

The PE42820 is a HaRP<sup>™</sup> technology-enhanced high power reflective SPDT RF switch designed for use in

This switch is a pin-compatible upgraded version of the

range, and external negative supply option. It maintains

exceptional linearity and power handling from 30 MHz

through 2.7 GHz. PE42820 also features low insertion

5x5 mm QFN package. In addition, no external blocking

UltraCMOS<sup>®</sup> process, a patented variation of silicon-oninsulator (SOI) technology on a sapphire substrate.

Peregrine's HaRP<sup>™</sup> technology enhancements deliver high linearity and excellent harmonics performance. It is an innovative feature of the UltraCMOS<sup>®</sup> process, offering

loss, high power handling, and is offered in a 32-lead

capacitors are required if 0V DC is present on the RF

The PE42820 is manufactured on Peregrine's

the performance of GaAs with the economy and

PE42510A with a wider frequency and power supply

mobile radio, relay replacement and other high

performance wireless applications.

**Product Description** 

ports.

# **Product Specification**

# PE42820

## UltraCMOS<sup>®</sup> SPDT RF Switch 30 - 2700 MHz

## Features

- High power handling
  - 45 dBm @ 850 MHz, 32W
  - 44 dBm @ 2 GHz, 25W
- Exceptional linearity
  - 85 dBm IIP3 @ 850 MHz
  - 81 dBm IIP3 @ 2.7 GHz
- Low insertion loss
  - 0.25 dB @ 850 MHz
  - 0.40 dB @ 2 GHz
- Wide supply range of 2.3V to 5.5V
- +1.8V control logic compatible
- ESD performance
  - 1.5kV HBM on all pins
- External negative supply option

#### Figure 2. Package Type 32-lead 5x5 mm QFN



## Figure 1. Functional Diagram

integration of conventional CMOS.



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## Table 1. Electrical Specifications @ 25°C ( $Z_S = Z_L = 50\Omega$ ) unless otherwise noted Normal mode<sup>1</sup>: $V_{DD} = 3.3V$ , $Vss_{EXT} = 0V$ or Bypass mode<sup>2</sup>: $V_{DD} = 3.3V$ , $Vss_{EXT} = -3.3V$

Parameter	Path	Condition	Min	Тур	Max	Unit
		30 MHz – 1 GHz		0.30	0.45	dB
Insertion loss <sup>3</sup>	RFC-RFX	1 GHz – 2 GHz		0.40	0.60	dB
		2 GHz – 2.7 GHz		0.70	0.95	dB
		30 MHz – 1 GHz	34	35		dB
Isolation	RFX-RFX	1 GHz – 2 GHz	27	28		dB
		2 GHz – 2.7 GHz	23	24		dB
Unbiased isolation	RFC-RFX	V <sub>DD</sub> , V1 = 0V, +27 dBm		6		dB
	RFX	30 MHz – 1 GHz		22		dB
Return loss <sup>3</sup>		1 GHz – 2 GHz		20		dB
		2 GHz – 2.7 GHz		14		dB
Harmonics	RFC-RFX	2fo: +45 dBm pulsed @ 1GHz, 50Ω		-94	-90	dBc
Harmonics		3fo: +45 dBm pulsed @ 1GHz, 50 $\Omega$		-84	-80	dBc
Input IP3	RFC-RFX	850 MHz 2700 MHz		85		dBm
				81		dBm
Input 0.1 dB compression point <sup>4</sup>	RFC-RFX	30 MHz – 2 GHz 2 GHz – 2.7 GHz		45.5 44.5		dBm dBm
Switching time		50% CTRL to 90% or 10% RF		15	25	μs
Settling time		50% CTRL to harmonics within specifications <sup>5</sup>		30	45	μs

Notes: 1. Normal mode: single external positive supply used

2. Bypass mode: both external positive supply and external negative supply used

3. Performance specified with external matching. Refer to Evaluation Kit section for additional information

4. The input 0.1dB compression point is a linearity figure of merit. Refer to *Table 3* for the operating RF input power (50Ω)

5. See harmonics specs above



#### Figure 3. Pin Configuration (Top View)



#### **Table 2. Pin Descriptions**

Pin #	Pin Name	Description
1, 3-11, 14, 15, 17-22, 24-27, 29-32	GND	Ground
2	RF1 <sup>1</sup>	RF port
12	V <sub>DD</sub>	Supply voltage (nominal 3.3V)
13	V1	Digital control logic input 1
16	Vss <sub>EXT</sub> <sup>2</sup>	External Vss negative voltage control
23	RF2 <sup>1</sup>	RF port
28	RFC <sup>1</sup>	RF common
Pad	GND	Exposed pad: ground for proper operation

1. RF pins 2, 23 and 28 must be at 0V DC. The RF pins do not require Notes: DC blocking capacitors for proper operation if the 0V DC requirement is met

2. Use  $Vss_{EXT}$  (pin 16,  $Vss_{EXT} = -V_{DD}$ ) to bypass and disable internal negative voltage generator. Connect  $Vss_{EXT}$  (pin 16,  $Vss_{EXT} = GND$ ) to enable internal negative voltage generator

#### **Table 3. Operating Ranges**

Parameter	Symbol	Min	Тур	Max	Unit
Normal mode <sup>1</sup>					
Supply voltage	V <sub>DD</sub>	2.3		5.5	V
Supply current	I <sub>DD</sub>		130	200	μA
Bypass mode <sup>2</sup>					
Supply voltage	V <sub>DD</sub>		3.3	5.5	V
Supply current	I <sub>DD</sub>		50	80	μA
Negative supply voltage	Vss <sub>ext</sub>	-3.6		-3.2	V
Negative supply current	I <sub>SS</sub>	-40	-16		μA
Normal or Bypass mode	9				
Digital input high (V1)	V <sub>IH</sub>	1.17		3.6 <sup>3</sup>	V
Digital input low (V1)	V <sub>IL</sub>	-0.3		0.6	V
RF input power, CW 30 MHz – 2 GHz >2 GHz – 2.7 GHz	P <sub>MAX,CW</sub>			43 42	dBm dBm
RF input power, pulsed <sup>4</sup> 30 MHz – 2 GHz >2 GHz – 2.7 GHz	P <sub>MAX,PULSED</sub>			45 44	dBm dBm
RF input power, unbiased	P <sub>MAX,UNB</sub>			27	dBm
Operating temperature range (Case)	T <sub>OP</sub>	-40		+85	°C
Operating junction temperature	TJ			+140	°C

Notes: 1. Normal mode: connect pin 16 to GND to enable internal negative voltage generator

2. Bypass mode: apply a negative voltage to Vss<sub>EXT</sub> (pin 16) to bypass and disable internal negative voltage generator 3. Maximum V<sub>IH</sub> voltage is limited to V<sub>DD</sub> and cannot exceed 3.6V

4. Pulsed, 10% duty cycle of 4620  $\mu s$  period,  $50\Omega$ 

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#### **Table 4. Absolute Maximum Ratings**

Parameter/Condition	Symbol	Min	Мах	Unit
Supply voltage	$V_{DD}$	-0.3	5.5	V
Digital input voltage (V1)	V <sub>CTRL</sub>	-0.3	3.6	V
Maximum input power 30 MHz – 2 GHz >2 GHz – 2.7 GHz	P <sub>MAX,ABS</sub>		45.5 44.5	dBm dBm
Storage temperature range	T <sub>ST</sub>	-65	+150	°C
Maximum case temperature	T <sub>CASE</sub>		+85	°C
Peak maximum junction temperature (10 seconds max)	$T_{\rm J}$		+200	°C
ESD voltage HBM <sup>1</sup> , all pins	V <sub>ESD</sub>		1500	V
ESD Voltage MM <sup>2</sup> , all pins	$V_{\text{ESD}}$		200	V

Notes: 1. Human Body Model (MIL-STD 883 Method 3015) 2. Machine Model (JEDEC JESD22-A115)

Exceeding absolute maximum ratings may cause permanent damage. Operation should be restricted to the limits in the Operating Ranges table. Operation between operating range maximum and absolute maximum for extended periods may reduce reliability.

## **Electrostatic Discharge (ESD) Precautions**

When handling this UltraCMOS<sup>®</sup> device, observe the same precautions that you would use with other ESD-sensitive devices. Although this device contains circuitry to protect it from damage due to ESD, precautions should be taken to avoid exceeding the rating specified.

## Latch-Up Avoidance

Unlike conventional CMOS devices, UltraCMOS<sup>®</sup> devices are immune to latch-up.

## **Moisture Sensitivity Level**

The Moisture Sensitivity Level rating for the 32lead 5x5 mm QFN package is MSL3.

#### Table 5. Control Logic Truth Table

Path	CTRL		
RFC – RF1	н		
RFC – RF2	L		

## **Optional External Vss Control (Vss<sub>EXT</sub>)**

For applications that require a faster switching rate or spur-free performance, this part can be operated in bypass mode. Bypass mode requires an external negative voltage in addition to an external V<sub>DD</sub> supply voltage.

As specified in *Table 3*, the external negative voltage ( $Vss_{EXT}$ ) when applied to pin 16 will disable and bypass the internal negative voltage generator.

## **Switching Frequency**

The PE42820 has a maximum 25 kHz switching rate in normal mode (pin 16 = GND). A faster switching rate is available in bypass mode (pin 16 =  $Vss_{EXT}$ ). The rate at which the PE42820 can be switched is then limited to the switching time as specified in *Table 1*.

Switching frequency describes the time duration between switching events. Switching time is the time duration between the point the control signal reaches 50% of the final value and the point the output signal reaches within 10% or 90% of its target value.

## **Spurious Performance**

The typical low-frequency spurious performance of the PE42820 in normal mode is -137 dBm (pin 16 = GND). If spur-free performance is desired, the internal negative voltage generator can be disabled by applying a negative voltage to Vss<sub>EXT</sub> (pin 16).

## Hot Switching Capability

The typical hot switching capability of the PE42820 is +30 dBm. Hot switching occurs when RF power is applied while switching between RF ports.

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## Typical Performance Data @ 25°C, $V_{DD}$ = 3.3V, $Vss_{EXT}$ = 0V unless otherwise noted

Figure 4. Insertion Loss vs. Temp (RFC-RFX)



## Figure 5. Insertion Loss vs. V<sub>DD</sub> (RFC–RFX)



## Figure 6. RFC Port Return Loss vs. Temp (RF1 Active)



Figure 7. RFC Port Return Loss vs. V<sub>DD</sub> (RF1 Active)



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## Typical Performance Data @ 25°C, $V_{DD}$ = 3.3V, $Vss_{EXT}$ = 0V unless otherwise noted

Figure 8. Active Port Return Loss vs. Temp (RF1 Active)



Figure 9. Active Port Return Loss vs. V<sub>DD</sub> (RF1 Active)



Figure 10. Isolation vs. Temp (RFC–RFX, RFX Active)



Figure 11. Isolation vs. V<sub>DD</sub> (RFC–RFX, RFX Active)



Figure 12. Isolation vs. Temp



# Typical Performance Data @ 25°C, V<sub>DD</sub> = 3.3V, Vss<sub>EXT</sub> = 0V unless otherwise noted



## Figure 13. Isolation vs. V<sub>DD</sub> (RFX–RFX, RFX Active)





## Thermal Data

Though the insertion loss for this part is very low, when handling high power RF signals, the junction temperature rises significantly.

VSWR conditions that present short circuit loads to the part can cause significantly more power dissipation than with proper matching.

Special consideration needs to be made in the design of the PCB to properly dissipate the heat away from the part and maintain the 85°C maximum case temperature. It is recommended to use best design practices for high power QFN packages: multi-layer PCBs with thermal vias in a thermal pad soldered to the slug of the package. Special care also needs to be made to alleviate solder voiding under the part.

## Table 6. Theta JC

Parameter	Min	Тур	Max	Unit
Theta JC (+85°C)		20		°C/W



## **Evaluation Kit**

The PE42820 Evaluation Kit board was designed to ease customer evaluation of the PE42820 RF switch.

The evaluation board in *Figure 14* was designed to test the part. DC power is supplied through J10, with VDD on pin 9, and GND on the entire lower row of even numbered pins. To evaluate a switch path, add or remove jumpers on V1 (pin 3) using *Table 5*.

The ANT port is connected through a  $50\Omega$  transmission line via the top SMA connector, J1. RF1 and RF2 paths are also connected through  $50\Omega$  transmission lines via SMA connectors as J2 and J3. A  $50\Omega$  through transmission line is available via SMA connectors J5 and J6. This transmission line can be used to estimate the loss of the PCB over the environmental conditions being evaluated. An open-ended  $50\Omega$  transmission line is also provided at J4 for calibration if needed.

Narrow trace widths are used near each part to improve impedance matching. The shunt C1 on RFC port is to provide for high frequency impedance matching.

#### Figure 14. Evaluation Board Layout



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## Figure 15. Evaluation Board Schematic





## Figure 16. Package Drawing



## Figure 17. Top Marking Specification





#### Figure 18. Tape and Reel Specs



#### Table 7. Ordering Information

Order Code	Description	Package	Shipping Method
PE42820MLBA-X	PE42820 SPDT RF switch	Green 32-lead 5x5 mm QFN	500 units/T&R
PE42820MLBA-Z	PE42820 SPDT RF switch	Green 32-lead 5x5 mm QFN	3000 units/T&R
EK42820-02	PE42820 Evaluation kit	Evaluation kit	1/Box

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