

Printers

Copiers PCs

**Cell Phones** 

**Cable Modems** 

**Computer Peripherals** 

## **DS1086L**

Features

## 3.3V Spread-Spectrum EconOscillator

66.6MHz

Output

## **General Description**

The DS1086L EconOscillator<sup>™</sup> is a 3.3V programmable clock generator that produces a spread-spectrum (dithered) square-wave output of frequencies from 130kHz to 66.6MHz. The selectable dithered output reduces radiated-emission peaks by dithering the frequency 0.5%,1%, 2%, 4%, or 8% below the programmed frequency. The DS1086L has a power-down mode and an output-enable control for power-sensitive applications. All the device settings are stored in non-volatile (NV) EEPROM memory allowing it to operate in stand-alone applications.

### Applications

Nonvolatile Settings
2.7V to 3.6V Supply

Adjustable Dither Rate

♦ 2-Wire Serial Interface

No External Timing Components Required

User-Programmable Square-Wave Generator

♦ Frequencies Programmable from 130kHz to

♦ 0.5%, 1%, 2%, 4%, or 8% Selectable Dithered

- Power-Down Mode
- ♦ 5kHz Master Frequency Step Size

Glitchless Output-Enable Control

- EMI Reduction
- ♦ Industrial Temperature Range: -40°C to +85°C

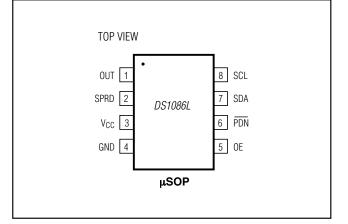
## \_Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
DS1086LU	-40°C to +85°C	8 µSOP (118 mils)
DS1086LU+	-40°C to +85°C	8 µSOP (118 mils)

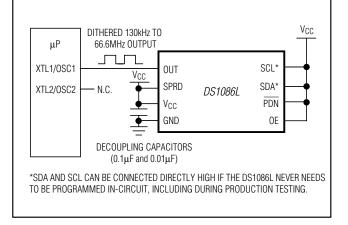
Note: Contact the factory for custom settings.

+Denotes a lead(Pb)-free/RoHS-compliant package.

## **Pin Configuration**



## Typical Operating Circuit



EconOscillator is a trademark of Maxim Integrated Products, Inc.

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

### **ABSOLUTE MAXIMUM RATINGS**

Voltage Range on V <sub>CC</sub> Relative to Ground	0.5V to +6.0V
Voltage Range on SPRD, PDN, OE, SDA, and	d SCL
Relative to Ground*	0.5 to (V <sub>CC</sub> + 0.5V)
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
µSOP (derate 4.5mW/°C above +70°C)	
	1000 1 0500

Operating	Temperature	Range	 40°C to	+85°C

Programming Temperature Range	
Junction Temperature	+150°C
Storage Temperature Range	55°C to +150°C
Soldering Temperature (reflow)	
Lead(Pb)-free	+260°C
Containing lead(Pb)	+240°C

\* This voltage must not exceed 6.0V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **RECOMMENDED DC OPERATING CONDITIONS**

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Supply Voltage	Vcc	(Note 1)	2.7	3.3	3.6	V
High-Level Input Voltage (SDA, SCL, SPRD, PDN, OE)	VIH		0.7 x V <sub>CC</sub>		V <sub>CC</sub> + 0.3	V
Low-Level Input Voltage (SDA, SCL SPRD, PDN, OE)	VIL		-0.3		0.3 x V <sub>CC</sub>	V

### **DC ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = 2.7V \text{ to } 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
High-Level Output Voltage (OUT)	VOH	$I_{OH} = -4mA$ , $V_{CC} = min$	2.4			V
Low-Level Output Voltage (OUT)	VOL	$I_{OL} = 4mA$	0		0.4	V
	V <sub>OL1</sub>	3mA sink current	0		0.4	V
Low-Level Output Voltage (SDA)	V <sub>OL2</sub>	6mA sink current	0		0.6	v
High-Level Input Current	IIН	$V_{CC} = 3.6V$			1	μA
Low-Level Input Current	١ <sub>IL</sub>	$V_{IL} = OV$	-1			μΑ
Supply Current (Active)	ICC	$C_L = 15 pF$ (output at default frequency)			10	mA
Standby Current (Power-Down)	ICCQ	Power-down mode			10	μA

## MASTER OSCILLATOR CHARACTERISTICS

 $(V_{CC} = 2.7V \text{ to } 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	COND	ITIONS	MIN	ТҮР	MAX	UNITS
Master Oscillator Frequency	fosc	(Note 2)		33.3		66.6	MHz
Default Master Oscillator Frequency	f <sub>0</sub>	Factory-programmed	default		48.65		MHz
Master Oscillator Frequency	Δ <u>fo</u> fo	V <sub>CC</sub> = 3.3V, T <sub>A</sub> = +25°C	Default frequency(f <sub>0</sub> )	-0.5		+0.5	%
Tolerance	fo	(Notes 3,17)	DAC step size	-0.5		+0.5	,
Voltage Frequency Variation	$\Delta \underline{f}_V$	Over voltage range,	Default frequency	-0.75		+0.75	%
voltage rrequency variation	fo	$T_A = +25^{\circ}C$ (Note 4)	DAC step size	-0.75		+0.75	/0
		Over temperature	Default frequency	-2.0		+0.75	
Temperature Frequency Variation	Δ <u>f</u> T f <sub>0</sub>	range, $V_{CC} = 3.3V$	66.6MHz	-2.0		+0.75	%
	10	(Note 5)	33.3MHz	-2.5		+0.75	
Dither Frequency Range (Note 6)		Prescaler bits JS2, JS	1, JS0 = 000		0.5		
		Prescaler bits JS2, JS1, JS0 = 001			1		
	<u>∆f</u> f <sub>0</sub>	Prescaler bits JS2, JS1, JS0 = 010			2		%
		Prescaler bits JS2, JS	1, JS0 = 100		4		
		Prescaler bits JS2, JS1, JS0 = 111			8		]
Integral Nonlinearity of Frequency	INL	Entire range (Note 7)		-0.6		+0.3	%
DAC Step Size		Δ between two consecutive DAC values (Note 8)			5		kHz
DAC Span		Frequency range for c (Table 2)	one offset setting		5.12		MHz
DAC Default		Factory default registe	er setting		500		decimal
Offset Step Size		$\Delta$ between two consecutive offset values (Table 2)			2.56		MHz
Offset Default	OS	Factory default OFFSET register setting (5 LSBs) (Table 2)		RANGE (5 LSBs of RANGE register)		of	hex
		Prescaler bits JS4, JS	3 = 00		f <sub>0</sub> /8192	2	
Dither Rate		Prescaler bits JS4, JS	3 = 01		f <sub>0</sub> /4096	6	Hz
		Prescaler bits JS4, JS3 = 10			f <sub>0</sub> /2048	3	1

## **AC ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = 2.7V \text{ to } 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Frequency Stable After Prescaler Change					1	period
Frequency Stable After DAC or Offset Change	<sup>t</sup> DACstab	(Note 9)		0.1	1	ms
Power-Up Time	t <sub>por</sub> + t <sub>stab</sub>	(Note 10)		0.1	0.5	ms
Enable of OUT After Exiting Power-Down Mode	t <sub>stab</sub>	(Note 18)			200	μs
OUT High-Z After Entering Power-Down Mode	t <sub>pdn</sub>				100	μs
Load Capacitance	CL	(Note 11)		15	50	рF
Output Duty Cycle (OUT)		Default frequency	45		55	%
Rise and Fall Time (OE, PDN)					1	μs

### AC ELECTRICAL CHARACTERISTICS—2-WIRE INTERFACE

 $(V_{CC} = 2.7V \text{ to } 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONE	DITIONS	MIN	ТҮР	MAX	UNITS
SCL Clock Frequency	fscl	Fast mode	(Note 12)			400	kHz
	ISCL	Standard mode				100	KI IZ
Bus Free Time Between a STOP	tour	Fast mode	(Note 12)	1.3			
and START Condition	tBUF	Standard mode		4.7			μs
Hold Time (Repeated) START		Fast mode	(Natao 10, 10)	0.6			
Condition	<sup>t</sup> HD:STA	Standard mode	(Notes 12, 13)	4.0			μs
LOW Period of SCL	ti ovu	Fast mode		1.3			
	tLOW	Standard mode	(Note 12)	4.7			μs
HIGH Pariad of SCI	IGH Period of SCI	Fast mode	(Note 12)	0.6			μs
HIGH FEIIOU OF SCL		Standard mode	(10016-12)	4.0			μο
Setup Time for a Repeated	tourort	Fast mode	(Note 12)	0.6			μs
START	<sup>t</sup> SU:STA	Standard mode	(NOLE 12)	4.7			
	1	Fast mode		0		0.0	
Data Hold Time	<sup>t</sup> HD:DAT	Standard mode	(Notes 12, 14, 15)	0		0.9	μs
Data Catura Tirra		Fast mode	(Nata 10)	100			
Data Setup Time	tsu:dat	Standard mode	(Note 12)	250			ns
Rise Time of Both SDA and SCL		Fast mode	(Note 16)	20 + 0.10	В	300	ns
Signals	tR	Standard mode		20 + 0.10	В	1000	
Fall Time of Both SDA and SCL	+	Fast mode		20 + 0.10	В	300	ns
Signals	t⊨	Standard mode	(Note 16)	20 + 0.10	В	1000	

### AC ELECTRICAL CHARACTERISTICS—2-WIRE INTERFACE (continued)

 $(V_{CC} = 2.7V \text{ to } 3.6V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
Setup Time for STOP	touro	Fast mode	0.6			
	tsu:sto	Standard mode	4.0			μs
Capacitive Load for Each Bus Line	CB	(Note 16)			400	pF
EEPROM Write Cycle Time	twR				10	ms
Input Capacitance	CI			5		рF

## NONVOLATILE MEMORY CHARACTERISTICS

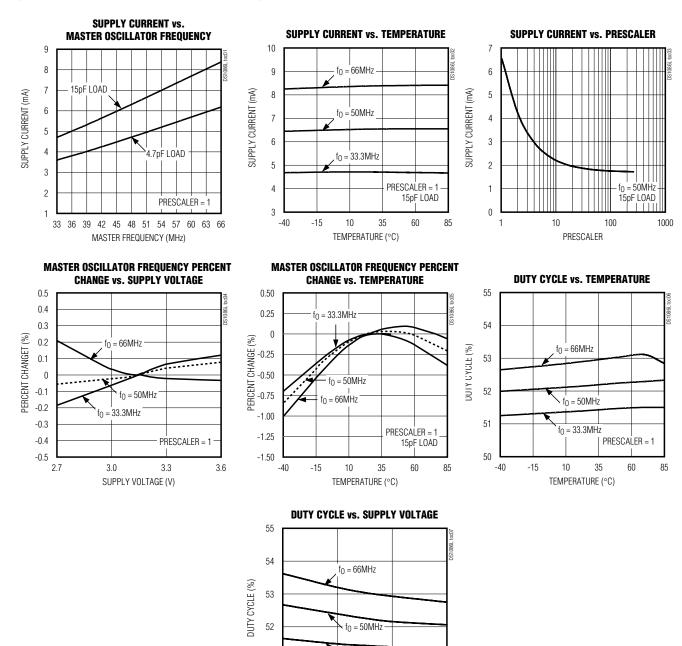
 $(V_{CC} = 2.7V \text{ to } 3.6V)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
EEPROM Writes		+70°C	10,000			

Note 1:	All voltages are referenced to ground.
Note 2:	DAC and OFFSET register settings must be configured to maintain the master oscillator frequency within this range.
	Correct operation of the device is not guaranteed if these limits are exceeded.
Note 3:	This is the absolute accuracy of the master oscillator frequency at the default settings.
Note 4:	This is the change that is observed in master oscillator frequency with changes in voltage from nominal voltage at
	$T_{A} = +25^{\circ}C.$
Note 5:	This is the percentage frequency change from the $+25^{\circ}$ C frequency due to temperature at V <sub>CC</sub> = 3.3V. The maximum temper-
	ature change varies with the master oscillator frequency setting. The minimum occurs at the default master oscillator frequen-
	cy (f <sub>default</sub> ). The maximum occurs at the extremes of the master oscillator frequency range (33.3MHz or 66.6MHz).
Note 6:	The dither deviation of the master oscillator frequency is unidirectional and lower than the undithered frequency.
Note 7:	The integral nonlinearity of the frequency is a measure of the deviation from a straight line drawn between the two end-
	points ( $f_{OSC}(MIN)$ to $f_{OSC}(MAX)$ ) of the range. The error is in percentage of the span.
Note 8:	This is true when the prescaler = 1.
Note 9:	Frequency settles faster for small changes in value. During a change, the frequency transitions smoothly from the original
	value to the new value.
Note 10:	This indicates the time elapsed between power-up and the output becoming active. An on-chip delay is intentionally
	introduced to allow the oscillator to stabilize. tstab is equivalent to approximately 512 master clock cycles and therefore
	depends on the programmed clock frequency.
Note 11:	
Note 12:	A fast-mode device can be used in a standard-mode system, but the requirement $t_{SU:DAT} > 250$ ns must then be met.
	This is automatically the case if the device does not stretch the LOW period of the SCL signal. If such a device does
	stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line at least $t_{RMAX} + t_{SU:DAT} =$
N	1000ns + 250ns = 1250ns before the SCL line is released.
Note 13:	
Note 14:	
Note 15:	signal) to bridge the undefined region of the falling edge of SCL.
Note 15: Note 16:	The maximum $t_{HD:DAT}$ need only be met if the device does not stretch the LOW period ( $t_{LOW}$ ) of the SCL signal. C <sub>B</sub> —total capacitance of one bus line, timing referenced to 0.9 x V <sub>CC</sub> and 0.1 x V <sub>CC</sub> .
Note 16: Note 17:	
Note 17:	Typical frequency shift due to aging is ±0.5%. Aging stressing includes Level 1 moisture reflow preconditioning (24hr +125°C bake, 168hr 85°C/85%RH moisture soak, and three solder reflow passes +240 +0/-5°C peak) followed by 1000hr
	max V <sub>CC</sub> biased 125°C HTOL, 1000 temperature cycles at -55°C to +125°C, 96hr 130°C/85%RH/3.6V HAST and 168hr
	121°C/2 ATM Steam/Unbiased Autoclave.
Note 18:	t <sub>stab</sub> is the time required after exiting power-down to the beginning of output oscillations. In addition, a delay of tDACstab
11010 10.	
	is required before the frequency will be within its specified tolerance.

## **Typical Operating Characteristics**

 $(V_{CC} = 3.3V, T_A = 25^{\circ}C, unless otherwise noted.)$ 



≏ f<sub>0</sub> = 33.3MHz

SUPPLY VOLTAGE (V)

3.0

PRESCALER = 1

3.6

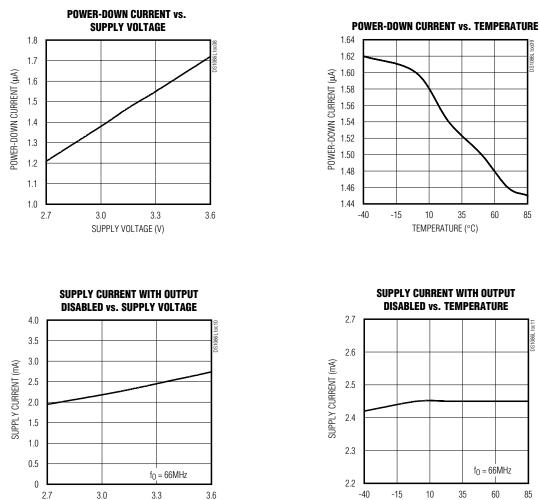
3.3

51

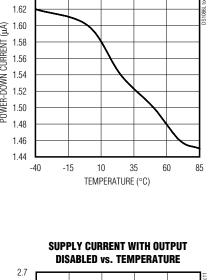
2.7

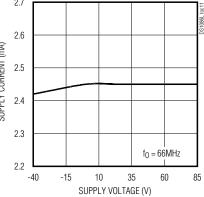
## **Typical Operating Characteristics (continued)**

(V<sub>CC</sub> = 3.3V, T<sub>A</sub> =  $25^{\circ}$ C, unless otherwise noted.)



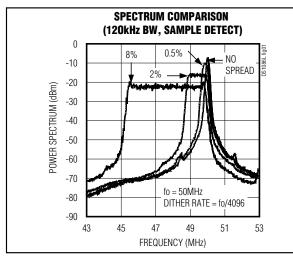
SUPPLY VOLTAGE (V)

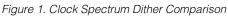


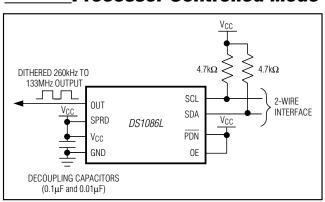


## Pin Description

PIN	NAME	FUNCTION
1	OUT	Oscillator Output. The output frequency is determined by the OFFSET, DAC, and prescaler registers.
2	SPRD	Dither Enable. When the pin is high, the dither is enabled. When the pin is low, the dither is disabled.
3	V <sub>CC</sub>	Power Supply
4	GND	Ground
5	OE	Output Enable. When the pin is high, the output buffer is enabled. When the pin is low, the output is disabled but the master oscillator is still on.
6	PDN	Power-Down. When the pin is high, the master oscillator is enabled. When the pin is low, the master oscillator is disabled (power-down mode).
7	SDA	2-Wire Serial Data. This pin is for serial data transfer to and from the device. The pin is open drain and can be wire-ORed with other open-drain or open-collector interfaces.
8	SCL	2-Wire Serial Clock. This pin is used to clock data into the device on rising edges and clock data out on falling edges.









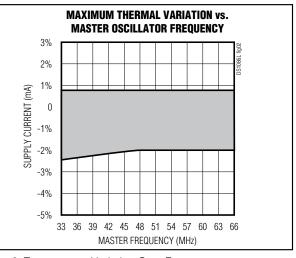
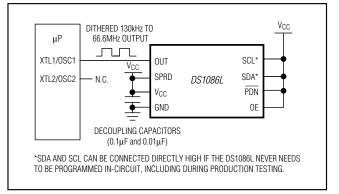


Figure 2. Temperature Variation Over Frequency

## Stand-Alone Mode



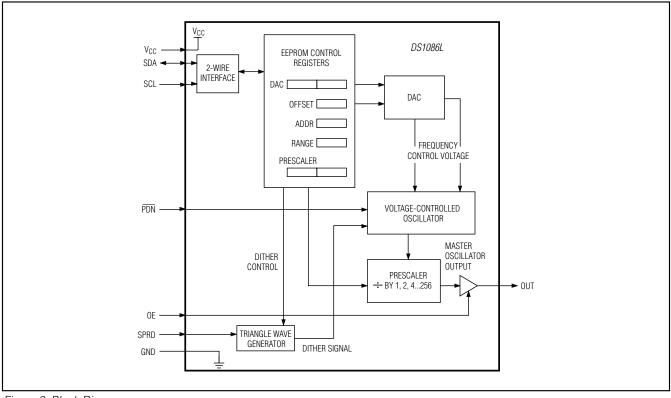


Figure 3. Block Diagram

#### **Detailed Description**

A block diagram of the DS1086L is shown in Figure 3. The internal master oscillator generates a square wave with a 33.3MHz to 66.6MHz frequency range. The frequency of the master oscillator can be programmed with the DAC register over a two-to-one range in 5kHz steps. The master oscillator range is larger than the range possible with the DAC step size, so the OFFSET register is used to select a smaller range of frequencies over which the DAC spans. The prescaler can then be set to divide the master oscillator frequency by 2<sup>x</sup> (where x equals 0 to 8) before routing the signal to the output (OUT) pin.

A programmable triangle-wave generator injects an offset element into the master oscillator to dither its output 0.5%, 1%, 2%, 4%, or 8%. The dither magnitude is controlled by the JS2, JS1, and JS0 bits in the PRESCALER word and enabled with the SPRD pin. Futhermore, the dither rate is controlled by the JS4 and JS3 bits in the PRESCALER word and determines the frequency of the dither. The maximum spectral attenuation occurs when the prescaler is set to 1 and is reduced by 2.7dB for every factor of 2 that is used in the prescaler. This happens because the prescaler's divider function tends to average the dither in creating the lower frequency. However, the most stringent spectral emission limits are imposed on the higher frequencies where the prescaler is set to a low divider ratio.

The external control input, OE, gates the clock output buffer. The PDN pin disables the master oscillator and turns off the clock output for power-sensitive applications\*. On power-up, the clock output is disabled until power is stable and the master oscillator has generated 512 clock cycles. Both controls feature a synchronous enable that ensures there are no output glitches when the output is enabled.

The control registers are programmed through a 2-wire interface and are used to determine the output frequency and settings. Once programmed into EEPROM, since the register settings are NV, the settings only need to be reprogrammed if it is desired to reconfigure the device.

\*The power-down command must persist for at least two output frequency cycles plus 10µs for deglitching purposes.

REGISTER	ADDR	MSB			BIN	ARY			LSB	FACTORY DEFAULT	ACCESS
PRESCALER	02h	JS4	JS3	JS2	JS1	JS0	LO/HiZ	P3	P2	01100000	R/W
PRESCALER	_	P1	P0	XX	XX	XX	Хх	Хх	XX	00XXXXX	R/W
DAC (MSB)	08h	b9	b8	b7	b6	b5	b4	b3	b2	01111101b	R/W
DAC (LSB)	_	b1	b0	X <sub>0</sub>	0000000b	R/W					
OFFSET	0Eh	X1	X1	X <sub>1</sub>	b4	b3	b2	b1	b0	111b	R/W
ADDR	0Dh	X1	X <sub>1</sub>	X1	X <sub>1</sub>	WC	A2	A1	A0	11110000b	R/W
RANGE	37h	Хх	Хх	Хх	b4	b3	b2	b1	b0	xxxb	R
WRITE EE	3Fh				NO [	DATA				—	_

#### Table 1. Register Summary

 $X_0 = Don't care, reads as zero.$ 

 $X_1 = Don't care, reads as one.$ 

Xx = Don't care, reads indeterminate.

X = Don't care.

### **Table 2. Offset Settings**

OFFSET	FREQUENCY RANGE (MHz)
OS - 6	30.74 to 35.86
OS - 5	33.30 to 38.42
OS - 4	35.86 to 40.98
OS - 3	38.42 to 43.54
OS - 2	40.98 to 46.10
OS - 1	43.54 to 48.66
OS*	46.10 to 51.22
OS + 1	48.66 to 53.78
OS + 2	51.22 to 56.34
OS + 3	53.78 to 58.90
OS + 4	56.34 to 61.46
OS + 5	58.90 to 64.02
OS + 6	61.46 to 66.58

\*Factory default setting. OS is the integer value of the five LSBs of the RANGE register.

The output frequency is determined by the following equation:

f	(MIN FREQUENCY OF SELECTED OFFSET RANGE)
OUTPUT =	+ (DAC VALUE $\times$ 5kHz STEP SIZE)
0011 01 -	PRESCALER

where: *min frequency of selected OFFSET range* is the lowest frequency (shown in Table 2 for the corresponding offset).

DAC value is the value of the DAC register (0 to 1023).

*Prescaler* is the value of  $2^x$  where x = 0 to 8.

See the *Example Frequency Calculations* section for a more in-depth look at using the registers.

### **Register Definitions**

The DS1086L registers are used to program the output frequency, dither percent, dither rate, and 2-wire address. Table 1 shows a summary of the registers and detailed descriptions follow below.

#### **PRESCALER** (02h)

The PRESCALER word is a two-byte value containing control bits for the prescaler (P3 to P0), output control (Lo/HiZ), the jitter rate (JS4 to JS3), as well as control bits for the jitter percentage (JS2 to JS0). The PRESCALER word is read and written using two-byte reads and writes beginning at address 02h.

**JS4 to JS3: Jitter Rate**. This is the frequency of the triangle wave generator and the modulation frequency that the output is dithered. It can be programmed to the master oscillator frequency, fOSC, divided by either 8192, 4096, or 2048.

JS4	JS3	JITTER RATE
0	0	f <sub>OSC</sub> /8192
0	1	f <sub>OSC</sub> /4096 (default)
1	0	f <sub>OSC</sub> /2048

## **DS1086L**

## 3.3V Spread-Spectrum EconOscillator

**JS2 to JS0: Jitter Percentage**. These three bits select the amount of jitter in percent. The SPRD pin must be a logic high for the jitter to be enabled. Bit combinations not shown are reserved.

JS2	JS1	JS0	JITTER %
0	0	0	0.5
0	0	1	1
0	1	0	2
1	0	0	4
1	1	1	8

**Lo/HiZ:** Output Low or High-Z. This bit determines the state of the output pin when the device is in power-down mode or when the output is disabled. If Lo/HiZ = 0, the output is HiZ when in power-down or disabled. If Lo/HiZ = 1, the output is held low when in power-down or disabled.

**P3 to P0: Prescaler Divider.** These bits divide the master oscillator frequency by  $2^x$ , where x is P3 to P0 and can be from 0 to 8. Any prescaler value entered greater than 8 decodes as 8.

**B9 to B0: DAC Setting.** The DAC word sets the master oscillator frequency to a specific value within the current offset range. Each step of the DAC changes the master oscillator frequency by 5kHz. The DAC word is read and written using two-byte reads and writes beginning at address 08h.

**OFFSET (OEh) B4 to B0: Offset.** This value selects the master oscillator frequency range that can be generated by varying the DAC word. Valid frequency ranges are shown in Table 2. Correct operation of the device is not guaranteed for values of OFFSET not shown in the table.

The default offset value (OS) is factory trimmed and can vary from device to device. Therefore, to change frequency range, OS must be read so the new offset value can be calculated relative to the default. For example, to generate a master oscillator frequency within the largest range (61.4MHz to 66.6MHz), Table 2 indicates that the OFFSET must be programmed to OS + 6. This is done by reading the RANGE register and adding 6 to the value of bits B4 to B0. The result is then written into bits B4 to B0 of the OFFSET register. Additional examples are provided in the *Example Frequency Calculations* section.

#### DAC (08h)

### RANGE (37h)

**B4 to B0: Range:** This read-only, factory programmed value is a copy of the factory default offset (OS). OS is required to program new master oscillator frequencies shown in Table 2. The read-only backup is important because the offset register is EEPROM and is likely to be overwritten.

### ADDR (0Dh)

WC: EEPROM Write Control Bit. The WC bit enables/disables the automatic writing of registers to EEPROM. This prevents EEPROM wear out and eliminates the EEPROM write cycle time. If WC = 0 (default), register writes are automatically written to EEPROM. If WC = 1, register writes are stored in SRAM and only written into EEPROM when the user sends a WRITE EE command. If power is cycled to the device, then the last value stored in EEPROM is recalled. WC = 1 is ideal for applications that frequently modify the frequency/registers.

Regardless of the value of the WC bit, the value of the ADDR register is always written immediately to EEPROM.

**A2 to A0: Device Address Bits.** These bits determine the 2-wire slave address of the device. They allow up to eight devices to be attached to the same 2-wire bus and to be addressed individually.

#### WRITE EE Command (3Fh)

This command can be used when WC = 1 (see the WC bit in ADDR register) to transfer all registers from SRAM into EEPROM. The time required to store the values is one EEPROM write cycle time. This command is not needed if WC = 0.

#### **Example Frequency Calculations**

**Example #1:** Calculate the register values needed to generate a desired output frequency of 11.0592MHz. Since the desired frequency is not within the valid master oscillator range of 33.3MHz to 66.6MHz, the prescaler must be used. Valid prescaler values are  $2^{x}$  where x equals 0 to 8 (and x is the value that is programmed into the P3 to P0 bits of the PRESCALER register). Equation 1 shows the relationship between the desired frequency, the master oscillator frequency, and the prescaler.

 $f_{\text{DESIRED}} = \frac{f_{\text{MASTER OSCILLATOR}}}{prescaler} = (1)$   $\frac{f_{\text{MASTER OSCILLATOR}}}{2^{X}}$ 

By trial and error, x is incremented from 0 to 8 in Equation 2, finding values of x that yield master oscillator frequencies within the range of 33.3MHz to 66.6MHz.

Equation 2 shows that a prescaler of 4 (x = 2) and a master oscillator frequency of 44.2368MHz generates our desired frequency. Writing 0080h to the PRESCALER register sets the PRESCALER to 4. Be aware that other settings also reside in the PRESCALER register.

 $fMASTER OSCILLATOR = fDESIRED \times prescaler = fDESIRED \times 2^{X}$ 

 $f_{MASTER OSCILLATOR} = 11.0592MHz \times 22 = 44.2368MHz$  (2)

Once the target master oscillator frequency has been calculated, the value of offset can be determined. Using Table 2, 44.2368MHz falls within both OS - 1 and OS - 2. However, choosing OS - 1 would be a poor choice since 44.2368MHz is so close to OS - 1's minimum frequency. On the other hand, OS - 2 is ideal since 44.2368MHz is close to the center of OS - 2's frequency span. Before the OFFSET register can be programmed, the default value of offset (OS) must be read from the RANGE register (last five bits). In this example, 12h (18 decimal) was read from the RANGE register. OS - 2 for this case is 10h (16 decimal). This is the value that is written to the OFFSET register.

Finally, the two-byte DAC value needs to be determined. Since OS - 2 only sets the range of frequencies, the DAC selects one frequency within that range as shown in Equation 3.

Valid values of DAC are 0 to 1023 (decimal) and 5kHz is the step size. Equation 4 is derived from rearranging Equation 3 and solving for the DAC value.

$$\frac{(f_{MASTER OSCILLATOR} - (4)}{MIN FREQUENCY OF SELECTED}$$
DAC VALUE = 
$$\frac{OFFSET RANGE}{5kHz STEP SIZE}$$

 $DAC VALUE = \frac{(44.2368MHz - 41.0MHz)}{5kHz STEP SIZE}$  $= 647.36 \approx 647 \text{ (decimal)}$ 

Since the two-byte DAC register is left justified, 647 is converted to hex (0287h) and bit-wise shifted left six places. The value to be programmed into the DAC register is A1C0h.

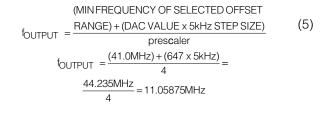
In summary, the DS1086L is programmed as follows:

PRESCALER = 0080h

OFFSET = OS - 2 or 10h (if range was read as 12h)

DAC = A1C0h

Notice that the DAC value was rounded. Unfortunately, this means that some error is introduced. To calculate how much error, a combination of Equation 1 and Equation 3 is used to calculate the expected output frequency. See Equation 5.



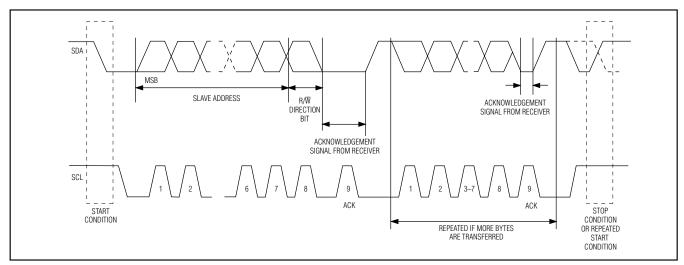


Figure 4. 2-Wire Data Transfer Protocol

The expected output frequency is not exactly equal to the desired frequency of 11.0592MHz. The difference is 450Hz. In terms of percentage, Equation 6 shows that the expected error is 0.004%. The expected error assumes typical values and does not include deviations from the typical as specified in the electrical tables.

$$\text{\%ERROR}_{\text{EXPECTED}} = \frac{\text{fbesired} - \text{fexpected}}{\text{fbesired}} \times 100$$
(6)
$$\text{\%ERROR}_{\text{EXPECTED}} = \frac{11.0592\text{MHz} - 11.05875\text{MHz}}{11.0592\text{MHz}} \times 100$$

$$= \frac{450\text{Hz}}{11.0592\text{MHz}} \times 100 = 0.004\%$$

## Example #2: Calculate the register values needed to generate a desired output frequency of 50MHz.

Since the desired frequency is already within the valid master oscillator frequency range, the prescaler is set to divide by 1, and hence, PRESCALER = 0000h (currently ignoring the other setting). (7)

$$f_{MASTER}$$
 OSCILLATOR = 50.0MHz x 2<sup>o</sup> = 50.0MHz

Next, looking at Table 2, OS + 1 provides a range of frequencies centered around the desired frequency. To determine what value to write to the OFFSET register, the RANGE register must first be read. Assuming 12h was read in this example, 13h (OS + 1) is written to the OFFSET register.

Finally, the DAC value is calculated as shown in Equation 8. (8)

$$DAC VALUE = \frac{(50.0MHz - 48.6MHz)}{5kHz STEP SIZE} = 280.00 \text{ (decimal)}$$

The result is then converted to hex (0118h) and then left-shifted, resulting in 4600h to be programmed into the DAC register.

In summary, the DS1086L is programmed as follows:

PRESCALER = 0000h

OFFSET = OS + 1 or 13h (if RANGE was read as 12h) DAC = 4600h

$$f_{\text{OUTPUT}} = \frac{(48.6\text{MHz}) + (280 \times 5\text{kHz})}{2^0} = \frac{50.0\text{MHz}}{1} = 50.0\text{MHz}$$
(9)

Since the expected output frequency is equal to the desired frequency, the calculated error is 0%.

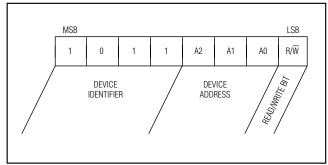


Figure 5. Slave Address

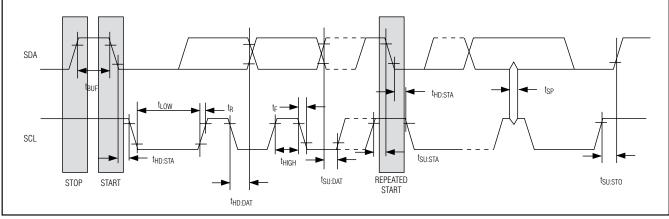


Figure 6. 2-Wire AC Characteristics

## 2-Wire Serial Port Operation

#### **2-Wire Serial Data Bus**

The DS1086L communicates through a 2-wire serial interface. A device that sends data onto the bus is defined as a transmitter, and a device receiving data as a receiver. The device that controls the message is called a "master." The devices that are controlled by the master are "slaves." A master device that generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions must control the bus. The DS1086L operates as a slave on the 2-wire bus. Connections to the bus are made through the open-drain I/O lines SDA and SCL.

The following bus protocol has been defined (see Figures 4 and 6):

- Data transfer can be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is HIGH. Changes

in the data line while the clock line is HIGH are interpreted as control signals.

Accordingly, the following bus conditions have been defined:

Bus not busy: Both data and clock lines remain HIGH.

**Start data transfer:** A change in the state of the data line, from HIGH to LOW, while the clock is HIGH, defines a START condition.

**Stop data transfer:** A change in the state of the data line, from LOW to HIGH, while the clock line is HIGH, defines the STOP condition.

**Data valid:** The state of the data line represents valid data when, after a START condition, the data line is stable for the duration of the HIGH period of the clock signal. The data on the line must be changed during the LOW period of the clock signal. There is one clock pulse per bit of data.

Each data transfer is initiated with a START condition and terminated with a STOP condition. The number of

data bytes transferred between START and STOP conditions is not limited, and is determined by the master device. The information is transferred byte-wise and each receiver acknowledges with a ninth bit.

Within the bus specifications a standard mode (100kHz clock rate) and a fast mode (400kHz clock rate) are defined. The DS1086L works in both modes.

**Acknowledge:** Each receiving device, when addressed, is obliged to generate an acknowledge after the byte has been received. The master device must generate an extra clock pulse that is associated with this acknowledge bit.

A device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge-related clock pulse. Of course, setup and hold times must be taken into account. When the DS1086L EEPROM is being written to, it is not able to perform additional responses. In this case, the slave DS1086L sends a not acknowledge to any data transfer request made by the master. It resumes normal operation when the EEPROM operation is complete.

A master must signal an end of data to the slave by not generating an acknowledge bit on the last byte that has been clocked out of the slave. In this case, the slave must leave the data line HIGH to enable the master to generate the STOP condition.

Figures 4, 5, 6, and 7 detail how data transfer is accomplished on the 2-wire bus. Depending upon the state of the R/W bit, two types of data transfer are possible:

- Data transfer from a master transmitter to a slave receiver. The first byte transmitted by the master is the slave address. Next follows a number of data bytes. The slave returns an acknowledge bit after each received byte.
- 2) Data transfer from a slave transmitter to a master receiver. The first byte (the slave address) is transmitted by the master. The slave then returns an acknowledge bit. Next follows a number of data bytes transmitted by the slave to the master. The master returns an acknowledge bit after all received bytes other than the last byte. At the end of the last received byte, a not acknowledge is returned.

The master device generates all the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the bus is not released.

The DS1086L can operate in the following two modes:

**Slave receiver mode:** Serial data and clock are received through SDA and SCL. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the slave address and direction bit.

**Slave transmitter mode:** The first byte is received and handled as in the slave receiver mode. However, in this mode, the direction bit indicates that the transfer direction is reversed. Serial data is transmitted on SDA by the DS1086L while the serial clock is input on SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer.

#### **Slave Address**

Figure 5 shows the first byte sent to the device. It includes the device identifier, device address, and the R/W bit. The device address is determined by the ADDR register.

#### **Registers/Commands**

See Table 1 for the complete list of registers/commands and Figure 7 for an example of using them.

#### **Applications Information**

#### **Power-Supply Decoupling**

To achieve the best results when using the DS1086L, decouple the power supply with  $0.01\mu$ F and  $0.1\mu$ F high-quality, ceramic, surface-mount capacitors. Surface-mount components minimize lead inductance, which improves performance, and ceramic capacitors tend to have adequate high-frequency response for decoupling applications. These capacitors should be placed as close to pins 3 and 4 as possible.

#### Stand-Alone Mode

SCL and SDA cannot be left unconnected when they are not used. If the DS1086L never needs to be programmed in-circuit, including during production testing, SDA and SCL can be tied high. The SPRD pin must be tied either high or low.

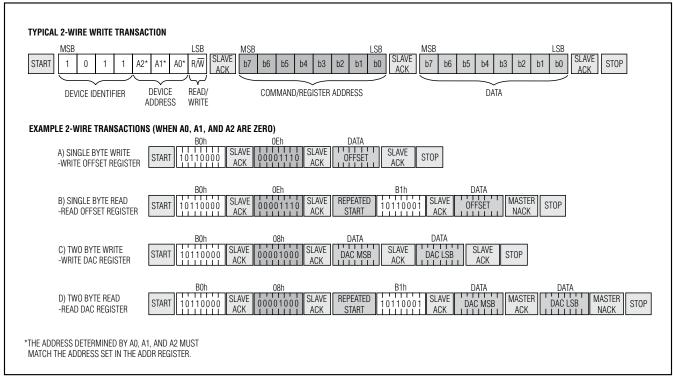


Figure 7. 2-Wire Transactions

## Chip Information

SUBSTRATE CONNECTED TO GROUND

### **Package Information**

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
8 µSOP	U8+1	<u>21-0036</u>	<u>90-0092</u>

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	11/03	Initial release	
1	9/07	Updated Table 2	10
2	3/12	Updated the Ordering Information, Absolute Maximum Ratings, and Package Information	1, 2, 16



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

#### Maxim Integrated 160 Rio Robles, San Jose, CA 95134 USA 1-408-601-1000

© 2012 Maxim Integrated

The Maxim logo and Maxim Integrated are trademarks of Maxim Integrated Products, Inc.



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



#### Как с нами связаться

**Телефон:** 8 (812) 309 58 32 (многоканальный) **Факс:** 8 (812) 320-02-42 **Электронная почта:** <u>org@eplast1.ru</u> **Адрес:** 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.