

4A, 20ns, MOSFET Driver

General Description

The MAX5078A/MAX5078B high-speed MOSFET drivers source and sink up to 4A peak current. These devices feature a fast 20ns propagation delay and 20ns rise and fall times while driving a 5000pF capacitive load. Propagation delay time is minimized and matched between the inverting and noninverting inputs. High sourcing/sinking peak currents, low propagation delay, and thermally enhanced packages make the MAX5078A/MAX5078B ideal for high-frequency and high-power circuits.

The MAX5078A/MAX5078B operate from a 4V to 15V single power supply and consume 40µA (typ) of supply current when not switching. These devices have an internal logic circuitry that prevents shoot-through during output state changes to minimize the operating current at a high switching frequency. The logic inputs are protected against voltage spikes up to +18V, regardless of the VDD voltage. The MAX5078A has CMOS input logic levels while the MAX5078B has TTL-compatible input logic levels.

The MAX5078A/MAX5078B feature both inverting and noninverting inputs for greater flexibility in controlling the MOSFET. They are available in a 6-pin TDFN (3mm x 3mm) package and operate over the automotive temperature range of -40°C to +125°C.

Applications

Power MOSFET Switching Switch-Mode Power Supplies Motor Control

Power-Supply Modules

DC-DC Converters

Features

- ♦ 4V to 15V Single Power Supply
- **♦** 4A Peak Source/Sink Drive Current
- ♦ 20ns (typ) Propagation Delay
- Matching Delay Between Inverting and Noninverting Inputs
- ♦ V_{DD} / 2 CMOS (MAX5078A)/TTL (MAX5078B) Logic Inputs
- ♦ 0.1 x V_{DD} (CMOS) and 0.3V (TTL) Logic-Input Hysteresis
- Up to +18V Logic Inputs (Regardless of V_{DD} Voltage)
- **♦** Low Input Capacitance: 2.5pF (typ)
- ♦ 40µA (typ) Quiescent Current
- **♦** -40°C to +125°C Operating Temperature Range
- ♦ 6-Pin TDFN Package

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX5078AATT+	-40°C to +125°C	6 TDFN-EP*
MAX5078AATT/V+	-40°C to +125°C	6 TDFN-EP*
MAX5078BATT+	-40°C to +125°C	6 TDFN-EP*
MAX5078BATT/V+	-40°C to +125°C	6 TDFN-EP*

^{*}EP = Exposed pad.

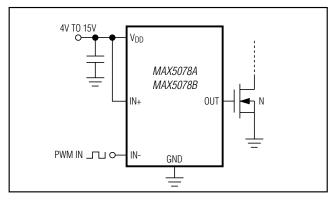
/V denotes an automotive qualified part.

Note: Devices are also available in a tape-and-reel package. Specify tape and reel by adding "T" to the part number when ordering.

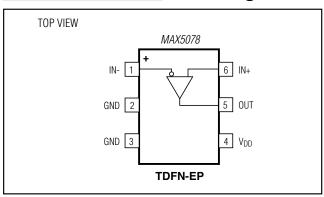
Selector Guide

PART	PIN-PACKAGE	LOGIC INPUT
MAX5078AATT	6 TDFN-EP	V _{DD} / 2 CMOS
MAX5078BATT	6 TDFN-EP	TTL

Typical Operating Circuit



Pin Configuration



For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

⁺Denotes a lead(Pb)-free/RoHS-compliant package.

^{*}EP = Exposed pad.

4A, 20ns, MOSFET Driver

ABSOLUTE MAXIMUM RATINGS

(Voltages referenced to GND.)		Continuous Power Dissipation (T _A = +70°C))
V _{DD}	0.3V to +18V	6-Pin TDFN-EP (derate 18.2mW/°C above	e +70°C)1454mW
IN+, IN	0.3V to +18V	Operating Temperature Range	40°C to +125°C
OUT	$0.3V$ to $(V_{DD} + 0.3V)$	Storage Temperature Range	65°C to +150°C
OUT Short-Circuit Duration	10ms	Junction Temperature	+150°C
Continuous Source/Sink Current at OUT_	_(P _D < P _{DMAX})200mA	Lead Temperature (soldering, 10s)	+300°C
		Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS (Note 1)

Junction-to-Ambient Thermal Resistance (θ_{JA})	42°C/W
Junction-to-Case Thermal Resistance (θ _{JC})	9°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = 4V \text{ to } 15V, T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}, \text{ unless otherwise noted.}$ Typical values are at $V_{DD} = 15V \text{ and } T_A = +25^{\circ}\text{C}.)$ (Note 1)

PARAMETER	SYMBOL	CONDIT	IONS	MIN	TYP	MAX	UNITS
POWER SUPPLY				•			
V _{DD} Operating Range	V_{DD}			4		15	V
V _{DD} Undervoltage Lockout	UVLO	V _{DD} rising		3.00	3.5	3.85	V
V _{DD} Undervoltage Lockout Hysteresis					200		mV
V _{DD} Undervoltage Lockout to Output Delay		V _{DD} rising			12		μs
		$V_{IN+} = 0V$, $IN- = V_{DD}$	$V_{DD} = 4V$		28	55	μΑ
V _{DD} Supply Current	IDD	(not switching)	$V_{DD} = 15V$		40	75	
	I _{DD-SW}	Switching at 250kHz, C	L = 0	0.5	1.2	2.2	mA
DRIVER OUTPUT (SINK)							
	R _{ON-N}	V _{DD} = 15V, I _{OUT} = -100mA	T _A = +25°C		1.1	1.8	Ω
Driver Output Resistance Pulling			$T_A = +125^{\circ}C$		1.5	2.4	
Down		$V_{DD} = 4.5V,$ $I_{OUT} = -100$ mA	$T_A = +25^{\circ}C$		2.2	3.3	
			$T_A = +125^{\circ}C$		3.0	4.5	
Peak Output Current (Sinking)	I _{PK-N}	$V_{DD} = 15V, C_L = 10,000$	V _{DD} = 15V, C _L = 10,000pF		4		А
Output-Voltage Low		HOUT = -100mA	$V_{DD} = 4.5V$			0.45	V
Output-voltage Low			$V_{DD} = 15V$			0.24	
Latchup Protection	ILUP	Reverse current I _{OUT} (N	lote 2)	400			mA
DRIVER OUTPUT (SOURCE)	1			. N			•
		V _{DD} = 15V,	T _A = +25°C		1.5	2.1	
Driver Output Resistance Pulling	D	$I_{OUT} = 100mA$	$T_A = +125^{\circ}C$		1.9	2.75] _
Up	R _{ON-P}	V _{DD} = 4.5V,	T _A = +25°C		2.75	4	Ω
		I _{OUT} = 100mA	T _A = +125°C		3.75	5.5	
Peak Output Current (Sourcing)	I _{PK-P}	$V_{DD} = 15V, C_L = 10,000$)pF		4		А

4A, 20ns, MOSFET Driver

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 4V \text{ to } 15V, T_A = -40^{\circ}\text{C to } + 125^{\circ}\text{C}, \text{ unless otherwise noted.}$ Typical values are at $V_{DD} = 15V \text{ and } T_A = +25^{\circ}\text{C}.)$ (Note 1)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS	
Output Valtage High		Louis 100mA	V _{DD} = 4.5V	V _{DD} - 0.55			\/	
Output-Voltage High		I _{OUT} = 100mA	V _{DD} = 15V	V _{DD} - 0.275			\ \ \ \	
LOGIC INPUT (Note 3)								
Logic 1 Input Voltage	VIH	MAX5078A		0.7 x V _{DD}			V	
		MAX5078B (Note 4)		2.1				
Logic 0 Input Voltage	V _{IL}	MAX5078A				0.3 x V _{DD}	V	
		MAX5078B				0.8		
Logic-Input Hysteresis	V _{HYS}	MAX5078A			0.1 x V _{DD}		V	
		MAX5078B			0.3			
Logic-Input-Current Leakage		$V_{IN+} = V_{IN-} = 0V \text{ or } V_I$	DD	-1	+0.1	+1	μΑ	
Input Capacitance	CIN				2.5		рF	
SWITCHING CHARACTERISTIC	S FOR V _{DD} = 1			1			1	
		C _L = 1000pF			4		_	
OUT Rise Time	tR	C _L = 5000pF			18		ns	
		C _L = 10,000pF			32			
		C _L = 1000pF			4		-	
OUT Fall Time	tF	C _L = 5000pF			15		ns	
Turn-On Delay Time	to ou	$C_L = 10,000pF$ $C_L = 10,000pF$ (Note	2)	10	26 20	34	ns	
Turn-Off Delay Time	tD-ON tD-OFF	$C_L = 10,000pF (Note)$		10	20	34	ns	
SWITCHING CHARACTERISTIC				10	20	0-1	110	
		$C_L = 1000pF$			7			
OUT Rise Time	t _R	C _L = 5000pF			37		ns	
		$C_L = 10,000pF$			85		1	
		C _L = 1000pF			7			
OUT Fall Time	t _F	C _L = 5000pF			30		ns	
		$C_L = 10,000pF$			75			
Turn-On Delay Time	t _{D-ON}	C _L = 10,000pF (Note	2)	18	35	70	ns	
Turn-Off Delay Time	t _{D-OFF}	C _L = 10,000pF (Note	2)	18	35	70	ns	

4A, 20ns, MOSFET Driver

ELECTRICAL CHARACTERISTICS (continued)

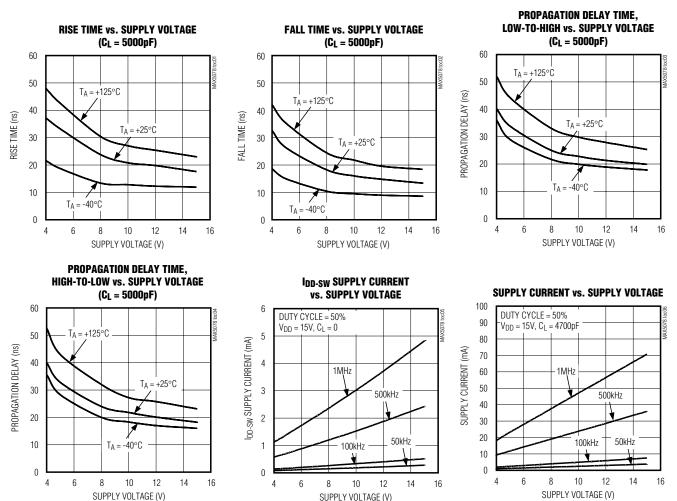
 $(V_{DD} = 4V \text{ to } 15V, T_A = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}, \text{ unless otherwise noted.}$ Typical values are at $V_{DD} = 15V \text{ and } T_A = +25^{\circ}\text{C.}$) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
MATCHING CHARACTERISTICS						
Mismatch Propagation Delays from	At	V _{DD} = 15V, C _L = 10,000pF		2		
Output	verting and Noninverting Inputs to	V _{DD} = 4.5V, C _L = 10,000pF		4		ns

- Note 1: All devices are 100% tested at $T_A = +25$ °C. Specifications over -40°C to +125°C are guaranteed by design.
- Note 2: Limits are guaranteed by design, not production tested.
- **Note 3:** The logic-input thresholds are tested at $V_{DD} = 4V$ and $V_{DD} = 15V$.
- Note 4: TTL compatible with reduced noise immunity.

Typical Operating Characteristics

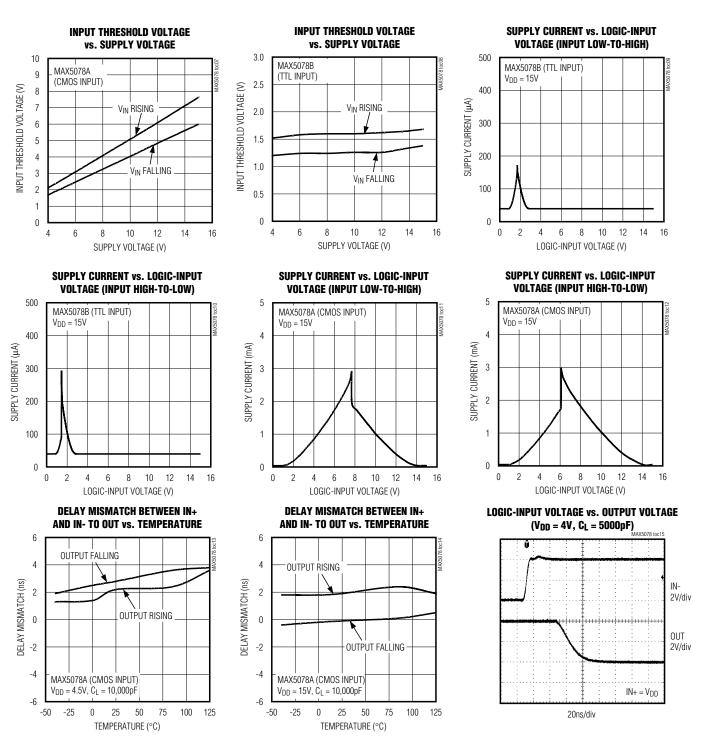
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



4A, 20ns, MOSFET Driver

Typical Operating Characteristics (continued)

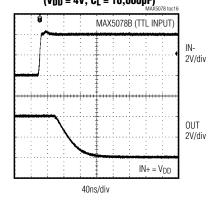
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



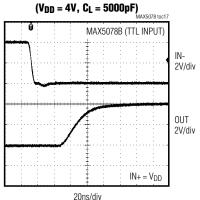
4A, 20ns, MOSFET Driver

Typical Operating Characteristics (continued)

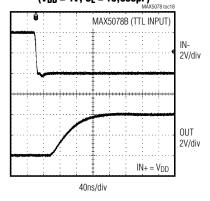
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$



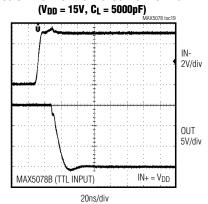
LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE



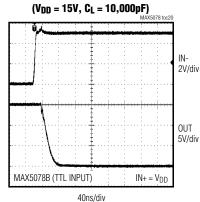
LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE $(V_{DD} = 4V, C_L = 10,000pF)$



LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE



LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE

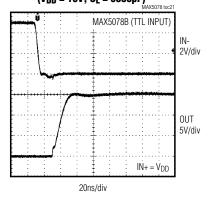


MAX5078 4A, 20ns, MOSFET Driver

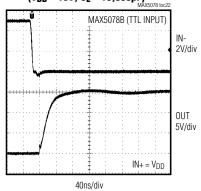
Typical Operating Characteristics (continued)

 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

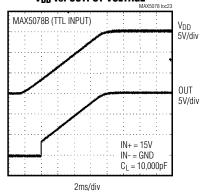
$\begin{array}{c} \text{LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE} \\ \text{(V}_{DD} = 15\text{V, C}_{L} = 5000\text{pF)} \end{array}$



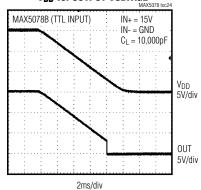
LOGIC-INPUT VOLTAGE vs. OUTPUT VOLTAGE $(V_{DD} = 15V, C_L = 10,000pF)$



VDD vs. OUTPUT VOLTAGE



VDD vs. OUTPUT VOLTAGE



4A, 20ns, MOSFET Driver

Pin Description

PIN	NAME	FUNCTION	
1	IN-	Inverting Logic-Input Terminal. Connect to GND when not used.	
2, 3 GND Ground			
4 V _{DD} Power Supply. Bypass to GND with one		Power Supply. Bypass to GND with one or more 0.1µF ceramic capacitors.	
5 OUT Driver Output. Sources or sinks current to turn		Driver Output. Sources or sinks current to turn the external MOSFET on or off.	
6 IN+ Noninverting Logic-Input Terminal		Noninverting Logic-Input Terminal. Connect to VDD when not used.	
_	EP	Exposed Pad. Internally connected to GND. Do not use the exposed pad as the only electrical ground connection.	

Detailed Description

V_{DD} Undervoltage Lockout (UVLO)

The MAX5078A/MAX5078B have internal undervoltage lockout (UVLO) for VDD. When VDD is below the UVLO threshold, OUT is pulled low independent of the state of the inputs. The undervoltage lockout is typically 3.5V with 200mV typical hysteresis to avoid chattering. When VDD rises above the UVLO threshold, the output goes high or low depending upon the logic-input levels. Bypass VDD using a low-ESR ceramic capacitor for proper operation (see the *Applications Information* section).

Logic Inputs

The MAX5078A has CMOS logic inputs while the MAX5078B has TTL-compatible logic inputs. The logic inputs are protected against the voltage spikes up to 18V, regardless of the V_{DD} voltage. The TTL and CMOS logic inputs have 300mV and 0.1 x V_{DD} hysteresis, respectively, to avoid double pulsing during transition. The low 2.5pF input capacitance reduces loading and increases switching speed.

The logic inputs are high impedance and must not be left floating. If the inputs are left open, OUT can go to an undefined state as soon as V_{DD} rises above the UVLO threshold. Therefore, the PWM output from the controller must assume proper state when powering up the device.

The MAX5078A/MAX5078B have two logic inputs, providing greater flexibility in controlling the MOSFET. Use IN+ for noninverting logic and IN- for inverting logic operation. Connect IN+ to V_{DD} and IN- to GND, if not used. Alternatively, the unused input can be used as an ON/OFF function. Use IN+ for active-low shutdown logic and IN- for active-high shutdown logic (see Figure 3). See Table 1 for all possible input combinations.

Driver Output

The MAX5078A/MAX5078B have low RDS(ON) p-channel and n-channel devices (totem pole) in the output stage for the fast turn-on/turn-off, high-gate-charge switching MOSFETs. The peak source or sink current is typically 4A. The output voltage (VOUT) is approximately equal to VDD when in high state and is ground when in low state. The driver RDS(ON) is lower at higher VDD resulting in higher source-/sink-current capability and faster switching speeds. The propagation delays from the noninverting and inverting logic inputs to OUT are matched to 2ns typically. The break-before-make logic avoids any cross-conduction between the internal p- and n-channel devices, and eliminates shoot-through, thus reducing the quiescent supply current.

Applications Information

RLC Series Circuit

The driver's RDS(ON) (RON), internal bond/lead inductance (LP), trace inductance (LS), gate inductance (LG), and gate capacitance (CG) form a series RLC circuit with a second-order characteristic equation. The series RLC circuit has an undamped natural frequency (ϖ_0) and a damping ratio (ζ) where:

$$\varpi_0 = \frac{1}{\sqrt{(L_P + L_S + L_G) \times C_G}}$$

$$\xi = \frac{R_{ON}}{2 \times \sqrt{\frac{(L_P + L_S + L_G)}{C_G}}}$$

The damping ratio needs to be greater than 0.5 (ideally 1) to avoid ringing. Add a small resistor (R_{GATE}) in series with the gate when driving a very low gate-charge MOSFET, or when the driver is placed away from the MOSFET.

4A, 20ns, MOSFET Driver

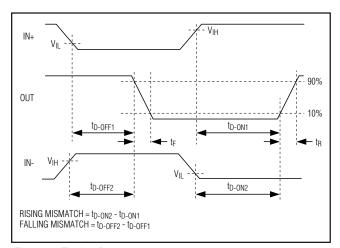


Figure 1. Timing Diagram

Use the following equation to calculate the series resistor:

$$R_{GATE} \ge \sqrt{\frac{(L_P + L_S + L_G)}{C_G}} - R_{ON}$$

Lp can be approximated as 2nH for the TDFN package. Ls is on the order of 20nH/in. Verify Lg with the MOSFET vendor.

Supply Bypassing and Grounding

Pay extra attention to bypassing and grounding the MAX5078A/MAX5078B. Peak supply and output currents may exceed 4A when driving large external capacitive loads. Supply voltage drops and ground shifts create negative feedback for inverters and may degrade the delay and transition times. Ground shifts due to poor device grounding may also disturb other circuits sharing the same AC ground return path. Any series inductance in the VDD, OUT, and/or GND paths can cause oscillations due to the very high di/dt when switching the MAX5078A/MAX5078B with any capacitive load. Place one or more 0.1µF ceramic capacitors in parallel as close to the device as possible to bypass Vnn to GND. Use a ground plane to minimize ground return resistance and series inductance. Place the external MOSFET as close as possible to the MAX5078A/MAX5078B to further minimize board inductance and AC path impedance.

Power Dissipation

Power dissipation of the MAX5078A/MAX5078B consists of three components: caused by the quiescent current, capacitive charge/discharge of internal nodes, and the output current (either capacitive or resistive load). Maintain the sum of these components below the maximum power dissipation limit.

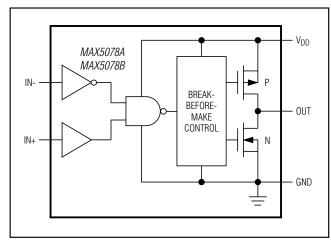


Figure 2. MAX5054 Simplified Diagram (1 Driver)

The current required to charge and discharge the internal nodes is frequency dependent (see the IDD-SW Supply Current vs. Supply Voltage graph in the *Typical Operating Characteristics*). The power dissipation (PQ) due to the quiescent switching supply current (IDD-SW) can be calculated as:

$$PQ = VDD \times IDD-SW$$

For capacitive loads, use the following equation to estimate the power dissipation:

$$PCLOAD = CLOAD \times (VDD)^2 \times fSW$$

where C_{LOAD} is the capacitive load, V_{DD} is the supply voltage, and fsw is the switching frequency.

Calculate the total power dissipation (PT) as follows:

$$PT = PQ + PCLOAD$$

Use the following equations to estimate the MAX5078A/MA5078B total power dissipation when driving a ground-referenced resistive load:

$$PRLOAD = D \times RON(MAX) \times ILOAD^2$$

where D is the fraction of the period the MAX5078A/ MA5078B's output pulls high, $R_{ON(MAX)}$ is the maximum on-resistance of the device with the output high, and I_{LOAD} is the output load current of the MAX5078A/ MAX5078B.

Layout Information

The MAX5078A/MAX5078B MOSFET drivers source and sink large currents to create very fast rising and falling edges at the gate of the switching MOSFET. The high di/dt can cause unacceptable ringing if the trace lengths and impedances are not well controlled.

4A, 20ns, MOSFET Driver

Table 1. MAX5078 Truth Table

IN+	IN-	OUT
Low	Low	Low
Low	High	Low
High	Low	High
High	High	Low

Use the following PC board layout guidelines when designing with the MAX5078A/MAX5078B:

- Place one or more 0.1µF decoupling ceramic capacitors from V_{DD} to GND as close to the device as possible. Connect V_{DD} and GND to large copper areas. Place one bulk capacitor of 10µF (min) on the PC board with a low resistance path to the V_{DD} input and GND of the MAX5078A/MAX5078B.
- Two AC current loops form between the device and the gate of the driven MOSFET. The MOSFET looks like a large capacitance from gate to source when the gate pulls low. The active current loop is from the MOSFET gate to OUT of the MAX5078A/MAX5078B, to GND of the MAX5078A/MAX5078B, and to the source of the MOSFET. When the gate of the MOSFET pulls high, the active current is from the VDD terminal of the decoupling capacitor, to VDD of the MAX5078A/MAX5078B, to OUT of the MAX5078A/MAX5078B, to the MOSFET gate, to the MOSFET source, and to the negative terminal of the decoupling capacitor. Both charging current and discharging current loops are important. Minimize the physical distance and the impedance in these AC current paths.
- Keep the device as close to the MOSFET as possible.
- In a multilayer PC board, the inner layers should consist of a GND plane containing the discharging and charging current loops.
- Pay extra attention to the ground loop and use a low-impedance source when using a TTL logicinput device. Fast fall time at OUT may corrupt the input during transition.

Exposed Pad

The TDFN-EP package has an exposed pad on the bottom of its package. This pad is internally connected to GND. For the best thermal conductivity, solder the exposed pad to the ground plane in order to dissipate 1.9W. Do not use the ground-connected pad as the only electrical ground connection or ground return. Use GND (pins 2 and 3) as the primary electrical ground connection.

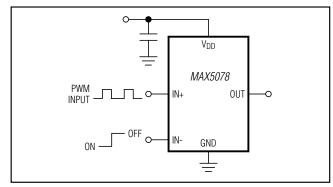


Figure 3. Unused Input as an ON/OFF Function

Additional Application Circuits

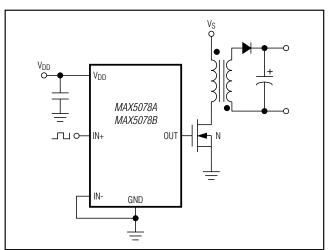


Figure 4. Noninverting Application

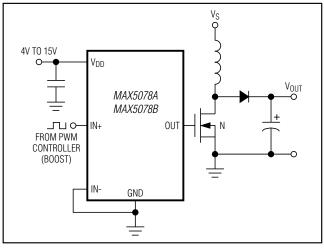


Figure 5. Boost Converter

4A, 20ns, MOSFET Driver

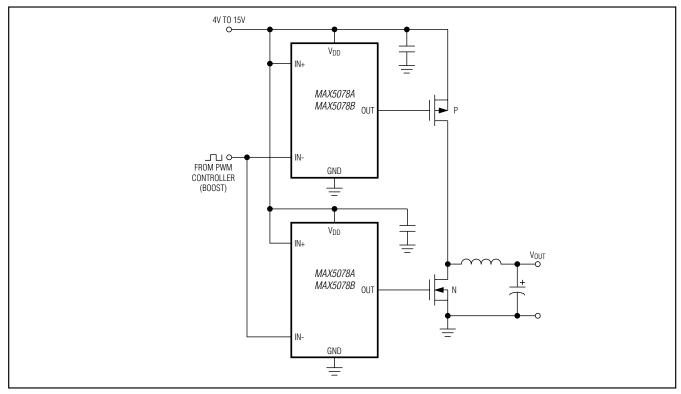


Figure 6. MAX5078A/MAX5078B In High-Power Synchronous Buck Converter

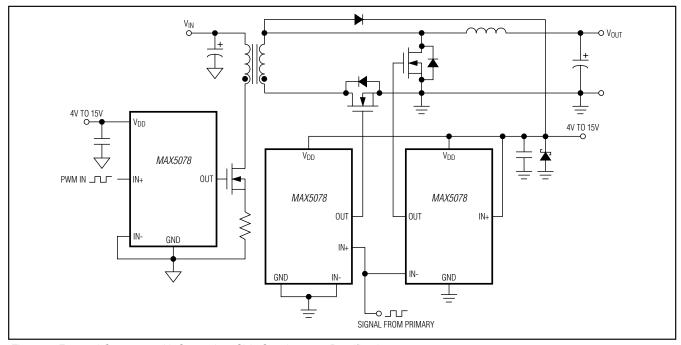


Figure 7. Forward Converter with Secondary-Side Synchronous Rectification

MAX5078 4A, 20ns, MOSFET Driver

PROCESS: CMOS

Chip Information

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE NO.	LAND
TYPE	CODE		PATTERN NO.
6 TDFN-EP	T633+2	21-0137	90-0058

MAX5078 4A, 20ns, MOSFET Driver

Revision History

REVISION	REVISION	DESCRIPTION	PAGES
NUMBER	DATE		CHANGED
2	9/12	Added automotive qualified parts to Ordering Information	1



Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов:
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001:
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)

Факс: 8 (812) 320-02-42

Электронная почта: <u>org@eplast1.ru</u>

Адрес: 198099, г. Санкт-Петербург, ул. Калинина,

дом 2, корпус 4, литера А.