## **NXP Semiconductors**

Technical Data

# **System basis chip (SBC) with low speed fault tolerant CAN interface**

The 33889 is an SBC having a fully protected, fixed 5.0 V low drop-out regulator, with current limit, overtemperature prewarning and reset. An SBC device is a monolithic IC combining many functions repeatedly found in standard microcontroller-based systems, e.g., protection, diagnostics, communication, power, etc.

An output drive with sense input is also provided to implement a second 5.0 V regulator using an external PNP. The 33889 has Normal, Standby, Stop and Sleep modes; an internally switched high-side power supply output with two wake-up inputs; programmable timeout or window watchdog, Interrupt, Reset, serial peripheral interface (SPI) input control, and a low-speed fault tolerant CAN transceiver, compatible with CAN 2.0 A and B protocols for module-tomodule communications. The combination is an economical solution for power management, high-speed communication, and control in MCU-based systems. This device is powered by SMARTMOS technology.

#### **Features**

- VDD1: 5.0 V low drop voltage regulator, current limitation, overtemperature detection, monitoring and reset function with total current capability 200 mA
- V2: tracking function of VDD1 regulator; control circuitry for external bipolar ballast transistor for high flexibility in choice of peripheral voltage and current supply
- Four operational modes
- Low standby current consumption in Stop and Sleep modes
- Built-in low speed 125 kbps fault tolerant CAN physical interface.
- External high voltage wake-up input, associated with HS1  $V_{BAT}$  switch
- 150 mA output current capability for HS1  $V_{BAT}$  switch allowing drive of external switches pull-up resistors or relays





\*Recommended for new designs



 **Figure 1. 33889 simplified application diagram**



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# **1 Device variations**

#### <span id="page-1-2"></span>**Table 1. Device variations between the 33889D and 33889B versions [\(1\)](#page-1-0)**





<span id="page-1-0"></span>Notes

1. This datasheet uses the term 33889 in the inclusive sense, referring to both the D version (33889D) and the B version (33689B).

<span id="page-1-1"></span>2. The 33889D and 33889B versions are nearly identical. However, where variations in characteristic occur, these items will be separated onto individual lines.

# **2 Internal block diagram**



# **3 Pin connections**



#### **Figure 3. 33889 pin connections**

A functional description of each pin can be found in the [Functional pin description](#page-19-0) section [page](#page-19-0) 20.

#### **Table 2. Pin definitions**



# **4 Electrical characteristics**

## **4.1 Maximum ratings**

#### **Table 3. Maximum ratings**

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device.

<span id="page-4-0"></span>

#### **Table 3. Maximum ratings (continued)**

All voltages are with respect to ground unless otherwise noted. Exceeding these ratings may cause a malfunction or permanent damage to the device.



<span id="page-5-1"></span>Notes:

3. Testing done in accordance with the Human Body Model (CzAP = 100 pF, RzAP = 1500 Ω), Machine Model (CzAP = 200 pF, RzAP = 0 Ω).

<span id="page-5-0"></span>4. ESD machine model (MM) is for MC33889B only. MM is now replaced by CDM (Charged Discharged model).

<span id="page-5-3"></span>5. Gnd pins 6,7,8,9,20, 21, 22, 23.



<span id="page-5-2"></span>Note: Waveform in accordance to ISO7637 part1, test pulses 1, 2, 3a and 3b.

#### **Figure 4. Transient test pulse for L0 and L1 inputs**

## **4.2 Static electrical characteristics**

#### **Table 4. Static electrical characteristics**

Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25$  °C under nominal conditions unless otherwise noted.



<span id="page-6-8"></span><span id="page-6-7"></span><span id="page-6-5"></span>Notes

6.  $V_{DD1}$  > 4.0 V, reset high, if R<sub>STTH-2</sub> selected and I<sub>OUT</sub> V<sub>DD1</sub> reduced, logic pin high level reduced, device is functional.

<span id="page-6-1"></span>7. Current measured at  $V_{\text{SUP}}$  pin.

<span id="page-6-0"></span>8. Device is fully functional. All modes available and operating, Watchdog, HS1 turn ON turn OFF, CAN cell operating, L0 and L1 inputs operating, SPI read write operation. Over temperature may occur.

<span id="page-6-3"></span>9. Measured in worst case condition with 5.0 V at V2 pin (V2 pin tied to VDD1).

<span id="page-6-2"></span>10. Oscillator running means "Forced Wake-up" or "Cyclic Sense" or "Software Watchdog" timer activated. Software Watchdog is available in stop mode only.

<span id="page-6-6"></span>11.  $V_{DD1}$  is ON with 2.0 mA typical output current capability.

<span id="page-6-4"></span>12. Guaranteed by design

Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A$  = 25 °C under nominal conditions unless otherwise noted.



Notes

<span id="page-7-1"></span>13. Guaranteed by design

<span id="page-7-2"></span>14. I<sub>DD1</sub> is the total regulator output current. VDD specification with external capacitor C  $\geq$  22 μF and ESR < 10 Ω.

<span id="page-7-0"></span>15. Selectable by SPI

Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25 \degree C$  under nominal conditions unless otherwise noted.



<span id="page-8-0"></span>16. V2 tracking voltage regulator - V2 specification with external capacitor

- option 1: C ≥ 22 μF and ESR < 10 Ω. Using a resistor of 2 kΩ or less between the base and emitter of the external PNP is recommended. - option2: 1.0  $\mu$ F < C < 22  $\mu$ F and ESR < 10  $\Omega$ . In this case depending on the ballast transistor gain an additional resistor and capacitor network between emitter and base of PNP ballast transistor might be required. Refer to NXP application information or contact your local technical support. - option 3: 10 uF < C < 22 μF ESR > 0.2 Ω: a resistor of 2 kΩ or less is required between the base and emitter of the external PNP.

<span id="page-8-1"></span>17. For  $I_{VDD1} > 10$  mA

Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A$  = 25 °C under nominal conditions unless otherwise noted.



Notes

<span id="page-9-0"></span>18. Refer to HS1 negative maximum rating voltage limitation of -0.2 V.

Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at T<sub>A</sub> = 25 °C under nominal conditions unless otherwise noted.



Characteristics noted under conditions - V<sub>SUP</sub> From 5.5 V to 18 V and T<sub>J</sub> from -40 to 125 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A$  = 25 °C under nominal conditions unless otherwise noted.





Notes

<span id="page-11-0"></span>19. For MC33889B, after 128 pulses on TX and no bus failure.

<span id="page-11-1"></span>20. Guaranteed by design

## **4.3 Dynamic electrical characteristics**

#### **Table 5. Dynamic electrical characteristics**

V<sub>SUP</sub> From 5.5 V to 18 V, V2INT from 4.75 to 5.25 V and T<sub>J</sub> from -40 to 150 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25^{\circ}$ C under nominal conditions, unless otherwise noted.



Notes

<span id="page-12-0"></span>21. Guaranteed by design

V<sub>SUP</sub> From 5.5 V to 18 V, V2INT from 4.75 to 5.25 V and T<sub>J</sub> from -40 to 150 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25^\circ C$  under nominal conditions, unless otherwise noted.



Notes

<span id="page-13-0"></span>22. State Machine Timing - Delay starts at rising edge of  $\overline{CS}$  (end of SPI command) and start of Turn on or Turn off of HS1 or V2.

V<sub>SUP</sub> From 5.5 V to 18 V, V2INT from 4.75 to 5.25 V and T<sub>J</sub> from -40 to 150 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25^{\circ}$ C under nominal conditions, unless otherwise noted.



Notes

<span id="page-14-0"></span>23. Guaranteed by design

<span id="page-14-1"></span>24. Dominant to recessive slew rate is dependant upon the bus load characteristics.

<span id="page-14-2"></span>25. AC Characteristics measured according to schematic [Figure](#page-16-0) 5

 $V_{SUP}$  From 5.5 V to 18 V, V2INT from 4.75 to 5.25 V and T<sub>J</sub> from -40 to 150 °C, unless otherwise noted. Typical values noted reflect the approximate parameter means at  $T_A = 25^{\circ}$ C under nominal conditions, unless otherwise noted.



Notes

<span id="page-15-0"></span>26. AC characteristic according to ISO11898-3, tested per figure 5 and 6. Guaranteed by design, room temperature only.

<span id="page-15-1"></span>27. AC characteristic according to ISO11898-3, tested per figure 7. Max. reported is the typical measurement under the worst condition (gnd shift, dominant/recessive edge, at source or destination node. ref to ISO test specification). Guaranteed by design, room temperature only.





<span id="page-16-0"></span>

 **Figure 6. ISO loop time without bus failure** RcanL = RcanH =  $125 \Omega$ 

<span id="page-16-1"></span>

CANL and CANH open

<span id="page-16-2"></span> **Figure 7. ISO loop time with bus failure**



 **Figure 8. Test set up for propagation delay with GND shift in a 5 node configuration**

## <span id="page-17-0"></span>**4.4 Timing diagrams**







 **Figure 10. Timing characteristic**

# **5 Functional description**

## **5.1 Introduction**

The MC33889 is an integrated circuit dedicated to automotive applications. It includes the following functions:

- One full protected voltage regulator with 200 mA total output current capability.
- Driver for external path transistor for V2 regulator function.
- Reset, programmable watchdog function
- Four operational modes
- Wake-up capabilities: Forced wake-up, cyclic sense and wake-up inputs, CAN and the SPI
- Can low speed fault tolerant physical interface.

## <span id="page-19-0"></span>**5.2 Functional pin description**

## **5.2.1 Receive and transmit data (RX and TX)**

The RX and TX pins (receive data and transmit data pins, respectively) are connected to a microcontroller's CAN protocol handler. TX is an input and controls the CANH and CANL line state (dominant when TX is LOW, recessive when TX is HIGH). RX is an output and reports the bus state (RX LOW when CAN bus is dominant, HIGH when CAN bus is recessive).

## **5.2.2 Voltage regulator one (VDD1)**

The VDD1 pin is the output pin of the 5.0 V internal regulator. It can deliver up to 200 mA. This output is protected against overcurrent and overtemperature. It includes an overtemperature prewarning flag, which is set when the internal regulator temperature exceeds 130 °C typical. When the temperature exceeds the overtemperature shutdown (170 °C typical), the regulator is turned off. VDD1 includes an undervoltage reset circuitry, which sets the RST pin LOW when VDD is below the undervoltage reset threshold.

## **5.2.3 Reset (RST)**

The Reset pin RST is an output that is set LOW when the device is in reset mode. The RST pin is set HIGH when the device is not in reset mode. RST includes an internal pull-up current source. When RST is LOW, the sink current capability is limited, allowing RST to be shorted to 5.0 V for software debug or software download purposes.

## **5.2.4 Interrupt (INT)**

The Interrupt pin INT is an output that is set LOW when an interrupt occurs. INT is enabled using the Interrupt Register (INTR). When an interrupt occurs, INT stays LOW until the interrupt source is cleared. INT output also reports a wake-up event by a 10 sec. typical pulse when the device is in Stop mode.

## **5.2.5 Ground (GND)**

This pin is the ground of the integrated circuit.

### **5.2.6 V2CTRL (V2CTRL)**

The V2CTRL pin is the output drive pin for the V2 regulator connected to the external series pass transistor.

### **5.2.7 Voltage supply (VSUP)**

The VSUP pin is the battery supply input of the device.

### **5.2.8 High-side output 1 (HS1)**

The HS pin is the internal high-side driver output. It is internally protected against overcurrent and overtemperature.

### **5.2.9 Level 0-1 inputs (L0: L1)**

The L0: L1 pins can be connected to contact switches or the output of other ICs for external inputs. The input states can be read by the SPI. These inputs can be used as wake-up events for the SBC when operating in the Sleep or Stop mode.

## **5.2.10 Voltage regulator two (V2)**

The V2 pin is the input sense for the V2 regulator. It is connected to the external series pass transistor. V2 is also the 5.0 V supply of the internal CAN interface. It is possible to connect V2 to an external 5.0 V regulator or to the VDD output when no external series pass transistor is used. In this case, the V2CTRL pin must be left open.

## **5.2.11 RTH (RTH)**

Pin for the connection of the bus termination resistor to CANH

## **5.2.12 RTL (RTL)**

Pin for the connection of the bus termination resistor to CANL

### **5.2.13 CAN high and CAN low outputs (CANH and CANL)**

The CAN High and CAN Low pins are the interfaces to the CAN bus lines. They are controlled by TXD input level, and the state of CANH and CANL is reported through RXD output.

### **5.2.14 System clock (SCLK)**

SCLK is the Serial Data Clock input pin of the serial peripheral interface.

### **5.2.15 Master in/slave out (MISO**

MISO is the Master In Slave Out pin of the serial peripheral interface. Data is sent from the SBC to the microcontroller through the MISO pin.

### **5.2.16 Master out/slave in (MOSI)**

MOSI is the Master Out Slave In pin of the serial peripheral interface. Control data from a microcontroller is received through this pin.

## **5.2.17 Chip select (CS)**

CS is the Chip Select pin of the serial peripheral interface. When this pin is LOW, the SPI port of the device is selected.

## **5.2.18 Watchdog (WDOG)**

The Watchdog output pin is asserted LOW to flag that the software watchdog has not been properly triggered.

## **5.3 Functional internal block description**

### **5.3.1 Device supply**

The device is supplied from the battery line through the VSUP pin. An external diode is required to protect against negative transients and reverse battery. It can operate from 4.5 V and under the jump start condition at 27 V DC. This pin sustains standard automotive voltage conditions such as load dump at 40 V. When  $V_{\text{SUP}}$  falls below 3.0 V typical, the MC33889 detects it and stores the information in the SPI register, in a bit called "BATFAIL". This detection is available in all operation modes.

### **5.3.2 VDD1 voltage regulator**

VDD1 Regulator is a 5.0 V output voltage with total current capability of 200 mA. It includes a voltage monitoring circuitry associated with a reset function. The VDD1 regulator is fully protected against overcurrent, short-circuit and has overtemperature detection warning flags and shutdown with hysteresis.

### **5.3.3 V2 regulator**

V2 Regulator circuitry is designed to drive an external path transistor in order to increase output current flexibility. Two pins are used: V2 and V2CTRL. Output voltage is 5.0 V and is realized by a tracking function of the VDD1 regulator. A recommended ballast transistor is the MJD32C. Other transistors might be used, however depending upon the PNP gain, an external resistor capacitor network might be connected between the emitter and base of the PNP. The use of external ballast is optional (refer to simplified typical application). The state of V2 is reported into the IOR register (if V2 is below 4.5 V typical, or in cases of overload or short-circuit).

### **5.3.4 HS1 VBAT switch output**

HS1 output is a 2.0 Ω typical switch from the VSUP pin. It allows the supply of external switches and their associated pull-up or pull-down circuitry, for example, in conjunction with the wake-up input pins. Output current is limited to 200 mA and HS1 is protected against shortcircuit and has an over temperature shutdown (reported into the IOR register). The HS1 output is controlled from the internal register and the SPI. It can be activated at regular intervals in sleep mode thanks to an internal timer. It can also be permanently turned on in normal or stand-by modes to drive external loads, such as relays or supply peripheral components. In case of inductive load drive, external clamp circuitry must be added.

### **5.3.5 SPI**

The complete device control as well as the status report is done through an 8 bit SPI interface. Refer to the SPI paragraph.

### **5.3.6 CAN**

The device incorporates a low speed fault tolerant CAN physical interface. The speed rate is up to 125 kBaud. The state of the CAN interface is programmable through the SPI. Reference the [CAN transceiver description on page 28.](#page-27-0)

### **5.3.7 Package and thermal consideration**

The device is proposed in a standard surface mount SO28 package. In order to improve the thermal performances of the SO28 package, 8 pins are internally connected to the lead frame and are used for heat transfer to the printed circuit board.

# **6 Functional device operation**

## **6.1 Operational modes**

#### **6.1.1 Introduction**

The device has four modes of operation, normal, stand-by, sleep and stop modes. All modes are controlled by the SPI. An additional temporary mode called "normal request mode" is automatically accessed by the device (refer to state machine) after wake-up events. Special mode and configurations are possible for software application debug and flash memory programming.

### **6.1.2 Normal mode**

In this mode both regulators are ON, and this corresponds to the normal application operation. All functions are available in this mode (watchdog, wake-up input reading through the SPI, HS1 activation, and CAN communication). The software watchdog is running and must be periodically cleared through the SPI.

### **6.1.3 Standby mode**

Only the Regulator 1 is ON. Regulator 2 is turned OFF by disabling the V2CTRL pin. The CAN cell is not available, as powered from V2. Other functions are available: wake-up input reading through the SPI and HS1 activation. The watchdog is running.

### **6.1.4 Sleep mode**

Regulators 1 and 2 are OFF. In this mode, the MCU is not powered. The device can be awakened internally by cyclic sense via the wakeup input pins and HS1 output, from the forced wake function, the CAN physical interface, and the SPI (CS pin).

### **6.1.5 Stop mode**

Regulator 2 is turned OFF by disabling the V2CTRL pin. Regulator 1 is activated in a special low power mode which allows it to deliver 2.0 mA. The objective is to supply the MCU of the application while it is turned into a power saving condition (i.e stop or wait mode).

Stop mode is entered through the SPI. Stop mode is dedicated to powering the Microcontroller when it is in low power mode (stop, pseudo stop, wait etc.). In these modes, the MCU supply current is less than 1.0 mA. The MCU can restart its software application very quickly without the complete power up and reset sequence.

When the application is in stop mode (both MCU and SBC), the application can wake-up from the SBC side (ex cyclic sense, forced wakeup, CAN message, wake-up inputs) or the MCU side (key wake-up etc.).

When Stop mode is selected by the SPI, stop mode becomes active 20 μs after end of the SPI message. The "go to stop" instruction must be the last instruction executed by the MCU before going to low power mode.

In Stop mode, the Software watchdog can be "running" or "not running" depending on the selection by the SPI. Refer to the SPI description, RCR register bit WDSTOP. If the W/D is enabled, the SBC must wake-up before the W/D time has expired, otherwise a reset is generated. In stop mode, the SBC wake-up capability is identical as in sleep mode.

### **6.1.5.1 Stop mode: wake-up from SBC side, INT pin activation**

When an application is in stop mode, it can wake-up from the SBC side. When a wake-up is detected by the SBC (CAN, Wake-up input, forced wake-up, etc.), the SBC turns itself into Normal request mode and activates the VDD1 main regulator. When the main regulator is fully active, then the wake-up is signalled to the MCU through the  $\overline{\text{INT}}$  pin. The  $\overline{\text{INT}}$  pin is pulled low for 10 us and then returns high. Wakeup events can be read through the SPI registers.

### **6.1.5.2 Stop mode: wake-up from MCU side**

When the application is in stop mode, the wake-up event may come to the MCU. In this case, the MCU has to signal to the SBC that it has to go into Normal mode in order for the VDD1 regulator to be able to deliver full current capability. This is done by a low to high transition of the CS pin. The CS pin low to high activation has to be done as soon as possible after the MCU. The SBC generates a pulse at the INT pin. Alternatively the L0 and L1 inputs can also be used as wake-up from the Stop mode.

#### **6.1.5.3 Stop mode current monitoring**

If the current in Stop mode exceeds the I<sub>DD1S-WU</sub> threshold, the SBC jumps into Normal request mode, activates the VDD1 main regulator, and generates an interrupt to the MCU. This interrupt is not maskable and a not bit are set into the INT register.

#### **6.1.5.4 Software watchdog in stop mode**

If the watchdog is enabled (register MCR, bit WDSTOP set), the MCU has to wake-up independently of the SBC before the end of the SBC watchdog time. In order to do this, the MCU has to signal the wake-up to the SBC through the SPI wake-up (CS pin low to high transition to activated the SPI wake-up). Then the SBC wakes up and jumps into the normal request mode. The MCU has to configure the SBC to go to either into normal or standby mode. The MCU can then choose to go back into stop mode.

If no MCU wake-up occurs within the watchdog timing, the SBC will activate the reset pin and jump into the normal request mode. The MCU can then be initialized.

#### **6.1.5.5 Normal request mode**

This is a temporary mode automatically accessed by the device after a wake-up event from sleep or stop mode, or after device power up. In this mode, the VDD1 regulator is ON, V2 is off, and the reset pin is high. As soon as the device enters the normal request mode, an internal 350 ms timer is started. During these 350 ms, the microcontroller of the application must address the SBC via the SPI and configure the watchdog register (TIM1 register). This is the condition for the SBC to leave the Normal request Mode and enter the Normal mode, and to set the watchdog timer according to the configuration done during the Normal Request mode.

The "BATFAIL flag" is a bit which is triggered when  $V_{\text{SLIP}}$  falls below 3.0 V. This bit is set into the MCR register. It is reset by the MCR register read.

### **6.1.6 Internal Clock**

This device has an internal clock used to generate all timings (reset, watchdog, cyclic wake-up, filtering time etc...).

### **6.1.7 Reset Pin**

A reset output is available in order to reset the microcontroller. Reset causes are:

- $V_{DD1}$  falling out of range: if  $V_{DD1}$  falls below the reset threshold (parameter  $R_{ST-TH}$ ), the reset pin is pulled low until  $V_{DD1}$  returns to the nominal voltage.
- Power on reset: at device power on or at device wake-up from sleep mode, the reset is maintained low until  $V_{DD1}$  is within its operation range.
- Watchdog timeout: if the watchdog is not cleared, the SBC will pull the reset pin low for the duration of the reset duration time (parameter: RESET-DUR).

For debug purposes at 25 °C, the reset pin can be shorted to 5.0 V.

### **6.1.8 Software watchdog (selectable window or timeout watchdog)**

The software watchdog is used in the SBC normal and stand-by modes for monitoring the MCU. The watchdog can be either a window or timeout. This is selectable by the SPI (register TIM, bit WDW). Default is the window watchdog. The period of the watchdog is selectable by the SPI from 5.0 to 350 ms (register TIM, bits WDT0 and WDT1). When the window watchdog is selected, the closed window is the first half of the selected period, and the open window is the second half of the period. The watchdog can only be cleared within the open window time. An attempt to clear the watchdog in the closed window will generate a reset. The Watchdog is cleared through the SPI by addressing the TIM register.

Refer to "table for reset pin operations" operation in mode 2.

### **6.1.9 Wake-up capabilities**

Several wake-up capabilities are available for the device when it is in sleep or stop mode. When a wake-up has occurred, the wake-up event is stored into the WUR or CAN registers. The MCU can then access the wake-up source. The wake-up options are selectable through the SPI while the device is in normal or standby mode, and prior to entering low power mode (Sleep or Stop mode).

### **6.1.10 Wake-up from wake-up inputs (L0, L1) without cyclic sense**

The wake-up lines are dedicated to sense external switch states, and when changes occur to wake-up the MCU (In sleep or stop modes). The wake-up pins are able to handle 40 V DC. The internal threshold is 3.0 V typical, and these inputs can be used as an input port expander. The wake-up inputs state can be read through the SPI (register WUR). L0 has a lower threshold than L1 in order to allow a connection and wake-up from a digital output such as a CAN physical interface.

## **6.1.11 Cyclic sense wake-up (cyclic sense timer and wake-up inputs L0, L1)**

The SBC can wake-up from a state change of one of the wake-up input lines (L0, L1), while the external pull-up or pull-down resistor of the switches associated to the wake-up input lines are biased with HS1 VSUP switch. The HS1 switch is activated in sleep or stop mode from an internal timer. Cyclic sense and forced wake-up are exclusive. If Cyclic sense is enabled, the forced wake-up can not be enabled.

### **6.1.12 Info for cyclic sense + dual edge selection**

In case the Cyclic sense and Lx both level sensitive conditions are use together, the initial value for Lx inputs are sampled in two cases:

1) When the register LPC[D3 and D0] are set and

2) At cyclic sense event, that is when device is in sleep or stop mode and HS1 is active.

The consequence is that when the device wake up by Lx transition, the new value is sampled as default, then when the device is set back into low power again, it will automatically wake up.

**The user should reset the LPC bits [D3 and D0] to 0 and set them again to the desired value prior to enter sleep or stop mode.**

### **6.1.13 Forced wake-up**

The SBC can wake-up automatically after a predetermined time spent in sleep or stop mode. Forced wake-up is enabled by setting bit FWU in the LPC register. Cyclic sense and forced wake-up are exclusive. If forced wake-up is enabled, the Cyclic sense can not be enabled.

## **6.1.14 CAN wake-up**

The device can wake-up from a CAN message. A CAN wake-up cannot be disabled.

### **6.1.15 SPI wake-up**

The device can wake-up by the CS pin in sleep or stop mode. Wake-up is detected by the CS pin transition from a low to high level. In stop mode this correspond to the condition where the MCU and SBC are both in Stop mode, and when the application wake-up events come through the MCU.

### **6.1.16 System power up**

At power up the device automatically wakes up.

### **6.1.17 Device power up, SBC wake up**

After device or system power up or a wake-up from sleep mode, the SBC enters into "reset mode" then into "normal request mode".

### **6.1.18 Battery fall early warning**

This function provides an interrupt when the VSUP voltage is below the 6.1 V typical. This interrupt is maskable. A hysteresis is included. Operation is only in Normal and Stand-by modes. VBAT low state reports in the IOR register.

### **6.1.19 Reset and Wdog operation**

The following figure shows the reset and watchdog output operations. Reset is active at device power up and wake-up. Reset is activated in case the VDD1 falls or the watchdog is not triggered. The WDOG output is active low as soon as the reset goes low and stays low for as long as the watchdog is not properly re-activated by the SPI.

The WDOG output pin is a push pull structure than can drive external components of the application, for instance to signal the MCU is in a wrong operation. Even if it is internally turned on (low-state), the reset pin can be forced to 5.0 V at 25 °C only, thanks to its internally limited current drive capability. The WDOG stays low until the Watchdog register is properly addressed through the SPI.



 **Figure 11. Reset and WDOG function diagram**

### **6.1.20 Debug mode Application hardware and software debug with the SBC.**

When the SBC is mounted on the same printed circuit board as the micro controller, it supplies both application software and the SBC with a dedicated routine that must be debugged. The following features allow the user to debug the software by disabling the SBC internal software watchdog timer.

### **6.1.21 Device power up, reset pin connected to VDD1**

At SBC power up, the VDD1 voltage is provided, but if no SPI communication occurs to configure the device, a reset occurs every 350 ms. In order to allow software debugging and avoid an MCU reset, the Reset pin can be connected directly to VDD1 by a jumper.

### **6.1.22 Debug modes with software watchdog disabled though SPI (normal debug, standby debug and stop debug)**

The software watchdog can be disabled through the SPI. In order to avoid unwanted watchdog disables, and to limit the risk of disabling the watchdog during an SBC normal operation, the watchdog disable has to be performed with the following sequence:

Step 1) Power down the SBC

Step 2) Power up the SBC (The BATFAIL bit is set, and the SBC enters normal request mode)

Step 3) Write to the TIM1 register to allow the SBC to enter Normal mode

Step 4) Write to the MCR register with data 0000 (this enables the debug mode). (Complete SPI byte: 000 1 0000)

Step 5) Write to the MCR register normal debug (0001 x101), stand-by debug (0001 x110), or Stop debug (0001 x111)

While in debug mode, the SBC can be used without having to clear the W/D on a regular basis to facilitate software and hardware debugging.

Step 6) To leave the debug mode, write 0000 to the MCR register.

To avoid entering the debug mode after a power up, first read the BATFAIL bit (MCR read) and write 0000 into the MCR. [Figure](#page-26-0) 12 illustrates entering the debug mode.





### <span id="page-26-0"></span>**6.1.23 MCU flash programming configuration**

To facilitate the possibility of down loading software into the application memory (MCU EEPROM or Flash), the SBC allows the following capabilities: The VDD1 can be forced by an external power supply to 5.0 V and the reset and WDOG output by external signal sources to zero or 5.0 V without damage. This supplies the complete application board with external power supply and applies the correct signal to the reset pin.

### <span id="page-27-0"></span>**6.1.24 CAN transceiver description**



 **Figure 13. Simplified block diagram of the CAN transceiver of the MC33889**

## **6.2 General description**

#### **6.2.1 CAN driver**

The CANH driver is a "high-side" switch to the V2 voltage (5.0 V). The CANL driver is a "low-side" switch to gnd.The turn on and turn off time is controlled in order to control the slew rate, and the CANH and CANL driver have a current limitation as well as an over temperature shutdown.

The CAN H or CANL driver can be disabled in case a failure is detected on the CAN bus (ex: CANH driver is disabled in case CANH is shorted to  $V_{DD}$ ). The disabling of one of the drivers is controlled by the CAN logic and the communication continues via the other drivers. When the failure is removed the logic detects a failure recovery and automatically reenables the associated driver.

The CAN drivers are also disabled in case of a Tx failure detection.

### **6.2.2 Bus termination**

The bus is terminated by pull-up and pull-down resistors, which are connected to GND, VDD, or VBAT through dedicated RTL and RTH pins and internal switches Srh, Srl, Stvbat. Each node must have a resistor connected between CANH and RTH and between CANL and RTL. The resistor value should be between 500 and 16000  $\Omega$ .

### **6.2.3 Transmitter function**

CAN bus levels are called Dominant and Recessive, and correspond respectively to Low and High states of the TX input pin.

Dominant state:

```
The CANH and CANL drivers are on. The voltage at CANL is <1.4 V, the voltage at CANH is >3.6 V, and the differential voltage between 
CANH and CANL line is >2.2 V (3.6 -1.4 V).
```
Recessive state:

This is a weak state, where the CANH and CANL drivers are off. The CANL line is pulled up to 5 V via the RTL pin and RTL resistor, and the CANH line is pull down via the RTH and RTH resistor. The resultant voltage at CANL is 5.0 V and 0V at CANH. The differential voltage is -5.0 V (0V - 5.0 V). The recessive state can be over written by any other node forcing a Dominant state.

## **6.2.4 Receiver function**

In normal operation (no bus failures), RX is the image of the differential bus voltage. The differential receiver inputs are connected to CANH and CANL.

The device incorporates single ended comparators connected to CANH and CANL in order to monitor the bus state as well as detect bus failures. Failures are reported via the SPI.

In normal operation when no failure is present, the differential comparator is active. Under a fault condition, one of the two CANH or CANL pins can be become non-operational. The single ended comparator of either CANH or CANL is activated and continues to report a bus state to Rx pin. The device permanently monitors the bus failure and recovery, and as soon as fault disappears, it automatically switches back to differential operation.

### **6.2.4.1 CAN interface operation mode**

The CAN has 3 operation modes: TxRx (Transmit-Receive), Receive Only, and Term-VBAT (Terminated to VBAT). The mode is selected by the SPI. As soon as the MC33889 mode is sleep or stop (selected via MCR register), the CAN interface automatically enters Tem-Vbat mode.

### **6.2.5 Tx Rx mode**

In this mode, the CAN drivers and receivers are enabled, and the device is able to send and receive messages. Bus failures are detected and managed, this means that in case of a bus failure, one of the CAN drivers can be disabled, but communication continues via the remaining drivers.

### **6.2.6 Receive only mode:**

In this mode, the transmitter path is disabled, so the device does not drive the bus. It maintains CANL and CANH in the recessive state. The receiver function operates normally.

### **6.2.7 TermVbat mode**

In this mode, the transmitter and receiver functions are disabled. The CANL pin is connected to  $V_{\text{SUP}}$  through the RTL resistor and internal pull up resistor of 12.5 kΩ. In this mode, the device monitors the bus activity and if a wake up conditions is encountered on the CAN bus, it will wakes up the MC33889.

The device will enter into a normal request mode if low power mode was in sleep, or generates an INT. It enters into Normal request mode if low power mode was in stop mode. If the device was in normal or stand by mode, the Rx pin will report a wake up (feature not available on the MC33889B). See Rx pin behavior.

## **6.2.8 Bus Failure Detection**

### **6.2.8.1 General description**

The device permanently monitors the bus lines and detects faults in normal and receive only modes. When a fault is detected, the device automatically takes appropriate actions to minimize the system current consumption and to allow communication on the network. Depending on the type of fault, the mode of operation, and the fault detected, the device automatically switches off one or more of the following functions: CANL or CANH line driver, RTL or RTH termination resistors, or internal switches. These actions are detailed in the following table.

The device permanently monitors the faults and in case of fault recovery, it automatically switches back to normal operation and reconnects the open functions. Fault detection and recovery circuitry have internal filters and delays timing, detailed in the AC characteristics parameters.

The failure list identification and the consequence on the device operation are described in following table. The failure detection, and recovery principle, the transceiver state after a failure detected, timing for failure detection and recovery can be found in the ISO11898-3 standard.

The following table is a summary of the failure identifications and of the consequences on the CAN driver and receiver when the CAN is in Tx Rx mode.



### **6.2.9 Open wire detection operation**

#### **6.2.9.1 Description**

The CANH and CANL open wire failures are not described in the ISO document. Open wire is only diagnostic information, as no CAN driver or receiver state will change in case of an open wire condition.

In case one of the CAN wires are open, the communication will continue through the remaining wire. In this situation the 33889 will receive information on one wire only and the consequences are as follows:

when the bus is set in dominant:

- The differential receiver will toggle
- Only one of the single ended receivers CANH or of CANL will toggle

The following figure illustrates the CAN signal during normal communication and in the example of a CANH open wire. The single ended receiver is sampled at the differential receiver switching event, in a window of 1.0 μs.







 **Figure 15. Open wire detection principle**

## **6.2.10 Open wire detection, MC33889B and D**

### **6.2.10.1 Failure detection**

The device will detect a difference in toggling counts between the differential receiver and one of the single ended receivers. Every time a difference in count is detected a counter is incremented. When the counter reaches 4, the device detects and reports an open wire condition. The open wire detection is performed only when the device receives a message and not when it send message.

Open wire recovery:

When the open wire failure has recovered, the difference in count is reduced and the device detects the open wire recovery.

#### <span id="page-31-0"></span>**6.2.10.2 MC33889B**

When detection is complete, the counter is no longer incremented. It can only be decremented by sampling of the dominant level on the S-H (S-L) (recovery pulse). When it reaches zero, the failure has recovered.

In application, with CAN communication, a recovery condition is detected after 4 acknowledge bits are sent by the MC33889B.

#### <span id="page-31-1"></span>**6.2.10.3 MC333889D**

When detection is complete, the counter is decremented by sampling the dominant pulse (recovery pulse) on S-H (S-L), and incremented (up to 4) by sampling the recessive pulse (detection pulses) on S-H (S-L). It is necessary to get 4 consecutive dominant samples (recovery pulse) to get to zero. When reaching zero, the failure is recovered.

In application with real CAN communication, a recovery condition will not be detected by a single acknowledge bit send by MC33889D, but requires a complete CAN message (at least 4 dominant bits) send in dual wire mode, without reception of any bit in single wire mode.

#### **6.2.10.4 Tx permanent dominant detection**

In addition to the previous list, the 33889 detects a permanent low state at the TX input which results in a permanent dominant bus state. If TX is low for more than 0.75-4.0 ms, the bus output driver is disabled. This avoids blocking communication between other nodes of the network. TXD is reported via the SPI (RCR register bit D1: TXFAILURE). Tx permanent dominant recovery is done with TX recessive for more than Typ. 32 μs.

### 6.2.10.5 Rx pin behavior while CAN interface is in TermV<sub>RAT</sub>

The MC33889D is able to signal bus activity on Rx while the CAN interface is in TermV $_{BAT}$  and the SBC in normal or standby mode. When the bus is driven into a dominant state by another sending node, each dominant state is reported at Rx by a low level, after a delay of t<sub>WAKE</sub>.

The bus state report is done through the CAN interface wake-up comparator on CANL and CANH, and thus operates also in case of bus failure. This is illustrated in [Figure](#page-32-0) 16.



t<sub>WAKE</sub>: duration of the CAN wake up filter, typ 16 μs. The MC33889D Rx dominant low level duration is the difference between the duration of the bus minus the  $t_{\text{WAKE}}$ , as illustrated below  $(t_{\text{RX\_DOM}} = t_{\text{BUS\_DOM}} - t_{\text{WAKE}})$ 



#### **Figure 16. Bus state report of the CAN interface wake-up comparator on CANL and CANH**

<span id="page-32-0"></span>The following table summarizes the device behavior when a CAN Wake-up event occurs.

**Table 6. Summary of RX pin operations for wake-up signaling**

<b>SBC</b> mode	<b>CAN state</b>	<b>MC33889B</b>	<b>MC33889D</b>
Normal	TermV <sub>BAT</sub>	no event on RX, no bit set	RX pulse (1), bit CANWU is not set
Standby	TermV <sub>BAT</sub>	no event on RX, no bit set	RX pulse (1), bit CANWU is not set
Sleep	TermV <sub>BAT</sub>	SBC mode transition to Normal request, bit CANWU set	SBC mode transition to Normal request, bit CANWU set
Stop	TermV <sub>BAT</sub>	INT pulse, bit CANWU set	Int pulse, bit CANWU set

Notes

28. pulse duration is bus dominant duration minus  $t_{\text{WAKE}}$ .

### **6.2.11 GND shift detection**

#### **6.2.11.1 General**

When normally working in two-wire operating mode, the CAN transmission can afford some ground shift between different nodes without trouble. Should a bus failure occur, the transceiver switches to single-wire operation, therefore working with less noise margin. The affordable ground shift is decreased.

The SBC provides a ground shift detection for diagnosis purpose. The four ground shift levels are selectable and the detection is stored in the IOR register which is accessible via the SPI.

#### **6.2.11.2 Detection principle**

The GND shift to detect is selected via the SPI from 4 different values (-0.3 V, -0.7 V, -1.2 V, -1.7 V). At each TX falling edge (end of recessive state), the CANH voltage is sensed. If it is detected to be below the selected gnd shift threshold, the bit SHIFT is set at 1 in the IOR register. No filter is implemented. Required filtering for reliable detection should be done by software (e.g. several trials).

### **6.2.12 Device state description**

#### **Table 7. 33889 table of operations**

The table below describes the SBC operation modes.





 **Figure 17. Simplified state machine**



 **Figure 18. Behavior at SBC power up**



 **Figure 19. Transitions to enter debug modes**





## **6.3 Logic commands and registers**

### **6.3.1 SPI interface**





The SPI is a 8 bit SPI. First 3 bits are used to identify the internal SBC register address, bit 4 is a read/write bit. The last 4 bits are data send from MCU to SBC or read back from SBC to MCU.

During write operation state of MISO has no signification.

During read operation only the last 4 bits at MISO have a meaning (content of the accessed register)

Following tables describe the SPI register list, and register bit meaning.

Registers "reset value" is also described, as well as the "reset condition". reset condition is the condition which cause the bit to be set at the "reset value".

Possible reset condition are:

Power On Reset: POR

SBC mode transition:

- NR2R Normal Request to Reset mode
- NR2N Normal Request to Normal mode
- N2R Normal to Reset mode
- STB2R Standby to Reset mode
- STO2R Stop to Reset mode
- SBC mode:RESET SBC in Reset mode

#### **Table 8. List of registers**



<span id="page-37-0"></span>NOTE: For SPI Operation

In case a low pulse is asserted by the device on the RST output pin during a SPI message, the SPI message can be corrupted. An RST low pulse is asserted in 2 cases:

Case 1: W/D refresh issue: The MCU does not perform the SPI watchdog refresh command before the expiration of the timeout (in Normal mode or Normal Request mode and if the "Timeout watchdog" option is selected), or the SPI watchdog refresh command is performed in the closed window (in Normal mode and if "Window watchdog" option is selected).

Case 2: VDD undervoltage condition: VDD falls below the VDD undervoltage threshold.

Message corruption means that the targeted register address can be changed, and another register is written. [Table 9](#page-38-0) shows the various cases and impacts on SPI register address:



#### <span id="page-38-0"></span>**Table 9. Possible corrupted registers in case of RST pulse during SPI communication**

Four registers can be corrupted: MCR, RCR, CAN, and IOR registers. As examples:

- write to CAN register can end up as write to MCR register, or
- write to TIM1 register can end up as write to RCR register

To avoid the previously described behavior, it is recommended to write into the MCR, RCR, CAN, and IOR registers with the expected configuration, after each RST assertion.

In the application, a RST low pulse leads to an MCU reset and a software restart. By applying this recommendation, all registers will be written with the expected configuration.

### **6.3.2 Register description**

#### **Table 10. MCR register**



#### **Table 11. Control bits**



#### Notes

<span id="page-38-1"></span>29. Bit BATFAIL cannot be set by SPI. BATFAIL is set when  $V_{\text{SUP}}$  falls below 3V.

<span id="page-38-2"></span>30. Watchdog ON or OFF depends on the RCR register bit D3.

- <span id="page-38-3"></span>31. Before entering sleep mode, bit NOSTOP in RCR register must be previously set to 1.
- <span id="page-38-4"></span>32. Stop command should be replaced by Stop Watchdog OFF. MCTR2=0, MCTR1= MCTR0=1

#### **Table 12. Status bits**



#### **RCR register**



#### **Table 13. Control bits**



#### **Table 14. CAN register**



### **6.3.3 Fault tolerant CAN transceiver standard modes**

The CAN transceiver standard mode can be programmed by setting CEXT to 0. The transceiver cell will then be behave as known from the 33889.

#### **Table 15. CAN transceiver modes**







Notes

<span id="page-40-0"></span>33. CEXT Bit should be set at 0. The CAN operation in extended mode is not recommended.

Fault tolerant CAN transceiver extended modes

By setting CEXT to 1 the transceiver cell supports sub bus communication



Note: if DFIS bit is set to 1, WUR register must be read before going into sleep or stop mode in order to clear the wake-up flag. During read out L0 must be at high level and should stay high when entering sleep or stop.

#### **Table 17. Status bits**



Comments:

CS2 bit at 0 = open failure. CS2 bit at 1 = short failure.

(CS3 bit at 0 and  $(CS1 = 1$  or  $CS2 = 1$ )) = CANH failure. CS3 bit at  $1 = CANL$  failure.

CS1 and CS0 bits: short type failure coding (GND, VDD or VBAT).

In case of multiple failures, the last failure is reported.

### . **Table 18. IOR register**



#### **Table 19. Control bits**



#### **Table 20. Gnd shift selection**



#### **Table 21. Status bits**



Notes

<span id="page-41-0"></span>34. Once the HS1 switch has been turned off because of overtemperature, it can be turned on again by setting the appropriate control bit to "1".

## **6.3.4 WUR register**

The local wake-up inputs L0 and L1 can be used in both normal and standby mode as port expander and for waking up the SBC in sleep or stop mode.

#### **Table 22. WUR register**



#### **Control bits**



#### **Table 23. Status bits**





### **6.3.5 TIM registers**

Description: This register is split into 2 sub registers, TIM1 and TIM2.

TIM1 controls the watchdog timing selection as well as the window or timeout option. TIM1 is selected when bit D3 is 0. TIM2 is used to define the timing for the cyclic sense and forced wake-up function. TIM2 is selected when bit D3 is 1. No read operation is allowed for registers TIM1 and TIM2

### **6.3.6 TIM register**

#### **Table 24. TIM register**



#### **Watchdog**







### **6.3.7 TIM2 register**

The purpose of TIM2 register is to select an appropriate timing for sensing the wake-up circuitry or cyclically supplying devices by switching on or off HS1

#### **Table 25. TIM2 register**



#### **Table 26. Cyclic sense timing**



j



### **6.3.8 LPC register**

Description: This register controls:

- The state of HS1 in Stop and Sleep mode (HS1 permanently off or HS1 cyclic)
- Enable or disable the forced wake-up function (SBC automatic wake-up after time spend in Sleep or Stop mode, time defined by the TIM2 register)
- Enable or disable the sense of the wake-up inputs (LX) at sampling point of the cyclic sense period (LX2HS1 bit).

#### **Table 27. LPC register**







#### **Table 28. INTR register**



#### **Table 29. Control bits**



When the mask bit has been set,  $\overline{\text{INT}}$  pin goes low if the appropriate condition occurs.

#### **Table 30. Status bits**



Notes:

If HS1OT-V2LOW interrupt is only selected (only bit D2 set in INTR register), reading INTR register bit D2 leads to two possibilities:

Bit D2 = 1: INT source is HS1OT

Bit D2 = 0: INT source is V2LOW.

Upon a wake-up condition from Stop mode due to overcurrent detection (I<sub>DD1S-WU1</sub> or I<sub>DD1S-WU2</sub>), an INT pulse is generated. However, the INTR register content remains at 0000 (not bit set into the INTR register).

# **7 Typical applications**



 **Figure 23. 33889D/33889B simplified typical application with ballast transistor**



 **Figure 24. 33889D/33889B simplified typical application without ballast transistor**

# **8 Packaging**

## **8.1 Package dimensions**

**Important** For the most current revision of the package, visit [www.nxp.com](http://www.nxp.com) and do a keyword search on the 98ASB42345B number listed below. Dimensions shown are provided for reference ONLY.







DETAIL A





NOTES:

- 1. DIMENSIONS ARE IN MILLIMETERS.
- 2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
- $\sqrt{3}$  this dimension does not include mold protrusion. Maximum mold protrusion 0.15 PER SIDE.
- 4. 751F-01 THRU -04 OBSOLETE. NEW STANDARD: 751F-05
- $\underbrace{\mathfrak{H}}$  This dimension does not include dam bar protrusion allowable dam bar protrusion shall be 0.13 total in excess of this dimension at maximum material condition.



# **9 Additional documentation**

## <span id="page-50-5"></span>**9.1 Thermal addendum (rev 2.0)**

#### **Introduction**

This thermal addendum is provided as a supplement to the MC33889 technical datasheet. The addendum provides thermal performance information that may be critical in the design and development of system applications. All electrical, application, and packaging information is provided in the datasheet.

#### **Packaging and thermal considerations**

The MC33889 is offered in a 28 pin SOICW, single die package. There is a single heat source (P), a single junction temperature  $(T_{\rm J})$ , and thermal resistance  $(R_{\rm BJA})$ .

$$
\{T_J\} = [R_{\theta JA}] \cdot \{P\}
$$

The stated values are solely for a thermal performance comparison of one package to another in a standardized environment. This methodology is not meant to and will not predict the performance of a package in an application-specific environment. Stated values were obtained by measurement and simulation according to the standards listed below.

#### **Standards**

#### **Table 31. Thermal performance comparison**



<span id="page-50-0"></span>Notes

- 1. Per JEDEC JESD51-2 at natural convection, still air condition.
- <span id="page-50-1"></span>2. 2s2p thermal test board per JEDEC JESD51-7.
- <span id="page-50-2"></span>3. Per JEDEC JESD51-8, with the board temperature on the center trace near the center lead.
- <span id="page-50-3"></span>4. Single layer thermal test board per JEDEC JESD51-3.
- <span id="page-50-4"></span>5. Thermal resistance between the die junction and the package top surface; cold plate attached to the package top surface and remaining surfaces insulated.





 **Figure 25. Surface mount for SOIC wide body non-exposed pad**





#### **Device on thermal test board**

Material:	Single layer printed circuit board FR4, 1.6 mm thickness Cu traces, 0.07 mm thickness
Outline:	80 mm x 100 mm board area, including edge connector for thermal testing
Area <b>A</b> :	Cu heat-spreading areas on board surface
<b>Ambient Conditions:</b>	Natural convection, still air

**Table 32. Thermal resistance performance**



 $R_{\theta JA}$  is the thermal resistance between die junction and ambient air.



 **Figure 27. Device on thermal test board R**θ**JA**



**Figure 28. Transient pin resistance**  $R_{\theta JA}$ **device on thermal test board area A = 600 (mm2)**

# **10 Revision history**



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