



Blackfin Embedded Processor with Codec

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

PROCESSOR FEATURES

- Up to 600 MHz high performance Blackfin processor
 - RISC-like register and instruction model for ease of programming and compiler-friendly support
 - Advanced debug, trace, and performance monitoring
- Accepts a wide range of supply voltages for internal and I/O operations. See operating conditions in the published ADSP-BF52x processor data sheet.
- Programmable on-chip voltage regulator (ADSP-BF523/ADSP-BF525/ADSP-BF527 processors only)
- Embedded low power audio codec
- 289-ball (12 mm x 12 mm) CSP_BGA package
- 132K bytes of on-chip memory
- External memory controller with glueless support for SDRAM and asynchronous 8-bit and 16-bit memories
- Flexible booting options from external flash, SPI and TWI memory or from host devices including SPI, TWI, and UART
- Code security with Lockbox Secure Technology one-time-programmable (OTP) memory
- Memory management unit providing memory protection
- 2 dual-channel memory DMA controllers

EMBEDDED CODEC FEATURES

- Stereo, 24-bit ADCs and DACs
 - DAC SNR: 100 dB (A-weighted), THD: -80 dB at 48 kHz, 3.3 V
 - ADC SNR: 90 dB (A-weighted), THD: -80 dB at 48 kHz, 3.3 V
- Highly efficient headphone amplifier
- Stereo line input and monaural microphone input
- Low power
 - 7 mW stereo playback (1.8 V supply)
 - 14 mW record and playback (1.8 V supply)
- Low supply voltages
 - Analog: 1.8 V to 3.6 V
 - Digital core: 1.8 V min
 - Digital I/O: 1.8 V to 3.6 V
- 256 × f_s / 384 × f_s oversampling rate in normal mode;
250 × f_s / 272 × f_s oversampling rate in USB mode
- Audio sampling rates: 8 kHz, 11.025 kHz, 12 kHz, 16 kHz, 22.05 kHz, 24 kHz, 32 kHz, 44.1 kHz, 48 kHz, 88.2 kHz, and 96 kHz

PERIPHERALS

See the published ADSP-BF52x processor data sheet for additional peripherals



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Rev. A

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REVISION HISTORY

3/10—Rev. 0 to Rev. A

Revised the following figures.

Recommended Application Circuit Using SPI Control	13
Recommended Application Circuit Using TWI Control ..	14
Added Sampling Rate = 48 kHz to all figures in	
Converter Filter Response	30
Revised Ordering Guide	36

GENERAL DESCRIPTION

This document describes the differences between the ADSP-BF52xC and the ADSP-BF52x standard Blackfin® product. Please refer to the published ADSP-BF52x data sheet for general description and specifications. This document only describes the differences from that data sheet.

The ADSP-BF52xC processors add a low power, high quality stereo audio codec for portable digital audio applications with one set of stereo programmable gain amplifier (PGA) line inputs and one monaural microphone input. It features two 24-bit analog-to-digital converter (ADC) channels and two 24-bit digital-to-analog (DAC) converter channels.

The codec can operate as a master or a slave. It supports various master clock frequencies, including 12 MHz or 24 MHz for USB devices; standard $256 \times f_s$ or $384 \times f_s$ based rates, such as 12.288 MHz and 24.576 MHz; and many common audio sampling rates, such as 96 kHz, 88.2 kHz, 48 kHz, 44.1 kHz, 32 kHz, 24 kHz, 22.05 kHz, 16 kHz, 12 kHz, 11.025 kHz, and 8 kHz.

The codec can operate at power supplies as low as 1.8 V for the analog circuitry and as low as 1.8 V for the digital circuitry. The maximum voltage supply is 3.6 V for all supplies.

The codec software-programmable stereo output options provide the programmer with many application possibilities because the device can be used as a headphone driver or as a speaker driver. Its volume control functions provide a large range of gain control of the audio signal.

CODEC DESCRIPTION

The ADSP-BF52xC codec contains a central clock source, called the codec master clock (CODEC_MCLK) that produces a reference clock for all internal audio data processing and synchronization. When using an external clock source to drive the CODEC_MCLK pin, care should be taken to select a clock source with less than 50 ps of jitter. Without careful generation of the CODEC_MCLK signal, the digital audio quality will suffer.

To enable the codec to generate the central reference clock in a system, connect a crystal oscillator between the XT1/ CODEC_MCLK input pin and the XTO output pin.

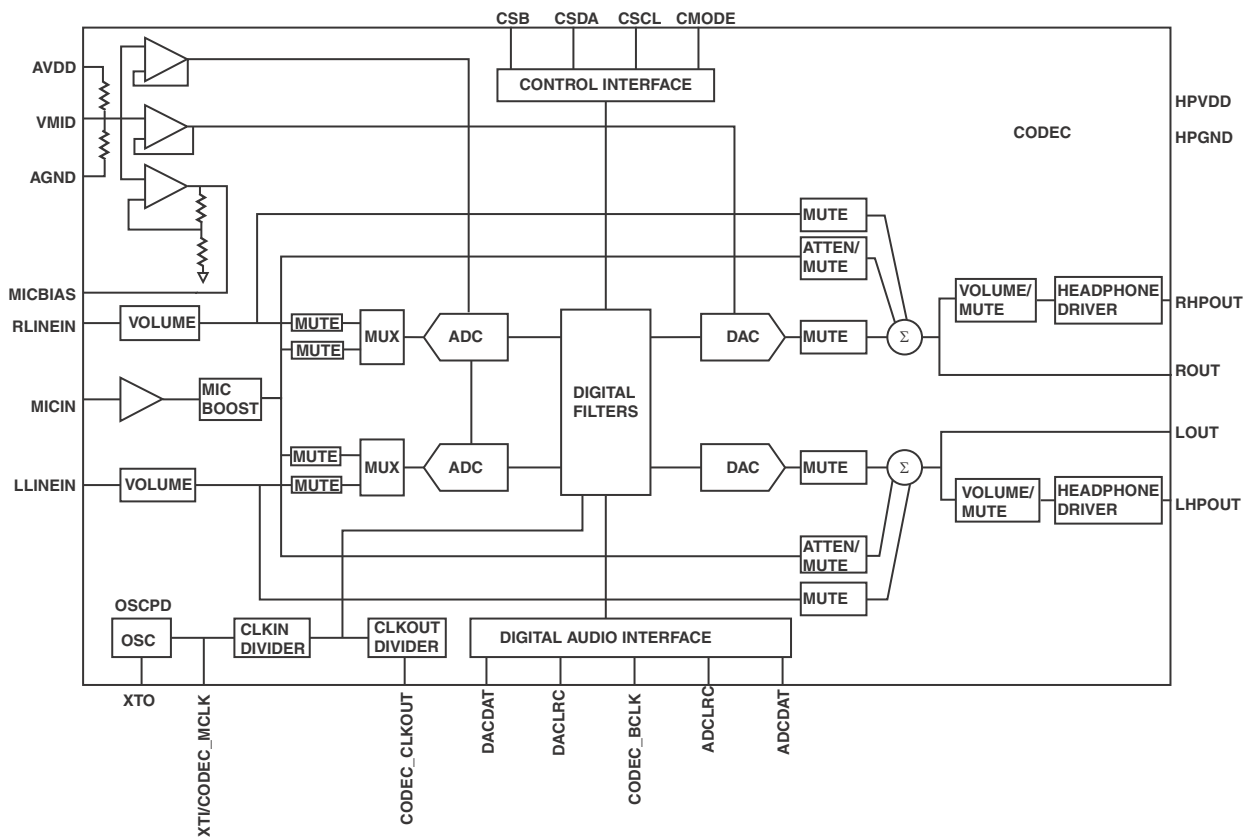


Figure 1. Codec Block Diagram

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To allow an external device to generate the central reference clock, apply the external clock signal directly through the XTI/ CODEC_MCLK input pin. In this configuration, the oscillator circuit of the codec can be powered down by using the OSCPDP bit (Register R6, Bit D5) to reduce power consumption.

To accommodate applications with very high frequency master clocks, the internal core reference clock of the codec can be set to either CODEC_MCLK or CODEC_MCLK divided by 2. This is enabled by adjusting the setting of the CLKDIV2 bit (Register R8, Bit D6). The CODEC_CLKOUT pin can also drive external clock sources with either the codec clock signal or codec clock divided by 2 by enabling the CLKODIV2 bit (Register R8, Bit D7).

ADC AND DAC

The codec contains a pair of oversampling Σ - Δ ADCs. The maximum ADC full-scale input level is $1.0 V_{rms}$ when $AVDD = 3.3 V$. If the input signal to the ADC exceeds this level, data overloading occurs and causes audible distortion.

The ADC can accept analog audio input from either the stereo line inputs or the monaural microphone input. Note that the ADC can only accept input from a single source, so the programmer must choose either the line inputs or the microphone input using the INSEL bit (Register R4, Bit D2). The digital data from the ADC output, once converted, is processed using the ADC filters.

Complementary to the Σ - Δ ADC channels, the codec contains a pair of oversampling DACs that convert the digital audio data from the internal DAC filters into an analog audio signal. The DAC output can also be muted by setting the DACMU bit (Register R5, Bit D3) in the control register.

ADC HIGH-PASS AND DAC DE-EMPHASIS FILTERS

The ADC and DAC employ separate digital filters that perform 24-bit signal processing. The digital filters are used for both record and playback modes and are optimized for each individual sampling rate used.

For recording mode operations, the unprocessed data from the ADC enters the ADC filters and is converted to the appropriate sampling frequency, then is output to the digital audio interface.

For playback mode operations, the DAC filters convert the digital audio interface data to oversampled data using a sampling rate selected by the programmer. The oversampled data is processed by the DAC and sent to the analog output mixer by enabling the DACSEL (Register R4, Bit D4).

Programmers have the option of setting up the device so that any dc offset in the input source signal is automatically detected and removed. To accomplish this, enable the digital high-pass filter (see Table 22 on Page 30 for characteristics) contained in the ADC digital filters by using the ADCHPD bit (Register R5, Bit D0).

In addition, programmers can implement digital de-emphasis by using the DEEMPH bits (Register R5, Bit D1 and Bit D2).

ANALOG AUDIO INTERFACES

The codec includes stereo single-ended line inputs and a monaural microphone input to the on-board ADC. Either the line inputs or the microphone input, but not both simultaneously, can be connected to the ADC by setting the INSEL bit (Register R4, Bit D2).

The codec also includes line and headphone outputs from the on-board DAC. The line or microphone inputs can be routed and mixed directly to the output terminals.

Stereo Line and Monaural Microphone Inputs

The single-ended stereo line inputs (RLINEIN and LLINEIN) are internally biased to VMID by way of a voltage divider between AVDD and AGND (see Figure 2). The line input signal can be connected to the internal ADC and, if desired, routed directly to the outputs via the bypass path by using the BYPASS bit (Register R4, Bit D3).



Figure 2. Line Input to ADC

The line input volume can be adjusted from -34.5 dB to $+33 \text{ dB}$ in steps of $+1.5 \text{ dB}$ by setting the LINVOL (Register R0, Bit D0 to Bit D5) and RINVOL (Register R1, Bit D0 to Bit D5) bits. By default the volume is independently adjustable for both right and left line inputs. However, if the LRINBOTH or RLINBOTH bit is programmed, both LINVOL and RINVOL are loaded with the same value. The programmer can also set the LINMUTE (Register R0, Bit D7) and RINMUTE (Register R1, Bit D7) bits to mute the line input signal to the ADC.

The high impedance monaural microphone input pin (MICIN, shown in Figure 3) has two gain stages and a microphone bias level (MICBIAS) that is internally biased to the VMID voltage level by way of a voltage divider between AVDD and AGND. The microphone input signal can be connected to the internal ADC and, if desired, routed directly to the outputs via the sidetone path by using the SIDETONE bit (Register R4, Bit D5).



Figure 3. Microphone Input to ADC

The first gain stage is composed of a low noise operational amplifier set to an inverting configuration with integrated 50 kΩ feedback and 10 kΩ input resistors. The default microphone input signal gain is 14 dB. An external resistor (R_{EXT}) can be connected in series with the MICIN pin to reduce the first-stage gain of the microphone input signal to as low as 0 dB by using the following equation:

$$\text{Microphone Input Gain} = 50 \text{ k}\Omega / (10 \text{ k}\Omega + R_{EXT})$$

The second-stage gain of the microphone signal path is derived from the internal microphone boost circuitry. The available settings are 0 dB, 20 dB, and 40 dB and are controlled by the MICBOOST (Register R4, Bit D0) and MICBOOST2 (Register R4, Bit D8) bits. To achieve 20 dB of secondary gain boost, the programmer can select either MICBOOST or MICBOOST2. To achieve 40 dB of secondary microphone signal gain, the programmer must select both MICBOOST and MICBOOST2.

The MUTEMIC bit (Register R4, Bit D1) mutes the microphone input signal to the ADC.

When using either the line or microphone inputs, the maximum full-scale input to the ADC is 1.0 V rms when AVDD = 3.3 V. Do not apply an input voltage larger than full-scale to avoid overloading the ADC, which causes distortion of sound and deterioration of audio quality. For best sound quality in both microphone and line inputs, gain should be carefully configured so that the ADC receives a signal equal to its full-scale. This maximizes the signal-to-noise ratio for best total audio quality.

Bypass and Sidetone Paths to Output

The line and microphone inputs can be routed and mixed directly to the output terminals by programming the SIDETONE (Register R4, Bit D5) and BYPASS (Register R4, Bit D3) registers. In both modes, the analog input signal is routed directly to the output terminals and is not digitally converted. The bypass signal at the output mixer is the same level as the output of the PGA associated with each line input.

The sidetone signal at the output mixer can be attenuated from -6 dB to -15 dB in steps of -3 dB by configuring the SIDEATT (Register R4, Bit D6 and Bit D7) control register bits. The

selected level of attenuation occurs after the initial microphone signal amplification from the microphone first and second stage gains.

Line and Headphone Outputs

The DAC outputs, the microphone (the sidetone path), and the line inputs (the bypass path) are summed at an output mixer (see Figure 4). This output signal is then applied to both the stereo line outputs and stereo headphone outputs.



Figure 4. Output Signal Chain

The codec has a set of efficient headphone amplifier outputs, LHPOUT and RHPOUT, that are able to drive 16 Ω or 32 Ω headphones (shown in Figure 5).



Figure 5. Headphone Output

Like the line inputs, the LHPOUT and RHPOUT volumes, by default, are independently adjusted by setting the LHPVOL (Register R2, Bit D0 to Bit D6) and RHPVOL (Register R3, Bit D0 to Bit D6) bits of the headphone output control registers. The headphone outputs can be muted by writing codes less than 0110000 to the LHPVOL and RHPVOL bits.

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The programmer can simultaneously load the volume control of both channels by writing to the LRHPBOTH (Register R2, Bit D8) and RLHPBOTH (Register R3, Bit D8) bits of the left- or right-channel DAC volume registers.

The maximum output level of the headphone outputs is 1.0 V rms when AVDD and HPVDD = 3.3 V. To suppress audible pops and clicks, the headphone and line outputs are held at the VMID dc voltage level when the device is set to standby mode or when the headphone outputs are muted.

The stereo line outputs of the codec, the LOUT and ROUT pins, can drive a load impedance of 10 kΩ and 50 pF. The line output signal levels are not adjustable at the output mixer, which has a fixed gain of 0 dB. The maximum output level of the line outputs is 1.0 V rms when AVDD = 3.3 V.

DIGITAL AUDIO INTERFACE

The digital audio input can support the following digital audio communication protocols: right-justified mode, left-justified mode, I²S mode, and frame sync mode. See [Figure 6 on Page 6](#) through [Figure 10 on Page 7](#).

The mode selection is performed by writing to the FORMAT bits of the digital audio interface register (Register R7, Bit D1 and Bit D0). All modes are MSB first and operate with data of 16 to 32 bits.

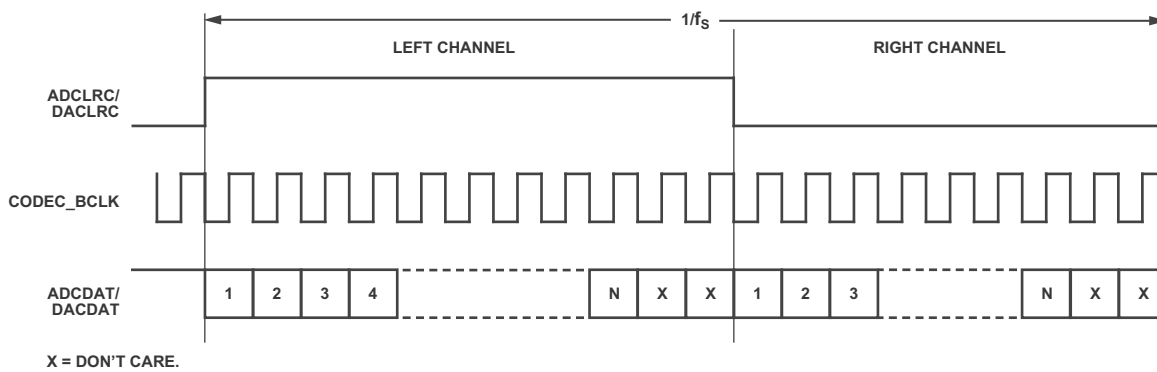


Figure 6. Left-Justified Audio Input Mode

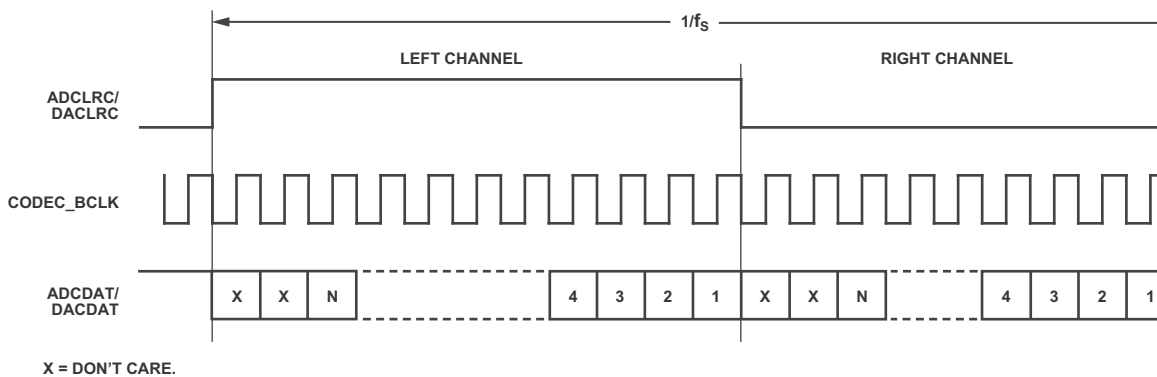


Figure 7. Right-Justified Audio Input Mode

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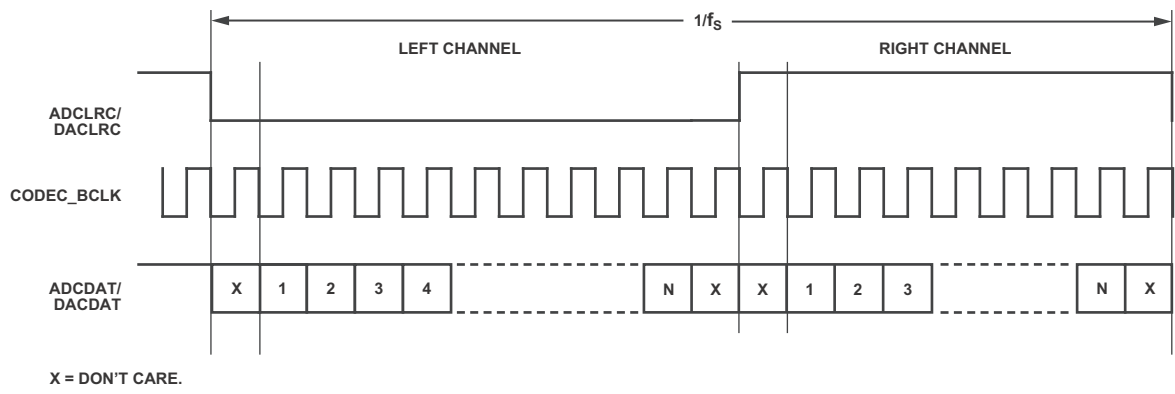


Figure 8. I²S Audio Input Mode

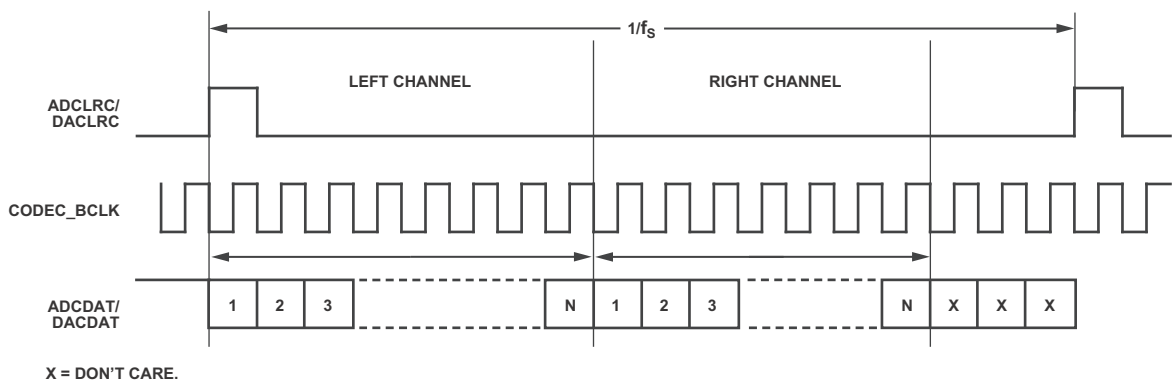


Figure 9. Frame Sync/PCM Mode Audio Input (Submode 1) [Bit LRP = 0]

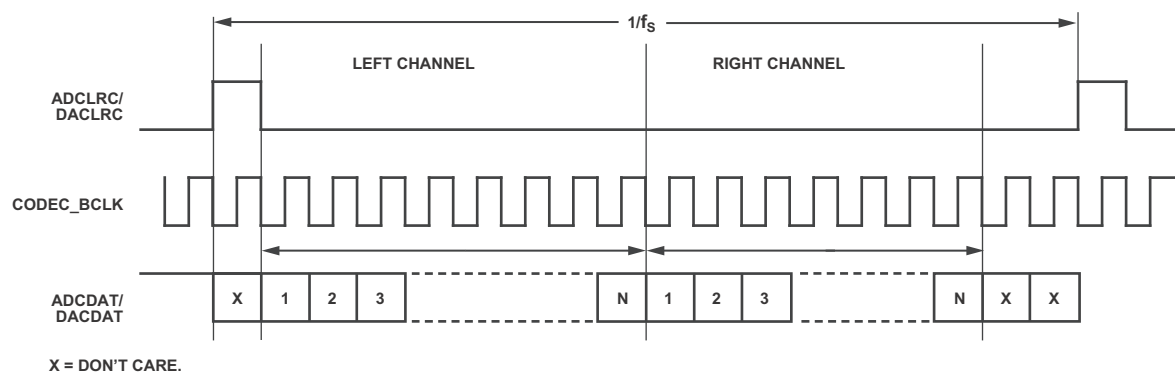


Figure 10. Frame Sync/PCM Mode Audio Input (Submode 2) [Bit LRP = 1]

Recording Mode

The digital audio interface sends the ADC digital filter data to the ADCDAT output pin for recording. The ADCDAT data stream multiplexes the left- and right-channel audio data in the time domain. The ADCLRC clock signal separates left- and right-channel digital audio frames on the ADCDAT lines.

The CODEC_BCLK signal clocks the digital audio data within the frames. The CODEC_BCLK signal is either an input or an output depending on whether the codec is in master or slave mode. During a recording operation, ADCDAT and ADCLRC must be synchronous to the CODEC_BCLK signal to avoid data corruption.

Playback Mode

The digital audio interface receives data on the DACDAT input pin for playback. The digital audio data stream on the DACDAT pin is time-domain-multiplexed left and right channel audio data. The DACLRC clock signal separates left and right channel digital audio frames on the DACDAT lines.

The CODEC_BCLK signal clocks the digital audio data within the frames. The CODEC_BCLK signal is either an input or an output depending on whether the codec is in master or slave mode. During a playback operation, DACDAT and DACLRC must be synchronous to the CODEC_BCLK signal to avoid data corruption.

Digital Audio Data Sampling Rate

To accommodate a wide variety of commonly used DAC and ADC sampling rates, the codec allows for two modes of operation, normal and USB, selected by the USB bit (Register R8, Bit D0).

The sampling rate is generated as a fixed divider from the CODEC_MCLK signal. Because all audio processing references the CODEC_MCLK signal, corruption of this signal will corrupt the quality of the audio at the codec output. The ADCLRC/ ADCDAT/CODEC_BCLK or DACLRC/DACDAT/ CODEC_BCLK signals must be synchronized with CODEC_MCLK in the digital audio interface circuit.

CODEC_MCLK must be faster or equal to the CODEC_BCLK frequency to guarantee that no data is lost during data synchronization. The CODEC_BCLK frequency should be greater than the sampling rate \times word length \times 2. Ensuring that the CODEC_BCLK frequency is greater than this, guarantees that all valid data bits are captured by the digital audio interface circuitry. For example, if a 32 kHz digital audio sampling rate with a 32-bit word length is desired, CODEC_BCLK = 2.048 MHz.

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Normal Mode

In normal mode, the codec supports digital audio sampling rates from 8 kHz to 96 kHz. Normal mode supports $256 \times f_s$ and $384 \times f_s$ based clocks. To select the desired sampling rate, the programmer must set the appropriate sampling rate register in

the SR control bits (Register R8, Bit D2 to Bit D5) and match this selection to the core clock frequency that is pulsed on the CODEC_MCLK pin. See [Table 1](#) for sampling rates in normal mode.

Table 1. Sampling Rate Lookup Table, Normal Mode (USB Disabled)

CODEC_MCLK (CLKDIV2 = 0)	CODEC_MCLK (CLKDIV2 = 1)	ADC Sampling Rate (ADCLRC)	DAC Sampling Rate (DACLRC)	USB	SR [3:0]	BOSR	CODEC_BCLK (MS = 1) ¹
12.288 MHz	24.576 MHz	8 kHz (CODEC_MCLK/1536)	8 kHz (CODEC_MCLK/1536)	0	0011	0	CODEC_MCLK/4
		8 kHz (CODEC_MCLK/1536)	48 kHz (CODEC_MCLK/256)	0	0010	0	CODEC_MCLK/4
		12 kHz (CODEC_MCLK/1024)	12 kHz (CODEC_MCLK/1024)	0	0100	0	CODEC_MCLK/4
		16 kHz (CODEC_MCLK/768)	16 kHz (CODEC_MCLK/768)	0	0101	0	CODEC_MCLK/4
		24 kHz (CODEC_MCLK/512)	24 kHz (CODEC_MCLK/512)	0	1110	0	CODEC_MCLK/4
		32 kHz (CODEC_MCLK/384)	32 kHz (CODEC_MCLK/384)	0	0110	0	CODEC_MCLK/4
		48 kHz (CODEC_MCLK/256)	8 kHz (CODEC_MCLK/1536)	0	0001	0	CODEC_MCLK/4
		48 kHz (CODEC_MCLK/256)	48 kHz (CODEC_MCLK/256)	0	0000	0	CODEC_MCLK/4
		96 kHz (CODEC_MCLK/128)	96 kHz (CODEC_MCLK/128)	0	0111	0	CODEC_MCLK/2
11.2896 MHz	22.5792 MHz	8.0182 kHz (CODEC_MCLK/1408)	8.0182 kHz (CODEC_MCLK/1408)	0	1011	0	CODEC_MCLK/4
		8.0182 kHz (CODEC_MCLK/1408)	44.1 kHz (CODEC_MCLK/256)	0	1010	0	CODEC_MCLK/4
		11.025 kHz (CODEC_MCLK/1024)	11.025 kHz (CODEC_MCLK/1024)	0	1100	0	CODEC_MCLK/4
		22.05 kHz (CODEC_MCLK/512)	22.05 kHz (CODEC_MCLK/512)	0	1101	0	CODEC_MCLK/4
		44.1 kHz (CODEC_MCLK/256)	8.0182 kHz (CODEC_MCLK/1408)	0	1001	0	CODEC_MCLK/4
		44.1 kHz (CODEC_MCLK/256)	44.1 kHz (CODEC_MCLK/256)	0	1000	0	CODEC_MCLK/4
		88.2 kHz (CODEC_MCLK/128)	88.2 kHz (CODEC_MCLK/128)	0	1111	0	CODEC_MCLK/2
18.432 MHz	36.864 MHz	8 kHz (CODEC_MCLK/2304)	8 kHz (CODEC_MCLK/2304)	0	0011	1	CODEC_MCLK/6
		8 kHz (CODEC_MCLK/2304)	48 kHz (CODEC_MCLK/384)	0	0010	1	CODEC_MCLK/6
		12 kHz (CODEC_MCLK/1536)	12 kHz (CODEC_MCLK/1536)	0	0100	1	CODEC_MCLK/6
		16 kHz (CODEC_MCLK/1152)	16 kHz (CODEC_MCLK/1152)	0	0101	1	CODEC_MCLK/6
		24 kHz (CODEC_MCLK/768)	24 kHz (CODEC_MCLK/768)	0	1110	1	CODEC_MCLK/6
		32 kHz (CODEC_MCLK/576)	32 kHz (CODEC_MCLK/576)	0	0110	1	CODEC_MCLK/6
		48 kHz (CODEC_MCLK/384)	48 kHz (CODEC_MCLK/384)	0	0000	1	CODEC_MCLK/6
		48 kHz (CODEC_MCLK/384)	8 kHz (CODEC_MCLK/2304)	0	0001	1	CODEC_MCLK/6
		96 kHz (CODEC_MCLK/192)	96 kHz (CODEC_MCLK/192)	0	0111	1	CODEC_MCLK/3
16.9344 MHz	33.8688 MHz	8.0182 kHz (CODEC_MCLK/2112)	8.0182 kHz (CODEC_MCLK/2112)	0	1011	1	CODEC_MCLK/6
		8.0182 kHz (CODEC_MCLK/2112)	44.1 kHz (CODEC_MCLK/384)	0	1010	1	CODEC_MCLK/6
		11.025 kHz (CODEC_MCLK/1536)	11.025 kHz (CODEC_MCLK/1536)	0	1100	1	CODEC_MCLK/6
		22.05 kHz (CODEC_MCLK/768)	22.05 kHz (CODEC_MCLK/768)	0	1101	1	CODEC_MCLK/6
		44.1 kHz (CODEC_MCLK/384)	8.0182 kHz (CODEC_MCLK/2112)	0	1001	1	CODEC_MCLK/6
		44.1 kHz (CODEC_MCLK/384)	44.1 kHz (CODEC_MCLK/384)	0	1000	1	CODEC_MCLK/6
		88.2 kHz (CODEC_MCLK/192)	88.2 kHz (CODEC_MCLK/192)	0	1111	1	CODEC_MCLK/3

¹ CODEC_BCLK frequency is for master mode and slave right-justified mode only.

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USB Mode

In USB mode, the codec supports digital audio sampling rates from 8 kHz to 96 kHz. USB mode is enabled on the codec to support the common universal serial bus (USB) clock rate of

12 MHz, or to support 24 MHz if the CLKDIV2 control register bit is activated. The programmer must set the appropriate sampling rate in the SR control bits (Register R8, Bit D2 to Bit D5). See [Table 2](#) for sampling rates in USB mode.

Table 2. Sampling Rate Lookup Table, USB Mode (USB Enabled)

CODEC_MCLK (CLKDIV2 = 0)	CODEC_MCLK (CLKDIV2 = 1)	ADC Sampling Rate (ADCLRC)	DAC Sampling Rate (DACLRC)	USB	SR [3:0]	BOSR	CODEC_BCLK (MS = 1) ¹
12.000 MHz	24.000 MHz	8 kHz (CODEC_MCLK/1500)	8 kHz (CODEC_MCLK/1500)	1	0011	0	CODEC_MCLK
		8 kHz (CODEC_MCLK/1500)	48 kHz (CODEC_MCLK/250)	1	0010	0	CODEC_MCLK
		8.0214 kHz (CODEC_MCLK/1496)	8.0214 kHz (CODEC_MCLK/1496)	1	1011	1	CODEC_MCLK
		8.0214 kHz (CODEC_MCLK/1496)	44.118 kHz (CODEC_MCLK/272)	1	1010	1	CODEC_MCLK
		11.0259 kHz (CODEC_MCLK/1088)	11.0259 kHz (CODEC_MCLK/1088)	1	1100	1	CODEC_MCLK
		12 kHz (CODEC_MCLK/1000)	12 kHz (CODEC_MCLK/1000)	1	1000	0	CODEC_MCLK
		16 kHz (CODEC_MCLK/750)	16 kHz (CODEC_MCLK/750)	1	1010	0	CODEC_MCLK
		22.0588 kHz (CODEC_MCLK/544)	22.0588 kHz (CODEC_MCLK/544)	1	1101	1	CODEC_MCLK
		24 kHz (CODEC_MCLK/500)	24 kHz (CODEC_MCLK/500)	1	1110	0	CODEC_MCLK
		32 kHz (CODEC_MCLK/375)	32 kHz (CODEC_MCLK/375)	1	0110	0	CODEC_MCLK
		44.118 kHz (CODEC_MCLK/272)	8.0214 kHz (CODEC_MCLK/1496)	1	1001	1	CODEC_MCLK
		44.118 kHz (CODEC_MCLK/272)	44.118 kHz (CODEC_MCLK/272)	1	1000	1	CODEC_MCLK
		48 kHz (CODEC_MCLK/250)	8 kHz (CODEC_MCLK/1500)	1	0001	0	CODEC_MCLK
		48 kHz (CODEC_MCLK/250)	48 kHz (CODEC_MCLK/250)	1	0000	0	CODEC_MCLK
		88.235 kHz (CODEC_MCLK/136)	88.235 kHz (CODEC_MCLK/136)	1	1111	1	CODEC_MCLK
		96 kHz (CODEC_MCLK/125)	96 kHz (CODEC_MCLK/125)	1	0111	0	CODEC_MCLK

¹ CODEC_BCLK frequency is for master mode and slave right-justified mode only.

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SOFTWARE CONTROL INTERFACE

The software control interface provides access to the programmer-selectable control registers and can operate with a 2-wire (TWI) or 3-wire (SPI) interface, depending on the setting of the CMODE pin. If the CMODE pin is set to 0, the 2-wire interface is selected; if 1, the 3-wire interface is selected.

Within each control register is a control data-word consisting of 16 bits, MSB first. Bit B15 to Bit B9 are the register map address, and Bit B8 to Bit B0 are register data for the associated register map.

When 2-wire (TWI) mode is selected, CSDA generates the serial control data-word; CSCL clocks the serial data; and CSB determines the TWI device address. If the CSB pin is set to 0, the address selected is 0011010; if 1, the address is 0011011.

When 3-wire (SPI) mode is selected, CSDA generates the control data-word, CSCL clocks the control data-word into the codec, and CSB latches in the control data-word.

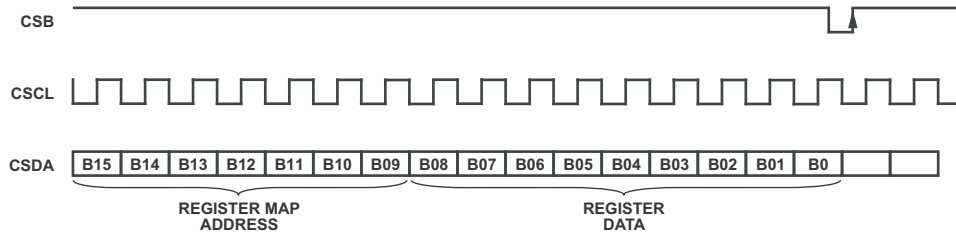


Figure 11. Codec SPI Serial Interface

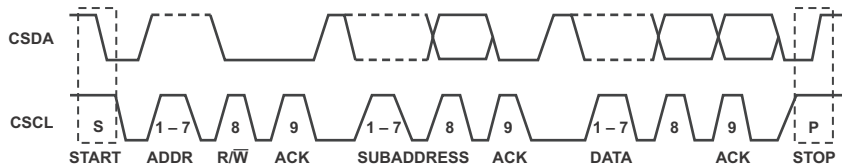


Figure 12. Codec TWI Serial Interface



S/P = START/STOP BIT.
 A0 = I²C R/W BIT.
 A(S) = ACKNOWLEDGE BY SLAVE.
 A(M) = ACKNOWLEDGE BY MASTER.
 A(M) = ACKNOWLEDGE BY MASTER (INVERSION).

Figure 13. Codec TWI Write and Read Sequences

CODEC PIN DESCRIPTIONS

Table 3 shows the signals added to the ADSP-BF52xC processor for the embedded codec. Please refer to the published ADSP-BF52x data sheet for descriptions of other signals for the processor.

Table 3. Codec Pin Descriptions

Pin Name	Type	Function	Pull-Up/Down
<i>Codec</i>			
CODEC_CLKOUT	O	Codec Clock Output	None
CODEC_BCLK	I/O	Codec Digital Audio Bit Clock	Internal Pull-down ¹
DACDAT	I	Codec Digital Audio Data (DAC) Input	None
DACLRC	I/O	Codec DAC Sample Rate Left/Right Clock	Internal Pull-down ¹
ADCDAT	O	Codec ADC Digital Audio Data Output	None
ADCLRC	I/O	Codec ADC Sample Rate Left/Right Clock	Internal Pull-down ¹
CMODE	I	Codec Control Interface Selection	Internal Pull-up ¹
CSB	I	Codec Chip Select Interface Address Selection	Internal Pull-up ¹
CSDA	I/O	Codec Data Input	None
CSCL	I/O	Codec Data Clock	None
XTI/CODEC_MCLK	I	Codec Crystal Input/ Clock Input	None
XTO	O	Codec Crystal Output	None
LHPOUT	O	Codec Left Channel Headphone Output (Analog Output)	None
RHPOUT	O	Codec Right Channel Headphone Output (Analog Output)	None
LOUT	O	Codec Left Channel Line Output (Analog Output)	None
ROUT	O	Codec Right Channel Line Output (Analog Output)	None
VMID	O	Codec Mid-rail Reference Decoupling Point (Analog Output)	None
MICBIAS	O	Codec Electret Microphone Bias (Analog Output)	None
MICIN	I	Codec Microphone Input; (Analog Input, AC Coupled)	None
RLINEIN	I	Codec Right Channel Line Input (Analog Input, AC Coupled)	None
LLINEIN	I	Codec Left Channel Line Input (Analog Input, AC Coupled)	None
AVDD	P	Codec Analog V _{DD}	N/A
AGND	P	Codec Analog Ground	N/A
CVDD	P	Codec Digital V _{DD}	N/A
HPVDD	P	Codec Analog Headphone V _{DD}	N/A
HPGND	P	Codec Headphone Ground	N/A

¹To conserve power, the pull-up/pull-down is only present when the control register interface is active (= 0).

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Figure 14 on Page 13 and Figure 15 on Page 14 describe alternative external connections for SPI or TWI control of the ADSP-BF52xC codec. The figures are the same except for the shaded area in each.

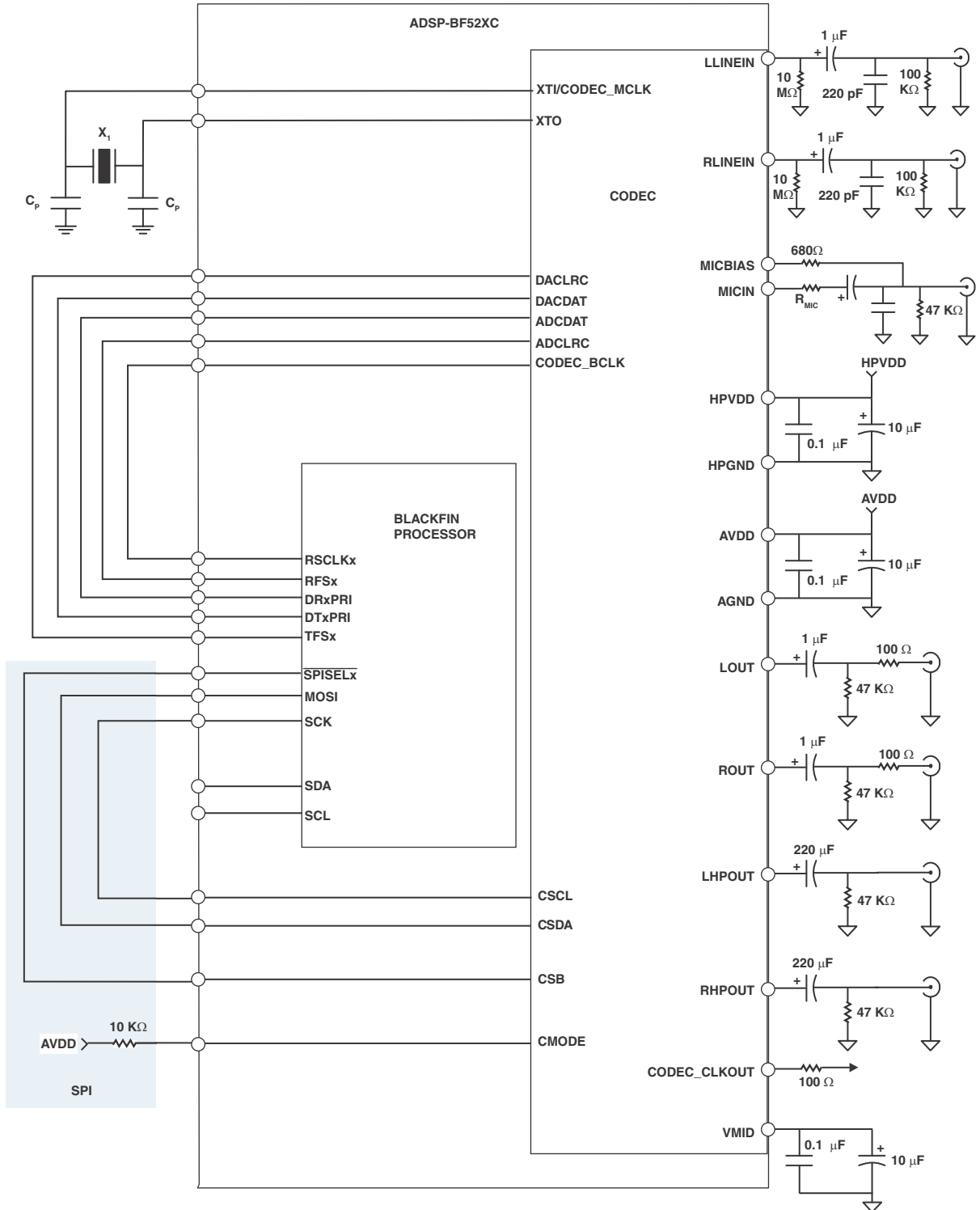


Figure 14. Recommended Application Circuit Using SPI Control

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C



Figure 15. Recommended Application Circuit Using TWI Control

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REGISTER DETAILS

Register	Address	B8	B7	B6	B5	B4	B3	B2	B1	B0
Register 0 Left-Channel ADC Input Volume on Page 16	0x00	LRINBOTH	LINMUTE	0	LINVOL					
Default = 010010111										
Register 1 Right-Channel ADC Input Volume on Page 17	0x01	RLINBOTH	RINMUTE	0	RINVOL					
Default = 010010111										
Register 2 Left-Channel DAC Volume on Page 17	0x02	LRHPBOTH	LZCEN	LHPVOL						
Default = 001111001										
Register 3 Right-Channel DAC Volume on Page 18	0x03	RLHPBOTH	RZCEN	RHPVOL						
Default = 001111001										
Register 4 Analog Audio Path on Page 18	0x04	MICBOOST2	SIDEATT[1:0]		SIDETONE	DACSEL	BYPASS	INSEL	MUTEMIC	MICBOOST
Default = 000001010										
Register 5 Digital Audio Path on Page 19	0x05	0	0	0	0	HPOR	DACMU	DEEMPH[1:0]		ADC HPD
Default = 000001000										
Register 6 Power Management on Page 19	0x06	0	POWEROFF	CLKOUTPD	OSCPD	OUTPD	DACPD	ADCPD	MICPD	LINEINPD
Default = 010011111										
Register 7 Digital Audio I/F on Page 20	0x07	0	BCLKINV	MS	LRSWAP	LRP	WL[1:0]		FORMAT[1:0]	
Default = 000001010										
Register 8 Sampling Rate on Page 20	0x08	0	CLKODIV2	CLKDIV2	SR[3:0]				BOSR	USB
Default = 000000000										
Register 9 Active on Page 20	0x09	0	0	0	0	0	0	0	0	ACTIVE
Default = 000000000										
Register 10 Software Reset on Page 20	0x0F	RESET								
Default = 000000000										

Figure 16. Register Mapping

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BIT DESCRIPTIONS

Table 4 through Table 14 on Page 20 describe each bit in the control registers.

Table 4. Register 0 Left-Channel ADC Input Volume

Bit Name	Bits	Description	Settings
LRINBOTH	B8	Left-to-right line input ADC data load control	0 = disable simultaneous loading of left-channel ADC data to right-channel register (default) 1 = enable simultaneous loading of left-channel ADC data to right-channel register
LINMUTE	B7	Left-channel input mute	0 = disable mute 1 = enable mute on data path to ADC (default)
LINVOL	B[5:0]	Left-channel PGA volume control	00 0000 = -34.5 dB ... 1.5 dB step up 01 0111 = 0 dB (default) ... 1.5 dB step up 01 1111 = 12 dB 10 0000 = 13.5 dB 10 0001 = 15 dB 10 0010 = 16.5 dB 10 0011 = 18 dB 10 0100 = 19.5 dB 10 0101 = 21 dB 10 0110 = 22.5 dB 10 0111 = 24 dB 10 1000 = 25.5 dB 10 1001 = 27 dB 10 1010 = 28.5 dB 10 1011 = 30 dB 10 1100 = 31.5 dB 10 1101 = 33 dB 11 1111 to 10 1101 = 33 dB

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Table 5. Register 1 Right-Channel ADC Input Volume

Bit Name	Bits	Description	Settings
RLINBOTH	B8	Right-to-left line input ADC data load control	0 = disable simultaneous loading of right-channel ADC data to left-channel register (default) 1 = enable simultaneous loading of right-channel ADC data to left-channel register
RINMUTE	B7	Right-channel input mute	0 = disable mute 1 = enable mute on data path to ADC (default)
RINVOL	B[5:0]	Right-channel PGA volume control	00 0000 = -34.5 dB ... 1.5 dB step up 01 0111 = 0 dB (default) ... 1.5 dB step up 01 1111 = 12 dB 10 0000 = 13.5 dB 10 0001 = 15 dB 10 0010 = 16.5 dB 10 0011 = 18 dB 10 0100 = 19.5 dB 10 0101 = 21 dB 10 0110 = 22.5 dB 10 0111 = 24 dB 10 1000 = 25.5 dB 10 1001 = 27 dB 10 1010 = 28.5 dB 10 1011 = 30 dB 10 1100 = 31.5 dB 10 1101 = 33 dB 11 1111 to 10 1101 = 33 dB

Table 6. Register 2 Left-Channel DAC Volume

Bit Name	Bits	Description	Settings
LRHPBOTH	B8	Left-to-right headphone volume load control	0 = disable simultaneous loading of left-channel headphone volume data to right-channel register (default) 1 = enable simultaneous loading of left-channel headphone volume data to right-channel register
LZCEN	B7	Left-channel zero cross detect enable	0 = disable (default) 1 = enable
LHPVOL	B[6:0]	Left-channel headphone volume control	000 0000 to 010 1111 = mute 011 0000 = -73 dB ... 111 1001 = 0 dB (default) ... 1 dB steps up to 111 1111 = +6 dB

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Table 7. Register 3 Right-Channel DAC Volume

Bit Name	Bits	Description	Settings
RLHPBOTH	B8	Right-to-left headphone volume load control	0 = disable simultaneous loading of right-channel headphone volume data to left-channel register (default) 1 = enable simultaneous loading of right-channel headphone volume data to left-channel register
RZCEN	B7	Right-channel zero cross detect enable	0 = disable (default) 1 = enable
RHPVOL [6:0]	B[6:0]	Right-channel headphone volume control	000 0000 to 010 1111 = mute 011 0000 = -73 dB ... 111 1001 = 0 dB (default) ... 1 dB steps up to 111 1111 = +6 dB

Table 8. Register 4 Analog Audio Path

Bit Name	Bits	Description	Settings
MICBOOST2	B8	Additional microphone amplifier gain booster control	0 = 0 dB (default) 1 = 20 dB
SIDEATT[1:0]	B[7:6]	Microphone sidetone gain control	00 = -6 dB (default) 01 = -9 dB 10 = -12 dB 11 = -15 dB
SIDETONE	B5	Sidetone enable. Allow attenuated microphone signal to be mixed at device output terminal	0 = sidetone disable (default) 1 = sidetone enable
DACSEL	B4	DAC select—allow DAC output to be mixed at device output terminal	0 = do not select DAC (default) 1 = select DAC
BYPASS	B3	Bypass select—allow line input signal to be mixed at device output terminal	0 = bypass disable 1 = bypass enable (default)
INSEL	B2	Line input or microphone input select to ADC	0 = line input select to ADC (default) 1 = microphone input select to ADC
MUTEMIC	B1	Microphone mute control to ADC	0 = mute on data path to ADC disable 1 = mute on data path to ADC enable (default)
MICBOOST	B0	Primary microphone amplifier gain booster control	0 = 0 dB (default) 1 = 20 dB

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Table 9. Register 5 Digital Audio Path

Bit Name	Bits	Description	Settings
HPOR	B4	Store dc offset when high-pass filter is disabled	0 = clear offset (default) 1 = store offset
DACMU	B3	DAC digital mute	0 = no mute (signal active) 1 = mute (default)
DEEMPH[1:0]	B[2:1]	De-emphasis control	00 = no de-emphasis (default) 01 = 32 kHz sampling rate 10 = 44.1 kHz sampling rate 11 = 48 kHz sampling rate
ADCHPD	B0	ADC high-pass filter control	0 = ADC high-pass filter enable (default) 1 = ADC high-pass filter disable

Table 10. Register 6 Power Management

Bit Name	Bits	Description	Settings
POWEROFF	B7	Whole chip power-down control	0 = power-up 1 = power-down (default)
CLKOUTPD	B6	Clock output power-down control	0 = power-up (default) 1 = power-down
OSCPD	B5	Crystal power-down control	0 = power-up (default) 1 = power-down
OUTPD	B4	Output power-down control	0 = power-up 1 = power-down (default)
DACPD	B3	DAC power-down control	0 = power-up 1 = power-down (default)
ADCPD	B2	ADC power-down control	0 = power-up 1 = power-down (default)
MICPD	B1	Microphone input power-down control	0 = power-up 1 = power-down (default)
LINEINPD	B0	Line input power-down control	0 = power-up 1 = power-down (default)

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Table 11. Register 7 Digital Audio I/F

Bit Name	Bits	Description	Settings
BCLKINV	B7	CODEC_BCLK inversion control	0 = CODEC_BCLK not inverted (default) 1 = CODEC_BCLK inverted
MS	B6	Master mode enable	0 = enable slave mode (default) 1 = enable master mode
LRSWAP	B5	Swap DAC data control	0 = output left- and right-channel data as normal (default) 1 = swap left- and right-channel DAC data in audio interface
LRP	B4	Polarity control for clocks in right-justified, left-justified, and I ² S modes	0 = normal DACLRC and ADCLRC (default), or processor Submode 1 1 = invert DACLRC and ADCLRC polarity, or processor Submode 2
WL [1:0]	B[3:2]	Data-word length control	00 = 16 bits 01 = 20 bits 10 = 24 bits (default) 11 = 32 bits
FORMAT [1:0]	B[1:0]	Digital audio input format control	00 = right justified 01 = left justified 10 = I ² S mode (default) 11 = processor mode

Table 12. Register 8 Sampling Rate

Bit Name	Bits	Description	Settings
CLKODIV2	B7	CODEC_CLKOUT divider select	0 = CODEC_CLKOUT is codec clock (default) 1 = CODEC_CLKOUT is codec clock divided by 2
CLKDIV2	B6	Codec clock divide select	0 = codec clock is CODEC_MCLK (default) 1 = codec clock is CODEC_MCLK divided by 2
SR [3:0]	B[5:2]	Clock setting condition	See Table 1 on Page 9 and Table 2 on Page 10
BOSR	B1	Base oversampling rate	USB mode: 0 = support for $250 \times f_s$ based clock (default) 1 = support for $272 \times f_s$ based clock Normal mode: 0 = support for $256 \times f_s$ based clock (default) 1 = support for $384 \times f_s$ based clock
USB	B0	USB mode select	0 = normal mode enable (default) 1 = USB mode enable

Table 13. Register 9 Active

Bit Name	Bit	Description	Settings
ACTIVE	B0	Digital core activation control	0 = disable digital core (default) 1 = activate digital core

Table 14. Register 10 Software Reset

Bit Name	Bit	Description	Settings
RESET [8:0]	B[8:0]	Write all 0s to this register to set all registers to their default settings. Other data written to this register has no effect.	0 = reset (default)

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SPECIFICATIONS

$T_{\text{Ambient}} = 25^{\circ}\text{C}$, $\text{AVDD} = \text{VDDEXT} = 3.3 \text{ V}$, $\text{HPVDD} = 3.3 \text{ V}$,
 1 kHz signal, $f_s = 48 \text{ kHz}$, PGA gain = 0 dB, 24-bit audio data,
 unless otherwise noted.

OPERATING CONDITIONS

See operating conditions in the published ADSP-BF52xC data sheet.

Parameter	Conditions	Min	Typical	Max	Unit
AVDD ¹		1.8	3.3	3.6	V
HPVDD		1.8	3.3	3.6	V

¹Note that AVDD must equal HPVDD.

CODEC ELECTRICAL CHARACTERISTICS

Parameter	Conditions	Min	Typical	Max	Unit
<i>Line Input</i>					
Input Signal Level (0 dB)			AVDD/3.3		V(rms)
Input Impedance	PGA gain = 0 dB		200		k Ω
	PGA gain = +33 dB		10		k Ω
	PGA gain = -34.5 dB		480		k Ω
Input Capacitance			10		pF
Signal-to-Noise Ratio (A-Weighted)	PGA gain = 0 dB, AVDD = 3.3 V	82	87		dB
	PGA gain = 0 dB, AVDD = 1.8 V		84		dB
Total Harmonic Distortion (THD)	-1 dBFS input, AVDD = 3.3 V	-80	-84		dB
	-1 dBFS input, AVDD = 1.8 V		-71	-60	dB
Channel Separation ¹			80		dB
Programmable Gain		-34.5	0	+33.5	dB
Gain Step			1.5		dB
Mute Attenuation			-80		dB
<i>Microphone Input</i>					
Input Signal Level			1		V(rms)
Signal-to-Noise Ratio (A-Weighted)	Microphone gain = 0 dB ($R_{\text{SOURCE}} = 40 \text{ k}\Omega$)		85		dB
Total Harmonic Distortion	-1 dBFS input, 0 dB gain, AVDD = 3.3 V		-75		dB
	-1 dBFS input, 0 dB gain, AVDD = 1.8 V		-65		dB
Power Supply Rejection Ratio			50		dB
Mute Attenuation			80		dB
Input Resistance			10		k Ω
Input Capacitance			10		pF
<i>Microphone Bias</i>					
Bias Voltage			$0.75 \times \text{AVDD}$		V
Bias Current Source				3	mA
Noise in the Signal Bandwidth	20 Hz to 20 kHz		40		nV/ $\sqrt{\text{Hz}}$

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Parameter	Conditions	Min	Typical	Max	Unit
<i>Line Output</i>					
DAC	-1 dBFS input DAC + line output				
Full-Scale Output			AVDD/3.3		V(rms)
Signal-to-Noise Ratio (A-Weighted)	AVDD = 3.3 V	90	95		dB
	AVDD = 1.8 V	85	88		dB
THD + N	AVDD = 3.3 V		-80	-70	dB
	AVDD = 1.8 V		-80	-70	dB
Power Supply Rejection Ratio			50		dB
Channel Separation			80		dB
<i>Headphone Output</i>					
Full-Scale Output Voltage			AVDD/3.3		V(rms)
Maximum Output Power	R _L = 32 Ω		30		mW
	R _L = 16 Ω		60		mW
Signal-to-Noise Ratio (A-Weighted)	AVDD = 3.3 V	90	94		dB
	AVDD = 1.8 V	80	85		dB
THD + N	HPOUT = 10 mW		-65		dB
	HPOUT = 20 mW		-60		dB
Power Supply Rejection Ratio			50		dB
Mute Attenuation			80		dB
<i>Line Input To Line Output</i>					
Full-Scale Output Voltage			AVDD/3.3		V(rms)
Signal-to-Noise Ratio (A-Weighted)	AVDD = 3.3 V		92		dB
	AVDD = 1.8 V		86		dB
Total Harmonic Distortion	AVDD = 3.3 V		-80		dB
	AVDD = 1.8 V		-80		dB
Power Supply Rejection			50		dB
<i>Microphone Input To Headphone Output</i>					
Full-Scale Output Voltage			AVDD/3.3		V(rms)
Signal-to-Noise Ratio (A-Weighted)	AVDD = 3.3 V		94		dB
	AVDD = 1.8 V		88		dB
Power Supply Rejection Ratio			50		dB
Programmable Attenuation		6		15	dB
Gain Step			3		dB
Mute Attenuation			80		dB

¹ Guaranteed but not tested.

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ABSOLUTE MAXIMUM RATINGS

See absolute maximum ratings in the published ADSP-BF52x processor data sheet.

ESD SENSITIVITY



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PACKAGE INFORMATION

The information presented in [Figure 17](#) and [Table 15](#) provides details about the package branding for the ADSP-BF52xC processor. For a complete listing of product availability, see [Ordering Guide on Page 36](#).



Figure 17. Product Information on Package

Table 15. Package Brand Information

Brand Key	Field Description
t	Temperature Range
pp	Package Type
Z	Lead Free Option
ccc	See Ordering Guide
vvvvv.x	Assembly Lot Code
n.n	Silicon Revision
yyww	Date Code

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POWER CONSUMPTION

These current consumption values are for the codec alone.
Please refer to the published ADSP-BF52x processor data sheet for the additional current consumption of the Blackfin processor.

Table 16. Power Consumption

Mode	POWEROFF	CLKOUTPD	OSCPD	OUTPD	DACPD	ADCPD	MICPD	LINEINPD	(1.8V)			(3.3V)			Unit
									AVDD	HPVDD	V _{DDEXT} ¹	AVDD	HPVDD	V _{DDEXT} ¹	
Record and Playback	0	0	0	0	0	0	0	0	7.4	1.5	6.3	14.8	2.0	12.0	mA
Playback Only															
Oscillator Enabled	0	0	0	0	0	1	1	1	3.1	1.30	3.0	4.7	2.0	6.1	mA
External Clock	0	1	1	0	0	1	1	1	2.9	1.2	3.0	4.7	2.0	6.1	mA
Record Only															
Line Oscillator	0	0	0	1	1	0	1	0	2.4	N/A	3.7	4.3	N/A	7.4	mA
Line Clock	0	0	1	1	1	0	1	0	2.5	N/A	3.8	4.3	N/A	7.4	mA
Microphone 1	0	0	0	1	1	0	0	1	3.6	N/A	1.9	9.4	N/A	3.6	mA
Microphone 2	0	0	1	1	1	0	0	1	3.6	N/A	1.8	9.4	N/A	3.6	mA
Sidetone (Microphone-to-Headphone Output)															
Internally Generated Clock	0	0	0	0	1	1	0	1	2.3	1.0	2.0	7.9	2.0	4.0	mA
External Clock	0	0	1	0	1	1	0	1	2.3	1.0	2.0	7.9	2.0	4.0	mA
Analog Bypass (Line Input or Line Output)															
Internally Generated Line	0	0	0	0	1	1	1	0	0.9	1.0	2.0	1.8	2.0	4.0	mA
External Line	0	0	1	0	1	1	1	0	0.9	1.0	2.0	1.8	2.0	4.0	mA
Power-Down															
Clock Stopped	1	1	1	1	1	1	1	1	3.1	6.3	3.8	9.4	6.3	12.3	μA

¹ V_{DDEXT} here refers to the total of the codec's DCVDD and DBVDD signals and does not include VDDEXT supplies in the Blackfin device.

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TIMING SPECIFICATIONS

TWI Timing

Table 17. TWI Timing

Parameter	Test Conditions ¹	Min	Max	Unit
t_{SCS}	Start condition setup time	600		ns
t_{SCH}	Start condition hold time	600		ns
t_{PH}	CSCL pulse width high	600		ns
t_{PL}	CSCL pulse width low	1.3		μ s
f_{SCL}	CSCL frequency	0	526	kHz
t_{DS}	Data setup time	100		ns
t_{DH}	Data hold time		900	ns
t_{RT}	CSDA and CSCL rise time		300	ns
t_{FT}	CSDA and CSCL fall time		300	ns
t_{HCS}	Stop condition setup time	600		ns

¹ AVDD, HPVDD, $V_{DDEXT} = 3.3$ V, AGND = 0 V, $T_A = +25^\circ\text{C}$, Slave Mode, $f_s = 48$ kHz, XTI/CODEC_MCLK = $256 \times f_s$ unless otherwise stated.

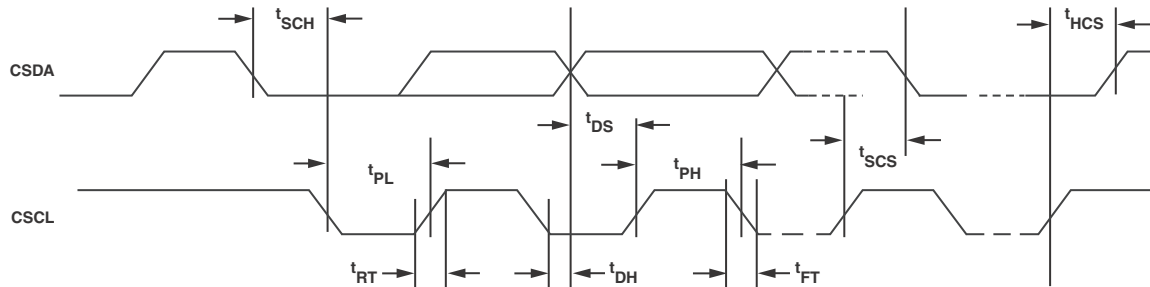


Figure 18. TWI Timing

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SPI Timing

Table 18. SPI Timing

Parameter	Test Conditions ¹	Min	Max	Unit
t_{DSU}	CSDA to CSCL setup time	20		ns
t_{DHO}	CSCL to CSDA hold time	20		ns
t_{SCH}	CSCL pulse width high	20		ns
t_{SCL}	CSCL pulse width low	20		ns
t_{SCS}	CSCL rising edge to CSB rising edge	60		ns
t_{CSS}	CSB rising to CSCL rising	20		ns
t_{CSH}	CSB pulse width high	20		ns
t_{CSL}	CSB pulse width low	20		ns
t_{PS}	Pulse width of spikes to be suppressed	0	5	ns

¹ AVDD, HPVDD, $V_{DDEXT} = 3.3$ V, AGND = 0 V, $T_A = +25^\circ\text{C}$, Slave Mode, $f_s = 48$ kHz, XTI/CODEC_MCLK = $256 \times f_s$ unless otherwise stated.



Figure 19. SPI Timing

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

Digital Audio Interface Slave Mode Timing

Table 19. Digital Audio Interface Slave Mode Timing

Parameter	Test Conditions ¹	Min	Max	Unit
t_{DS}	DACDAT setup time from CODEC_BCLK rising edge	10		ns
t_{DH}	DACDAT hold time from CODEC_BCLK rising edge	10		ns
t_{LRSU}	ADCLRC/DACLRC setup time to CODEC_BCLK rising edge	10		ns
t_{LRH}	ADCLRC/DACLRC hold time to CODEC_BCLK rising edge	10		ns
t_{DD}	ADCDAT propagation delay from CODEC_BCLK falling edge (external load of 70 pF)		30	ns
t_{BCH}	CODEC_BCLK pulse width high	25		ns
t_{BCL}	CODEC_BCLK pulse width low	25		ns
t_{BCY}	CODEC_BCLK cycle time	50		ns

¹ AVDD, HPVDD, $V_{DDEXT} = 3.3$ V, AGND = 0 V, $T_A = +25^\circ\text{C}$, Slave Mode, $f_s = 48$ kHz, XTI/CODEC_MCLK = $256 \times f_s$ unless otherwise stated.



Figure 20. Digital Audio Interface Slave Mode Timing

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

Digital Audio Interface Master Mode Timing

Table 20. Digital Audio Interface Master Mode Timing

Parameter	Test Conditions ¹	Min	Max	Unit
t_{DST}	DACDAT setup time to CODEC_BCLK rising edge	30		ns
t_{DHT}	DACDAT hold time to CODEC_BCLK rising edge	10		ns
t_{DL}	ADCLRC/DACLRC propagation delay from CODEC_BCLK falling edge		10	ns
t_{DDA}	ADCDAT propagation delay from CODEC_BCLK falling edge		10	ns
t_{BCLKR}	CODEC_BCLK rising time (10 pF load)	10		ns
t_{BCLKF}	CODEC_BCLK falling time (10 pF load)	10		ns
t_{BCLKDS}	CODEC_BCLK duty cycle (normal and USB mode)	45:55	55:45	

¹ AVDD, HPVDD, $V_{DDEXT} = 3.3$ V, AGND = 0 V, $T_A = +25^\circ\text{C}$, Slave Mode, $f_s = 48$ kHz, XTI/CODEC_MCLK = $256 \times f_s$ unless otherwise stated.

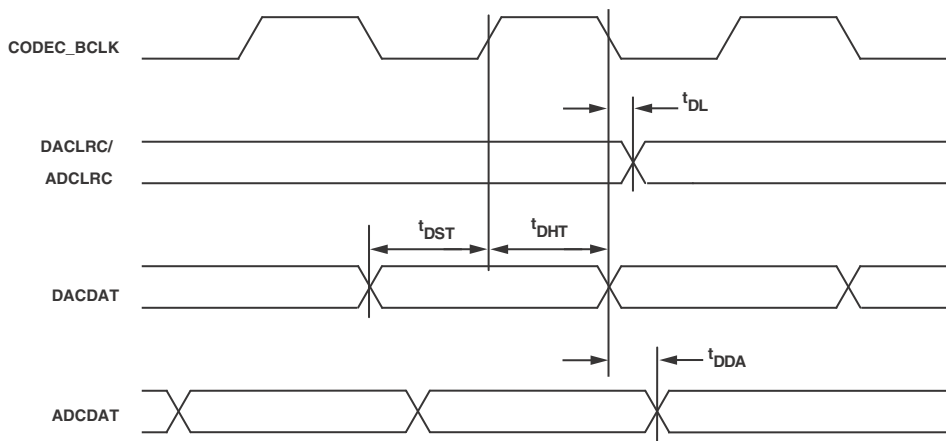


Figure 21. Digital Audio Interface Master Mode Timing

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

System Clock Timing

Table 21. System Clock Timing

Parameter	Test Conditions ¹	Min	Max	Unit
t_{XTIY}	XTI/CODEC_MCLK system clock cycle time	72		ns
t_{MCLKDS}	XTI/CODEC_MCLK duty cycle	40:60	60:40	
t_{XTIH}	XTI/CODEC_MCLK system clock pulse width high	32		ns
t_{XTIL}	XTI/CODEC_MCLK system clock pulse width low	32		ns
t_{COP}	CODEC_CLKOUT propagation delay from XTI/CODEC_MCLK falling edge	20		ns
$t_{COPDIV2}$	CLKODIV2 propagation delay from XTI/CODEC_MCLK falling edge	20		ns

¹ AVDD, HPVDD, V_{DDEXT} = 3.3 V, AGND = 0 V, T_A = +25°C, Slave Mode, f_s = 48 kHz, XTI/CODEC_MCLK = 256 × f_s unless otherwise stated.

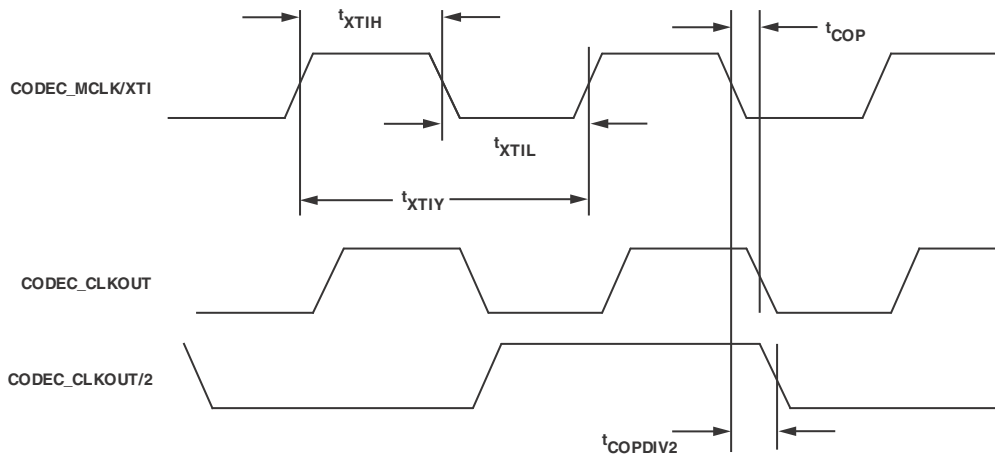


Figure 22. System (CODEC_MCLK) Clock Timing

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

Digital Filter Characteristics

Table 22. Digital Filter Characteristics

Parameter	Conditions	Min	Typical	Max	Unit
ADC FILTER					
Pass Band	± 0.04 dB	0		$0.445 \times f_s$	Hz
	-6 dB		$0.5 \times f_s$		Hz
Pass Band Ripple				± 0.04	dB
Stop Band		$0.555 \times f_s$			Hz
Stop Band Attenuation	$f > 0.567 \times f_s$	-61			dB
High-Pass Filter Corner Frequency	-3 dB		3.7		Hz
	-0.5 dB		10.4		Hz
	-0.1 dB		21.6		Hz
DAC FILTER					
Pass Band	± 0.04 dB	0		$0.445 \times f_s$	Hz
	-6 dB		$0.5 \times f_s$		Hz
Pass Band Ripple				± 0.04	dB
Stop Band		$0.555 \times f_s$			Hz
Stop Band Attenuation	$f > 0.565 \times f_s$	-61			dB
Codec Clock Tolerance					
Frequency Range		8.0		13.8	MHz
Jitter Tolerance			50		pS

CONVERTER FILTER RESPONSE

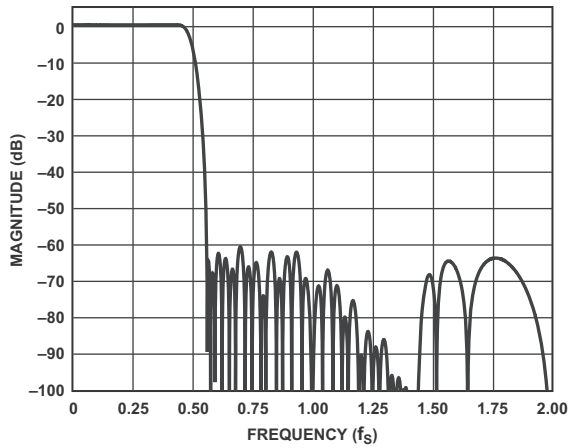


Figure 23. ADC Digital Filter Frequency Response, Sampling Rate = 48 kHz

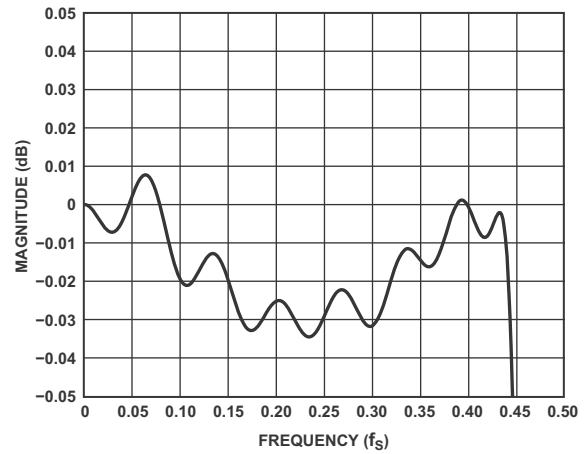


Figure 24. ADC Digital Filter Ripple, Sampling Rate = 48 kHz

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C



Figure 25. DAC Digital Filter Frequency Response, Sampling Rate = 48 kHz

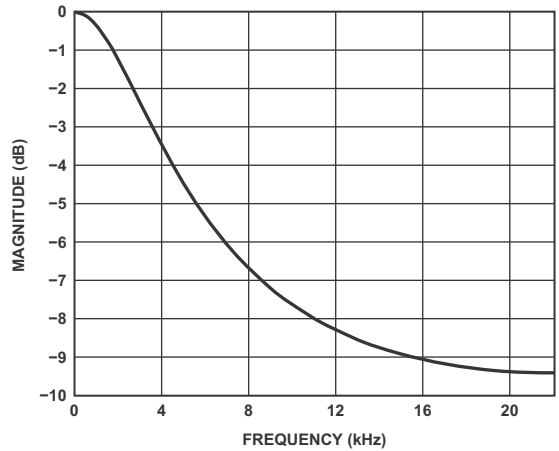


Figure 28. De-Emphasis Frequency Response, Sampling Rate = 44.1 kHz

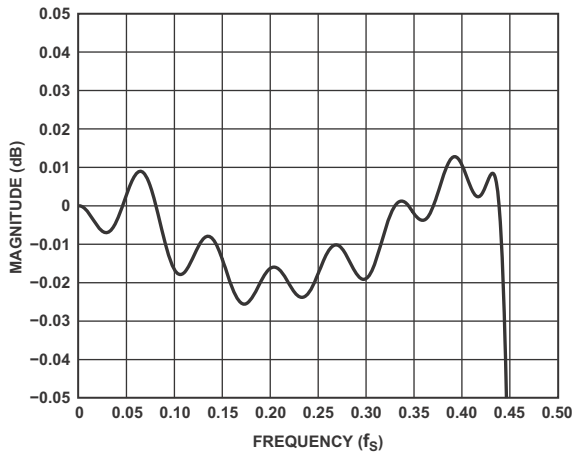


Figure 26. DAC Digital Filter Ripple, Sampling Rate = 48 kHz

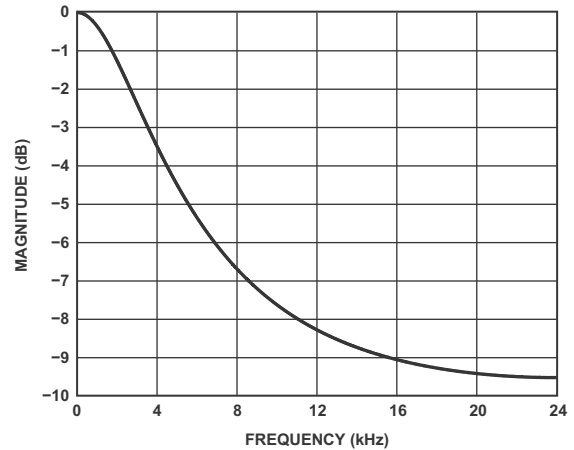


Figure 29. De-Emphasis Frequency Response, Sampling Rate = 48 kHz

DIGITAL DE-EMPHASIS

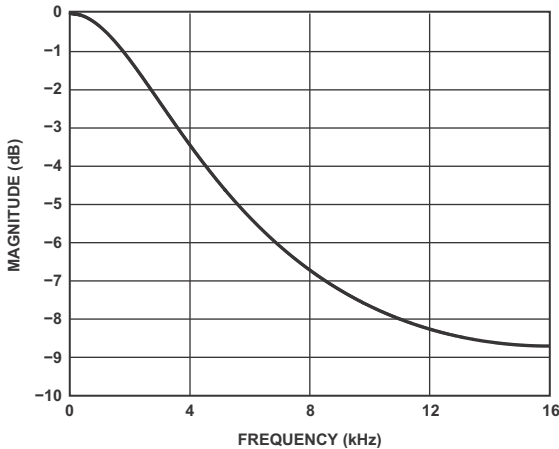


Figure 27. De-Emphasis Frequency Response, Sampling Rate = 32 kHz

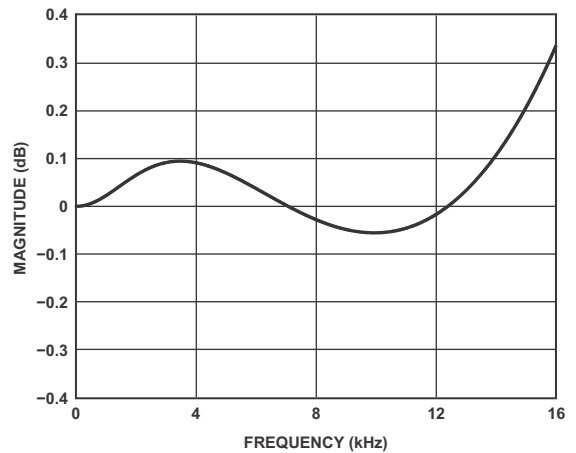


Figure 30. De-Emphasis Error, Sampling Rate = 32 kHz

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

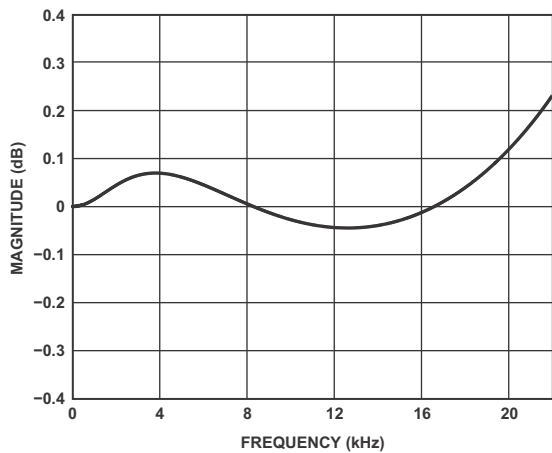


Figure 31. De-Emphasis Error, Sampling Rate = 44.1 kHz



Figure 32. De-Emphasis Error, Sampling Rate = 48 kHz

289-BALL CSP_BGA BALL ASSIGNMENT

Signals added or changed to the ADSP-BF52xC processor for the embedded codec are shown in Table 23 and Table 24. Please refer to the published ADSP-BF52x processor data sheet for descriptions of additional signals for the processor.

Table 23. 289-Ball CSP_BGA Ball Assignment (Alphabetically)

Signal	Ball No.	Signal	Ball No.
ADCDAT	A16	HPGND	G17
ADCLRC	A15	HPVDD	G16 ¹
AGND	H22	LHPOUT	B20
AVDD	J22 ¹	LLINEIN	E23
CMODE	E22	LOUT	F22
CODEC_BCLK	A19	MICBIAS	H23
CODEC_CLKOUT	D22	MICIN	J23
CSB	D23	RHPOUT	B21
CSCL	B23	RLINEIN	F23
CSDA	C23	ROUT	G22
CVDD	H17 ¹	VMID	G23
DACDAT	A18	XTI/CODEC_MCLK	A22
DACLRC	A17	XTO	A21

¹ For ADSP-BF52x processor (without internal codec) compatibility, connect this ball to V_{DDEXT}.

Table 24. 289-Ball CSP_BGA Ball Assignment (Numerically)

Ball No.	Signal	Ball No.	Signal
A15	ADCLRC	E22	CMODE
A16	ADCDAT	E23	LLINEIN
A17	DACLRC	F22	LOUT
A18	DACDAT	F23	RLINEIN
A19	CODEC_BCLK	G16 ¹	HPVDD
A21	XTO	G17	HPGND
A22	XTI/CODEC_MCLK	G22	ROUT
B20	LHPOUT	G23	VMID
B21	RHPOUT	H17 ¹	CVDD
B23	CSCL	H22	AGND
C23	CSDA	H23	MICBIAS
D22	CODEC_CLKOUT	J22 ¹	AVDD
D23	CSB	J23	MICIN

¹ For ADSP-BF52x processor (without internal codec) compatibility, connect this ball to V_{DDEXT}.

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

Figure 33 shows the top view of the ADSP-BF52xC processor ball configuration.



KEY:

- ⊖ V_{DDINT} ● GND
- ⊕ V_{DDEXT} ○ I/O ⊕ V_{DDMEM}

BALLS THAT HAVE CHANGED USAGE ON THE ADSP-BF522C/523C/524C/525C/526C/527C:

- CODEC I/O
- ⊕ AVDD ⊗ HPVDD
- ⊕ AGND ⊗ HPGND

Figure 33. ADSP-BF52xC Processor Ball Configuration (Top View)

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

Figure 34 shows the bottom view of the ADSP-BF52xC processor ball configuration.



KEY:

- ⊖ V_{DDINT} ● GND
- ⊕ V_{DDEXT} ○ I/O ⊕ V_{DDMEM}

BALLS THAT HAVE CHANGED USAGE ON THE ADSP-BF522C/523C/524C/525C/526C/527C:

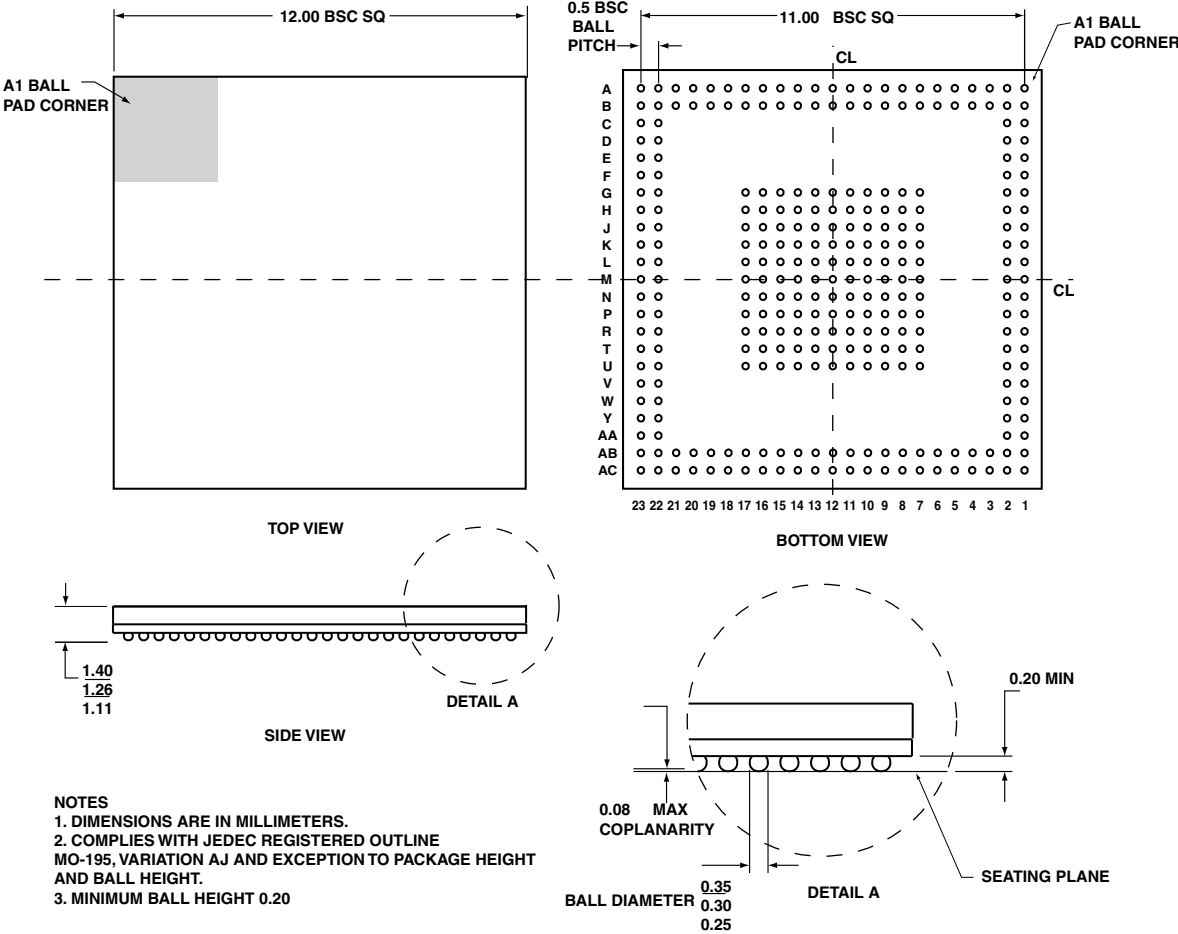
- CODEC I/O
- ⊕ AVDD ⊗ HPVDD
- ⊕ AGND ⊗ HPGND

Figure 34. ADSP-BF52xC Processor Ball Configuration (Bottom View)

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

OUTLINE DIMENSIONS

Dimensions in Figure 35, 289-Ball CSP_BGA (BC-289-2) are shown in millimeters.



- NOTES**
1. DIMENSIONS ARE IN MILLIMETERS.
 2. COMPLIES WITH JEDEC REGISTERED OUTLINE MO-195, VARIATION AJ AND EXCEPTION TO PACKAGE HEIGHT AND BALL HEIGHT.
 3. MINIMUM BALL HEIGHT 0.20

Figure 35. 289-Ball CSP_BGA (BC-289-2)

ADSP-BF522C/ADSP-BF523C/ADSP-BF524C/ADSP-BF525C/ADSP-BF526C/ADSP-BF527C

ORDERING GUIDE

Model ¹	Temperature Range ²	Instruction Rate (Max)	Package Description	Package Option
ADSP-BF522KBCZ-3C2	0°C to +70°C	300 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF522KBCZ-4C2	0°C to +70°C	400 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF523KBCZ-5C2	0°C to +70°C	533 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF523KBCZ-6C2	0°C to +70°C	600 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF524KBCZ-3C2	0°C to +70°C	300 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF524KBCZ-4C2	0°C to +70°C	400 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF525KBCZ-5C2	0°C to +70°C	533 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF525KBCZ-6C2	0°C to +70°C	600 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF526KBCZ-3C2	0°C to +70°C	300 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF526KBCZ-4C2	0°C to +70°C	400 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF527KBCZ-5C2	0°C to +70°C	533 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2
ADSP-BF527KBCZ-6C2	0°C to +70°C	600 MHz	289-Ball Chip Scale Package Ball Grid Array (CSP_BGA)	BC-289-2

¹Z = RoHS Compliant Part.

²Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see [Operating Conditions on Page 21](#) for junction temperature (T_j) specification which is the only temperature specification.



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- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
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- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



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