# 2.35 V to 5.25 V, 1 MSPS, 12-/10-/8-Bit ADCs in 6-Lead SC70

# AD7476A/AD7477A/AD7478A

## FEATURES

- Fast throughput rate: 1 MSPS
- Specified for $V_{DD}$ of 2.35 V to 5.25 V
- Low power
  - 3.6 mW at 1 MSPS with 3 V supplies
  - 12.5 mW at 1 MSPS with 5 V supplies
- Wide input bandwidth
  - 71 dB SNR at 100 kHz input frequency
- Flexible power/serial clock speed management
- No pipeline delays
- High speed serial interface
  - SPI®/QSPI™/MICROWIRE™/DSP compatible
- Standby mode: 1 µA maximum
- 6-lead SC70 package
- 8-lead MSOP package
- Qualified for automotive applications

## APPLICATIONS

- Battery-powered systems
  - Personal digital assistants
  - Medical instruments
  - Mobile communications
- Instrumentation and control systems
- Data acquisition systems
- High speed modems
- Optical sensors

## FUNCTIONAL BLOCK DIAGRAM

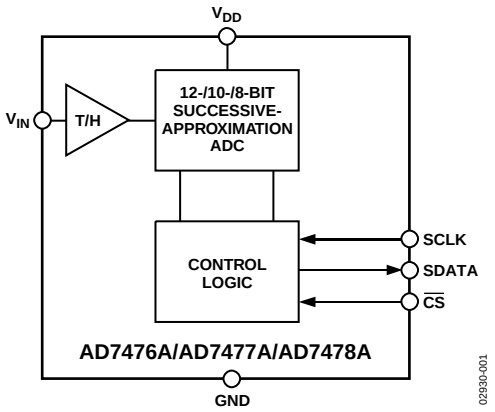



*Figure 1.*

## GENERAL DESCRIPTION

The AD7476A/AD7477A/AD7478A are 12-bit, 10-bit, and 8-bit high speed, low power, successive-approximation analog-to-digital converters (ADCs), respectively. The parts operate from a single 2.35 V to 5.25 V power supply and feature throughput rates up to 1 MSPS. The parts contain a low noise, wide bandwidth track-and-hold amplifier that can handle input frequencies in excess of 13 MHz. The conversion process and data acquisition are controlled using $\overline{CS}$ and the serial clock, allowing the devices to interface with microprocessors or DSPs. The input signal is sampled on the falling edge of $\overline{CS}$, and the conversion is also initiated at this point. There are no pipeline delays associated with the parts. The AD7476A/AD7477A/AD7478A use advanced design techniques to achieve low power dissipation at high throughput rates. The reference for the part is taken internally from $V_{DD}$ to allow the widest dynamic input range to the ADC. Thus, the analog input range for the part is 0 V to $V_{DD}$. The conversion rate is determined by the SCLK.

## PRODUCT HIGHLIGHTS

1. First 12-/10-/8-bit ADCs in a SC70 package.
2. High throughput with low power consumption.
3. Flexible power/serial clock speed management. The conversion rate is determined by the serial clock, allowing the conversion time to be reduced through the serial clock speed increase. This allows the average power consumption to be reduced when a power-down mode is used while not converting. The parts also feature a power-down mode to maximize power efficiency at lower throughput rates. Current consumption is 1 µA maximum and 50 nA typically when in power-down mode.
4. Reference derived from the power supply.
5. No pipeline delay. The parts feature a standard successive approximation ADC with accurate control of the sampling instant via a $\overline{CS}$ input and once-off conversion control.

**Rev. F**

Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties that may result from its use. Specifications subject to change without notice. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices. Trademarks and registered trademarks are the property of their respective owners.

One Technology Way, P.O. Box 9106, Norwood, MA 02062-9106, U.S.A.
Tel: 781.329.4700 www.analog.com
Fax: 781.461.3113 ©2002–2011 Analog Devices, Inc. All rights reserved.

# TABLE OF CONTENTS

Features ... 1
Applications ... 1
Functional Block Diagram ... 1
General Description ... 1
Product Highlights ... 1
Revision History ... 2
Specifications ... 3
AD7476A Specifications ... 3
AD7477A Specifications ... 5
AD7478A Specifications ... 6
Timing Specifications ... 8
Absolute Maximum Ratings ... 10
ESD Caution ... 10
Pin Configurations and Function Descriptions ... 11
Typical Performance Characteristics ... 12
Terminology ... 14
Theory of Operation ... 15
Circuit Information ... 15
The Converter Operation ... 15
ADC Transfer Function ... 15
Typical Connection Diagram ... 16
Analog Input ... 16
Digital Inputs ... 17
Modes of Operation ... 18
Normal Mode ... 18
Power-Down Mode ... 18
Power-Up Time ... 18
Power vs. Throughput Rate ... 20
Serial Interface ... 21
AD7478A in a 12 SCLK Cycle Serial Interface ... 22
Microprocessor Interfacing ... 23
AD7476A/AD7477A/AD7478A to TMS320C541 Interface 23
AD7476A/AD7477A/AD7478A to ADSP-218x Interface ... 23
AD7476A/AD7477A/AD7478A to DSP563xx Interface ... 24
Application Hints ... 25
Grounding and Layout ... 25
Evaluating the AD7476A/AD7477A Performance ... 25
Outline Dimensions ... 26
Ordering Guide ... 26
Automotive Products ... 27

## REVISION HISTORY

**1/11—Rev. E to Rev. F**

Changes to Features Section ... 1
Changes to Ordering Guide ... 26
Added Automotive Products Section ... 27

**2/09—Rev. D to Rev. E**

Changes to Features ... 1
Changes to Ordering Guide ... 26

**4/06—Rev. C to Rev. D**

Updated Format ... Universal
Changes to Ordering Guide ... 26

# SPECIFICATIONS

## AD7476A SPECIFICATIONS

$V_{DD}$ = 2.35 V to 5.25 V, $f_{SCLK}$ = 20 MHz, $f_{SAMPLE}$ = 1 MSPS, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted.[1]

**Table 1.**

| Parameter | A Grade[2] | B Grade[2] | Y Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|---|---|
| DYNAMIC PERFORMANCE | | | | | $f_{IN}$ = 100 kHz sine wave |
| Signal-to-Noise + Distortion (SINAD)[3] | 70 | 70 | 70 | dB min | $V_{DD}$ = 2.35 V to 3.6 V, $T_A$ = 25°C |
| | 69 | 69 | 69 | dB min | $V_{DD}$ = 2.4 V to 3.6 V |
| | 71.5 | 71.5 | 71.5 | dB typ | $V_{DD}$ = 2.35 V to 3.6 V |
| | 69 | 69 | 69 | dB min | $V_{DD}$ = 4.75 V to 5.25 V, $T_A$ = 25°C |
| | 68 | 68 | 68 | dB min | $V_{DD}$ = 4.75 V to 5.25 V |
| Signal-to-Noise Ratio (SNR)[3] | 71 | 71 | 71 | dB min | $V_{DD}$ = 2.35 V to 3.6 V, $T_A$ = 25°C |
| | 70 | 70 | 70 | dB min | $V_{DD}$ = 2.4 V to 3.6 V |
| | 70 | 70 | 70 | dB min | $V_{DD}$ = 4.75 V to 5.25 V, $T_A$ = 25°C |
| | 69 | 69 | 69 | dB min | $V_{DD}$ = 4.75 V to 5.25 V |
| Total Harmonic Distortion (THD)[3] | –80 | –80 | –80 | dB typ | |
| Peak Harmonic or Spurious Noise (SFDR)[3] | –82 | –82 | –82 | dB typ | |
| Intermodulation Distortion (IMD)[3] | | | | | |
| Second-Order Terms | –84 | –84 | –84 | dB typ | fa = 100.73 kHz, fb = 90.72 kHz |
| Third-Order Terms | –84 | –84 | –84 | dB typ | fa = 100.73 kHz, fb = 90.72 kHz |
| Aperture Delay | 10 | 10 | 10 | ns typ | |
| Aperture Jitter | 30 | 30 | 30 | ps typ | |
| Full Power Bandwidth | 13.5 | 13.5 | 13.5 | MHz typ | @ 3 dB |
| | 2 | 2 | 2 | MHz typ | @ 0.1 dB |
| DC ACCURACY | | | | | B and Y grades[4] |
| Resolution | 12 | 12 | 12 | Bits | |
| Integral Nonlinearity[3] | | ±1.5 | ±1.5 | LSB max | |
| | ±0.75 | | | LSB typ | |
| Differential Nonlinearity | | –0.9/+1.5 | –0.9/+1.5 | LSB max | Guaranteed no missed codes to 12 bits |
| | ±0.75 | | | LSB typ | |
| Offset Error[3, 5] | | ±1.5 | ±1.5 | LSB max | |
| | ±1.5 | ±0.2 | ±0.2 | LSB typ | |
| Gain Error[3, 5] | | ±1.5 | ±1.5 | LSB max | |
| | ±1.5 | ±0.5 | ±0.5 | LSB typ | |
| Total Unadjusted Error (TUE)[3, 5] | | ±2 | ±2 | LSB max | |
| ANALOG INPUT | | | | | |
| Input Voltage Range | 0 to $V_{DD}$ | 0 to $V_{DD}$ | 0 to $V_{DD}$ | V | |
| DC Leakage Current | ±0.5 | ±0.5 | ±0.5 | µA max | |
| Input Capacitance | 20 | 20 | 20 | pF typ | Track-and-hold in track; 6 pF typ when in hold |
| LOGIC INPUTS | | | | | |
| Input High Voltage, $V_{INH}$ | 2.4 | 2.4 | 2.4 | V min | |
| | 1.8 | 1.8 | 1.8 | V min | $V_{DD}$ = 2.35 V |
| Input Low Voltage, $V_{INL}$ | 0.8 | 0.8 | 0.8 | V max | $V_{DD}$ = 5 V |
| | 0.4 | 0.4 | 0.4 | V max | $V_{DD}$ = 3 V |
| Input Current, $I_{IN}$, SCLK Pin | ±0.5 | ±0.5 | ±0.5 | µA max | Typically 10 nA, $V_{IN}$ = 0 V or $V_{DD}$ |
| Input Current, $I_{IN}$, $\overline{CS}$ Pin | ±10 | ±10 | ±10 | nA typ | |
| Input Capacitance, $C_{IN}$[6] | 5 | 5 | 5 | pF max | |

| Parameter | A Grade[2] | B Grade[2] | Y Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|---|---|
| LOGIC OUTPUTS | | | | | |
| Output High Voltage, $V_{OH}$ | $V_{DD}$ – 0.2 | $V_{DD}$ – 0.2 | $V_{DD}$ – 0.2 | V min | $I_{SOURCE}$ = 200 µA; $V_{DD}$ = 2.35 V to 5.25 V |
| Output Low Voltage, $V_{OL}$ | 0.4 | 0.4 | 0.4 | V max | $I_{SINK}$ = 200 µA |
| Floating-State Leakage Current | ±1 | ±1 | ±1 | µA max | |
| Floating-State Output Capacitance[6] | 5 | 5 | 5 | pF max | |
| Output Coding | Straight (Natural) Binary | | | | |
| CONVERSION RATE | | | | | |
| Conversion Time | 800 | 800 | 800 | ns max | 16 SCLK cycles |
| Track-and-Hold Acquisition Time[3] | 250 | 250 | 250 | ns max | |
| Throughput Rate | 1 | 1 | 1 | MSPS max | See Serial Interface section |
| POWER REQUIREMENTS | | | | | |
| $V_{DD}$ | 2.35/5.25 | 2.35/5.25 | 2.35/5.25 | V min/max | |
| $I_{DD}$ | | | | | Digital I/Ps = 0 V or $V_{DD}$ |
| Normal Mode (Static) | 2.5 | 2.5 | 2.5 | mA typ | $V_{DD}$ = 4.75 V to 5.25 V, SCLK on or off |
| | 1.2 | 1.2 | 1.2 | mA typ | $V_{DD}$ = 2.35 V to 3.6 V, SCLK on or off |
| Normal Mode (Operational) | 3.5 | 3.5 | 3.5 | mA max | $V_{DD}$ = 4.75 V to 5.25 V, $f_{SAMPLE}$ = 1 MSPS |
| | 1.7 | 1.7 | 1.7 | mA max | $V_{DD}$ = 2.35 V to 3.6 V, $f_{SAMPLE}$ = 1 MSPS |
| Full Power-Down Mode (Static) | 1 | 1 | 1 | µA max | Typically 50 nA |
| Full Power-Down Mode (Dynamic) | 0.6 | 0.6 | 0.6 | mA typ | $V_{DD}$ = 5 V, $f_{SAMPLE}$ = 100 kSPS |
| Power Dissipation[7] | 0.3 | 0.3 | 0.3 | mA typ | $V_{DD}$ = 3 V, $f_{SAMPLE}$ = 100 kSPS |
| Normal Mode (Operational) | 17.5 | 17.5 | 17.5 | mW max | $V_{DD}$ = 5 V, $f_{SAMPLE}$ = 1 MSPS |
| | 5.1 | 5.1 | 5.1 | mW max | $V_{DD}$ = 3 V, $f_{SAMPLE}$ = 1 MSPS |
| Full Power-Down Mode | 5 | 5 | 5 | µW max | $V_{DD}$ = 5 V |
| | 3 | 3 | 3 | µW max | $V_{DD}$ = 3 V |

[1] Temperature ranges are as follows: A, B grades from –40°C to +85°C, Y grade from –40°C to +125°C.
[2] Operational from $V_{DD}$ = 2.0 V, with input low voltage ($V_{INL}$) 0.35 V maximum.
[3] See the Terminology section.
[4] B and Y grades, maximum specifications apply as typical figures when $V_{DD}$ = 4.75 V to 5.25 V.
[5] SC70 values guaranteed by characterization.
[6] Guaranteed by characterization.
[7] See the Power vs. Throughput Rate section.

## AD7477A SPECIFICATIONS

$V_{DD}$ = 2.35 V to 5.25 V, $f_{SCLK}$ = 20 MHz, $f_{SAMPLE}$ = 1 MSPS, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted.[1]

**Table 2.**

| Parameter | A Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|
| DYNAMIC PERFORMANCE | | | $f_{IN}$ = 100 kHz sine wave |
| Signal-to-Noise + Distortion (SINAD)[3] | 61 | dB min | |
| Total Harmonic Distortion (THD)[3] | −72 | dB max | |
| Peak Harmonic or Spurious Noise (SFDR)[3] | −73 | dB max | |
| Intermodulation Distortion (IMD)[3] | | | |
| Second-Order Terms | −82 | dB typ | fa = 100.73 kHz, fb = 90.7 kHz |
| Third-Order Terms | −82 | dB typ | fa = 100.73 kHz, fb = 90.7 kHz |
| Aperture Delay | 10 | ns typ | |
| Aperture Jitter | 30 | ps typ | |
| Full Power Bandwidth | 13.5 | MHz typ | @ 3 dB |
| | 2 | MHz typ | @ 0.1 dB |
| DC ACCURACY | | | |
| Resolution | 10 | Bits | |
| Integral Nonlinearity | ±0.5 | LSB max | |
| Differential Nonlinearity | ±0.5 | LSB max | Guaranteed no missed codes to 10 bits |
| Offset Error[3, 4] | ±1 | LSB max | |
| Gain Error[3, 4] | ±1 | LSB max | |
| Total Unadjusted Error (TUE)[3, 4] | ±1.2 | LSB max | |
| ANALOG INPUT | | | |
| Input Voltage Range | 0 to $V_{DD}$ | V | |
| DC Leakage Current | ±0.5 | μA max | |
| Input Capacitance | 20 | pF typ | Track-and-hold in track; 6 pF typ when in hold |
| LOGIC INPUTS | | | |
| Input High Voltage, $V_{INH}$ | 2.4 | V min | |
| | 1.8 | V min | $V_{DD}$ = 2.35 V |
| Input Low Voltage, $V_{INL}$ | 0.8 | V max | $V_{DD}$ = 5 V |
| | 0.4 | V max | $V_{DD}$ = 3 V |
| Input Current, $I_{IN}$, SCLK Pin | ±0.5 | μA max | Typically 10 nA, $V_{IN}$ = 0 V or $V_{DD}$ |
| Input Current, $I_{IN}$, $\overline{CS}$ Pin | ±10 | nA typ | |
| Input Capacitance, $C_{IN}$[5] | 5 | pF max | |
| LOGIC OUTPUTS | | | |
| Output High Voltage $V_{OH}$ | $V_{DD}$ − 0.2 | V min | $I_{SOURCE}$ = 200 μA, $V_{DD}$ = 2.35 V to 5.25 V |
| Output Low Voltage, $V_{OL}$ | 0.4 | V max | $I_{SINK}$ = 200 μA |
| Floating-State Leakage Current | ±1 | μA max | |
| Floating-State Output Capacitance[5] | 5 | pF max | |
| Output Coding | Straight (Natural) Binary | | |
| CONVERSION RATE | | | |
| Conversion Time | 700 | ns max | 14 SCLK cycles with SCLK at 20 MHz |
| Track-and-Hold Acquisition Time[3] | 250 | ns max | |
| Throughput Rate | 1 | MSPS max | |

| Parameter | A Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|
| POWER REQUIREMENTS | | | |
| $V_{DD}$ | 2.35/5.25 | V min/max | |
| $I_{DD}$ | | | Digital I/Ps = 0 V or $V_{DD}$ |
| Normal Mode (Static) | 2.5 | mA typ | $V_{DD}$ = 4.75 V to 5.25 V, SCLK on or off |
| | 1.2 | mA typ | $V_{DD}$ = 2.35 V to 3.6 V, SCLK on or off |
| Normal Mode (Operational) | 3.5 | mA max | $V_{DD}$ = 4.75 V to 5.25 V, $f_{SAMPLE}$ = 1 MSPS |
| | 1.7 | mA max | $V_{DD}$ = 2.35 V to 3.6 V, $f_{SAMPLE}$ = 1 MSPS |
| Full Power-Down Mode (Static) | 1 | µA max | Typically 50 nA |
| Full Power-Down Mode (Dynamic) | 0.6 | mA typ | $V_{DD}$ = 5 V, $f_{SAMPLE}$ = 100 kSPS |
| Power Dissipation[6] | 0.3 | mA typ | $V_{DD}$ = 3 V, $f_{SAMPLE}$ = 100 kSPS |
| Normal Mode (Operational) | 17.5 | mW max | $V_{DD}$ = 5 V, $f_{SAMPLE}$ = 1 MSPS |
| | 5.1 | mW max | $V_{DD}$ = 3 V, $f_{SAMPLE}$ = 1 MSPS |
| Full Power-Down Mode | 5 | µW max | $V_{DD}$ = 5 V |

[1] Temperature range is from –40°C to +85°C.
[2] Operational from $V_{DD}$ = 2.0 V, with input high voltage ($V_{INH}$) 1.8 V minimum.
[3] See the Terminology section.
[4] SC70 values guaranteed by characterization.
[5] Guaranteed by characterization.
[6] See the Power vs. Throughput Rate section.

## AD7478A SPECIFICATIONS

$V_{DD}$ = 2.35 V to 5.25 V, $f_{SCLK}$ = 20 MHz, $f_{SAMPLE}$ = 1 MSPS, $T_A$ = $T_{MIN}$ to $T_{MAX}$, unless otherwise noted.[1]

**Table 3.**

| Parameter | A Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|
| DYNAMIC PERFORMANCE | | | $f_{IN}$ = 100 kHz sine wave |
| Signal-to-Noise + Distortion (SINAD)[3] | 49 | dB min | |
| Total Harmonic Distortion (THD)[3] | –65 | dB max | |
| Peak Harmonic or Spurious Noise (SFDR)[3] | –65 | dB max | |
| Intermodulation Distortion (IMD)[3] | | | |
| Second-Order Terms | –76 | dB typ | fa = 100.73 kHz, fb = 90.7 kHz |
| Third-Order Terms | –76 | dB typ | fa = 100.73 kHz, fb = 90.7 kHz |
| Aperture Delay | 10 | ns typ | |
| Aperture Jitter | 30 | ps typ | |
| Full Power Bandwidth | 13.5 | MHz typ | @ 3 dB |
| | 2 | MHz typ | @ 0.1 dB |
| DC ACCURACY | | | |
| Resolution | 8 | Bits | |
| Integral Nonlinearity[3] | ±0.3 | LSB max | |
| Differential Nonlinearity[3] | ±0.3 | LSB max | Guaranteed no missed codes to eight bits |
| Offset Error[3, 4] | ±0.3 | LSB max | |
| Gain Error[3, 4] | ±0.3 | LSB max | |
| Total Unadjusted Error (TUE)[3, 4] | ±0.5 | LSB max | |
| ANALOG INPUT | | | |
| Input Voltage Range | 0 to $V_{DD}$ | V | |
| DC Leakage Current | ±0.5 | µA max | |
| Input Capacitance | 20 | pF typ | Track-and-hold in track; 6 pF typ when in hold |

| Parameter | A Grade[2] | Unit | Test Conditions/Comments |
|---|---|---|---|
| LOGIC INPUTS | | | |
| Input High Voltage, $V_{INH}$ | 2.4 | V min | |
| | 1.8 | V min | $V_{DD}$ = 2.35 V |
| Input Low Voltage, $V_{INL}$ | 0.8 | V max | $V_{DD}$ = 5 V |
| | 0.4 | V max | $V_{DD}$ = 3 V |
| Input Current, $I_{IN}$, SCLK Pin | ±0.5 | µA max | Typically 10 nA, $V_{IN}$ = 0 V or $V_{DD}$ |
| Input Current, $I_{IN}$, $\overline{CS}$ Pin | ±10 | nA typ | |
| Input Capacitance, $C_{IN}$[5] | 5 | pF max | |
| LOGIC OUTPUTS | | | |
| Output High Voltage, $V_{OH}$ | $V_{DD}$ – 0.2 | V min | $I_{SOURCE}$ = 200 µA, $V_{DD}$ = 2.35 V to 5.25 V |
| Output Low Voltage, $V_{OL}$ | 0.4 | V max | $I_{SINK}$ = 200 µA |
| Floating-State Leakage Current | ±1 | µA max | |
| Floating-State Output Capacitance[5] | 5 | pF max | |
| Output Coding | Straight (Natural) Binary | | |
| CONVERSION RATE | | | |
| Conversion Time | 600 | ns max | 12 SCLK cycles with SCLK at 20 MHz |
| Track-and-Hold Acquisition Time[3] | 225 | ns max | |
| Throughput Rate | 1.2 | MSPS max | |
| POWER REQUIREMENTS | | | |
| $V_{DD}$ | 2.35/5.25 | V min/max | |
| $I_{DD}$ | | | Digital I/Ps = 0 V or $V_{DD}$ |
| Normal Mode (Static) | 2.5 | mA typ | $V_{DD}$ = 4.75 V to 5.25 V, SCLK on or off |
| | 1.2 | mA typ | $V_{DD}$ = 2.35 V to 3.6 V, SCLK on or off |
| Normal Mode (Operational) | 3.5 | mA max | $V_{DD}$ = 4.75 V to 5.25 V |
| | 1.7 | mA max | $V_{DD}$ = 2.35 V to 3.6 V |
| Full Power-Down Mode (Static) | 1 | µA max | Typically 50 nA |
| Full Power-Down Mode (Dynamic) | 0.6 | mA typ | $V_{DD}$ = 5 V, $f_{SAMPLE}$ = 100 kSPS |
| Power Dissipation[6] | 0.3 | mA typ | $V_{DD}$ = 3 V, $f_{SAMPLE}$ = 100 kSPS |
| Normal Mode (Operational) | 17.5 | mW max | $V_{DD}$ = 5 V |
| | 5.1 | mW max | $V_{DD}$ = 3 V |
| Full Power-Down Mode | 5 | µW max | $V_{DD}$ = 5 V |

[1] Temperature range is from –40°C to +85°C.
[2] Operational from $V_{DD}$ = 2.0 V, with input high voltage ($V_{INH}$) 1.8 V minimum.
[3] See the Terminology section.
[4] SC70 values guaranteed by characterization.
[5] Guaranteed by characterization.
[6] See the Power vs. Throughput Rate section.

## TIMING SPECIFICATIONS

$V_{DD}$ = 2.35 V to 5.25 V; $T_A = T_{MIN}$ to $T_{MAX}$, unless otherwise noted.[1]

**Table 4.**

| Parameter | Limit at $T_{MIN}$, $T_{MAX}$ | Unit | Description |
|---|---|---|---|
| $f_{SCLK}$[2] | 10 | kHz min[3] | A, B grades |
| | 20 | kHz min[3] | Y grade |
| | 20 | MHz max | |
| $t_{CONVERT}$ | 16 × $t_{SCLK}$ | | AD7476A |
| | 14 × $t_{SCLK}$ | | AD7477A |
| | 12 × $t_{SCLK}$ | | AD7478A |
| $t_{QUIET}$ | 50 | ns min | Minimum quiet time required between bus relinquish and start of next conversion |
| $t_1$ | 10 | ns min | Minimum $\overline{CS}$ pulse width |
| $t_2$ | 10 | ns min | $\overline{CS}$ to SCLK setup time |
| $t_3$[4] | 22 | ns max | Delay from $\overline{CS}$ until SDATA three-state disabled |
| $t_4$[4] | 40 | ns max | Data access time after SCLK falling edge |
| $t_5$ | 0.4 $t_{SCLK}$ | ns min | SCLK low pulse width |
| $t_6$ | 0.4 $t_{SCLK}$ | ns min | SCLK high pulse width |
| $t_7$[5] | | | SCLK to data valid hold time |
| | 10 | ns min | $V_{DD} \leq 3.3$ V |
| | 9.5 | ns min | 3.3 V < $V_{DD} \leq 3.6$ V |
| | 7 | ns min | $V_{DD} > 3.6$ V |
| $t_8$[6] | 36 | ns max | SCLK falling edge to SDATA high impedance |
| | $t_7$ values also apply to $t_8$ minimum values | ns min | SCLK falling edge to SDATA high impedance |
| $t_{POWER-UP}$[7] | 1 | µs max | Power-up time from full power-down |

[1] Guaranteed by characterization. All input signals are specified with tr = tf = 5 ns (10% to 90% of $V_{DD}$) and timed from a voltage level of 1.6 V.
[2] Mark/space ratio for the SCLK input is 40/60 to 60/40.
[3] Minimum $f_{SCLK}$ at which specifications are guaranteed.
[4] Measured with the load circuit shown in Figure 2, and defined as the time required for the output to cross 0.8 V or 1.8 V when $V_{DD}$ = 2.35 V, and 0.8 V or 2.0 V for $V_{DD}$ > 2.35 V.
[5] Measured with a 50 pF load capacitor.
[6] $t_8$ is derived from the measured time taken by the data outputs to change 0.5 V when loaded with the circuit shown in Figure 2. The measured number is then extrapolated back to remove the effects of charging or discharging the 50 pF capacitor. Therefore, the time, $t_8$, quoted in the timing characteristics is the true bus relinquish time of the part and is independent of the bus loading.
[7] See the Power-Up Time section.

### *Timing Diagrams*

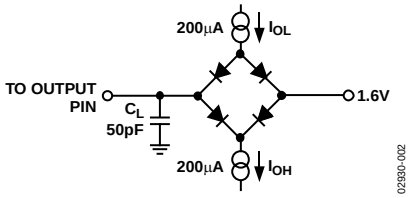



*Figure 2. Load Circuit for Digital Output Timing Specifications*

**Timing Example 1**

Having $f_{SCLK}$ = 20 MHz and a throughput of 1 MSPS, a cycle time of

$$t_2 + 12.5\ (1/f_{SCLK}) + t_{ACQ} = 1\ \mu s$$

where:

$t_2$ = 10 ns min, leaving $t_{ACQ}$ to be 365 ns. This 365 ns satisfies the requirement of 250 ns for $t_{ACQ}$.

From Figure 4, $t_{ACQ}$ is comprised of

$$2.5\ (1/f_{SCLK}) + t_8 + t_{QUIET}$$

where:

$t_8$ = 36 ns maximum. This allows a value of 204 ns for $t_{QUIET}$, satisfying the minimum requirement of 50 ns.

**Timing Example 2**

Having $f_{SCLK}$ = 5 MHz and a throughput is 315 kSPS yields a cycle time of

$$t_2 + 12.5\ (1/f_{SCLK}) + t_{ACQ} = 3.174\ \mu s$$

where:

$t_2$ = 10 ns min, this leaves $t_{ACQ}$ to be 664 ns. This 664 ns satisfies the requirement of 250 ns for $t_{ACQ}$.

From Figure 4, $t_{ACQ}$ is comprised of

$$2.5\ (1/f_{SCLK}) + t_8 + t_{QUIET},\ t_8 = 36\ ns\ maximum$$

This allows a value of 128 ns for $t_{QUIET}$, satisfying the minimum requirement of 50 ns.

In this example and with other, slower clock values, the signal may already be acquired before the conversion is complete, but it is still necessary to leave 50 ns minimum $t_{QUIET}$ between conversions. In Example 2, acquire the signal fully at approximately Point C in Figure 4.

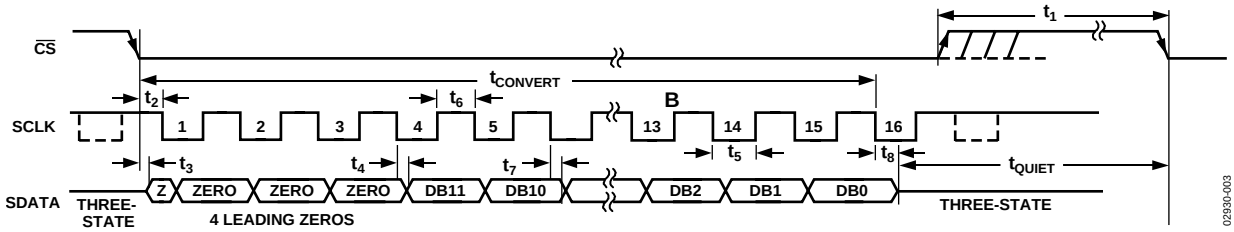



*Figure 3. AD7476A Serial Interface Timing Diagram*

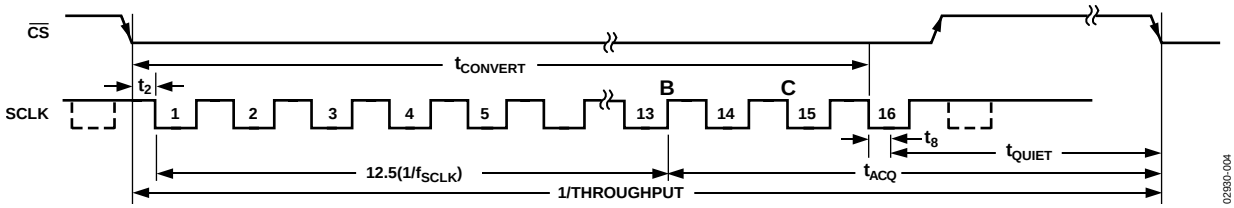



*Figure 4. Serial Interface Timing Example*

# ABSOLUTE MAXIMUM RATINGS

$T_A$ = 25°C, unless otherwise noted.[1]

**Table 5.**

| Parameter | Ratings |
|---|---|
| $V_{DD}$ to GND | −0.3 V to +7 V |
| Analog Input Voltage to GND | −0.3 V to $V_{DD}$ + 0.3 V |
| Digital Input Voltage to GND | −0.3 V to +7 V |
| Digital Output Voltage to GND | −0.3 V to $V_{DD}$ + 0.3 V |
| Input Current to Any Pin Except Supplies | 10 mA |
| Operating Temperature Range | |
| Commercial (A and B Grades) | −40°C to +85°C |
| Industrial (Y Grade) | −40°C to +125°C |
| Storage Temperature Range | −65°C to +150°C |
| Junction Temperature | 150°C |
| MSOP Package | |
| $\theta_{JA}$ Thermal Impedance | 205.9°C/W |
| $\theta_{JC}$ Thermal Impedance | 43.74°C/W |
| SC70 Package | |
| $\theta_{JA}$ Thermal Impedance | 340.2°C/W |
| $\theta_{JC}$ Thermal Impedance | 228.9°C/W |
| Lead Temperature, Soldering | |
| Reflow (10 sec to 30 sec) | 235 (0/+5)°C |
| Pb-Free Temperature Soldering | |
| Reflow | 255 (0/+5)°C |
| ESD | 3.5 kV |

[1] Transient currents of up to 100 mA do not cause SCR latch-up.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

# PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

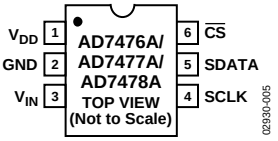



*Figure 5. 6-Lead SC70 Pin Configuration*

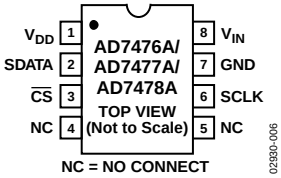



*Figure 6. 8-Lead MSOP Pin Configuration*

**Table 6. Pin Function Descriptions**

| Mnemonic | Description |
|---|---|
| $\overline{CS}$ | Chip Select. Active low logic input. This input provides the dual function of initiating conversions on the AD7476A/AD7477A/AD7478A and also frames the serial data transfer. |
| $V_{DD}$ | Power Supply Input. The $V_{DD}$ range for AD7476A/AD7477A/AD7478A is from 2.35 V to 5.25 V. |
| GND | Analog Ground. Ground reference point for all circuitry on AD7476A/AD7477A/AD7478A. Refer all analog input signals to this GND voltage. |
| $V_{IN}$ | Analog Input. Single-ended analog input channel. The input range is 0 V to $V_{DD}$. |
| SDATA | Data Out. Logic output. The conversion result from AD7476A/AD7477A/AD7478A is provided on this output as a serial data stream. The bits are clocked out on the falling edge of the SCLK input. The data stream from the AD7476A consists of four leading zeros followed by 12 bits of conversion data that are provided MSB first. The data stream from the AD7477A consists of four leading zeros followed by 10 bits of conversion data followed by two trailing zeros, provided MSB first. The data stream from the AD7478A consists of four leading zeros followed by 8 bits of conversion data followed by four trailing zeros that are provided MSB first. |
| SCLK | Serial Clock. Logic input. SCLK provides the serial clock for accessing data from the part. This clock input is also used as the clock source for the conversion process of AD7476A/AD7477A/AD7478A. |
| NC | No Connect. |

# TYPICAL PERFORMANCE CHARACTERISTICS

Figure 7, Figure 8, and Figure 9 each show a typical FFT plot for the AD7476A, AD7477A, and AD7478A, respectively, at a 1 MSPS sample rate and 100 kHz input frequency. Figure 10 shows the signal-to-(noise + distortion) ratio performance vs. the input frequency for various supply voltages while sampling at 1 MSPS with an SCLK frequency of 20 MHz for the AD7476A.

Figure 11 and Figure 12 show INL and DNL performance for the AD7476A. Figure 13 shows a graph of the total harmonic distortion vs. the analog input frequency for different source impedances when using a supply voltage of 3.6 V and sampling at a rate of 1 MSPS (see the Analog Input section). Figure 14 shows a graph of the total harmonic distortion vs. the analog input signal frequency for various supply voltages while sampling at 1 MSPS with an SCLK frequency of 20 MHz.

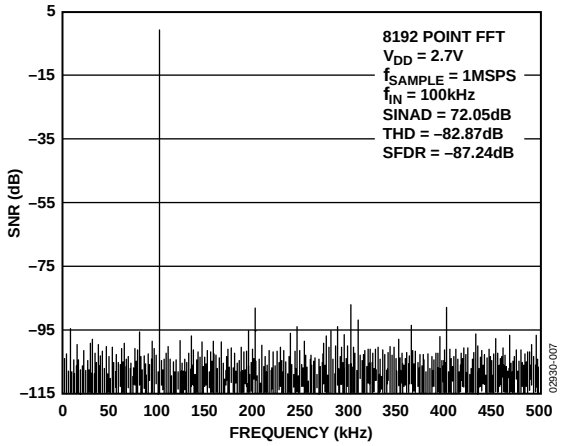



*Figure 7. AD7476A Dynamic Performance at 1 MSPS*

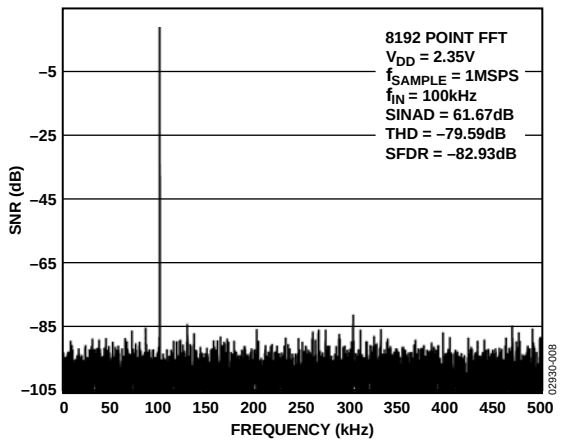



*Figure 8. AD7477A Dynamic Performance at 1 MSPS*

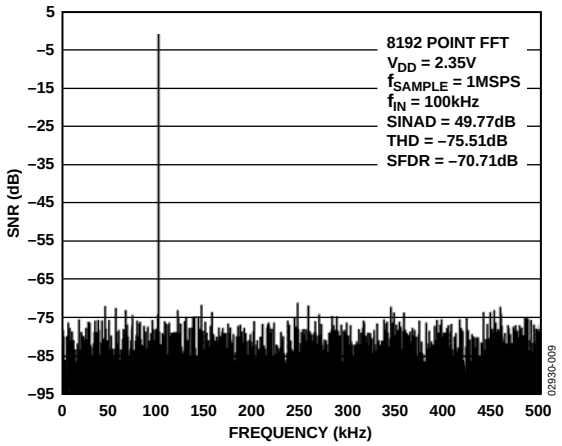



*Figure 9. AD7478A Dynamic Performance at 1 MSPS*

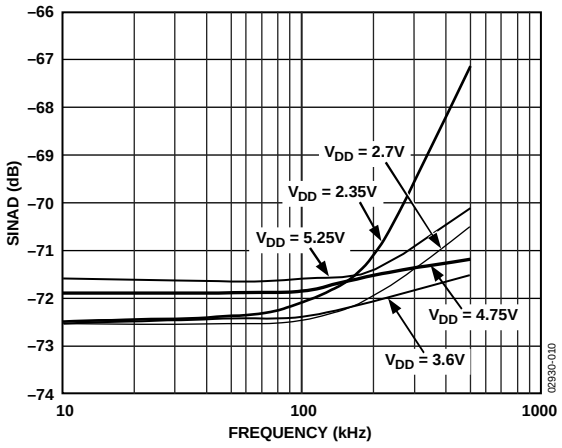



*Figure 10. AD7476A SINAD vs. Input Frequency at 1 MSPS*

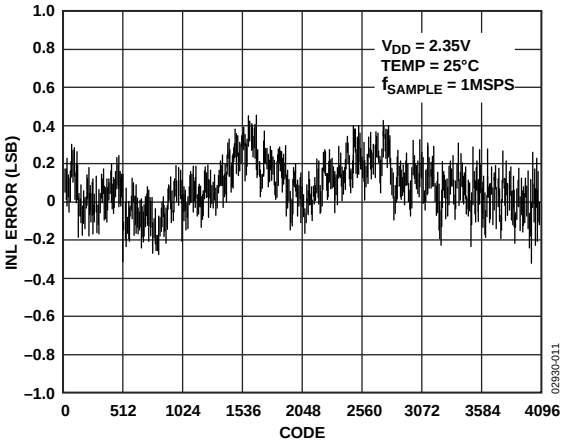



*Figure 11. AD7476A INL Performance*

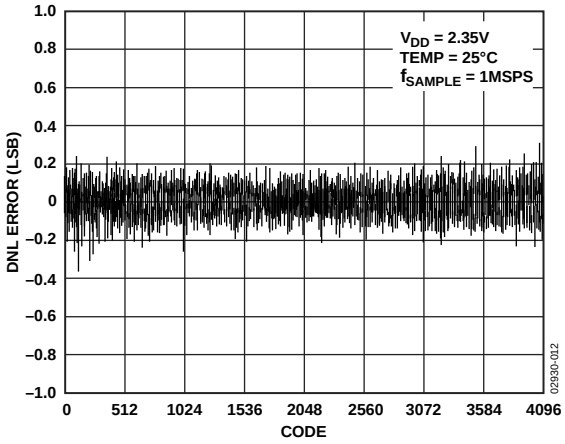



*Figure 12. AD7476A DNL Performance*

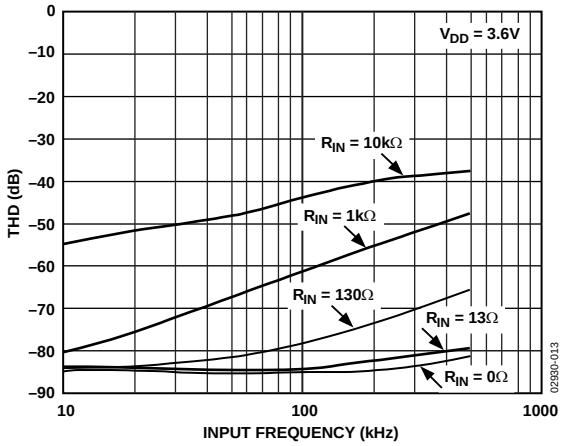



*Figure 13. THD vs. Analog Input Frequency for Various Source Impedances*

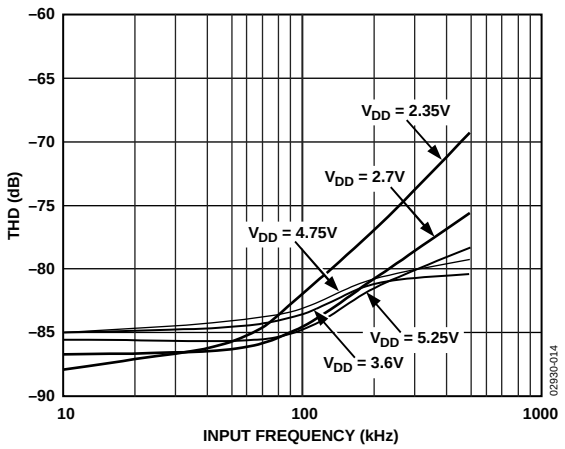



*Figure 14. THD vs. Analog Input Frequency for Various Supply Voltages*

# TERMINOLOGY

**Integral Nonlinearity (INL)**
INL is the maximum deviation from a straight line passing through the endpoints of the ADC transfer function. For the AD7476A/AD7477A/AD7478A, the endpoints of the transfer function are zero scale (1 LSB below the first code transition), and full scale (1 LSB above the last code transition).

**Differential Nonlinearity (DNL)**
DNL is the difference between the measured and the ideal 1 LSB change between any two adjacent codes in the ADC.

**Offset Error**
This is the deviation of the first code transition (00 . . . 000) to (00 . . . 001) from the ideal, that is, AGND + 1 LSB.

**Gain Error**
This is the deviation of the last code transition (111 . . . 110) to (111 . . . 111) from the ideal, that is, $V_{REF}$ – 1 LSB after the offset error has been adjusted out.

**Track-and-Hold Acquisition Time**
The track-and-hold amplifier returns to track mode at the end of a conversion. The track-and-hold acquisition time is the time required for the output of the track-and-hold amplifier to reach its final value, within 0.5 LSB, after the end of conversion. See the Serial Interface section for more details.

**Signal-to-(Noise + Distortion) Ratio (SINAD)**
This is the measured ratio of signal-to-(noise + distortion) at the output of the ADC. The signal is the rms amplitude of the fundamental. Noise is the sum of all nonfundamental signals up to half the sampling frequency ($f_S/2$), excluding dc. The ratio is dependent on the number of quantization levels in the digitization process; the more levels, the smaller the quantization noise. The theoretical signal-to-(noise + distortion) ratio for an ideal N-bit converter with a sine wave input is given by signal-to-(noise + distortion) = (6.02 N + 1.76) dB. Thus, it is 74 dB for a 12-bit converter, 62 dB for a 10-bit converter, and 50 dB for an 8-bit converter.

**Total Unadjusted Error (TUE)**
This is a comprehensive specification that includes the gain, linearity, and offset errors.

**Total Harmonic Distortion (THD)**
Total harmonic distortion is the ratio of the rms sum of harmonics to the fundamental. It is defined as

$$THD(\text{dB}) = 20\log\frac{\sqrt{V_2^2 + V_3^2 + V_4^2 + V_5^2 + V_6^2}}{V_1}$$

where $V_1$ is the rms amplitude of the fundamental, and $V_2$, $V_3$, $V_4$, $V_5$, and $V_6$ are the rms amplitudes of the second through the sixth harmonics.

**Peak Harmonic or Spurious Noise (SFDR)**
Peak harmonic or spurious noise is defined as the ratio of the rms value of the next largest component in the ADC output spectrum (up to $f_S/2$ and excluding dc) to the rms value of the fundamental. Normally, the value of this specification is determined by the largest harmonic in the spectrum. For ADCs where the harmonics are buried in the noise floor, it is a noise peak.

**Intermodulation Distortion (IMD)**
With inputs consisting of sine waves at two frequencies, fa and fb, any active device with nonlinearities create distortion products at sum and difference frequencies of mfa, nfb, where m and n = 0, 1, 2, 3, and so on. Intermodulation distortion terms are those for which neither m nor n are equal to zero. For example, the second-order terms include (fa + fb) and (fa – fb), and the third-order terms include (2fa + fb), (2fa – fb), (fa + 2fb), and (fa – 2fb).

The AD7476A/AD7477A/AD7478A are tested using the CCIF standard where two input frequencies are used (see fa and fb in the Specifications section). In this case, the second-order terms are usually distanced in frequency from the original sine waves, while the third-order terms are usually at a frequency close to the input frequencies. As a result, the second- and third-order terms are specified separately. The calculation of the intermodulation distortion is per the THD specification, where it is the ratio of the rms sum of the individual distortion products to the rms amplitude of the sum of the fundamentals expressed in decibels.

# THEORY OF OPERATION

## CIRCUIT INFORMATION

The AD7476A/AD7477A/AD7478A are fast, micropower, 12-/10-/8-bit, single-supply analog-to-digital converters (ADCs), respectively. The parts can be operated from a 2.35 V to 5.25 V supply. When operated from either a 5 V supply or a 3 V supply, the AD7476A/AD7477A/AD7478A are capable of throughput rates of 1 MSPS when provided with a 20 MHz clock. The AD7476A/AD7477A/AD7478A provide the user with an on-chip, track-and-hold ADC and a serial interface housed in a tiny 6-lead SC70 or 8-lead MSOP package, offering the user considerable space-saving advantages over alternative solutions. The serial clock input accesses data from the part but also provides the clock source for the successive-approximation ADC. The analog input range is 0 V to $V_{DD}$. The ADC does not require an external reference or an on-chip reference. The reference for the AD7476A/AD7477A/AD7478A is derived from the power supply and, thus, gives the widest dynamic input range. The AD7476A/AD7477A/AD7478A also feature a power-down option to allow power saving between conversions. The power-down feature is implemented across the standard serial interface, as described in the Modes of Operation section.

## THE CONVERTER OPERATION

AD7476A/AD7477A/AD7478A are successive approximation, analog-to-digital converters based around a charge redistribution DAC. Figure 15 and Figure 16 show simplified schematics of the ADC. Figure 15 shows the ADC during its acquisition phase. SW2 is closed and SW1 is in Position A, the comparator is held in a balanced condition, and the sampling capacitor acquires the signal on $V_{IN}$.

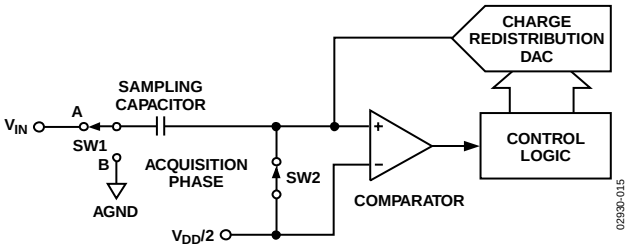



*Figure 15. ADC Acquisition Phase*

When the ADC starts a conversion (see Figure 16), SW2 opens and SW1 moves to Position B, causing the comparator to become unbalanced. The control logic and the charge redistribution DAC are used to add and subtract fixed amounts of charge from the sampling capacitor to bring the comparator back into a balanced condition. When the comparator is rebalanced, the conversion is complete. The control logic generates the ADC output code. Figure 17 shows the ADC transfer function.

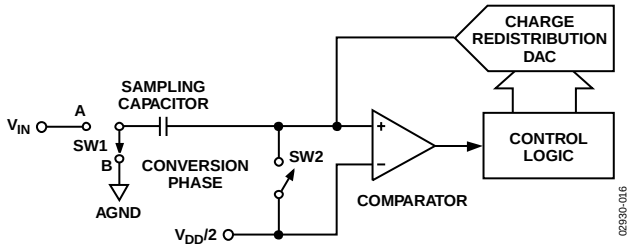



*Figure 16. ADC Conversion Phase*

## ADC TRANSFER FUNCTION

The output coding of the AD7476A/AD7477A/AD7478A is straight binary. The designed code transitions occur at the successive integer LSB values, that is, 1 LSB, 2 LSB, and so on. The LSB size is $V_{DD}/4096$ for the AD7476A, $V_{DD}/1024$ for the AD7477A, and $V_{DD}/256$ for the AD7478A. The ideal transfer characteristic for the AD7476A/AD7477A/AD7478A is shown in Figure 17.

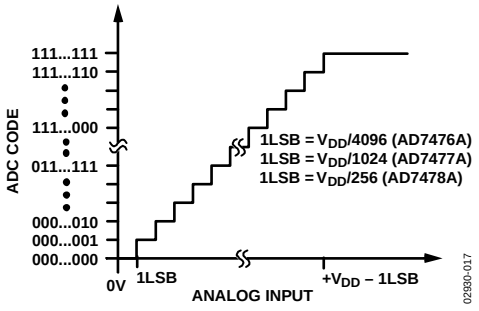



*Figure 17. AD7476A/AD7477A/AD7478A Transfer Characteristic*

# TYPICAL CONNECTION DIAGRAM

Figure 18 shows a typical connection diagram for the AD7476A/AD7477A/AD7478A. $V_{REF}$ is taken internally from $V_{DD}$ and, as such, $V_{DD}$ should be well decoupled. This provides an analog input range of 0 V to $V_{DD}$. The conversion result is output in a 16-bit word with four leading zeros followed by the MSB of the 12-bit, 10-bit, or 8-bit result. The 10-bit result from the AD7477A is followed by two trailing zeros, and the 8-bit result from the AD7478A is followed by four trailing zeros. Alternatively, because the supply current required by the AD7476A/AD7477A/AD7478A is so low, a precision reference can be used as the supply source to the AD7476A/AD7477A/AD7478A. A REF19x voltage reference (REF195 for 5 V or REF193 for 3 V) can be used to supply the required voltage to the ADC (see Figure 18). This configuration is especially useful if the power supply is quite noisy, or if the system supply voltages are at some value other than 5 V or 3 V (for example, 15 V).

The REF19x outputs a steady voltage to the AD7476A/AD7477A/AD7478A. If the low dropout REF193 is used, the current it needs to supply to the AD7476A/AD7477A/ AD7478A is typically 1.2 mA. When the ADC is converting at a rate of 1 MSPS, the REF193 needs to supply a maximum of 1.7 mA to the AD7476A/AD7477A/AD7478A. The load regulation of the REF193 is typically 10 ppm/mA ($V_S$ = 5 V), resulting in an error of 17 ppm (51 µV) for the 1.7 mA drawn from it. This corresponds to a 0.069 LSB error for the AD7476A with $V_{DD}$ = 3 V from the REF193, a 0.017 LSB error for the AD7477A, and a 0.0043 LSB error for the AD7478A.

For applications where power consumption is a concern, use the power-down mode of the ADC and the sleep mode of the REF19x reference to improve power performance. See the Modes of Operation section.

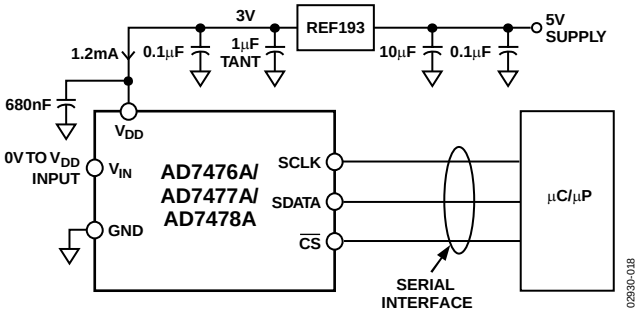



*Figure 18. REF193 as Power Supply to AD7476A/AD7477A/AD7478A*

Table 7 provides typical performance data with various references used as a $V_{DD}$ source for a 100 kHz input tone at room temperature under the same setup conditions.

**Table 7. AD7476A Typical Performance for Various Voltage References**

| Reference Tied to $V_{DD}$ | AD7476A SNR Performance (dB) |
|---|---|
| AD780 @ 3 V | 72.65 |
| REF193 | 72.35 |
| AD780 @ 2.5 V | 72.5 |
| REF192 | 72.2 |
| REF43 | 72.6 |

## ANALOG INPUT

Figure 19 shows an equivalent circuit of the analog input structure of the AD7476A/AD7477A/AD7478A. The two diodes, D1 and D2, provide ESD protection for the analog input. Care must be taken to ensure that the analog input signal never exceeds the supply rails by more than 300 mV. This causes the diodes to become forward-biased and start conducting current into the substrate. The maximum current these diodes can conduct without causing irreversible damage to the part is 10 mA. The Capacitor C1 in Figure 19 is typically about 6 pF and can primarily be attributed to pin capacitance. The Resistor R1 is a lumped component made up of the on resistance of a switch. This resistor is typically about 100 Ω. The Capacitor C2 is the ADC sampling capacitor and has a capacitance of 20 pF typically.

For ac applications, removing high frequency components from the analog input signal is recommended by use of a band-pass filter on the relevant analog input pin. In applications where harmonic distortion and signal-to-noise ratio are critical, drive the analog input from a low impedance source. Large source impedances significantly affect the ac performance of the ADC, necessitating the use of an input buffer amplifier. The choice of the op amp is a function of the particular application.

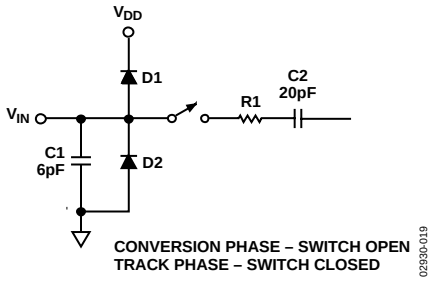



*Figure 19. Equivalent Analog Input Circuit*

Table 8 provides typical performance data with various op amps used as the input buffer for a 100 kHz input tone at room temperature under the same setup conditions.

**Table 8. AD7476A Typical Performance with Various Input Buffers, $V_{DD}$ = 3 V**

| Op Amp in the Input Buffer | AD7476A SNR Performance (dB) |
|---|---|
| AD711 | 72.3 |
| AD797 | 72.5 |
| AD845 | 71.4 |

When no amplifier is used to drive the analog input, limit the source impedance to low values. The maximum source impedance depends on the amount of total harmonic distortion (THD) that can be tolerated. The THD increases as the source impedance increases, degrading the performance (see Figure 13).

## DIGITAL INPUTS

The digital inputs applied to the AD7476A/AD7477A/AD7478A are not limited by the maximum ratings that limit the analog input. Instead, the digital inputs applied can reach 7 V and are not restricted by the $V_{DD}$ + 0.3 V limit as on the analog input. For example, if operating the AD7476A/AD7477A/AD7478A with a $V_{DD}$ of 3 V, use 5 V logic levels on the digital inputs. However, note that the data output on SDATA still has 3 V logic levels when $V_{DD}$ = 3 V. Another advantage of SCLK and $\overline{CS}$ not being restricted by the $V_{DD}$ + 0.3 V limit is that power supply sequencing issues are avoided. If $\overline{CS}$ or SCLK are applied before $V_{DD}$, there is no risk of latch-up as there would be on the analog input if a signal greater than 0.3 V were applied prior to $V_{DD}$.

# MODES OF OPERATION

The modes of operation for the AD7476A/AD7477A/AD7478A are selected by controlling the (logic) state of the $\overline{CS}$ signal during a conversion. There are two possible modes of operation: normal and power-down. The point at which $\overline{CS}$ is pulled high after the conversion has been initiated determines whether the AD7476A/AD7477A/AD7478A enters power-down mode. Similarly, if already in power-down, $\overline{CS}$ can control whether the device returns to normal operation or remains in power-down. These modes of operation are designed to provide flexible power management options. These options can be chosen to optimize the power dissipation/throughput rate ratio for different application requirements.

## NORMAL MODE

This mode is intended for the fastest throughput rate performance. In normal mode, the user does not have to worry about any power-up times because AD7476A/AD7477A/AD7478A remain fully powered at all times. Figure 20 shows the general diagram of the operation of the AD7476A/AD7477A/AD7478A in this mode. The conversion is initiated on the falling edge of $\overline{CS}$ as described in the Serial Interface section. To ensure that the part remains fully powered up at all times, $\overline{CS}$ must remain low until at least 10 SCLK falling edges have elapsed after the falling edge of $\overline{CS}$. If $\overline{CS}$ is brought high any time after the 10th SCLK falling edge but before the end of the $t_{CONVERT}$, the part remains powered up, but the conversion is terminated and SDATA goes back into three-state. For the AD7476A, 16 serial clock cycles are required to complete the conversion and access the complete conversion results. For the AD7477A and AD7478A, a minimum of 14 and 12 serial clock cycles are required to complete the conversion and access the complete conversion results, respectively. $\overline{CS}$ can idle high until the next conversion or idle low until $\overline{CS}$ returns high sometime prior to the next conversion (effectively idling $\overline{CS}$ low). Once a data transfer is complete (SDATA has returned to three-state), another conversion can be initiated after the quiet time, $t_{QUIET}$, has elapsed by bringing $\overline{CS}$ low again.

## POWER-DOWN MODE

This mode is intended for use in applications where slower throughput rates are required; either the ADC is powered down between each conversion, or a series of conversions is performed at a high throughput rate and the ADC is then powered down for a relatively long duration between these bursts of several conversions. When the AD7476A/AD7477A/AD7478A are in power-down, all analog circuitry is powered down. To enter power-down, the conversion process must be interrupted by bringing $\overline{CS}$ high anywhere after the second falling edge of SCLK and before the 10th falling edge of SCLK, as shown in Figure 22.

Once $\overline{CS}$ has been brought high in this window of SCLKs, the part enters power-down, the conversion that was initiated by the falling edge of $\overline{CS}$ is terminated, and SDATA goes back into three-state. If $\overline{CS}$ is brought high before the second SCLK falling edge, the part remains in normal mode and does not power down. This avoids accidental power-down due to glitches on the $\overline{CS}$ line. In order to exit this mode of operation and power up the AD7476A/AD7477A/AD7478A again, a dummy conversion is performed. On the falling edge of $\overline{CS}$, the device begins to power up and continues to power up as long as $\overline{CS}$ is held low until after the falling edge of the 10th SCLK. The device is fully powered up once 16 SCLKs have elapsed, and valid data results from the next conversion, as shown in Figure 24. If $\overline{CS}$ is brought high before the 10th falling edge of SCLK, then the AD7476A/AD7477A/AD7478A go back into power-down. This avoids accidental power-up due to glitches on the $\overline{CS}$ line or an inadvertent burst of eight SCLK cycles while $\overline{CS}$ is low. Although the device can begin to power up on the falling edge of $\overline{CS}$, it powers down again on the rising edge of $\overline{CS}$ as long as it occurs before the 10th SCLK falling edge.

## POWER-UP TIME

The power-up time of the AD7476A/AD7477A/AD7478A is 1 µs, meaning that with any frequency of SCLK up to 20 MHz, one dummy cycle is always sufficient to allow the device to power up. Once the dummy cycle is complete, the ADC is fully powered up and the input signal is acquired properly. The quiet time, $t_{QUIET}$, must still be allowed from the point where the bus goes back into three-state after the dummy conversion to the next falling edge of $\overline{CS}$. When running at a 1 MSPS throughput rate, the AD7476A/AD7477A/AD7478A power up and acquire a signal within 0.5 LSB in one dummy cycle, that is, 1 µs.

When powering up from the power-down mode with a dummy cycle, as in Figure 22, the track-and-hold that was in hold mode while the part was powered down returns to track mode after the first SCLK edge the part receives after the falling edge of $\overline{CS}$. This is shown as Point A in Figure 22. Although at any SCLK frequency, one dummy cycle is sufficient to power up the device and acquire $V_{IN}$, it does not necessarily mean that a full dummy cycle of 16 SCLKs must always elapse to power up the device and acquire $V_{IN}$ fully; 1 µs is sufficient to power up the device and acquire the input signal. If, for example, a 5 MHz SCLK frequency is applied to the ADC, the cycle time becomes 3.2 µs. In one dummy cycle, 3.2 µs, the part powers up and $V_{IN}$ acquires fully. However, after 1 µs with a 5 MHz SCLK, only five SCLK cycles would have elapsed. At this stage, the ADC would fully power up and acquire the signal. In this case, the $\overline{CS}$ can be brought high after the 10th SCLK falling edge and brought low again after a time, $t_{QUIET}$, to initiate the conversion.

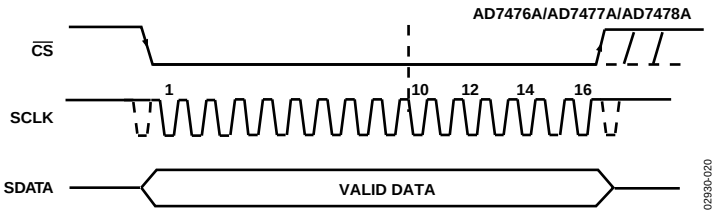



*Figure 20. Normal Mode Operation*

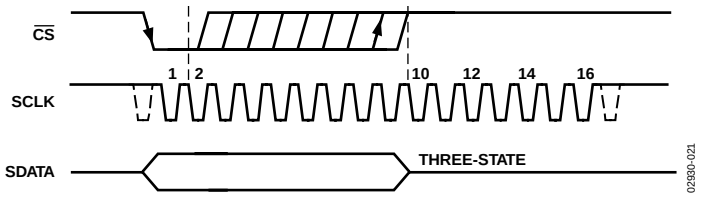



*Figure 21. Entering Power-Down Mode*

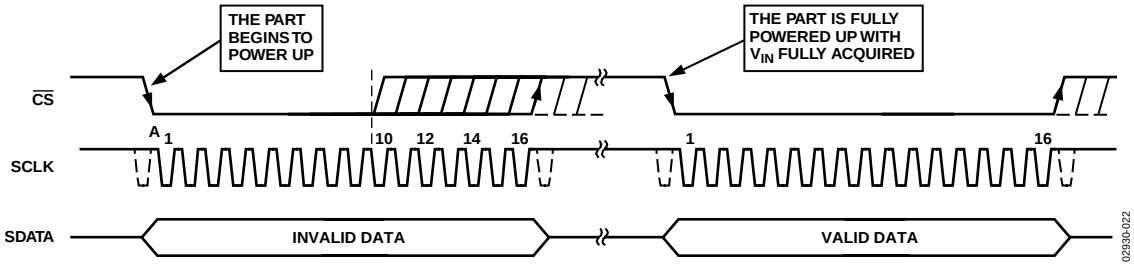



*Figure 22. Exiting Power-Down Mode*

When power supplies are first applied to the AD7476A/AD7477A/AD7478A, the ADC can power up in either the power-down or normal modes. Because of this, it is best to allow a dummy cycle to elapse to ensure that the part is fully powered up before attempting a valid conversion. Likewise, if it is intended to keep the part in the power-down mode while not in use and the user wishes the part to power up in power-down mode, the dummy cycle can be used to ensure that the device is in power-down by executing a cycle such as that shown in Figure 22. Once supplies are applied to the AD7476A/AD7477A/AD7478A, the power-up time is the same as that when powering up from the power-down mode. It takes approximately 1 μs to power up fully if the part powers up in normal mode. It is not necessary to wait 1 μs before executing a dummy cycle to ensure the desired mode of operation.

Instead, a dummy cycle can occur directly after power is supplied to the ADC. If the first valid conversion is performed directly after the dummy conversion, care must be taken to ensure that an adequate acquisition time has been allowed. As mentioned earlier, when powering up from the power-down mode, the part returns to track upon the first SCLK edge applied after the falling edge of $\overline{CS}$. However, when the ADC initially powers up after supplies are applied, the track-and-hold is already in track. This means, assuming one has the facility to monitor the ADC supply current, if the ADC powers up in the desired mode of operation and thus a dummy cycle is not required to change the mode, a dummy cycle is not required to place the track-and-hold into track.

# POWER VS. THROUGHPUT RATE

By using the power-down mode on the AD7476A/AD7477A/AD7478A when not converting, the average power consumption of the ADC decreases at lower throughput rates. Figure 23 shows that as the throughput rate is reduced, the device remains in its power-down state longer and the average power consumption over time drops accordingly.

For example, if the AD7476A/AD7477A/AD7478A operate in a continuous sampling mode with a throughput rate of 100 kSPS and an SCLK of 20 MHz ($V_{DD}$ = 5 V) and the devices are placed in the power-down mode between conversions, the power consumption is calculated as follows:

The power dissipation during normal operation is 17.5 mW ($V_{DD}$ = 5 V). If the power-up time is one dummy cycle, that is, 1 µs, and the remaining conversion time is another cycle, that is, 1 µs, then the AD7476A/AD7477A/AD7478A dissipate 17.5 mW for 2 µs during each conversion cycle.

If the throughput rate is 100 kSPS, the cycle time is 10 µs, then the average power dissipated during each cycle is (2/10) × (17.5 mW) = 3.5 mW.

If $V_{DD}$ = 3 V, SCLK = 20 MHz, and the devices are again in power-down mode between conversions, then the power dissipation during normal operation is 5.1 mW. Thus, the AD7576A/AD7477A/AD8478A dissipate 5.1 mW for 2 µs during each conversion cycle. With a throughput rate of 100 kSPS, the average power dissipated during each cycle is (2/10) × (5.1 mW) = 1.02 mW.

Figure 23 shows the power vs. the throughput rate when using the power-down mode between conversions with both 5 V and 3 V supplies. The power-down mode is intended for use with throughput rates of approximately 333 kSPS or less, because at higher sampling rates, the power-down mode produces no power savings.

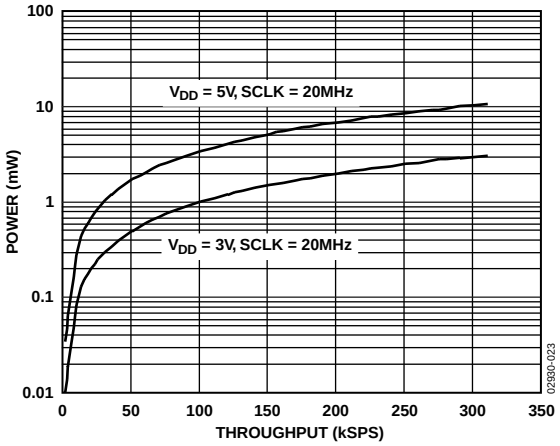



*Figure 23. Power vs. Throughput*

# SERIAL INTERFACE

Figure 24, Figure 25, and Figure 26 show the detailed timing diagrams for serial interfacing to the AD7476A, AD7477A, and AD7478A, respectively. The serial clock provides the conversion clock and also controls the transfer of information from the AD7476A/AD7477A/AD7478A during conversion.

The $\overline{CS}$ signal initiates the data transfer and conversion process. The falling edge of $\overline{CS}$ puts the track-and-hold into hold mode and takes the bus out of three-state; the analog input is sampled at this point. Also, the conversion is initiated at this point.

For the AD7476A, the conversion requires 16 SCLK cycles to complete. Once 13 SCLK falling edges have elapsed, the track-and-hold goes back into track on the next SCLK rising edge, as shown in Figure 24 at Point B. On the 16th SCLK falling edge, the SDATA line goes back into three-state. If the rising edge of $\overline{CS}$ occurs before 16 SCLKs have elapsed, the conversion is terminated and the SDATA line goes back into three-state; otherwise, SDATA returns to three-state on the 16th SCLK falling edge, as shown in Figure 24. Sixteen serial clock cycles are required to perform the conversion process and to access data from the AD7476A.

For the AD7477A, the conversion requires 14 SCLK cycles to complete. Once 13 SCLK falling edges have elapsed, the track-and-hold goes back into track on the next rising edge as shown at Point B in Figure 25. If the rising edge of $\overline{CS}$ occurs before 14 SCLKs have elapsed, the conversion is terminated and the SDATA line goes back into three-state. If 16 SCLKs are considered in the cycle, SDATA returns to three-state on the 16th SCLK falling edge, as shown in Figure 25.

For the AD7478A, the conversion requires 12 SCLK cycles to complete. The track-and-hold goes back into track on the rising edge after the 11th falling edge, as shown in Figure 26 at Point B. If the rising edge of $\overline{CS}$ occurs before 12 SCLKs have elapsed, the conversion is terminated and the SDATA line goes back into three-state. If 16 SCLKs are considered in the cycle, SDATA returns to three-state on the 16th SCLK falling edge, as shown in Figure 26.

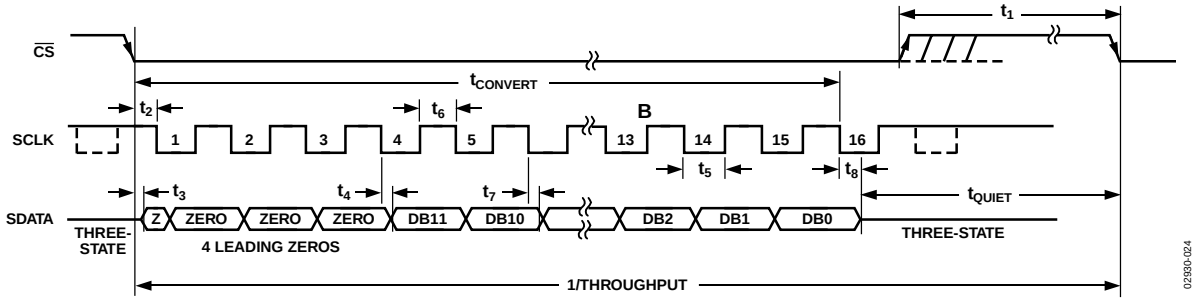



*Figure 24. AD7476A Serial Interface Timing Diagram*

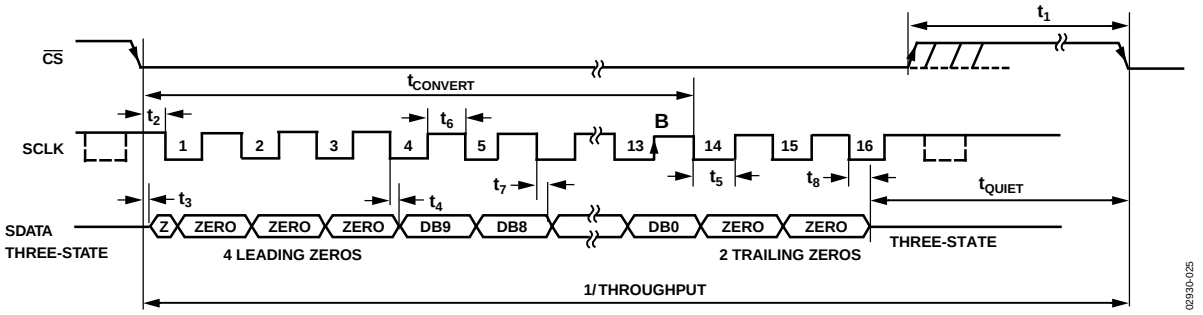



*Figure 25. AD7477A Serial Interface Timing Diagram*

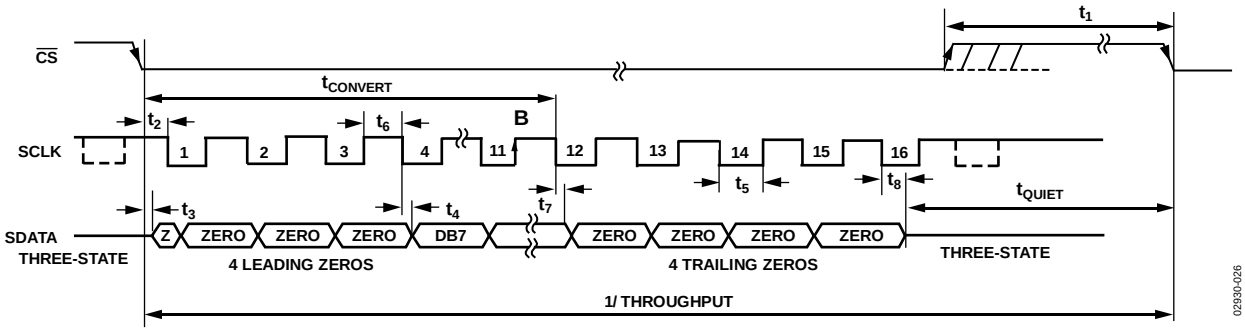



*Figure 26. AD7478A Serial Interface Timing Diagram*

$\overline{CS}$ going low clocks out the first leading zero to be read in by the microcontroller or DSP. The remaining data is then clocked out by subsequent SCLK falling edges beginning with the second leading zero. Thus, the first falling clock edge on the serial clock has the first leading zero provided and also clocks out the second leading zero. For the AD7476A, the final bit in the data transfer is valid on the 16th falling edge, having been clocked out on the previous (15th) falling edge.

In applications with a slower SCLK, it is possible to read in data on each SCLK rising edge. In this case, the first falling edge of SCLK clocks out the second leading zero, which can be read in the first rising edge. However, the first leading zero that was clocked out when $\overline{CS}$ went low will be missed, unless it was not read in the first falling edge. The 15th falling edge of SCLK clocks out the last bit and it can be read in the 15th rising SCLK edge.

If $\overline{CS}$ goes low just after one SCLK falling edge has elapsed, $\overline{CS}$ clocks out the first leading zero as it did before, and it can be read in the SCLK rising edge. The next SCLK falling edge clocks out the second leading zero, and it can be read in the following rising edge.

## AD7478A IN A 12 SCLK CYCLE SERIAL INTERFACE

For the AD7478A, if $\overline{CS}$ is brought high in the 12th rising edge after four leading zeros and eight bits of the conversion have been provided, the part can achieve a 1.2 MSPS throughput rate. For the AD7478A, the track-and-hold goes back into track in the 11th rising edge. In this case, a $f_{SCLK}$ = 20 MHz and a throughput of 1.2 MSPS give a cycle time of

$$t_2 + 10.5(1/f_{SCLK}) + t_{ACQ} = 833 \text{ ns}$$

With $t_2$ = 10 ns min, this leaves $t_{ACQ}$ to be 298 ns. This 298 ns satisfies the requirement of 225 ns for $t_{ACQ}$.

From Figure 27, $t_{ACQ}$ is comprised of

$$0.5\ (1/f_{SCLK}) + t_8 + t_{QUIET}$$

where $t_8$ = 36 ns maximum.

This allows a value of 237 ns for $t_{QUIET}$, satisfying the minimum requirement of 50 ns.

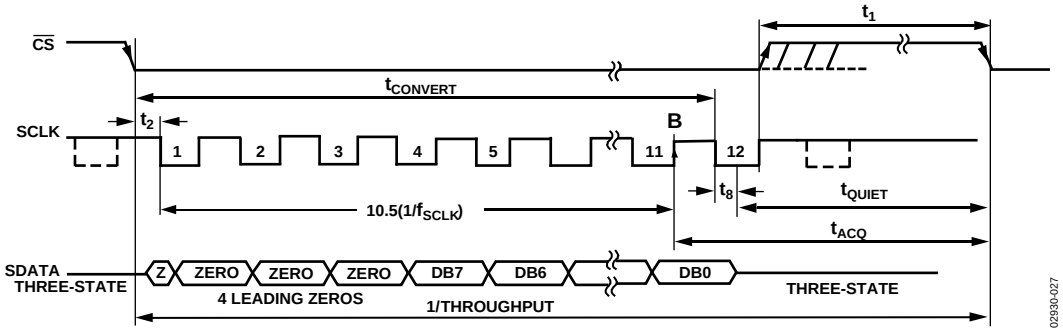



*Figure 27. AD7478A in a 12 SCLK Cycle Serial Interface*

# MICROPROCESSOR INTERFACING

The serial interface on the AD7476A/AD7477A/AD7478A allows the part to be directly connected to a range of different microprocessors. This section explains how to interface the AD7476A/AD7477A/AD7478A with some of the more common microcontroller and DSP serial interface protocols.

## AD7476A/AD7477A/AD7478A TO TMS320C541 INTERFACE

The serial interface on the TMS320C541 uses a continuous serial clock and frame synchronization signals to synchronize the data transfer operations with peripheral devices, such as the AD7476A/AD7477A/AD7478A. The $\overline{CS}$ input allows easy interfacing between the TMS320C541 and the AD7476A/AD7477A/AD7478A without any glue logic required. The serial port of the TMS320C541 is set up to operate in burst mode (FSM = 1 in the serial port control register, SPC) with Internal Serial Clock CLKX (MCM = 1 in the SPC register) and internal frame signal (TXM = 1 in the SPC register), so both pins are configured as outputs. For the AD7476A, set the word length to 16 bits (FO = 0 in the SPC register). This DSP only allows frames with a word length of 16 bits or 8 bits. Therefore, in the case of the AD7477A and AD7478A where 14 bits and 12 bits are required, the FO bit is set up to 16 bits. This means that to obtain the conversion result, 16 SCLKs are needed. In both situations, the remaining SCLKs clock out trailing zeros. For the AD7477A, two trailing zeros are clocked out in the last two clock cycles; for the AD7478A, four trailing zeros are clocked out.

To summarize, the values in the SPC register are

FO = 0
FSM = 1
MCM = 1
TXM = 1

The format bit, FO, can be set to 1 to set the word length to eight bits in order to implement the power-down mode on the AD7476A/AD7477A/AD7478A.

The connection diagram is shown in Figure 28. For signal processing applications, it is imperative that the frame synchronization signal from the TMS320C541 provide equidistant sampling.

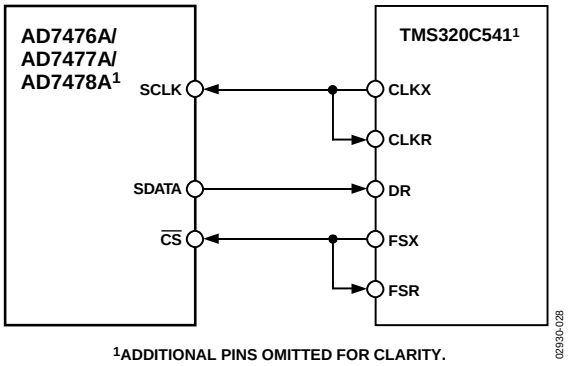



*Figure 28. Interfacing to the TMS320C541*

## AD7476A/AD7477A/AD7478A TO ADSP-218x INTERFACE

The ADSP-218x family of DSPs are interfaced directly to the AD7476A/AD7477A/AD7478A without any glue logic required. Set up the SPORT control register as follows:

TFSW = RFSW = 1, alternate framing
INVRFS = INVTFS = 1, active low frame signal
DTYPE = 00, right justify data
ISCLK = 1, internal serial clock
TFSR = RFSR = 1, frame every word
IRFS = 0, sets up RFS as an input
ITFS = 1, sets up TFS as an output
SLEN = 1111, 16 bits for the AD7476A
SLEN = 1101, 14 bits for the AD7477A
SLEN = 1011, 12 bits for the AD7478A

To implement the power-down mode, set SLEN to 0111 to issue an 8-bit SCLK burst. The connection diagram is shown in Figure 29. The ADSP-218x has the TFS and RFS of the SPORT tied together, with TFS set as an output and RFS set as an input. The DSP operates in alternate framing mode, and the SPORT control register is set up as described. The frame synchronization signal generated on the TFS is tied to $\overline{CS}$, and, as with all signal processing applications, equidistant sampling is necessary. However, in this example, the timer interrupt is used to control the sampling rate of the ADC and, under certain conditions, equidistant sampling may not be achieved.

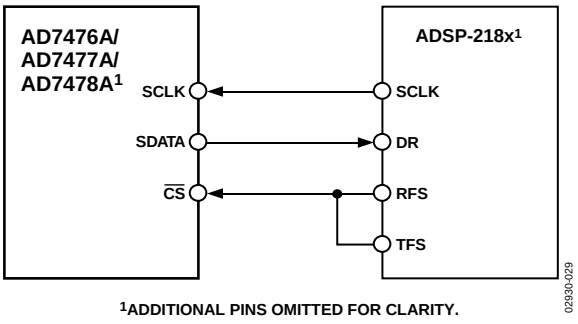



*Figure 29. Interfacing to the ADSP-218x*

The timer registers, for example, are loaded with a value that provides an interrupt at the required sample interval. When an interrupt is received, a value is transmitted with TFS/DT (ADC control word). The TFS controls the RFS and, thus, the reading of data. The frequency of the serial clock is set in the SCLKDIV register. When the instruction to transmit with TFS is given, that is, TX0 = AX0, the state of the SCLK is checked. The DSP waits until the SCLK has gone high, low, and high before transmission starts. If the timer and SCLK values are chosen such that the instruction to transmit occurs on or near the rising edge of SCLK, the data can be transmitted or it can wait until the next clock edge. For example, the ADSP-2111 has a master clock frequency of 16 MHz. If the SCLKDIV register is loaded with the Value 3, an SCLK of 2 MHz is obtained and eight master clock periods will elapse for every one SCLK period. If the timer registers are loaded with the Value 803, 100.5 SCLKs occur between interrupts and, subsequently, between transmit instructions. This situation results in nonequidistant sampling as the transmit instruction is occurring on an SCLK edge. If the number of SCLKs between interrupts is a whole integer figure of N, equidistant sampling is implemented by the DSP.

## AD7476A/AD7477A/AD7478A TO DSP563xx INTERFACE

The connection diagram in Figure 30 shows how the AD7476A/AD7477A/AD7478A can be connected to the SSI (synchronous serial interface) of the DSP563xx family of DSPs from Motorola. The SSI is operated in synchronous and normal mode (SYN 1 = and MOD = 0 in Control Register B, CRB) with internally generated word length frame sync for both Tx and Rx (Bit FSL1 = 0 and Bit FSL0 = 0 in CRB). Set the word length in Control Register A (CRA) to 16 by setting Bit WL2 = 0, Bit WL1 = 1, and Bit WL0 = 0 for the AD7476A. The word length for the AD7478A can be set to 12 bits (WL2 = 0, WL1 = 0, and WL0 = 1). This DSP does not offer the option for a 14-bit word length, so the AD7477A word length is set up to 16 bits, the same as the AD7476A. For the AD7477A, the conversion process uses 16 SCLK cycles, with the last two clock periods clocking out two trailing zeros to fill the 16-bit word.

To implement the power-down mode on the AD7476A/AD7477A/AD7478A, the word length can be changed to eight bits by setting Bit WL2 = 0, Bit WL1 = 0, and Bit WL0 = 0 in CRA. The FSP bit in the CRB register can be set to 1, meaning the frame goes low and a conversion starts. Likewise, by means of the Bit SCD2, Bit SCKD, and Bit SHFD in the CRB register, it establishes that Pin SC2 (the frame sync signal) and Pin SCK in the serial port are configured as outputs and the MSB is shifted first.

In summary:

MOD = 0
SYN = 1
WL2, WL1, and WL0 depend on the word length
FSL1 = 0 and FSL0 = 0
FSP = 1, negative frame sync
SCD2 = 1
SCKD = 1
SHFD = 0

Note that for signal processing applications, it is imperative that the frame synchronization signal from the DSP563xx provide equidistant sampling.

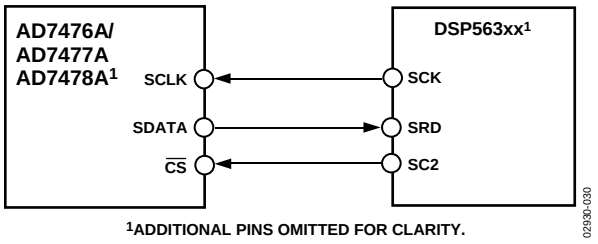



*Figure 30. Interfacing to the DSP563xx*

# APPLICATION HINTS

## GROUNDING AND LAYOUT

Design the printed circuit board that houses the AD7476A/AD7477A/AD7478A such that the analog and digital sections are separated and confined to certain areas of the board. This facilitates the use of ground planes that can be separated easily. A minimum etch technique is generally best for ground planes because it gives the best shielding. Join digital and analog ground planes at only one place. If the AD7476A/AD7477A/AD7478A is in a system where multiple devices require an AGND to DGND connection, make the connection at one point only, a star ground point that is established as close as possible to the AD7476A/AD7477A/AD7478A.

Avoid running digital lines under the device as these couple noise onto the die. Allow the analog ground plane to run under the AD7476A/AD7477A/AD7478A in order to avoid noise coupling. Use as large a trace as possible on the power supply lines to the AD7476A/AD7477A/AD7478A to provide low impedance paths and reduce the effects of glitches on the power supply line. Shield fast switching signals like clocks with digital grounds to avoid radiating noise to other sections of the board, and never run clock signals near the analog inputs. Avoid crossover of digital and analog signals. Run traces on opposite sides of the board at right angles to each other. This reduces the effects of feedthrough through the board. A microstrip technique is by far the best but is not always possible with a double-sided board. In this technique, the component side of the board is dedicated to ground planes while signals are placed on the solder side.

Good decoupling is also very important. Decouple the supply with, for instance, a 680 nF 0805 capacitor to GND. When using the SC70 package in applications where the size of the components is of concern, a 220 nF 0603 capacitor, for example, can be used instead. However, in that case, the decoupling may not be as effective, resulting in an approximate SINAD degradation of 0.3 dB. To achieve the best performance from these decoupling components, the user should endeavor to keep the distance between the decoupling capacitor and the $V_{DD}$ and GND pins to a minimum with short track lengths connecting the respective pins. Figure 31 and Figure 32 and show the recommended positions of the decoupling capacitor for the SC70 and MSOP packages, respectively.

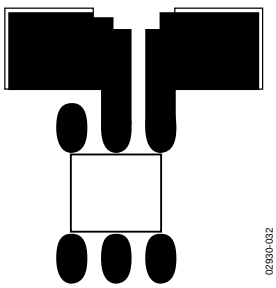



*Figure 31. Recommended Supply Decoupling Scheme for the SC70 Package*

As can be seen in Figure 32, for the MSOP package, the decoupling capacitor has been placed as close as possible to the IC with short track lengths to $V_{DD}$ and GND pins. The decoupling capacitor can also be placed on the underside of the PCB directly underneath the IC, between the $V_{DD}$ and GND pins attached by vias. This method is not recommended on PCBs above a standard 1.6 mm thickness. The best performance is realized with the decoupling capacitor on the top of the PCB next to the IC.

Similarly, for the SC70 package, locate the decoupling capacitor as close as possible to the $V_{DD}$ and the GND pins. Because of its pinout, that is, $V_{DD}$ being next to GND, the decoupling capacitor can be placed extremely close to the IC. The decoupling capacitor can be placed on the underside of the PCB directly under the $V_{DD}$ and GND pins, but the best performance is achieved with the decoupling capacitor on the same side as the IC.

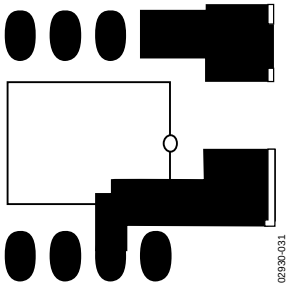



*Figure 32. Recommended Supply Decoupling Scheme for the AD7476A/AD7477A/AD7478A MSOP Package*

## EVALUATING THE AD7476A/AD7477A PERFORMANCE

The evaluation board package includes a fully assembled and tested evaluation board, documentation, and software for controlling the board from the PC via the EVAL-BOARD CONTROLLER. The EVAL-BOARD CONTROLLER can be used in conjunction with the AD7476ACB/AD7477ACB evaluation board, as well as many other Analog Devices evaluation boards ending in the CB designator, to demonstrate/evaluate the ac and dc performance of the AD7476A/AD7477A. The software allows the user to perform ac (fast Fourier transform) and dc (histogram of codes) tests on the AD7476A/AD7477A. See the evaluation board application note for more information.

# OUTLINE DIMENSIONS

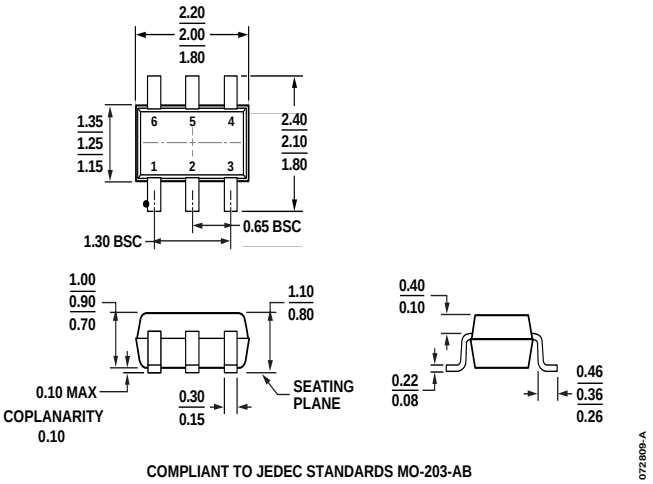



*Figure 33. 6-Lead Thin Shrink Small Outline Transistor Package [SC70]*
*(KS-6)*
*Dimensions shown in millimeters*

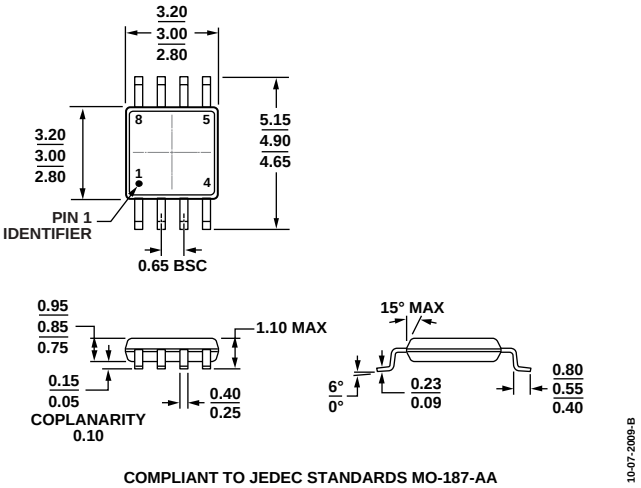



*Figure 34. 8-Lead Mini Small Outline Package [MSOP]*
*(RM-8)*
*Dimensions shown in millimeters*

## ORDERING GUIDE

| Model[1, 2, 3, 4] | Temperature Range | Linearity Error (LSB)[5] | Package Description | Package Option[6] | Branding |
|---|---|---|---|---|---|
| AD7476AAKSZ-500RL7 | –40°C to +85°C | ±0.75 typical | 6-Lead SC70 | KS-6 | C3V |
| AD7476AAKSZ-REEL | –40°C to +85°C | ±0.75 typical | 6-Lead SC70 | KS-6 | C3V |
| AD7476AAKSZ-REEL7 | –40°C to +85°C | ±0.75 typical | 6-Lead SC70 | KS-6 | C3V |
| AD7476ABKSZ-500RL7 | –40°C to +85°C | ±1.5 maximum | 6-Lead SC70 | KS-6 | C3W |
| AD7476ABKSZ-REEL | –40°C to +85°C | ±1.5 maximum | 6-Lead SC70 | KS-6 | C3W |
| AD7476ABKSZ-REEL7 | –40°C to +85°C | ±1.5 maximum | 6-Lead SC70 | KS-6 | C3W |
| AD7476ABRM | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | CEY |
| AD7476ABRM-REEL | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | CEY |
| AD7476ABRM-REEL7 | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | CEY |
| AD7476ABRMZ | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C3W |
| AD7476ABRMZ-REEL | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C3W |
| AD7476ABRMZ-REEL7 | –40°C to +85°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C3W |
| AD7476AWYRMZ | –40°C to +125°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C45 |
| AD7476AWYRMZ-RL7 | –40°C to +125°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C45 |
| AD7476AYKSZ-500RL7 | –40°C to +125°C | ±1.5 maximum | 6-Lead SC70 | KS-6 | C45 |
| AD7476AYKSZ-REEL7 | –40°C to +125°C | ±1.5 maximum | 6-Lead SC70 | KS-6 | C45 |
| AD7476AYRMZ | –40°C to +125°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C45 |
| AD7476AYRMZ-REEL7 | –40°C to +125°C | ±1.5 maximum | 8-Lead MSOP | RM-8 | C45 |
| AD7477AAKSZ-500RL7 | –40°C to +85°C | ±0.5 maximum | 6-Lead SC70 | KS-6 | C3X |
| AD7477AAKSZ-REEL | –40°C to +85°C | ±0.5 maximum | 6-Lead SC70 | KS-6 | C3X |
| AD7477AARM-REEL | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | CFZ |
| AD7477AARMZ | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | C3X |
| AD7477AARMZ-REEL | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | C3X |
| AD7477AARMZ-REEL7 | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | C3X |
| AD7477AWARMZ | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | C3X |
| AD7477AWARMZ-RL | –40°C to +85°C | ±0.5 maximum | 8-Lead MSOP | RM-8 | C3X |
| AD7478AAKSZ-500RL7 | –40°C to +85°C | ±0.3 maximum | 6-Lead SC70 | KS-6 | C48 |
| AD7478AAKSZ-REEL | –40°C to +85°C | ±0.3 maximum | 6-Lead SC70 | KS-6 | C48 |
| AD7478AAKSZ-REEL7 | –40°C to +85°C | ±0.3 maximum | 6-Lead SC70 | KS-6 | C48 |
| AD7478AARM | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | CJZ |
| AD7478AARMZ | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | C48 |

| Model[1, 2, 3, 4] | Temperature Range | Linearity Error (LSB)[5] | Package Description | Package Option[6] | Branding |
|---|---|---|---|---|---|
| AD7478AARMZ-REEL | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | C48 |
| AD7478AARMZ-REEL7 | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | C48 |
| AD7478AWARMZ | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | C48 |
| AD7478AWARMZ-RL | –40°C to +85°C | ±0.3 maximum | 8-Lead MSOP | RM-8 | C48 |
| EVAL-AD7476ACBZ | | | Evaluation Board | | |
| EVAL-CONTROL BRD2 | | | Evaluation Control | | |

[1] Z = RoHS Compliant Part.
[2] W = Qualified for Automotive Applications.
[3] EVAL-AD7476ACBZ can be used as a standalone evaluation board or in conjunction with the EVAL-CONTROL BOARD for evaluation/demonstration purposes.
[4] EVAL-CONTROL BRD2 is a complete unit, allowing a PC to control and communicate with all Analog Devices evaluation boards ending in the CB designator. To order a complete evaluation kit, you will need to order the particular ADC evaluation board, for example, EVAL-AD7476ACB, the EVAL-CONTROLBRD2, and a 12 V ac transformer. See relevant evaluation board application note for more information.
[5] Linearity error here refers to integral nonlinearity.
[6] KS = SC70; RM = MSOP.

## AUTOMOTIVE PRODUCTS

The AD7476AWYRMZ, AD7476AWYRMZ-RL7, AD7477AWARMZ, AD7477AWARMZ-RL, AD7478AWARMZ, and AD7478AWARMZ-RL models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

## NOTES

©2002–2011 Analog Devices, Inc. All rights reserved. Trademarks and registered trademarks are the property of their respective owners.
D02930-0-1/11(F)

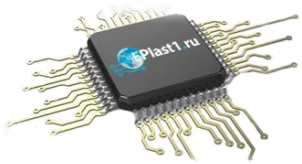
Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits,General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.




**Как с нами связаться**

**Телефон:** 8 (812) 309 58 32 (многоканальный)
**Факс:** 8 (812) 320-02-42
**Электронная почта:** org@eplast1.ru
**Адрес:** 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.