MOSFET.

Place the bypass capacitor between VDD and GND as close as possible to the driver to improve the noise filtering. It is preferred to use low inductance components such as chip capacitor and chip resistor. If vias are used, connect several paralleled vias to reduce the inductance of the vias.

Minimize the turn-on/sourcing current and turn-off/sinking current paths in order to minimize stray inductance. Otherwise high di/dt established in these loops with stray inductance can induce significant voltage spikes on the output of the driver and MOSFET Gate terminal.

Keep power loops as short as possible by paralleling the source and return traces (flux cancellation).

Keep low level signal lines away from high level power lines with a lot of switching noise.

Place a ground plane for better noise shielding. Beside noise shielding, ground plane is also useful for heat dissipation.

NCP81071 DFN and MSOP package have thermal pad for: 1) quiet GND for all the driver circuits; 2) heat sink for the driver. This pad must be connected to a ground plane and no switching currents from the driven MOSFET should pass through the ground plane under the driver. To maximize the heatsinking capability, it is recommended several ground layers are added to connect to the ground plane and thermal pad. A via array within the area of package can conduct the heat from the package to the ground layers and the whole PCB board. The number of vias and the size of ground plane are determined by the power dissipation of NCP81071 (VDD voltage, switching frequency and load condition), the air flow condition and its maximum junction temperature.

PACKAGE DIMENSIONS



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
 2. CONTROLLING DIMENSION: MILLIMETER.
 3. DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
 4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
 5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
 6. 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

	MILLIN	IETERS	INCHES				
DIM	MIN	MAX	MIN	MAX			
Α	4.80	5.00	0.189	0.197			
в	3.80	4.00	0.150	0.157			
С	1.35	1.75	0.053	0.069			
D	0.33	0.51	0.013	0.020			
G	1.27	7 BSC	0.050 BSC				
н	0.10	0.25	0.004	0.010			
J	0.19	0.25	0.007	0.010			
Κ	0.40	1.27	0.016	0.050			
М	0 °	8 °	0 °	8 °			
Ν	0.25	0.50	0.010	0.020			
S	5.80	6.20	0.228	0.244			

SOLDERING FOOTPRINT*



*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

PACKAGE DIMENSIONS



NOTES:

- NOTES: 1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M, 1994. 2. CONTROLLING DIMENSIONS: MILLIMETERS. 3. DIMENSION & DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.10 MM IN EXCESS OF MAXIMUM MATERIAL CONDITION. 4. DIMENSION D DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS. MOLD FLASH, PROTRUSIONS, OR GATE BURRS. MOLD FLASH, PROTRUSIONS, OR GATE BURRS SHALL NOT EXCEED 0.15 MM PER SIDE. DIMENSION E DOES NOT INCLUDE INTER-LEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 MM PER SIDE. DIMENSIONS DAND E ARE DETERMINED AT DATUM F. 5. DATUMS A AND B TO BE DETERMINED AT DATUM F. 6. A1 IS DEFINED AS THE VERTICAL DISTANCE FROM THE SEATING PLANE TO THE LOWEST POINT ON THE PACKAGE BODY.
- BODY.

	MILLIMETERS				
DIM	MIN MAX				
Α		1.10			
A1	0.05	0.15			
b	0.25	0.40			
с	0.13	0.23			
D	2.90	3.10			
D2	1.78	REF			
E	4.75	5.05			
E1	2.90	3.10			
E2	1.42	REF			
е	0.65	BSC			
L	0.40	0.70			
L2	0.254 BSC				

DIMENSIONS: MILLIMETERS

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

PACKAGE DIMENSIONS



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