



TLC5955 48-Channel, 16-Bit, PWM LED Driver with DC, BC, LED Open-Short Detection, and Internal Current Setting

1 Features

- 48 Constant-Current Sink Output Channels
- Sink Current Capability with Maximum MC, DC, and BC Data:
 - 23.9 mA ($V_{CC} \leq 3.6$ V, MC = 5)
 - 31.9 mA ($V_{CC} > 3.6$ V, MC = 7)
- Grayscale (GS) Control:
 - 16-Bit (65,536 Steps) with Enhanced Spectrum or Conventional PWM
- Maximum Current (MC) Control:
 - 3 Bits (8 Steps) with a 3-mA to 30-mA Range
 - 3 MC Sets for Each Color Group
- Dot Correction (DC) Control:
 - 7 Bits (128 Steps) with a 26.2% to 100% Range
- Global Brightness Control (BC):
 - 7 Bits (128 Steps) with a 10% to 100% Range
 - 3 BC Sets for Each Color Group
- LED Power-Supply Voltage: Up to 10 V
- VCC: 3.0 V to 5.5 V
- Constant-Current Accuracy:
 - Channel-to-Channel: $\pm 2\%$ (typ), $\pm 5\%$ (max)
 - Device-to-Device: $\pm 2\%$ (typ), $\pm 4\%$ (max)
- Data Transfer Rate: 25 MHz
- Grayscale Control Clock: 33 MHz
- Auto Display Repeat
- Display Timing Reset
- Auto Data Refresh (GS and DC Only)
- LED Open Detection (LOD)
- LED Short Detection (LSD)
- UVLO Sets Default Data
- Delay Switching to Prevent Inrush Current
- Operating Temperature: -40°C to $+85^{\circ}\text{C}$

2 Applications

- LED Video Displays
- Variable Message Signs (VMS)
- Illumination

3 Description

The TLC5955 is a 48-channel, constant-current sink driver. Each channel has an individually-adjustable, pulse width modulation (PWM), grayscale (GS) brightness control with 65,536 steps and 128 steps of constant-current dot correction (DC). DC adjusts brightness deviation between channels. All channels have a 128-step global brightness control (BC). BC adjusts brightness deviation between the R, G, B color group. The eight-step maximum current control (MC) selects the maximum output current range for all channels of each color group. GS, DC, BC, and MC data are accessible with a serial interface port.

The TLC5955 has two error flags: LED open detection (LOD) and LED short detection (LSD). The error detection results can be read with a serial interface port.

Device Information

ORDER NUMBER	PACKAGE	BODY SIZE
TLC5955DCA	HTSSOP (56)	14,0 mm x 6,1 mm
TLC5955RTQ	QFN (56)	8,0 mm x 8,0 mm

Application Circuit

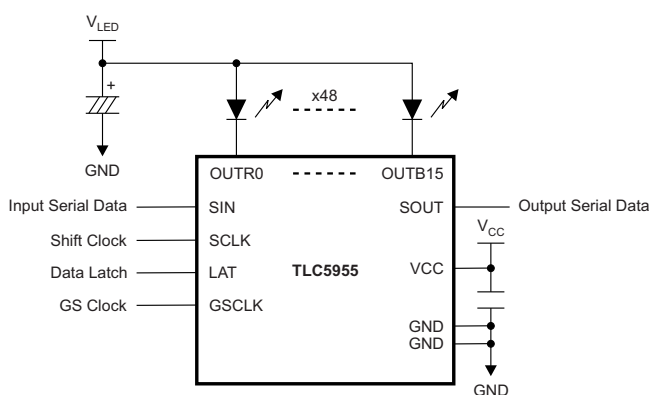


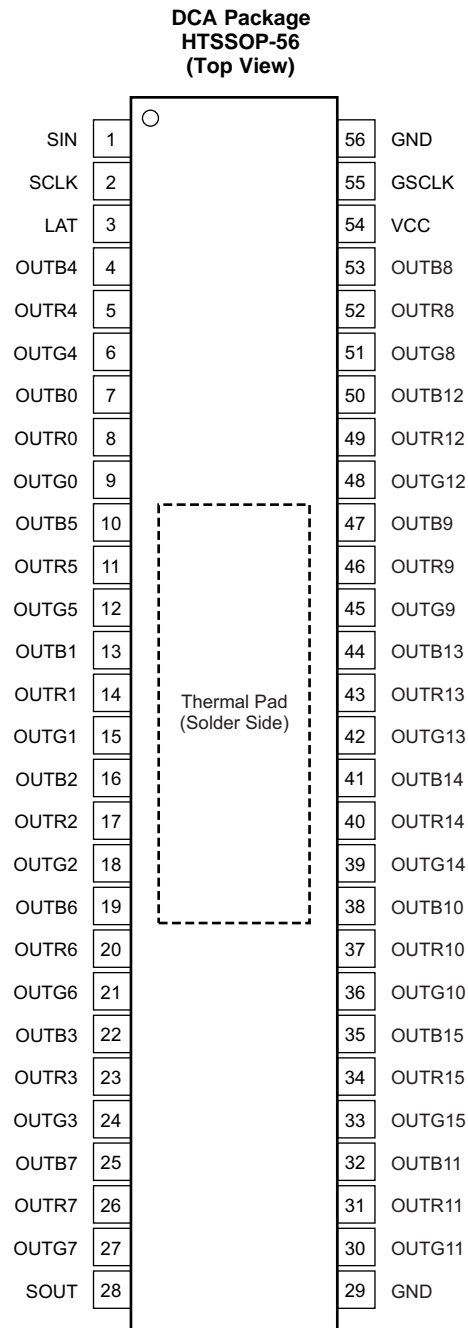
Table of Contents

1 Features	1	8.1 Overview	15
2 Applications	1	8.2 Functional Block Diagram	16
3 Description	1	8.3 Feature Description	17
4 Revision History	2	8.4 Device Functional Modes	28
5 Terminal Configurations and Functions	3	9 Applications and Implementation	38
6 Specifications	6	9.1 Application Information	38
6.1 Absolute Maximum Ratings	6	9.2 Typical Application	38
6.2 Handling Ratings	6	10 Power Supply Recommendations	41
6.3 Recommended Operating Conditions	7	11 Layout	41
6.4 Thermal Information	7	11.1 Layout Guidelines	41
6.5 Electrical Characteristics	8	11.2 Layout Example	42
6.6 Switching Characteristics	9	12 Device and Documentation Support	43
6.7 Typical Characteristics	10	12.1 Device Support	43
7 Parameter Measurement Information	12	12.2 Documentation Support	43
7.1 Terminal-Equivalent Input and Output Schematic Diagrams	12	12.3 Trademarks	43
7.2 Test Circuits	12	12.4 Electrostatic Discharge Caution	43
7.3 Timing Diagrams	13	12.5 Glossary	43
8 Detailed Description	15	13 Mechanical, Packaging, and Orderable Information	43

4 Revision History

DATE	REVISION	NOTES
March 2014	*	Initial release.

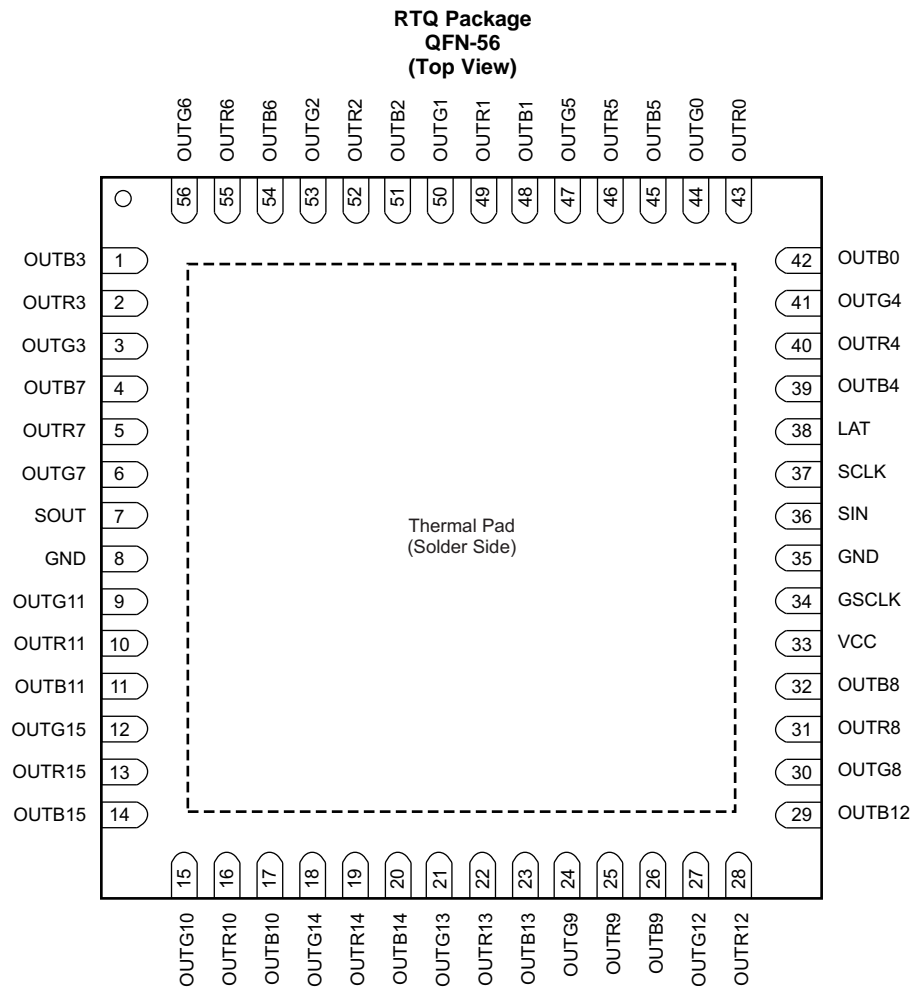
5 Terminal Configurations and Functions



TLC5955

SBVS237 – MARCH 2014

www.ti.com



Terminal Functions

TERMINAL			I/O	DESCRIPTION
NAME	DCA NUMBER	RTQ NUMBER		
GND	29, 56	8, 35	—	Power ground
GSCLK	55	34	I	Reference clock for the grayscale (GS) pulse width modulation (PWM) control for all outputs. Each GSCLK rising edge increments the grayscale counter for PWM control. When the LAT signal is input for a GS data write with the timing reset mode enabled, all constant-current outputs (OUTX0-OUTX15, where X = R, G, or B) are forced off, the grayscale counter is reset to 0, and the grayscale PWM timing controller is initialized.
LAT	3	38	I	The LAT rising edge either latches the data from the common shift register into the GS data latch when the MSB of the common shift register is 0 or latches the data into the control data latch when the MSB of the common shift register is 1. When the display timing reset bit (TMGRST) in the control data latch is 1, the grayscale counter initialized at the LAT signal is input for a grayscale data write. Dot correction (DC) data in the control data latch are copied to DC data latch at the same time.
OUTB0 to OUTB15	4, 7, 10, 13, 16, 19, 22, 25, 32, 35, 38, 41, 44, 47, 50, 53	1, 4, 11, 14, 17, 20, 23, 26, 29, 32, 39, 42, 45, 48, 51, 54	O	Constant-current outputs for the blue color group. Multiple outputs can be configured in parallel to increase the constant-current capability. Different voltages can be applied to each output.
OUTG0 to OUTG15	6, 9, 12, 15, 18, 21, 24, 27, 30, 33, 36, 39, 42, 45, 48, 51	3, 6, 9, 12, 15, 18, 21, 24, 27, 30, 41, 44, 47, 50, 53, 56	O	Constant-current outputs for the green color group. Multiple outputs can be configured in parallel to increase the constant-current capability. Different voltages can be applied to each output.
OUTR0 to OUTR15	5, 8, 11, 14, 17, 20, 23, 26, 31, 34, 37, 40, 43, 46, 49, 52	2, 5, 10, 13, 16, 19, 22, 25, 28, 31, 40, 43, 46, 49, 52, 55	O	Constant-current outputs for the red color group. Multiple outputs can be configured in parallel to increase the constant-current capability. Different voltages can be applied to each output.
SCLK	2	37	I	Serial data shift clock. Data present on SIN are shifted to the LSB of the common shift register with the SCLK rising edge. Data in the shift register are shifted toward the MSB at each SCLK rising edge. The MSB data of the common shift register appears on SOUT.
SIN	1	36	I	Serial data input for the 769-bit common shift register.
SOUT	28	7	O	This bit is the serial data output of the 769-bit common shift register. LED open detection (LOD) and LED short detection (LSD) can be read out with SOUT in the form of status information data (SID) after the LAT falling edge is input for a GS data write. SOUT is connected to the MSB of the 769-bit common shift register. Data are clocked out at the SCLK rising edge.
VCC	54	33	—	Power-supply voltage
Thermal pad			—	The thermal pad is not connected to GND internally. The thermal pad must be connected to GND via the PCB.

6 Specifications

6.1 Absolute Maximum Ratings⁽¹⁾

Over operating free-air temperature range, unless otherwise noted.

				MIN	MAX	UNIT
V _{CC}	Voltage ⁽²⁾	Supply		−0.3	+6.0	V
V _{IN}		Input range	SIN, SCLK, LAT, GSCLK	−0.3	V _{CC} + 0.3	V
V _{OUT}		Output range	SOUT	−0.3	V _{CC} + 0.3	V
			OUTR0 to OUTR15, OUTG0 to OUTG15, OUTB0 to OUTB15	−0.3	+11	V
T _J (max)	Maximum operating junction temperature				+150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages are with respect to device ground terminal.

6.2 Handling Ratings

		MIN	MAX	UNIT
T _{STG}	Storage temperature range	–55	+150	°C
V _{ESD} ⁽¹⁾	Human body model (HBM) ESD stress voltage ⁽²⁾		4000	V
	Charged device model (CDM) ESD stress voltage ⁽³⁾		2000	V

- (1) Electrostatic discharge (ESD) to measure device sensitivity and immunity to damage caused by assembly line electrostatic discharges in to the device.
- (2) Level listed above is the passing level per ANSI, ESDA, and JEDEC JS-001. JEDEC document JEP155 states that 4000-V HBM allows safe manufacturing with a standard ESD control process.
- (3) Level listed above is the passing level per EIA-JEDEC JESD22-C101. JEDEC document JEP157 states that 2000-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

PARAMETER		TEST CONDITIONS	MIN	NOM	MAX	UNIT
DC CHARACTERISTICS						
V _{CC}	Supply voltage		3.0		5.5	V
V _O	Voltage applied to output	OUTX0 to OUTX15 ⁽¹⁾			10	V
V _{IH}	High-level input voltage	SIN, SCLK, LAT, GSCLK	0.7 × V _{CC}		V _{CC}	V
V _{IL}	Low-level input voltage	SIN, SCLK, LAT, GSCLK	GND		0.3 × V _{CC}	V
I _{OH}	High-level output current	SOUT			−2	mA
I _{OL}	Low-level output current	SOUT			2	mA
I _{OLC}	Constant output sink current	OUTX0 to OUTX15 ⁽¹⁾ , 3 V ≤ V _{CC} ≤ 3.6 V			23.9	mA
		OUTX0 to OUTX15 ⁽¹⁾ , 3.6 V < V _{CC} ≤ 5.5 V			31.9	mA
T _A	Operating free-air temperature range		−40		+85	°C
T _J	Operating junction temperature range		−40		+125	°C
AC CHARACTERISTICS						
f _{CLK} (SCLK)	Data shift clock frequency	SCLK			25	MHz
f _{CLK} (GSCLK)	Grayscale control clock frequency	GSCLK			33	MHz
t _{WH0}	Pulse duration	SCLK	10			ns
t _{WL0}		SCLK	10			ns
t _{WH1}		GSCLK	10			ns
t _{WL1}		GSCLK	10			ns
t _{WH2}		LAT	30			ns
t _{SU0}	Setup time	SIN to SCLK↑	5			ns
t _{SU1}		LAT↓ to SCLK↑ (auto data refresh is disabled ⁽²⁾)	30			ns
t _{SU2}		LAT↑ for GS data written to GSCLK↑ when display time reset mode is disabled	50			ns
t _{SU3}		LAT↑ for GS data written to GSCLK↑ when display time reset mode is enabled	70			ns
t _{H0}	Hold time	SCLK↑ to SIN	2			ns
t _{H1}		SCLK↑ to LAT↑	5			ns

(1) X = R, G, or B.

(2) When auto data refresh is enabled, the first SCLK rising edge after the LAT signal input must be input after the first GSCLK is input.

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		DCA (HTSSOP)	RTQ (QFN)	UNIT
		56 TERMINALS	56 TERMINALS	
θ _{JA}	Junction-to-ambient thermal resistance	32.2	27.9	°C/W
θ _{JCTop}	Junction-to-case (top) thermal resistance	16.8	14.9	
θ _{JB}	Junction-to-board thermal resistance	16.1	6.5	
ψ _{JT}	Junction-to-top characterization parameter	0.8	0.3	
ψ _{JB}	Junction-to-board characterization parameter	16.0	6.4	
θ _{JCbot}	Junction-to-case (bottom) thermal resistance	0.9	2.0	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

6.5 Electrical Characteristics

At $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$ and $V_{CC} = 3\text{ V}$ to 5.5 V , unless otherwise noted. Typical values at $T_A = +25^\circ\text{C}$ and $V_{CC} = 3.3\text{ V}$.

PARAMETER		CONDITION	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage (SOUT)	$I_{OH} = -2\text{ mA}$	$V_{CC} - 0.4$		V_{CC}	V
V_{OL}	Low-level output voltage (SOUT)	$I_{OL} = 2\text{ mA}$			0.4	V
I_{IN}	Input current (SIN, SCLK, LAT, GSCLK)	$V_{IN} = V_{CC}$ or GND	-1		1	μA
I_{CC0}	Supply current (V_{CC})	SIN, SCLK, and LAT = GND, all $OUTX_n$ = off, GSCLK = GND, GSX n = 0000h, DCX n and BCX = 7Fh, $V_{OUTX_n} = 0.8\text{ V}$, MCX = 0 (3.2-mA target) ⁽¹⁾⁽²⁾		15	20	mA
I_{CC1}		SIN, SCLK, and LAT = GND, all $OUTX_n$ = off, GSCLK = GND, GSX n = 0000h, DCX n and BCX = 7Fh, $V_{OUTX_n} = 0.8\text{ V}$, MCX = 4 (19.1-mA target)		16	22	mA
I_{CC2}		SIN, SCLK, and LAT = GND, auto display repeat enabled, GSCLK = 33 MHz, GSX n = FFFFh, DCX n and BCX = 7Fh, $V_{OUTX_n} = 0.8\text{ V}$, MCX = 4 (19.1-mA target)		18	26	mA
I_{CC3}		$V_{CC} = 5.0\text{ V}$, SIN, SCLK, and LAT = GND, auto display repeat enabled, GSCLK = 33 MHz, GSX n = FFFFh, DCX n and BCX = 7Fh, $V_{OUTX_n} = 0.8\text{ V}$, MCX = 7 (31.9-mA target)		20	29	mA
I_{OLC0}	Constant output sink current (OUTX0 to OUTX15)	All $OUTX_n$ = on, DCX n and BCX = 7Fh, $V_{OUTX_n} = V_{OUTfix} = 0.8\text{ V}$, MCX = 4	17.4	19.1	20.8	mA
I_{OLC1}		$V_{CC} = 5.0\text{ V}$, all $OUTX_n$ = on, DCX n and BCX = 7Fh, $V_{OUTX_n} = V_{OUTfix} = 0.8\text{ V}$, MCX = 7	29.1	31.9	34.7	mA
I_{OLKG0}	Output leakage current (OUTX0 to OUTX15)	All OUT_n = off, $V_{OUTX_n} = V_{OUTfix} = 10\text{ V}$, MCX = 7	$T_J = +25^\circ\text{C}$		0.1	μA
I_{OLKG1}			$T_J = +85^\circ\text{C}$		0.2	μA
I_{OLKG2}			$T_J = +125^\circ\text{C}$		0.3	μA
ΔI_{OLC0}	Constant-current error (channel-to-channel, OUTX0 to OUTX15) ⁽³⁾	All $OUTX_n$ = on, DCX n and BCX = 7Fh, $V_{OUTX_n} = V_{OUTfix} = 0.8\text{ V}$, MCX = 4		$\pm 2\%$	$\pm 5\%$	
ΔI_{OLC1}	Constant-current error (device-to-device, OUTX0 to OUTX15) ⁽⁴⁾	All $OUTX_n$ = on, DCX n and BCX = 7Fh, $V_{OUTX_n} = V_{OUTfix} = 0.8\text{ V}$, MCX = 4		$\pm 2\%$	$\pm 4\%$	
ΔI_{OLC2}	Line regulation (OUTX0 to OUTX15) ⁽⁵⁾	$V_{CC} = 3.0\text{ V}$ to 5.5 V , all $OUTX_n$ = on, DCX n and BCX = 7Fh, $V_{OUTX_n} = V_{OUTfix} = 0.8\text{ V}$, MCX = 4		± 0.1	± 1	%/V

(1) X = R, G, or B. For example, MCX = MCR, MCG, and MCB.

(2) n = 0 to 15.

(3) The deviation of each output from the OUTX0 to OUTX15 constant-current average of the same color group. Deviation is calculated by the formula:

$$\Delta (\%) = \left[\frac{\frac{I_{OLCX_n}}{(I_{OLCX0} + I_{OLCX1} + \dots + I_{OLCX14} + I_{OLCX15})} - 1}{16} \right] \times 100$$

where X = R, G, or B; n = 0 to 15.

(4) Deviation of the OUTX0 to OUTX15 constant-current average from the ideal constant-current value. Deviation is calculated by the formula:

$$\Delta (\%) = \left[\frac{\left(\frac{I_{OLCX0} + I_{OLCX1} + \dots + I_{OLCX14} + I_{OLCX15}}{16} \right) - (\text{Ideal Output Current})}{\text{Ideal Output Current}} \right] \times 100$$

where X = R, G, or B; n = 0 to 15.

Ideal current is the target current when MC is 4.

(5) Line regulation is calculated by the formula:

$$\Delta (\%/V) = \left[\frac{(I_{OLCX_n} \text{ at } V_{CC} = 5.5\text{ V}) - (I_{OLCX_n} \text{ at } V_{CC} = 3.0\text{ V})}{(I_{OLCX_n} \text{ at } V_{CC} = 3.0\text{ V})} \right] \times \frac{100}{5.5\text{ V} - 3.0\text{ V}}$$

where X = R, G, or B; n = 0 to 15.

Electrical Characteristics (continued)

At $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ and $V_{CC} = 3\text{ V}$ to 5.5 V , unless otherwise noted. Typical values at $T_A = +25^{\circ}\text{C}$ and $V_{CC} = 3.3\text{ V}$.

PARAMETER	CONDITION	MIN	TYP	MAX	UNIT
ΔI_{OLC3}	Load regulation (OUTx0 to OUTx15) ⁽⁶⁾ All OUTXn = on, DCXn and BCX = 7Fh, $V_{OUTXn} = 0.8\text{ V}$ to 3.0 V , $V_{OUTfix} = 0.8\text{ V}$, MCX = 4		± 0.1	± 1	%/V
V_{LOD}	LED open-detection threshold All OUTXn = on	0.25	0.30	0.35	V
V_{LSD0}	LED short-detection threshold All OUTXn = on, LSDVLT = 0	$0.65 \times V_{CC}$	$0.70 \times V_{CC}$	$0.75 \times V_{CC}$	V
V_{LSD1}	All OUTXn = on, LSDVLT = 1	$0.85 \times V_{CC}$	$0.90 \times V_{CC}$	$0.95 \times V_{CC}$	V

(6) Load regulation is calculated by the equation:

$$\Delta (\%/V) = \left(\frac{(I_{OLCXn} \text{ at } V_{OUTXn} = 3\text{ V}) - (I_{OLCXn} \text{ at } V_{OUTXn} = 0.8\text{ V})}{I_{OLCXn} \text{ at } V_{OUTXn} = 0.8\text{ V}} \right) \times \frac{100}{3\text{ V} - 0.8\text{ V}}$$

where X = R, G, or B; n = 0 to 15.

6.6 Switching Characteristics

At $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$, $V_{CC} = 3\text{ V}$ to 5.5 V , $C_L = 15\text{ pF}$, $R_L = 120\ \Omega$, MCX = 7, and $V_{LED} = 4.5\text{ V}$, unless otherwise noted.

Typical values at $T_A = +25^{\circ}\text{C}$ and $V_{CC} = 3.3\text{ V}$.

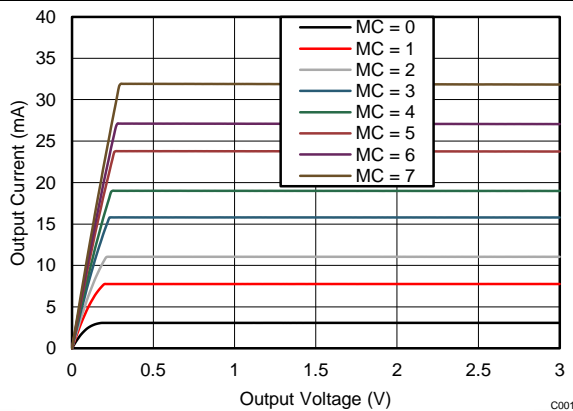
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{R0}	Rise time SOUT		3	5	ns
t_{R1}			30		ns
t_{F0}	Fall time SOUT		3	5	ns
t_{F1}			40		ns
t_{D0}	SCLK↑ to SOUT↑↓		20	30	ns
t_{D1}	Propagation delay $V_{CC} = 3.6\text{ V}$, GSCLK = 33 MHz, DCXn and BCX = 7Fh, $T_A = +25^{\circ}\text{C}$		40		ns
t_{D2}			43		ns
t_{D3}			46		ns
t_{D4}			49		ns
t_{D5}			52		ns
t_{D6}			55		ns
t_{D7}			58		ns
t_{D8}			61		ns
t_{ON_ERR}	Output on-time error ⁽²⁾ $t_{OUTON} - t_{GSCLK}$, $V_{CC} = 3.6\text{ V}$ to 5.5 V , GSXn = 0001h, GSCLK = 33 MHz, DCXn and BCX = 7Fh	-20		20	ns

(1) X = R, G, or B; n = 0 to 15.

(2) Output on-time error (t_{ON_ERR}) is calculated by the formula: $t_{ON_ERR} = t_{OUT_ON} - t_{GSCLK}$. t_{OUTON} is the actual on-time of the constant-current driver. t_{GSCLK} is the GSCLK period.

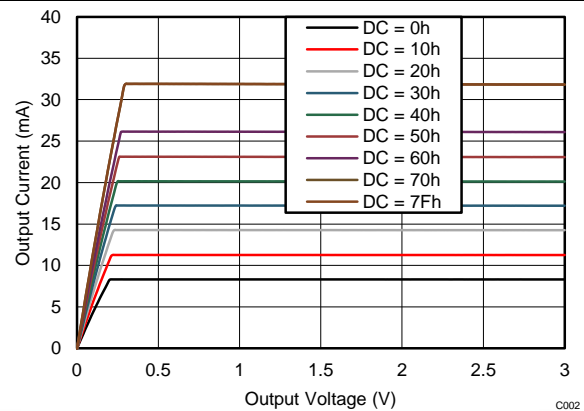
6.7 Typical Characteristics

At $T_A = +25^\circ\text{C}$ and $V_{CC} = 5.0\text{ V}$, unless otherwise noted.



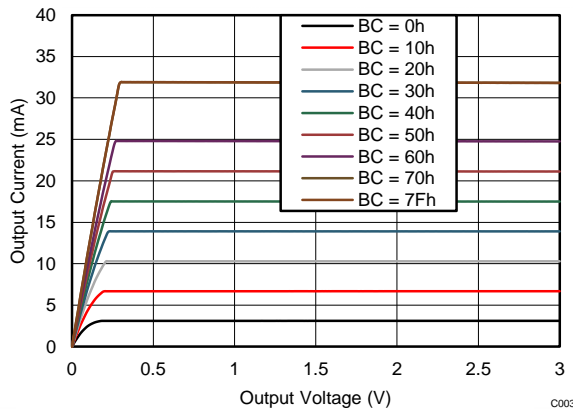
BCX = DCXn = 7Fh

Figure 1. Output Current vs Output Voltage (MCX Changing)



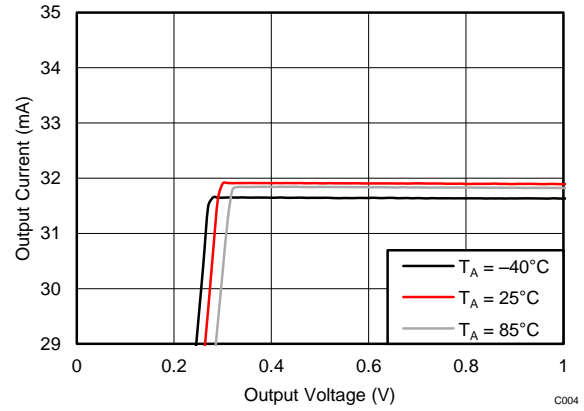
BCX = DCXn = 7Fh

Figure 2. Output Current vs Output Voltage (DCXn Changing)



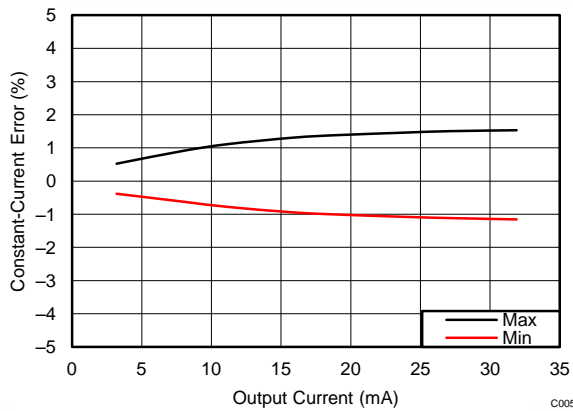
BCX = DCXn = 7Fh

Figure 3. Output Current vs Output Voltage (BCX Changing)



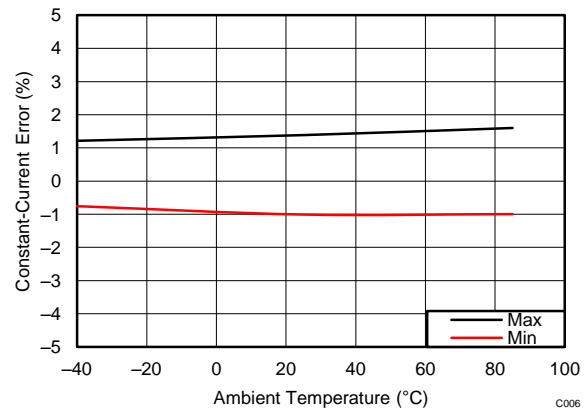
BCX = DCXn = 7Fh

Figure 4. Output Current vs Output Voltage (Temperature Changing)



BCX = DCXn = 7Fh $V_{OUTXn} = 0.8\text{ V}$

Figure 5. Constant-Current Error vs Output Current (Channel-to-Channel in Each Color Group)



MCX = 4 BCX = DCXn = 7Fh $V_{OUTXn} = 0.8\text{ V}$

Figure 6. Constant-Current Error vs Ambient Temperature (Channel-to-Channel in Each Color Group)

Typical Characteristics (continued)

At $T_A = +25^\circ\text{C}$ and $V_{CC} = 5.0\text{ V}$, unless otherwise noted.

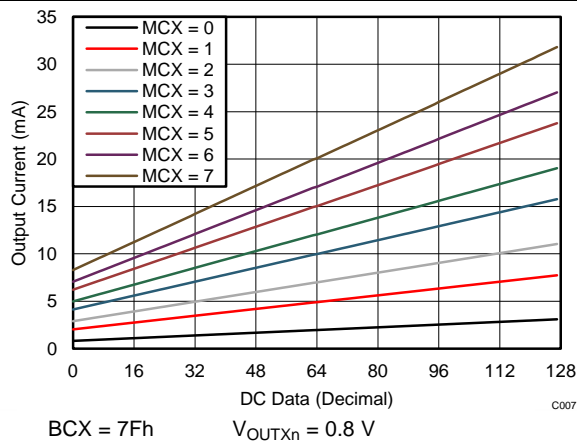


Figure 7. Dot Correction (DC) Linearity

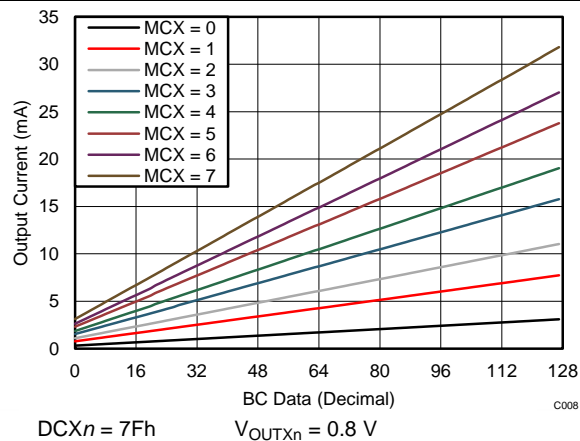


Figure 8. Global Brightness Control (BC) Linearity

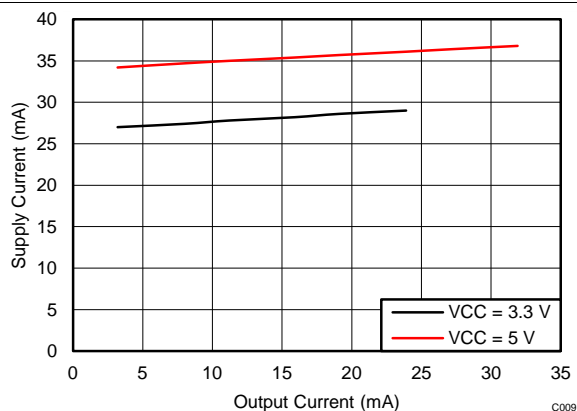


Figure 9. Supply Current vs Output Current

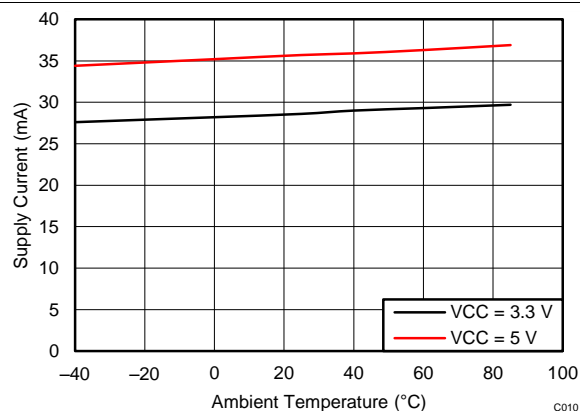


Figure 10. Supply Current vs Ambient Temperature

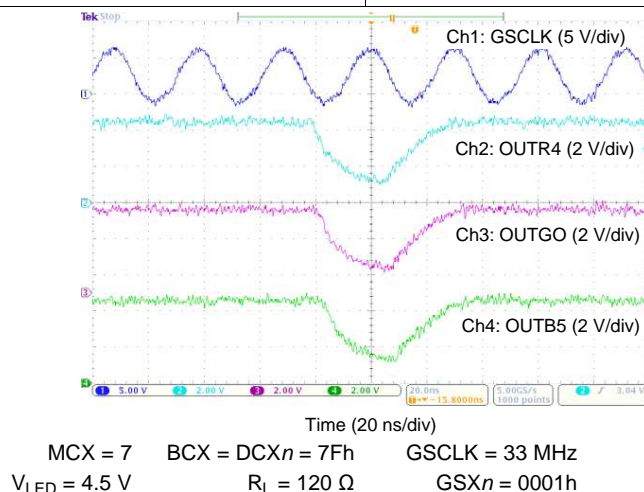


Figure 11. Constant-Current Output Voltage Waveform

7 Parameter Measurement Information

7.1 Terminal-Equivalent Input and Output Schematic Diagrams

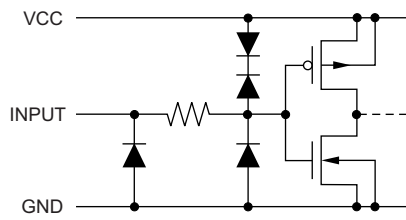


Figure 12. SIN, SCLK, LAT, GSKL

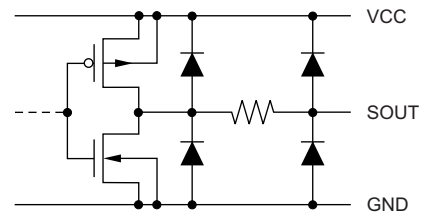
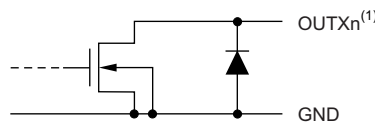


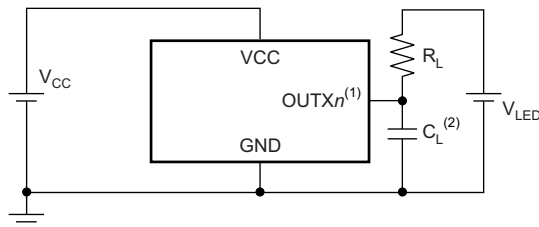
Figure 13. SOUT



(1) X = R, G, or B; n = 0 to 15.

Figure 14. OUTX0 Through OUTX15

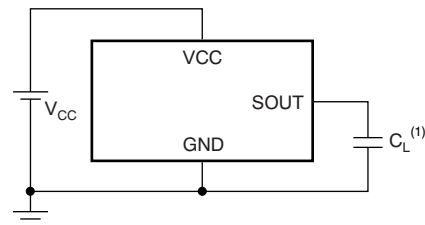
7.2 Test Circuits



(1) X = R, G, or B; n = 0 to 15.

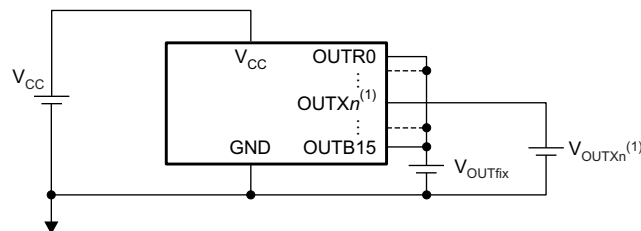
(2) CL includes measurement probe and jig capacitance.

Figure 15. Rise Time and Fall Time Test Circuit for OUTXn



(1) CL includes measurement probe and jig capacitance.

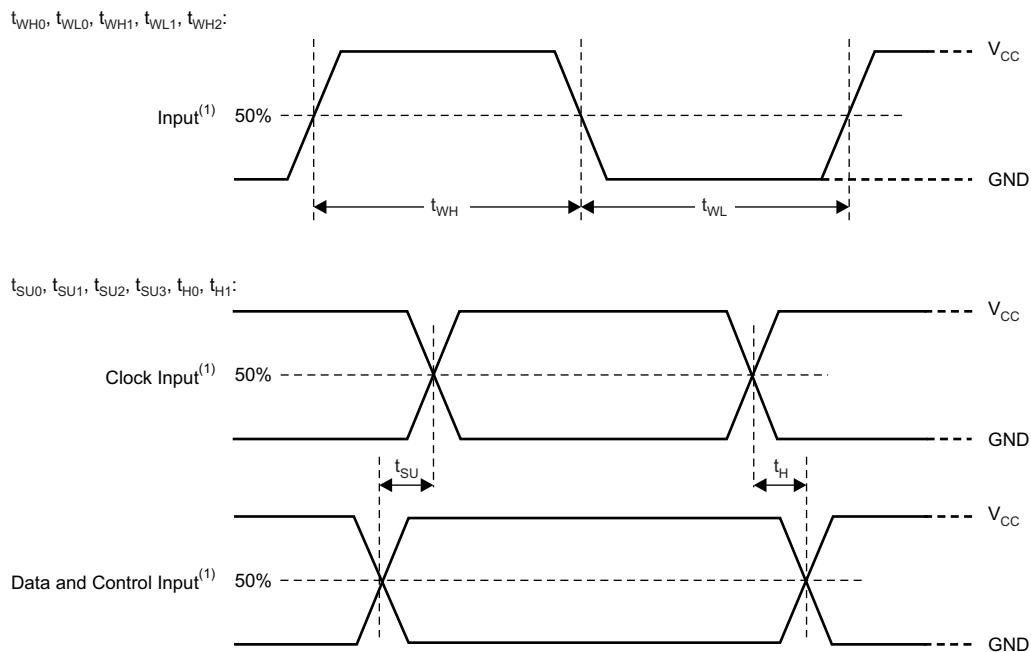
Figure 16. Rise Time and Fall Time Test Circuit for SOUT



(1) X = R, G, or B; n = 0 to 15.

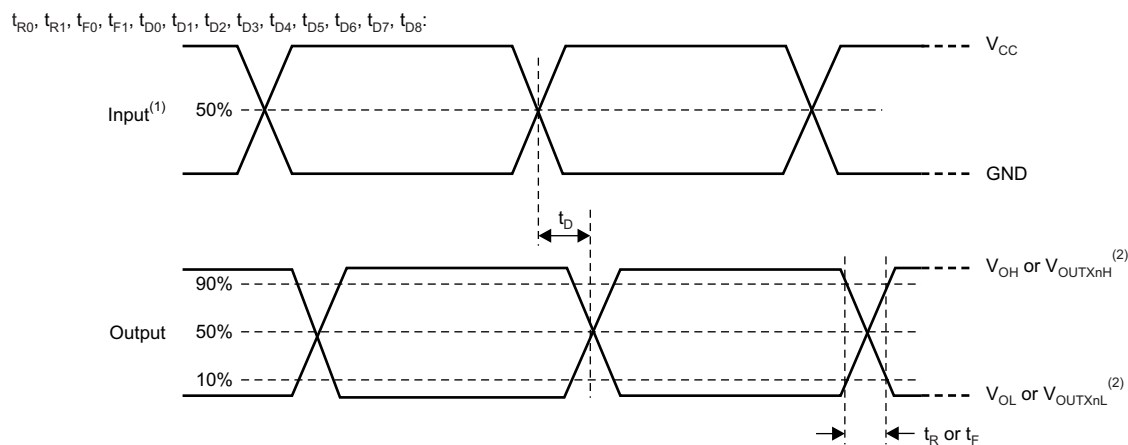
Figure 17. Constant-Current Test Circuit for OUTXn

7.3 Timing Diagrams



(1) Input pulse rise and fall time is 1 ns to 3 ns.

Figure 18. Input Timing



(1) Input pulse rise and fall time is 1 ns to 3 ns.

(2) X = R, G, or B; n = 0 to 15.

Figure 19. Output Timing

Timing Diagrams (continued)

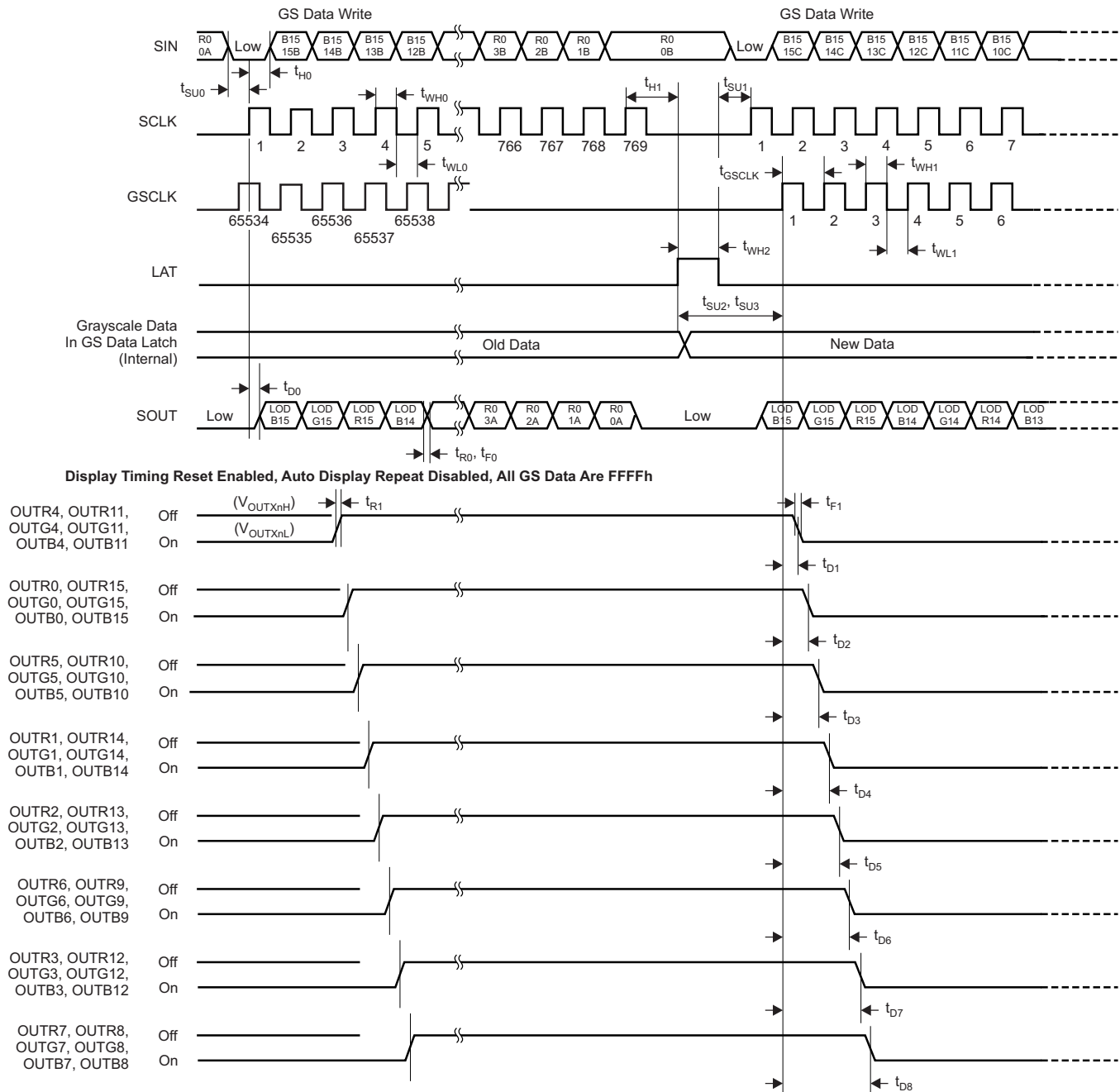


Figure 20. Data Input, Output, and Constant Output Timing

8 Detailed Description

8.1 Overview

The TLC5955 is 48-channel, 30-mA, constant-current LED driver that can control LED on-time with pulse width modulation (PWM) in 65,536 steps for grayscale (GS) control. A maximum of 281 trillion colors can be generated with red, green, and blue LEDs connected to the constant-current outputs.

The device has a 128-step, 7-bit, output current control function called *dot correction* (DC) that can control each constant-current output. Inherently, LED lamps have different intensities resulting from manufacturing differences. The DC function can reduce the inherent differences in intensity and improve LED lamp brightness uniformity.

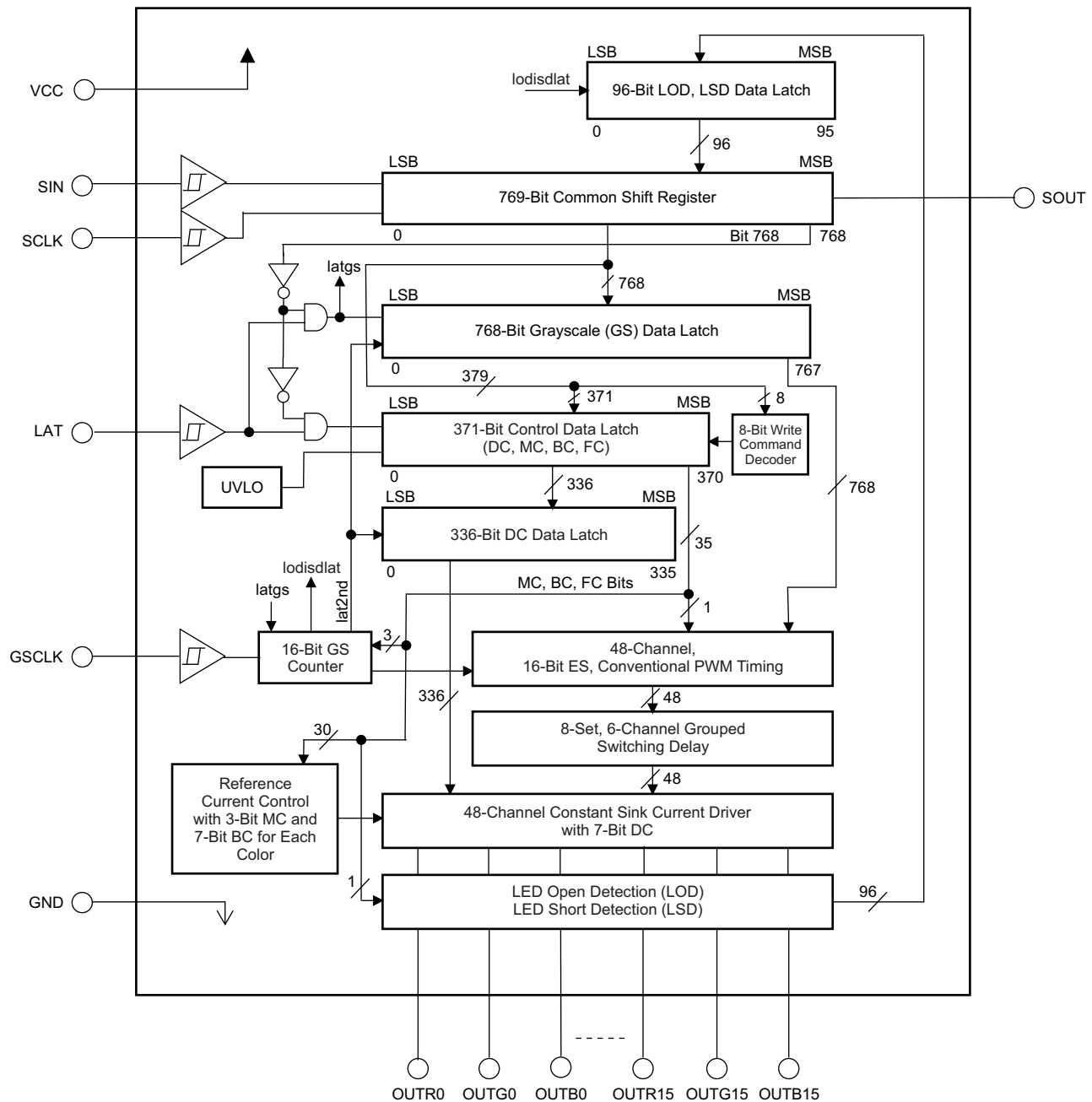
The device also has a 128-step, 7-bit, output current control function called *global brightness control* (BC) that can control each color group output. The BC function can adjust the red, green, and blue LED intensity for true white with constant-current control. The device contributes higher image quality to LED displays with fine white balance tuning by using these GS, DC, and BC functions.

The display controller can locate LED lamp failures via the device because the controller can detect LED lamp failures with the LED open detection (LOD) and LED short detection (LSD) functions and the reliability of the display can be improved by the LOD, LSD function.

The device maximum constant-current output value can be set by internal register data instead of the general method of using an external resistor setting. Thus, any failure modes that occur from the external resistor can be eliminated and one resistor can be eliminated.

The device constant-current output can drive approximately 19 mA at a 0.25-V output voltage and a +25°C ambient temperature. This voltage is called *knee* voltage. This 0.25-V, low-knee voltage can contribute to the design of a lower-power display system. The total number of LED drivers on one LED display panel can be reduced because 48 LED lamps can be driven by one LED driver. Therefore, designing fine-pitch LED displays is simplified.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Output Current Calculation

The output current value controlled by MC, DC, and BC can be calculated by [Equation 1](#).

$$I_{OUTn} \text{ (mA)} = I_{OLCMax} \text{ (mA)} \times \left[0.262 + 0.738 \times \frac{DCXn}{127} \right] \times \left[0.10 + 0.90 \times \frac{BCX}{127} \right]$$

where:

- I_{OLCMax} = the maximum constant-current value for all $OUTXn$ for each color group programmed by MC data,
- $DCXn$ = the dot correction value for each channel (0h to 7Fh),
- BCX = the global brightness control value (0h to 7Fh),
- $X = R, G, \text{ or } B$ for the red, green, or blue color group, and
- $n = 0 \text{ to } 15$.

(1)

Each output sinks the I_{OLCMax} current when they turn on and the dot correction (DC) data and the global brightness control (BC) data are set to the maximum value of 7Fh (127d). Each output sink current can be reduced by lowering the DC and BC values.

When I_{OUT} is set lower than 1 mA by both MC and BC or BC only, the output may be unstable. Output currents lower than 1 mA can be achieved by setting I_{OUT} to 1 mA with MC and BC or BC only and then using DC to lower the output current.

8.3.2 Register and Data Latch Configuration

The TLC5955 has one common shift register and three data latches: the grayscale (GS) data latch, the control data latch, and the dot correction (DC) data latch. The common shift register is 769 bits long, the GS data latch is 768 bits long, the control data latch is 371 bits long, and the DC data latch is 336 bits long.

If the common shift register MSB is 0, the least significant 768 bits from the common shift register are latched into the GS data latch. If the MSB is 1, and bits 767 to 760 are 96h (10010110b), the data are latched into the control data latch. Refer to [Figure 21](#) for the common shift register, GS data latch, control data latch, and DC data latch configurations.

Feature Description (continued)

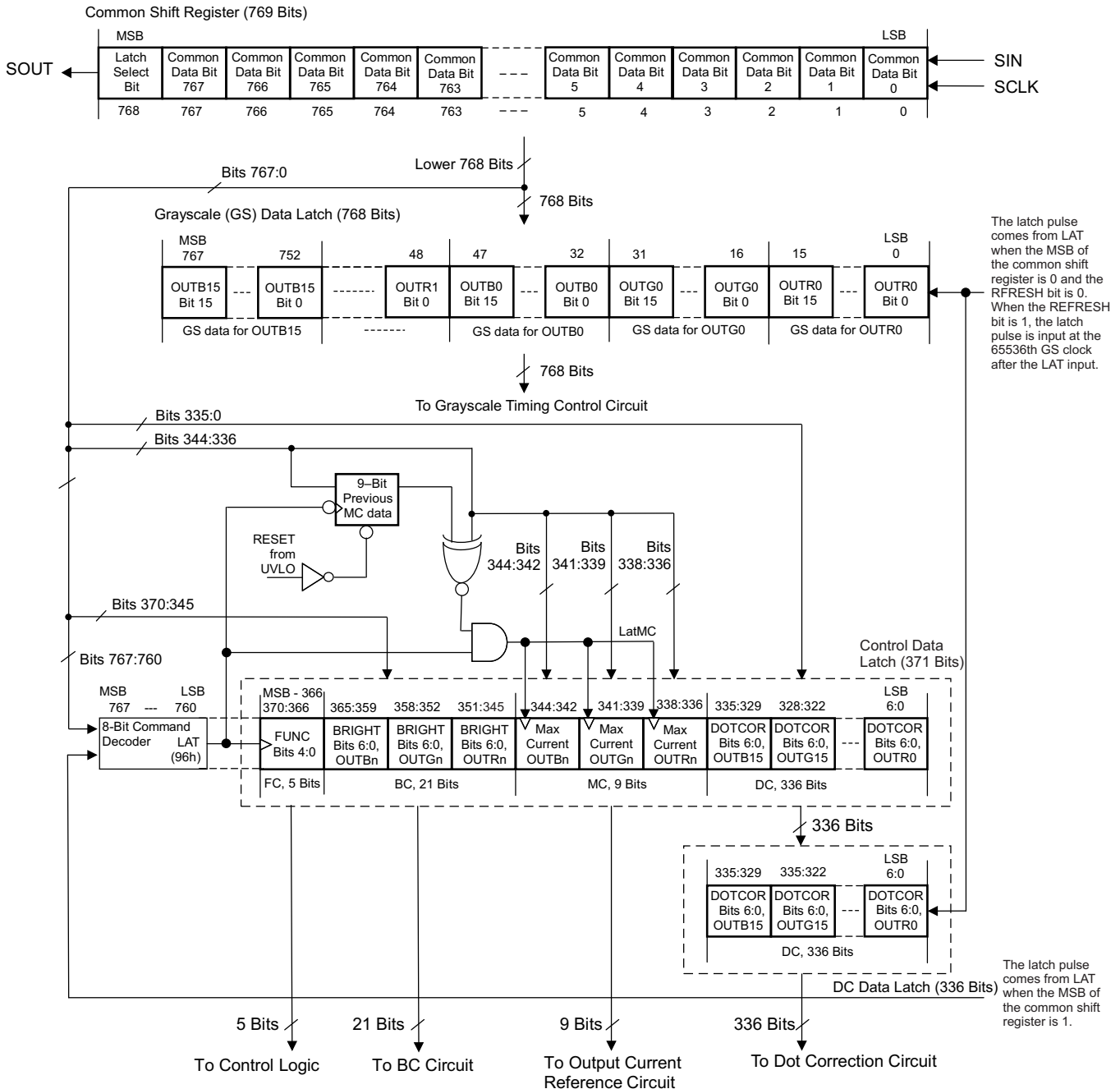


Figure 21. Common Shift Register and Data Latches Configuration

8.3.2.1 769-Bit Common Shift Register

The 769-bit common shift register is used to shift data from the SIN terminal into the TLC5955. The data shifted into the register are used for GS, DC, maximum output current, global BC functions, and function control data write operations. The common shift register LSB is connected to SIN and the MSB is connected to SOUT. On each SCLK rising edge, the data on SIN are shifted into the LSB and all 769 bits are shifted towards the MSB. The register MSB is always connected to SOUT. When the device is powered up, the data in the 769-bit common shift register are random.

Feature Description (continued)

8.3.2.2 Grayscale (GS) Data Latch

The GS data latch is 768 bits long, and sets the PWM timing for each constant-current output. The on-time of all constant-current outputs is controlled by the data in this data latch. The 768-bit GS data in the common shift register are copied to the data latch at a LAT rising edge when the common shift register MSB is 0.

When the device is powered up, the data are random and all constant-current outputs are forced off. However, no outputs turn on until GS data are written to the GS data latch even if a GSCLK is input. The data bit assignment is shown in [Table 1](#). Refer to [Figure 22](#) for a GS data write timing diagram.

Table 1. Grayscale Data Latch Bit Description

GS DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL	GS DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL
15-0	GSR0[15:0]	N/A (no default value)	Bits[15:0] for OUTR0	399-384	GSR8[15:0]	N/A (no default value)	Bits[15:0] for OUTR8
31-16	GSG0[15:0]		Bits[15:0] for OUTG0	415-400	GSG8[15:0]		Bits[15:0] for OUTG8
47-32	GSB0[15:0]		Bits[15:0] for OUTB0	431-416	GSB8[15:0]		Bits[15:0] for OUTB8
63-48	GSR1[15:0]		Bits[15:0] for OUTR1	447-432	GSR9[15:0]		Bits[15:0] for OUTR9
79-64	GSG1[15:0]		Bits[15:0] for OUTG1	463-448	GSG9[15:0]		Bits[15:0] for OUTG9
95-80	GSB1[15:0]		Bits[15:0] for OUTB1	479-464	GSB9[15:0]		Bits[15:0] for OUTB9
111-96	GSR2[15:0]		Bits[15:0] for OUTR2	495-480	GSR10[15:0]		Bits[15:0] for OUTR10
127-112	GSG2[15:0]		Bits[15:0] for OUTG2	511-496	GSG10[15:0]		Bits[15:0] for OUTG10
143-128	GSB2[15:0]		Bits[15:0] for OUTB2	527-512	GSB10[15:0]		Bits[15:0] for OUTB10
159-144	GSR3[15:0]		Bits[15:0] for OUTR3	543-528	GSR11[15:0]		Bits[15:0] for OUTR11
175-160	GSG3[15:0]		Bits[15:0] for OUTG3	559-544	GSG11[15:0]		Bits[15:0] for OUTG11
191-176	GSB3[15:0]		Bits[15:0] for OUTB3	575-560	GSB11[15:0]		Bits[15:0] for OUTB11
207-192	GSR4[15:0]		Bits[15:0] for OUTR4	591-576	GSR12[15:0]		Bits[15:0] for OUTR12
223-208	GSG4[15:0]		Bits[15:0] for OUTG4	607-592	GSG12[15:0]		Bits[15:0] for OUTG12
239-224	GSB4[15:0]		Bits[15:0] for OUTB4	623-608	GSB12[15:0]		Bits[15:0] for OUTB12
255-240	GSR5[15:0]		Bits[15:0] for OUTR5	639-624	GSR13[15:0]		Bits[15:0] for OUTR13
271-256	GSG5[15:0]		Bits[15:0] for OUTG5	655-640	GSG13[15:0]		Bits[15:0] for OUTG13
287-272	GSB5[15:0]		Bits[15:0] for OUTB5	671-656	GSB13[15:0]		Bits[15:0] for OUTB13
303-288	GSR6[15:0]		Bits[15:0] for OUTR6	687-672	GSR14[15:0]		Bits[15:0] for OUTR14
319-304	GSG6[15:0]		Bits[15:0] for OUTG6	703-688	GSG14[15:0]		Bits[15:0] for OUTG14
335-320	GSB6[15:0]		Bits[15:0] for OUTB6	719-704	GSB14[15:0]		Bits[15:0] for OUTB14
351-336	GSR7[15:0]		Bits[15:0] for OUTR7	735-720	GSR15[15:0]		Bits[15:0] for OUTR15
367-352	GSG7[15:0]		Bits[15:0] for OUTG7	751-736	GSG15[15:0]		Bits[15:0] for OUTG15
383-368	GSB7[15:0]		Bits[15:0] for OUTB7	767-752	GSB15[15:0]		Bits[15:0] for OUTB15

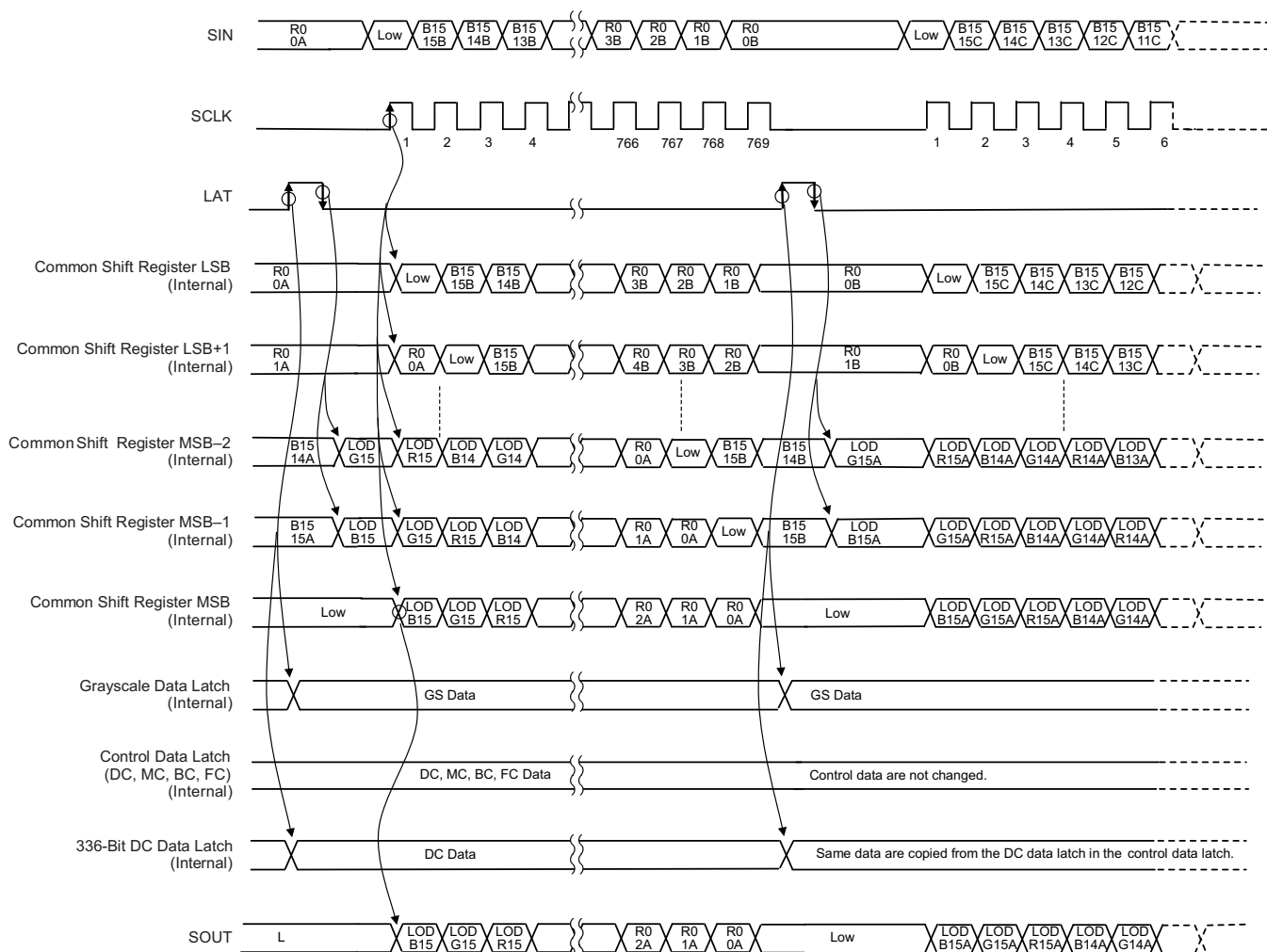


Figure 22. Grayscale Data Write Timing Diagram (RFRESH = 0)

8.3.2.3 Control Data Latch

The control data latch is 371 bits long. The data latch contains dot correction (DC) data, maximum current (MC) data, global brightness control (BC) data, and function control (FC) data. The DC for each constant-current output are controlled by the data in the DC data latch. The control data in the data latch are updated with the lower 371 bits of the common shift register at the LAT rising edge when the common shift register MSB is 1. The 336 bits of DC data are copied from the control data latch when the 65,536th GSCLK is input with RFRESH set to 1 in the control data latch after the GS data are written or the LAT rising edge for GS data writes is input when the RFRESH bit is 0.

When the device is powered up, the data in the control data latch (except the MC bits) are random. Therefore, DC, BC, and FC data must be written to the control data latch before turning on the constant-current outputs. Furthermore, MC data should be set appropriately for the application. Refer to [Figure 23](#) for a control data write timing diagram.

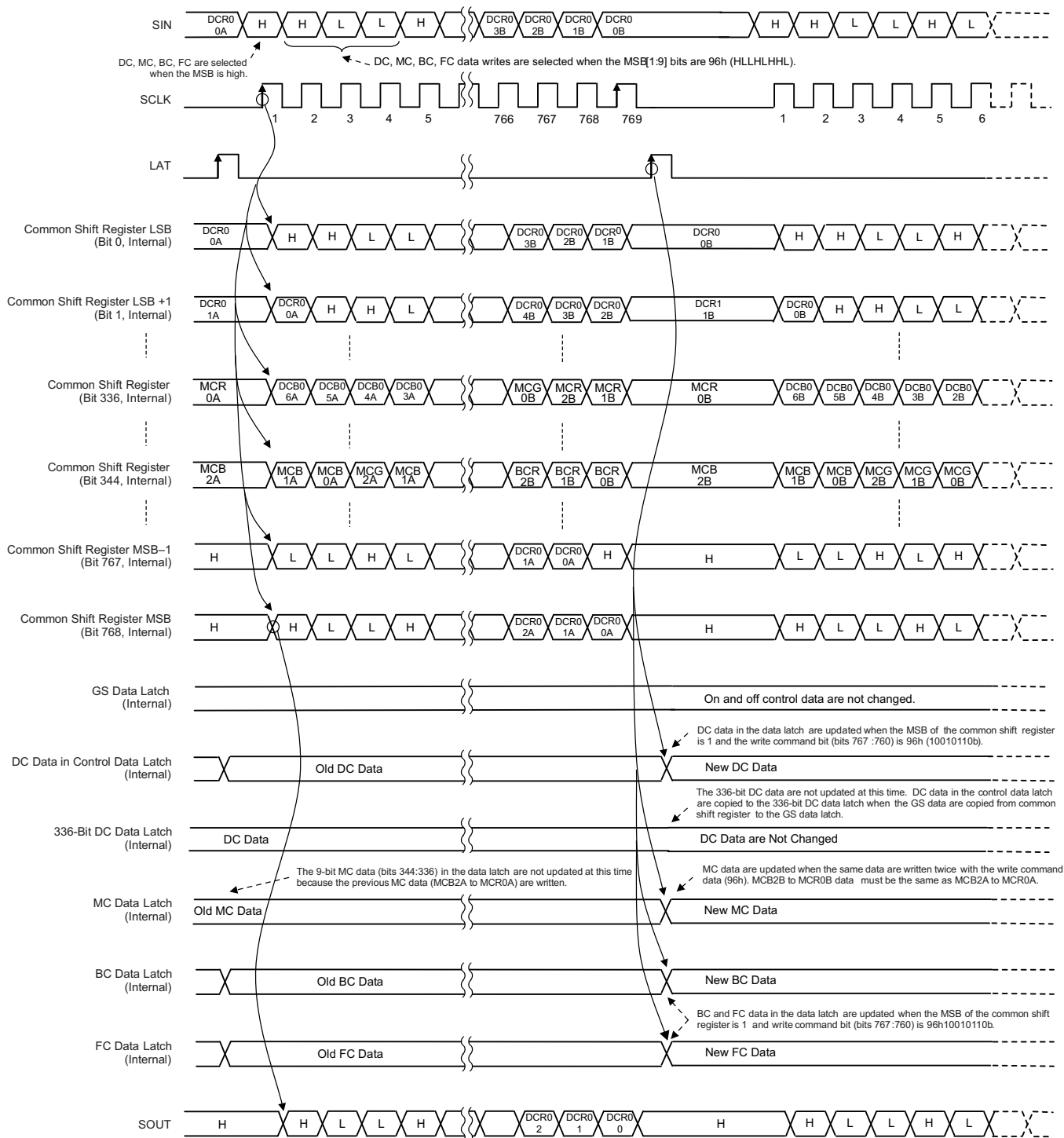


Figure 23. Control Data Write Timing Diagram for DC, MC, BC, and FC

8.3.2.4 Dot Correction (DC) Data Latch

DC data are 336 bits long; the data for each constant-current output are controlled by seven bits. Each constant-current output DC is controlled by the DC data latch. Each DC value individually adjusts the output current for each constant-current output. As explained in the [Dot Correction \(DC\) Function](#) section, the DC values are used to adjust the output current from 26.2% to 100% of the current value set by MC and BC data. When the device is powered on, the data in the DC data latch are random.

The DC data bit assignment is shown in [Table 2](#). See [Table 9](#) for a summary of the DC data value versus set current value.

Table 2. Dot Correction Data Bit Description

CONTROL DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL	CONTROL DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL
6-0	DCR0[6:0]	N/A (no default value)	DC bits[6:0] for OUTR0	174-168	DCR8[6:0]	N/A (no default value)	DC bits[6:0] for OUTR8
13-7	DCG0[6:0]		DC bits[6:0] for OUTG0	181-175	DCG8[6:0]		DC bits[6:0] for OUTG8
20-14	DCB0[6:0]		DC bits[6:0] for OUTB0	188-182	DCB8[6:0]		DC bits[6:0] for OUTB8
27-21	DCR1[6:0]		DC bits[6:0] for OUTR1	195-189	DCR9[6:0]		DC bits[6:0] for OUTR9
34-28	DCG1[6:0]		DC bits[6:0] for OUTG1	202-196	DCG9[6:0]		DC bits[6:0] for OUTG9
41-35	DCB1[6:0]		DC bits[6:0] for OUTB1	209-203	DCB9[6:0]		DC bits[6:0] for OUTB9
48-42	DCR2[6:0]		DC bits[6:0] for OUTR2	216-210	DCR10[6:0]		DC bits[6:0] for OUTR10
55-49	DCG2[6:0]		DC bits[6:0] for OUTG2	223-217	DCG10[6:0]		DC bits[6:0] for OUTG10
62-56	DCB2[6:0]		DC bits[6:0] for OUTB2	230-224	DCB10[6:0]		DC bits[6:0] for OUTB10
69-63	DCR3[6:0]		DC bits[6:0] for OUTR3	237-231	DCR11[6:0]		DC bits[6:0] for OUTR11
76-70	DCG3[6:0]		DC bits[6:0] for OUTG3	244-238	DCG11[6:0]		DC bits[6:0] for OUTG11
83-77	DCB3[6:0]		DC bits[6:0] for OUTB3	251-245	DCB11[6:0]		DC bits[6:0] for OUTB11
90-84	DCR4[6:0]		DC bits[6:0] for OUTR4	258-252	DCR12[6:0]		DC bits[6:0] for OUTR12
97-91	DCG4[6:0]		DC bits[6:0] for OUTG4	265-259	DCG12[6:0]		DC bits[6:0] for OUTG12
104-98	DCB4[6:0]		DC bits[6:0] for OUTB4	272-266	DCB12[6:0]		DC bits[6:0] for OUTB12
111-105	DCR5[6:0]		DC bits[6:0] for OUTR5	279-273	DCR13[6:0]		DC bits[6:0] for OUTR13
118-112	DCG5[6:0]		DC bits[6:0] for OUTG5	286-280	DCG13[6:0]		DC bits[6:0] for OUTG13
125-119	DCB5[6:0]		DC bits[6:0] for OUTB5	293-287	DCB13[6:0]		DC bits[6:0] for OUTB13
132-126	DCR6[6:0]		DC bits[6:0] for OUTR6	300-294	DCR14[6:0]		DC bits[6:0] for OUTR14
139-133	DCG6[6:0]		DC bits[6:0] for OUTG6	307-301	DCG14[6:0]		DC bits[6:0] for OUTG14
146-140	DCB6[6:0]		DC bits[6:0] for OUTB6	314-308	DCB14[6:0]		DC bits[6:0] for OUTB14
153-147	DCR7[6:0]		DC bits[6:0] for OUTR7	321-315	DCR15[6:0]		DC bits[6:0] for OUTR15
160-154	DCG7[6:0]		DC bits[6:0] for OUTG7	328-322	DCG15[6:0]		DC bits[6:0] for OUTG15
167-161	DCB7[6:0]		DC bits[6:0] for OUTB7	335-329	DCB15[6:0]		DC bits[6:0] for OUTB15

8.3.2.5 Maximum Current (MC) Data Latch

The maximum output current per channel, I_{OLCMax} , is programmed by MC data and can be set with the serial interface. I_{OLCMax} is the largest current for each output. Each output sinks the I_{OLCMax} current when they turn on with DC and BC data set to the maximum value of 7Fh (127d). MC data must have the same data continuously written twice in order to change the data. When the device is powered on, the MC data are set to 0.

The MC data bit assignment is shown in [Table 3](#). See [Table 8](#) for a summary of the MC data value for each color group versus the set current value.

Table 3. Maximum Current Data Bit Assignment in the Control Data Latch

CONTROL DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL
338-336	MCR[2:0]	0	MC bits[2:0] for red color group channels (OUTR0 to OUTR15)
341-339	MCG[2:0]	0	MC bits[2:0] for green color group channels (OUTG0 to OUTG15)
344-342	MCB[2:0]	0	MC bits[2:0] for blue color group channels (OUTB0 to OUTB15)

8.3.2.6 Global Brightness Control (BC) Data Latch

Global BC data are seven bits long. The global brightness for all outputs is controlled by the data in the control data latch. The data are used to adjust the constant-current values for the 48-channel constant-current outputs. As explained in the [Global Brightness Control \(BC\) Function](#) section, the BC values are used to adjust the output current from 10% to 100% of the maximum value. When the device is powered on, the BC data are random.

The global BC data bit assignment in the control data latch is shown in [Table 4](#). See [Table 10](#) for a summary of the BC data value versus set current value.

Table 4. Global Brightness Control Data Bit Assignment in the Control Data Latch

CONTROL DATA LATCH BIT NUMBER	BIT NAME	DEFAULT VALUE	CONTROLLED CHANNEL
351-345	BCR[6:0]	N/A (no default value)	BC bits[6:0] for red color group channels (OUTR0 to OUTR15)
358-352	BCG[6:0]		BC bits[6:0] for green color group channels (OUTG0 to OUTG15)
365-359	BCB[6:0]		BC bits[6:0] for blue color group channels (OUTB0 to OUTB15)

8.3.2.7 Function Control (FC) Data Latch

The FC data latch is 5 bits long. This latch enables the auto display repeat and display timing reset functions, and sets the DC data auto refresh, PWM control mode, and the LSD detection voltage. Each function is selected by the data in the control data latch. When the device is powered on, the FC data are random. The FC data bit assignment in the control data latch is shown in [Table 5](#).

Table 5. Function Control Data Latch Bit Description

BIT NUMBER	BIT NAME	DEFAULT VALUE (Binary)	DESCRIPTION
366	DSRPT	N/A (no default value)	Auto display repeat mode enable bit 0 = Disabled, 1 = Enabled When this bit is 0, the auto display repeat function is disabled. Each constant-current output is turned on and off for one display period. When this bit is 1, each output repeats the PWM control every 65,536 GSCLKs.
367	TMGRST		Display timing reset mode enable bit 0 = Disabled, 1 = Enabled When this bit is 0, the GS counter is not reset and the outputs are not forced off even when a LAT rising edge is input for a GS data write. When this bit is 1, the GS counter is reset to 0 and all outputs are forced off at the LAT rising edge for a GS data write. Afterwards, PWM control resumes from the next GSCLK rising edge.
368	RFRESH		Auto data refresh mode enable bit 0 = Disabled, 1 = Enabled When this bit is 0, the auto data refresh function is disabled. The data in the common shift register are copied to the GS data latch at the next LAT rising edge for a GS data write. DC data in the control data latch are copied to the DC data latch at the same time. When this bit is 1, the auto data refresh function is enabled. The data in the common shift register are copied to the GS data latch at the 65,536th GSCLK after the LAT rising edge for a GS data write. DC data in the control data latch are copied to the DC data latch at the same time.
369	ESPWM		ES-PWM mode enable bit 0 = Disabled, 1 = Enabled When this bit is 0, the conventional PWM control mode is selected. If the TLC5955 is used for multiplexing a drive, the conventional PWM mode should be selected to prevent excess on or off switching. When this bit is 1, ES-PWM control mode is selected.
370	LSDVLT		LSD detection voltage selection bit LED short detection (LSD) detects a fault caused by a shorted LED by comparing the OUTX _n voltage to the LSD detection threshold voltage. The threshold voltage is selected by this bit. When this bit is 0, the LSD voltage is VCC × 70%. When this bit is 1, the LSD voltage is VCC × 90%.

8.3.3 Status Information Data (SID)

The status information data (SID) contains the status of the LED open detection (LOD) and LED short detection (LSD). When the MSB of the common shift register is set to 0 and the RFRESH bit in the control data latch is 0, the SID are loaded to the common shift register at the LAT falling edge after the data in the common shift register are loaded to the grayscale data latch. If the common shift register MSB is 1, the SID are not loaded to the common shift register.

When the MSB of the common shift register is set to 0 and the RFRESH bit in the control data latch is 1, the SID are loaded to the common shift register at the GS counter 0000h just after LAT when the GS data are input. If the common shift register MSB is 1, the SID are not loaded to the common shift register. When the RFRESH bit is 1, the SCLK rising edge must be input with a low-level LAT signal after 65,538 GSCLKs (or more) are input from the LAT rising signal input.

After being loaded into the common shift register, new SID data cannot be loaded until at least one new bit of data is written into the common shift register. To recheck SID without changing the GS data, reprogram the common shift register with the same data currently programmed into the GS latch. When LAT goes high, the GS data do not change, but new SID data are loaded into the common shift register. LOD and LSD are shifted out of SOUT with each SCLK rising edge. The SID load configuration is shown in Figure 24 and Table 6.

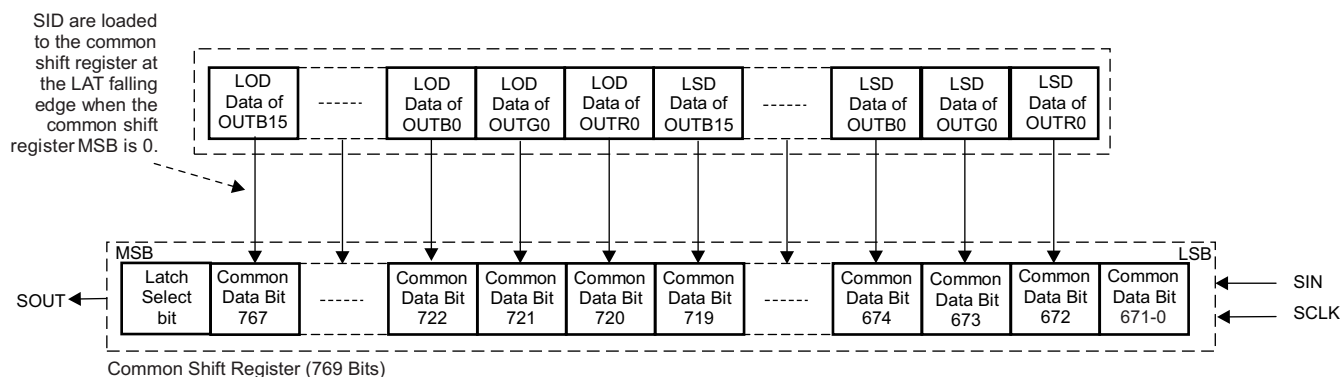


Figure 24. SID Load Configuration

Table 6. SID Load Description

COMMON SHIFT REGISTER BIT NUMBER	LOADED SID	COMMON SHIFT REGISTER BIT NUMBER	LOADED SID
671-0	No data loaded	720	OUTR0 LOD data
672	OUTR0 LSD data (0 = No error, 1 = Error)	721	OUTG0 LOD data
673	OUTG0 LSD data	722	OUTB0 LOD data
674	OUTB0 LSD data	723	OUTR1 LOD data
675	OUTR1 LSD data	724	OUTG1 LOD data
676	OUTG1 LSD data	725	OUTB1 LOD data
677	OUTB1 LSD data	726	OUTR2 LOD data
678	OUTR2 LSD data	727	OUTG2 LOD data
679	OUTG2 LSD data	728	OUTB2 LOD data
680	OUTB2 LSD data	729	OUTR3 LOD data
681	OUTR3 LSD data	730	OUTG3 LOD data
682	OUTG3 LSD data	731	OUTB3 LOD data
683	OUTB3 LSD data	732	OUTR4 LOD data
684	OUTR4 LSD data	733	OUTG4 LOD data
685	OUTG4 LSD data	734	OUTB4 LOD data
686	OUTB4 LSD data	735	OUTR5 LOD data
687	OUTR5 LSD data	736	OUTG5 LOD data
688	OUTG5 LSD data	737	OUTB5 LOD data
689	OUTB5 LSD data	738	OUTR6 LOD data
690	OUTR6 LSD data	739	OUTG6 LOD data
691	OUTG6 LSD data	740	OUTB6 LOD data
692	OUTB6 LSD data	741	OUTR7 LOD data
693	OUTR7 LSD data	742	OUTG7 LOD data
694	OUTG7 LSD data	743	OUTB7 LOD data
695	OUTB7 LSD data	744	OUTR8 LOD data
696	OUTR8 LSD data	745	OUTG8 LOD data
697	OUTG8 LSD data	746	OUTB8 LOD data
698	OUTB8 LSD data	747	OUTR9 LOD data
699	OUTR9 LSD data	748	OUTG9 LOD data
700	OUTG9 LSD data	749	OUTB9 LOD data
701	OUTB9 LSD data	750	OUTR10 LOD data
702	OUTR10 LSD data	751	OUTG10 LOD data
703	OUTG10 LSD data	752	OUTB10 LOD data
704	OUTB10 LSD data	753	OUTR11 LOD data
705	OUTR11 LSD data	754	OUTG11 LOD data
706	OUTG11 LSD data	755	OUTB11 LOD data
707	OUTB11 LSD data	756	OUTR12 LOD data
708	OUTR12 LSD data	757	OUTG12 LOD data
709	OUTG12 LSD data	758	OUTB12 LOD data
710	OUTB12 LSD data	759	OUTR13 LOD data
711	OUTR13 LSD data	760	OUTG13 LOD data
712	OUTG13 LSD data	761	OUTB13 LOD data
713	OUTB13 LSD data	762	OUTR14 LOD data
714	OUTR14 LSD data	763	OUTG14 LOD data
715	OUTG14 LSD data	764	OUTB14 LOD data
716	OUTB14 LSD data	765	OUTR15 LOD data
717	OUTR15 LSD data	766	OUTG15 LOD data
718	OUTG15 LSD data	767	OUTB15 LOD data
719	OUTB15 LSD data	768	No data loaded

8.3.4 LED Open Detection (LOD)

LOD detects a fault caused by an LED open circuit or a short from $OUTX_n$ to ground with low resistance by comparing the $OUTX_n$ voltage to the LOD detection threshold voltage (0.3 V, typically). If the $OUTX_n$ voltage is lower than the threshold voltage when $OUTX_n$ is on, that output LOD bit is set to 1 to indicate an open LED. Otherwise, the LOD bit is set to 0. LOD data are only valid for outputs that are programmed to be on. LOD data are latched into the LOD, LSD data latch at the 33rd GSCLK. LOD data for outputs programmed to be off at the 33rd GSCLK are always 0. The LED open detection circuit is shown in Figure 25 and Table 7 lists an LOD truth table. Refer to Figure 26 for an LOD read timing diagram.

8.3.5 LED Short Detection (LSD)

LSD data detect a fault caused by a shorted LED between LED terminals by comparing the $OUTX_n$ voltage to the LSD detection threshold voltage level set by LSDVLT in the control data latch. If the $OUTX_n$ voltage is higher than the programmed voltage when $OUTX_n$ is on, the corresponding output LSD bit is set to 1 to indicate a shorted LED. Otherwise, the LSD bit is set to 0. LSD data are only valid for outputs that are programmed to be on. LSD data are latched into the LOD, LSD data latch at the 33rd GSCLK. LSD data for outputs programmed to be off at the 33rd GSCLK are always 0. The LSD open detection circuit is shown in Figure 25 and Table 7 lists an LSD truth table. Refer to Figure 26 for an LSD read timing diagram.

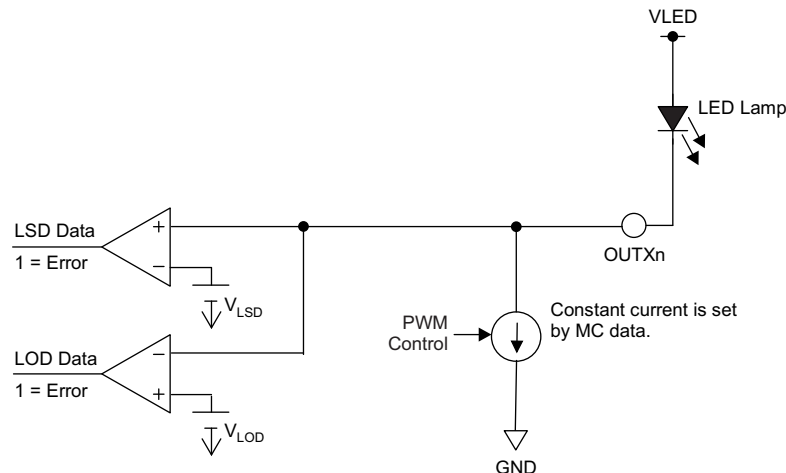


Figure 25. LOD and LSD Circuit

Table 7. LOD and LSD Truth Table

SID DATA	CONDITION	
	LOD	LSD
0	LED is not opened ($V_{OUTX_n} > V_{LOD}$)	LED is not shorted ($V_{OUTX_n} \leq V_{LSD}$)
1	LED is open or shorted to GND ($V_{OUTX_n} \leq V_{LOD}$)	LED is shorted between anode and cathode, or shorted to higher voltage side ($V_{OUTX_n} > V_{LSD}$)

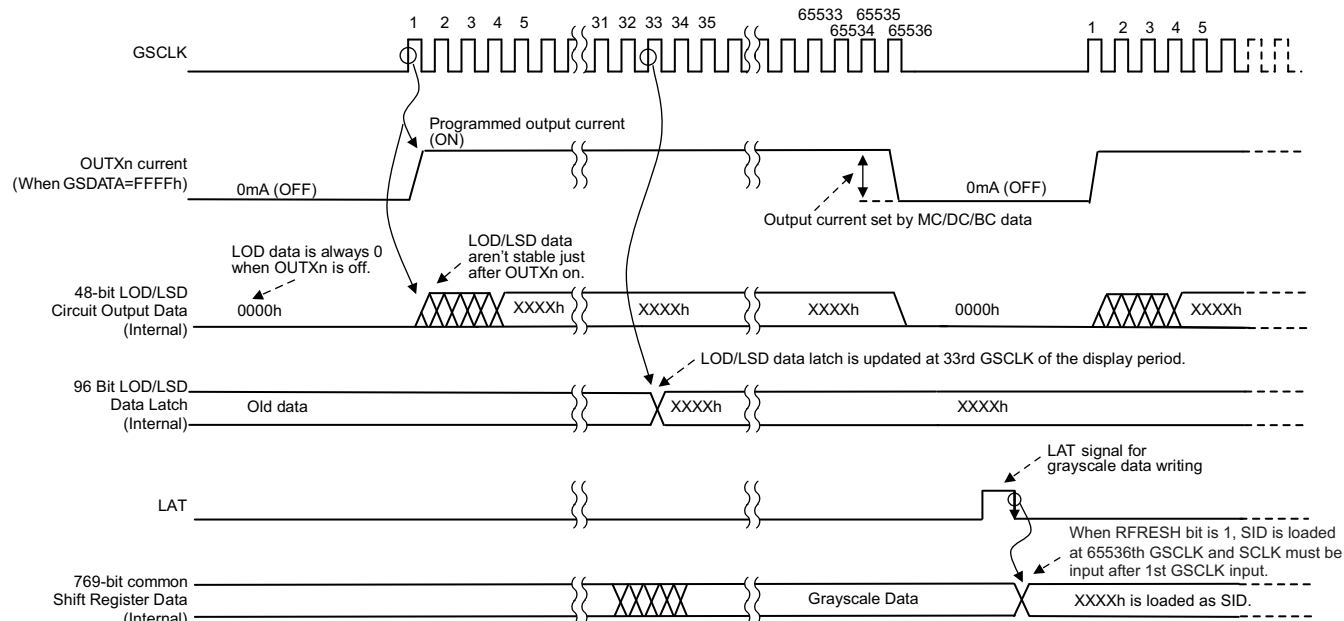


Figure 26. LOD and LSD Read and Load Timing Diagram

8.3.6 Noise Reduction

Large surge currents may flow through the device and the board on which the device is mounted if all 48 outputs turn on simultaneously at the start of each GS cycle. These large current surges can introduce detrimental noise and electromagnetic interference (EMI) into other circuits. The TLC5955 independently turns the outputs on with a series delay for each group to provide a soft-start feature. The output current sinks are grouped into eight groups. The first output group that is turned on or off are OUTR4, OUTG4, OUTB4, OUTR11, OUTG11, and OUTB11; the second output group is OUTX0 and OUTX15; the third output group is OUTX5 and OUTX10; the fourth output group is OUTX1 and OUTX14; the fifth output group is OUTX2 and OUTX13; the sixth output group is OUTX6 and OUTX9; the seventh output group is OUTX3 and OUTX12; and the eighth output group is OUTX7 and OUTX8. Each output group is turned on and off sequentially with a small delay between groups.

8.4 Device Functional Modes

8.4.1 Maximum Current Control (MC) Function

The maximum output current per channel, I_{OLCMax} , is programmed by the MC data and is set with the serial interface. I_{OLCMax} is the largest current for each output. Each $OUTXn$ sinks the I_{OLCMax} current when they turn on and the dot correction and global brightness control data are set to the maximum value of 7Fh (127d).

When the device is powered on, the MC data are set to 0. MC data should be changed when all constant-current outputs ($OUTXn$, where $X = R, G, \text{ or } B$; $n = 0$ to 7) are off. $MCX = 6$ and $MCX = 7$ are used when V_{CC} is greater than 3.6 V. The same MC data must be written twice to change the maximum constant-current output. Table 8 shows the characteristics of the constant-current sink versus the maximum current (MC) control data.

Table 8. Maximum Constant-Current Output versus MC Data

MCX ⁽¹⁾ DATA			I_{OLCMax} (mA), $OUTXn$ ⁽²⁾
BINARY	DECIMAL	HEX	
000 (default)	0 (default)	0 (default)	3.2
001	1	1	8.0
010	2	2	11.2
011	3	3	15.9
100	4	4	19.1
101	5	5	23.9
110 ⁽³⁾	6	6	27.1
111 ⁽³⁾	7	7	31.9

(1) $X = R, G, \text{ or } B$.

(2) $X = R, G, \text{ or } B$. $n = 0$ to 15.

(3) $MCX7$ and $MCX6$ can be used when V_{CC} is greater than 3.6 V.

8.4.2 Dot Correction (DC) Function

The TLC5955 can individually adjust the output current of each channel ($OUTx0$ to $OUTx15$, where x is $R, G, \text{ or } B$) by using DC. The DC function allows the brightness deviations of the LEDs connected to each output to be individually adjusted. Each output DC is programmed with a 7-bit word, so the value is adjusted with 128 steps within the range of 26.2% to 100% of I_{OLCMax} . DC data are programmed into the TLC5955 with the serial interface. When the device is powered on, the DC data in the control latch contains random data. Therefore, DC data must be written to the DC data latch before turning the constant-current outputs on. Table 9 summarizes the DC data value versus the set current value.

Table 9. DC Data versus Current Ratio and Set Current Value

DCXn ⁽¹⁾ DATA			BC DATA (Hex)	RATIO OF OUTPUT CURRENT TO I_{OLCMax} (%)	I_{OUT} (mA) (MC = 7, typical)	I_{OUT} (mA) (MC = 0, typical)
BINARY	DECIMAL	HEX				
000 0000	0	00	7F	26.2	8.36	0.84
000 0001	1	01	7F	26.7	8.54	0.86
000 0010	2	02	7F	27.3	8.73	0.88
—	—	—	—	—	—	—
111 1101	125	7D	7F	98.8	31.5	3.16
111 1110	126	7E	7F	99.4	31.7	3.18
111 1111	127	7F	7F	100.0	31.9	3.20

(1) $X = R, G, \text{ or } B$. $n = 0$ to 15.

8.4.3 Global Brightness Control (BC) Function

The TLC5955 has the ability to adjust the output current of all constant-current outputs of each color group (OUTR0 to OUTR15, OUTG0 to OUTG15, and OUTB0 to OUTB15) simultaneously to the same current ratio. This function is called *global brightness control* (BC). The BC function allows the global brightness of LEDs connected to the output to be adjusted. All outputs of each color group can be adjusted in 128 steps from 10% to 100% of the maximum output current, I_{OLCMax} . BC data are programmed into the TLC5955 with the serial interface. When the BC data change, the output current also changes immediately. When the device is powered on, the BC data contain random data. [Table 10](#) summarizes the BC data versus the set current value.

Table 10. BC Data versus Constant-Current Ratio and Set Current Value

BCX ⁽¹⁾ DATA			DCXn ⁽²⁾ DATA (Hex)	RATIO OF OUTPUT CURRENT TO I_{OLCMax} (%)	I_{OUT} (mA) (MC = 7, typical)	I_{OUT} (mA) (MC = 0, typical)
BINARY	DECIMAL	HEX				
000 0000	0	00	7F	10.0	3.19	0.32
000 0001	1	01	7F	10.7	3.42	0.34
000 0010	2	02	7F	11.4	3.64	0.37
—	—	—	—	—	—	—
111 1101	125	7D	7F	98.6	31.5	3.15
111 1110	126	7E	7F	99.3	31.7	3.18
111 1111	127	7F	7F	100.0	31.9	3.20

(1) X = R, G, or B.

(2) X = R, G, or B. n = 0 to 15.

8.4.4 Grayscale (GS) Function (PWM Control)

The TLC5955 can adjust the brightness of each output channel using a pulse width modulation (PWM) control scheme. The architecture of 16 bits per channel results in 65,536 brightness steps, from 0% up to 100% brightness.

The PWM operation for OUTn is controlled by a 16-bit grayscale (GS) counter. The GS counter increments on each GS reference clock (GSCLK) rising edge. The GS counter resets to 0000h when the LAT rising signal for a GS data write is input with the display timing reset mode enabled.

The TLC5955 has two types of PWM control: conventional PWM control and enhanced spectrum (ES) PWM control. The conventional PWM control can be selected when the ESPWM bit in the control data latch is 0. The ES PWM control is selected when the ESPWM bit is 1. The conventional PWM control should be selected for multiplexing a drive. The ES-PWM control should be selected for a static drive.

The on-time (t_{OUT_ON}) of each output (OUTn) can be calculated by [Equation 2](#).

$$t_{OUT_ON} (ns) = t_{GSCLK} (ns) \times GSXn$$

where:

- TGSKLK = one GS clock period,
- GSXn = the programmed GS value for OUTXn (GSXn = 0d to 65535d),
- X = R, G, or B for the red, green, or blue color group, and
- n = 0 to 15.

(2)

Table 11 summarizes the GS data values versus the output on-time duty cycle. When the device powers up, all OUTX n are forced off, the GS counter initializes to 0000h, and the status remains the same until GS data are written. After that, each OUTX n on and off status can be controlled by GS data and GSCLK.

Table 11. Output Duty Cycle and On-Time versus GS Data

GS DATA		ON-TIME DUTY (%)	GS DATA		ON-TIME DUTY (%)
DECIMAL	HEX		DECIMAL	HEX	
0	0	0	32768	8000	50.000
1	1	0.002	32769	8001	50.002
2	2	0.003	32770	8002	50.003
3	3	0.005	32771	8003	50.005
—	—	—	—	—	—
8191	1FFF	12.498	40959	9FFF	62.498
8192	2000	12.500	40960	A000	62.500
8193	2001	12.502	40961	A001	62.502
—	—	—	—	—	—
16381	3FFD	24.996	49149	BFFD	74.995
16382	3FFE	24.997	49150	BFFE	74.997
16383	3FFF	24.998	49151	BFFF	74.998
16384	4000	25.000	49152	C000	75.000
16385	4001	25.002	49153	C001	75.002
16386	4002	25.003	49154	C002	75.003
16387	4003	25.005	49155	C003	75.005
—	—	—	—	—	—
24575	5FFF	37.498	57343	DFFF	87.498
24576	6000	37.500	57344	E000	87.500
24577	6001	37.502	57345	E001	87.502
—	—	—	—	—	—
32765	7FFD	49.995	65533	FFFD	99.995
32766	7FFE	49.997	65534	FFFE	99.997
32767	7FFF	49.998	65535	FFFF	99.998

8.4.4.1 Conventional PWM Control

The first GS clock rising edge increments the GS counter by one and switches on all outputs with a non-zero GS value programmed into the GS data latch. Each additional GS clock rising edge increases the corresponding GS counter by one.

The GS counter keeps track of the number of clock pulses from the respective GS clock inputs. Each output stays on while the counter is less than or equal to the programmed GS value. Each output turns off at the GS counter value rising edge when the counter becomes greater than the output GS latch value. Figure 27 illustrates the conventional PWM operation.

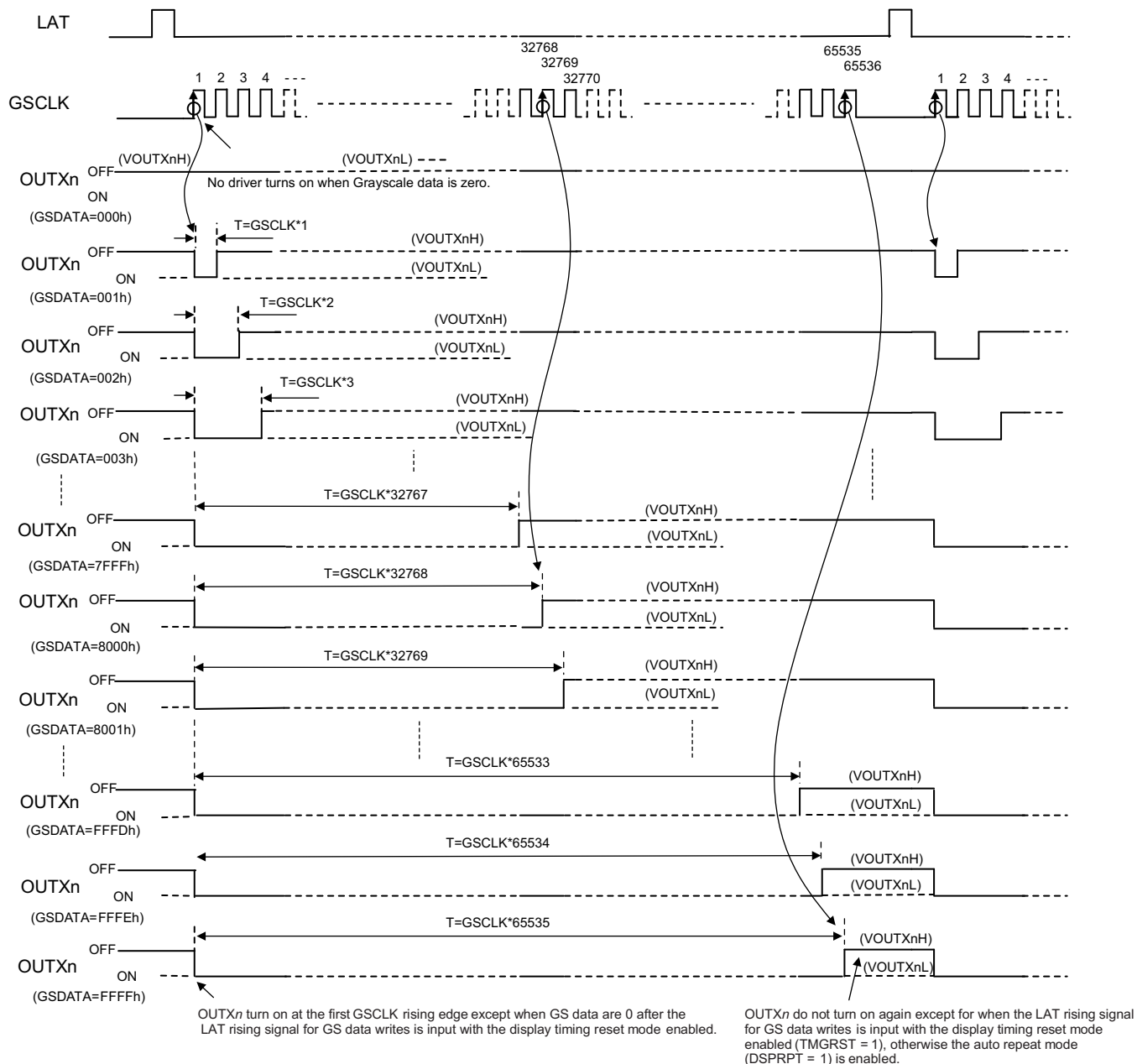


Figure 27. Conventional PWM Operation

8.4.4.2 Enhanced Spectrum (ES) PWM Control

In this PWM control, the total display period is divided into 128 display segments. The total display period is the time from the first GS clock (GSCLK) to the 65,536th GSCLK input. Each display segment has a maximum of 512 GSCLKs. The OUT_n on-time changes, depending on the 16-bit GS data. Refer to [Table 12](#) for the sequence of information and to [Figure 28](#) for the timing information.

Table 12. ES PWM Drive Turn On-Time Length

GS DATA		OUT _n DRIVER OPERATION
DECIMAL	HEX	
0	0000h	Does not turn on
1	0001h	Turns on for one GSCLK period in the first display segment
2	0002h	Turns on for one GSCLK period in the first and 65th display segments
3	0003h	Turns on for one GSCLK period in the first, 65th, and 33rd display segments
4	0004h	Turns on for one GSCLK period in the first, 65th, 33rd, and 97th display segments
5	0005h	Turns on for one GSCLK period in the first, 65th, 33rd, 97th, and 17th display segments
6	0006h	Turns on for one GSCLK period in the first, 65th, 33rd, 97th, 17th, and 81st display segments
—	—	The number of display segments where OUT _n is turned on for one GSCLK is incremented by increasing GS data in the following order: 1 > 65 > 33 > 97 > 17 > 81 > 49 > 113 > 9 > 73 > 41 > 105 > 25 > 89 > 57 > 121 > 5 > 69 > 37 > 101 > 21 > 85 > 53 > 117 > 13 > 77 > 45 > 109 > 29 > 93 > 61 > 125 > 3 > 67 > 35 > 99 > 19 > 83 > 51 > 115 > 11 > 75 > 43 > 107 > 27 > 91 > 59 > 123 > 7 > 71 > 39 > 103 > 23 > 87 > 55 > 119 > 15 > 79 > 47 > 111 > 31 > 95 > 63 > 127 > 2 > 66 > 34 > 98 > 18 > 82 > 50 > 114 > 10 > 74 > 42 > 106 > 26 > 90 > 58 > 122 > 6 > 70 > 38 > 102 > 22 > 86 > 54 > 118 > 14 > 78 > 46 > 110 > 30 > 94 > 62 > 126 > 4 > 68 > 36 > 100 > 20 > 84 > 52 > 116 > 12 > 76 > 44 > 108 > 28 > 92 > 60 > 124 > 8 > 72 > 40 > 104 > 24 > 88 > 56 > 120 > 16 > 80 > 48 > 112 > 32 > 96 > 64 > 128.
127	007Fh	Turns on for one GSCLK period in the first to 127th display segments, but does not turn on in the 128th display segment
128	0080h	Turns on for one GSCLK period in all display segments (first to 128th)
129	0081h	Turns on for two GSCLK periods in the first display period and for one GSCLK period in all other display periods
—	—	The number of display segments where OUT _n is turned on for one GSCLK is incremented by increasing GS data in the following order: 1 > 65 > 33 > 97 > 17 > 81 > 49 > 113 > 9 > 73 > 41 > 105 > 25 > 89 > 57 > 121 > 5 > 69 > 37 > 101 > 21 > 85 > 53 > 117 > 13 > 77 > 45 > 109 > 29 > 93 > 61 > 125 > 3 > 67 > 35 > 99 > 19 > 83 > 51 > 115 > 11 > 75 > 43 > 107 > 27 > 91 > 59 > 123 > 7 > 71 > 39 > 103 > 23 > 87 > 55 > 119 > 15 > 79 > 47 > 111 > 31 > 95 > 63 > 127 > 2 > 66 > 34 > 98 > 18 > 82 > 50 > 114 > 10 > 74 > 42 > 106 > 26 > 90 > 58 > 122 > 6 > 70 > 38 > 102 > 22 > 86 > 54 > 118 > 14 > 78 > 46 > 110 > 30 > 94 > 62 > 126 > 4 > 68 > 36 > 100 > 20 > 84 > 52 > 116 > 12 > 76 > 44 > 108 > 28 > 92 > 60 > 124 > 8 > 72 > 40 > 104 > 24 > 88 > 56 > 120 > 16 > 80 > 48 > 112 > 32 > 96 > 64 > 128.
255	00FFh	Turns on for two GSCLK periods in the first to 127th display segments and turns on one GSCLK period in the 128th display segment
256	0100h	Turns on for two GSCLK periods in all display segments (first to 128th)
257	0101h	Turns on for three GSCLK periods in the first display segments and for two GSCLK periods in all other display segments
—	—	The number of display segments where OUT _n is turned on for one GSCLK is incremented by increasing GS data in the following order: 1 > 65 > 33 > 97 > 17 > 81 > 49 > 113 > 9 > 73 > 41 > 105 > 25 > 89 > 57 > 121 > 5 > 69 > 37 > 101 > 21 > 85 > 53 > 117 > 13 > 77 > 45 > 109 > 29 > 93 > 61 > 125 > 3 > 67 > 35 > 99 > 19 > 83 > 51 > 115 > 11 > 75 > 43 > 107 > 27 > 91 > 59 > 123 > 7 > 71 > 39 > 103 > 23 > 87 > 55 > 119 > 15 > 79 > 47 > 111 > 31 > 95 > 63 > 127 > 2 > 66 > 34 > 98 > 18 > 82 > 50 > 114 > 10 > 74 > 42 > 106 > 26 > 90 > 58 > 122 > 6 > 70 > 38 > 102 > 22 > 86 > 54 > 118 > 14 > 78 > 46 > 110 > 30 > 94 > 62 > 126 > 4 > 68 > 36 > 100 > 20 > 84 > 52 > 116 > 12 > 76 > 44 > 108 > 28 > 92 > 60 > 124 > 8 > 72 > 40 > 104 > 24 > 88 > 56 > 120 > 16 > 80 > 48 > 112 > 32 > 96 > 64 > 128.
65479	FEFFh	Turns on for 511 GSCLK periods in the first to 127th display segments, but only turns on for 510 GSCLK periods in the 128th display segment
65480	FF00h	Turns on for 511 GSCLK periods in all display segments (first to 128th)
65481	FF01h	Turns on for 512 GSCLK periods in the first display period and for 511 GSCLK periods in the second to 128th display segments
—	—	—
65534	FFFEh	Turns on for 512 GSCLK periods in the first to 63rd and 65th to 127th display segments; also turns on for 511 GSCLK periods in the 64th and 128th display segments
65535	FFFFh	Turns on for 512 GSCLK periods in the first to 127th display segments but only turns on for 511 GSCLK periods in the 128th display segment



8.4.4.3 Auto Display Repeat Function

This function can repeat the total display period as long as GSCLK is present, as shown in Figure 29. This function is switched on or off by the content of the DSPRPT bit in the control data latch.

When the DSPRPT bit is 1, auto display repeat is enabled and the entire display period repeats. When the DSPRPT bit is 0, auto display repeat is disabled and the entire display period only executes one time after a LAT signal rising edge is input for GS data writes when the display timing reset is enabled.

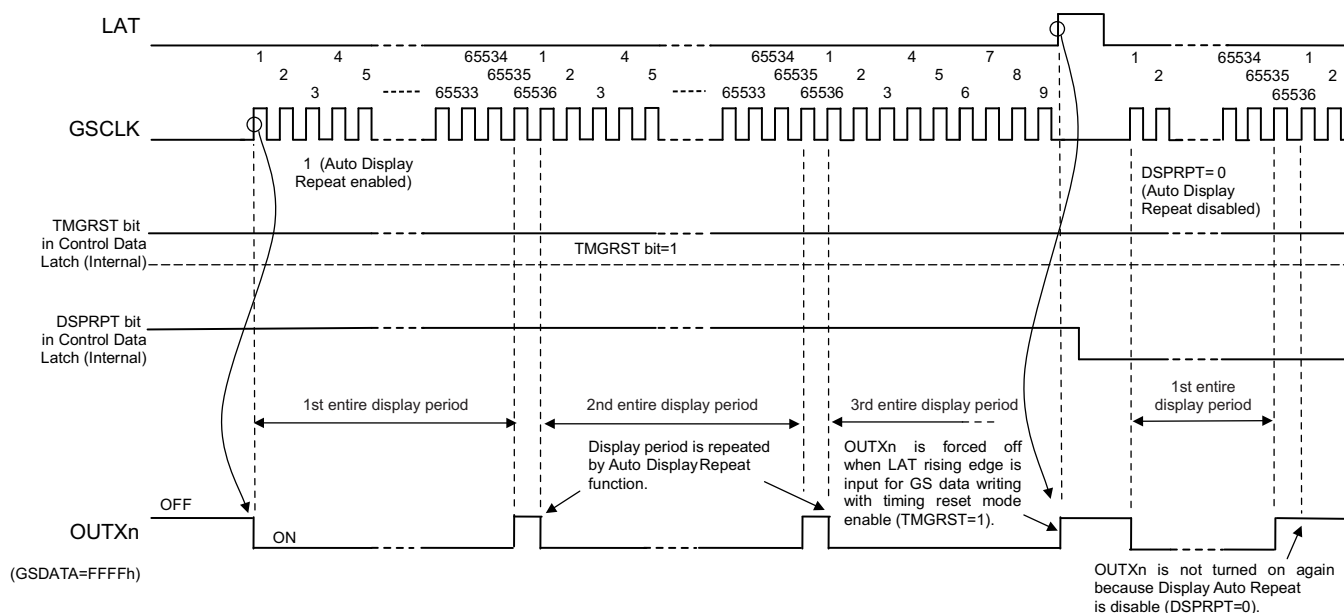


Figure 29. Auto Display Repeat Function

8.4.4.4 Display Timing Reset Function

The display timing reset function allows initializing the display timing with a LAT rising edge. This function can be switched on or off with the TMGRST bit in the control data latch. When the TMGRST bit is 1, the GS counter is reset to 0 and all outputs are forced off at the LAT rising edge for a GS write. Furthermore, the 768-bit GS data latch is updated with the data from the common shift register and the 336-bit DC data latch is updated with the DC data in the 371-bit control data latch. When the TMGRST bit is 0, the GS counter is not reset and the outputs are not forced off, even if a LAT rising edge is input. A timing diagram for this function is shown in Figure 30.

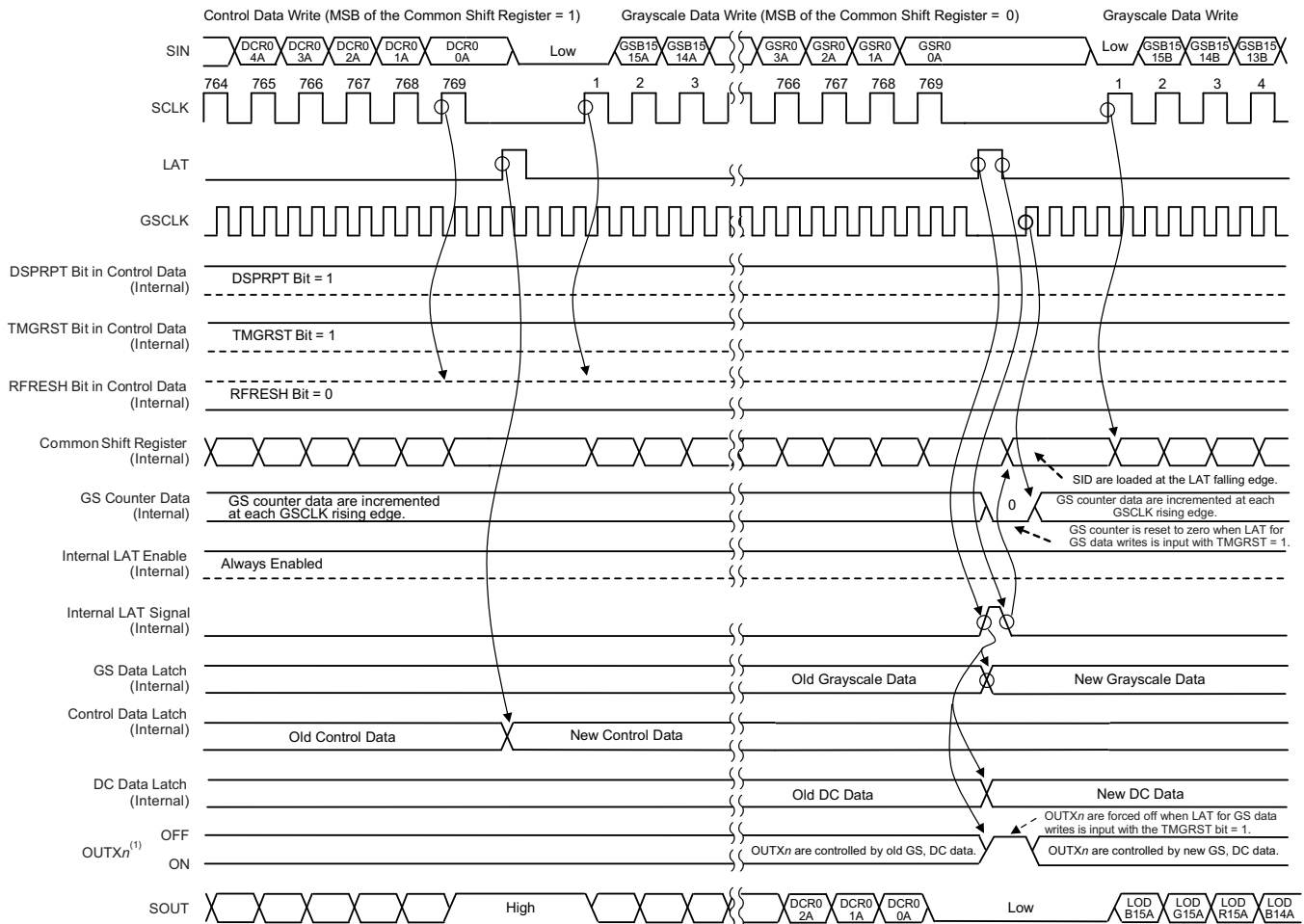


Figure 30. Display Timing Reset Function (DSPRPT = 1, TMGRST = 1, and RFRESH = 0)

8.4.4.5 Auto Data Refresh Function

This function delays updating the grayscale (GS) and dot correction (DC) data until the end of one entire display period. If both DC data and GS data are written by the end of an entire display period, the input DC data are held in the control data latch and the GS data are held in the common shift register. Both DC and GS data are copied to the 336-bit DC data latch and 768-bit GS data latch at the end of an entire display period. The data latches are used for the next display period. GS data are directly copied from the common shift register to the GS data latch. Therefore, GS data must be written after the DC data are written. Furthermore, the GS data in the common shift resistor must not be changed until all data are copied to the GS data latch. [Figure 31](#) and [Figure 32](#) show timing diagrams for this function.

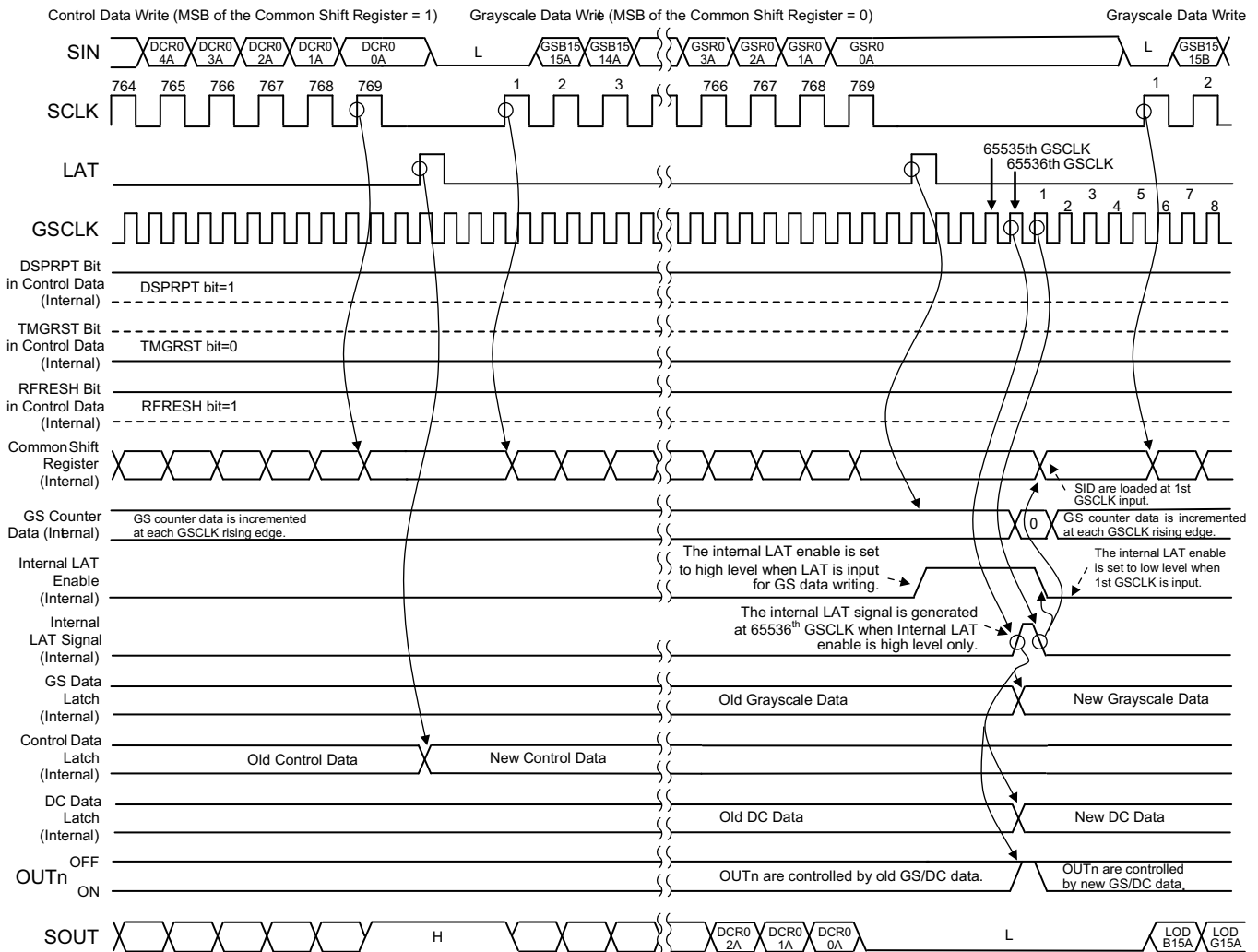


Figure 31. Auto Data Refresh Function 1 (DSPRPT = 1, TMGRST = 0, and RFRESH = 1)

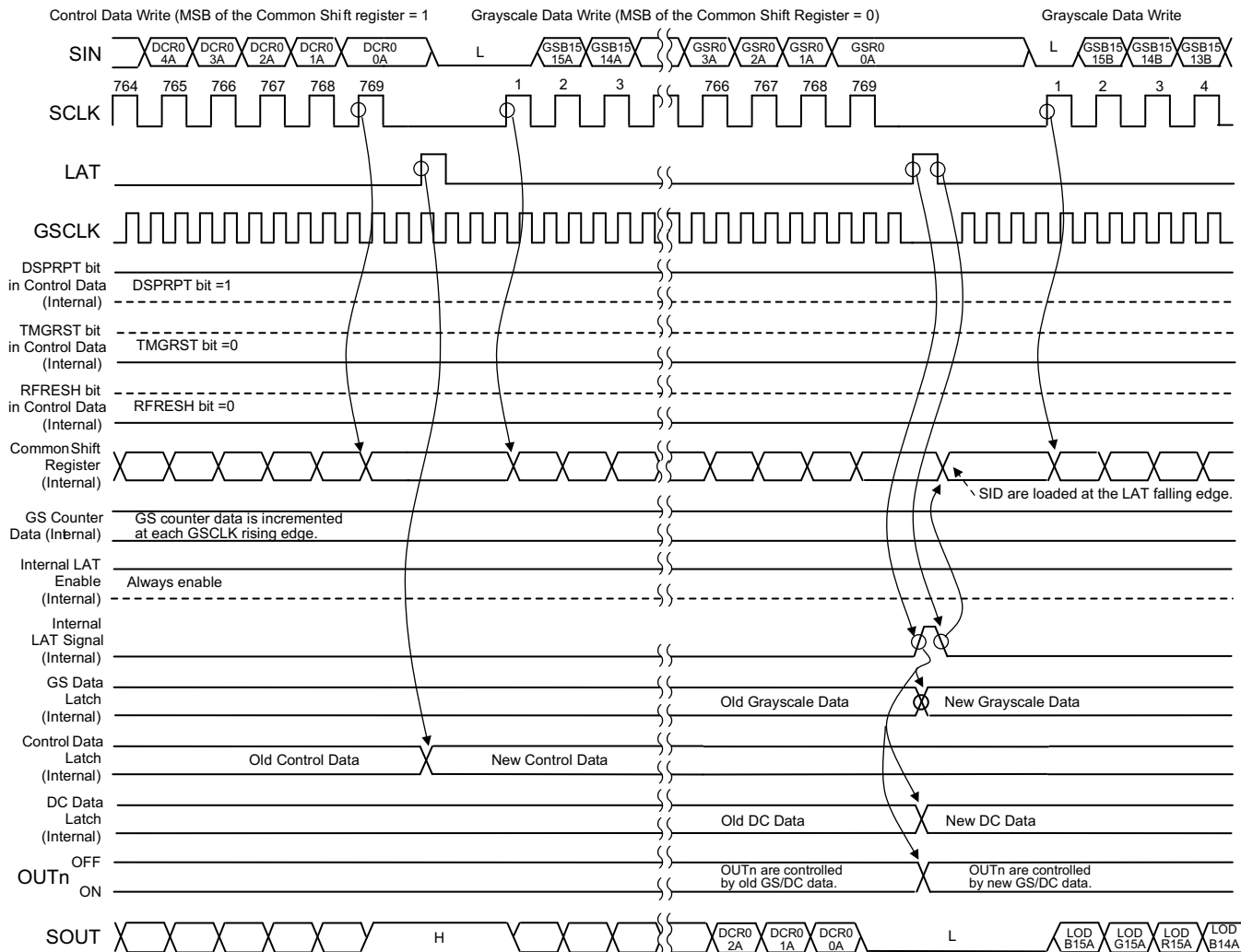


Figure 32. Auto Data Refresh Function 2 (DSPRPT = 1, TMGRST = 0, and RFRESH = 0)

9 Applications and Implementation

9.1 Application Information

The device is a 48-channel, constant sink current, LED driver. This device is typically connected in series to drive many LED lamps with only a few controller ports. Output current control data and PWM control data can be written from the SIN input terminal. The PWM timing reference clock can be supplied from the GSCLK input terminal. Also, the LED open and short error flag can be read out from the SOUT output terminal. Furthermore, the device maximum GSCLK clock frequency is 33 MHz and can reduce flickering discernable by the human eye.

9.2 Typical Application

9.2.1 Daisy-Chain Application

In this application, the device VCC and LED lamp anode voltages are supplied from different power supplies.

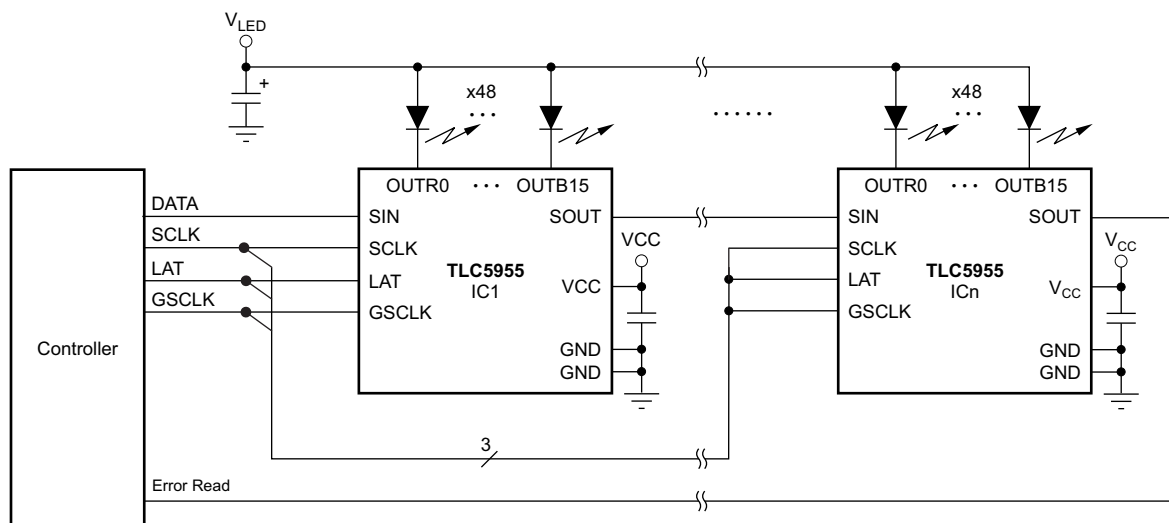


Figure 33. Multiple Daisy-Chained TLC5955 Devices

9.2.1.1 Design Requirements

For this design example, use the following as the input parameters.

Table 13. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
VCC input voltage range	3.0 V to 5.5 V
LED lamp (V_{LED}) input voltage range	Maximum LED forward voltage (V_F) + 0.3 V (knee voltage)
SIN, SCLK, LAT, and GSCLK voltage range	Low level = GND, High level = VCC

9.2.1.2 Detailed Design Procedure

9.2.1.2.1 Step-by-Step Design Procedure

To begin the design process, a few parameters must be decided upon. The designer needs to know the following:

- Maximum output constant-current value for each color LED ramp.
- Maximum LED forward voltage (V_F).
- Current ratio of red, green, and blue LED lamps for the best white balance.
- Are the auto display repeat function, display timing reset function, or auto data refresh function used?
- Which PWM control method is used: ES-PWM or conventional PWM?
- Is the LED short detect (LSD) function used? If so, which detection level (70% VCC or 90% VCC) is used?

9.2.1.2.2 Maximum Current (MC) Data

There are a total of nine bits of MC data for the red, green, and blue LED ramp. Select the MC data to be greater than each LED ramp current and write the data with other control data.

9.2.1.2.3 Global Brightness Control (BC) Data

There are a total of three sets of 7-bit BC data for the red, green, and blue LED ramp. Select the BC data for the best white balance of the red, green, and blue LED ramp and write the data with other control data.

9.2.1.2.4 Dot Correction (DC) Data

There are a total of 48 sets of 7-bit DC data for each current adjustment. Select the DC data for the best uniformity of each color LED ramp and write the data with other control data.

9.2.1.2.5 Grayscale (GS) Data

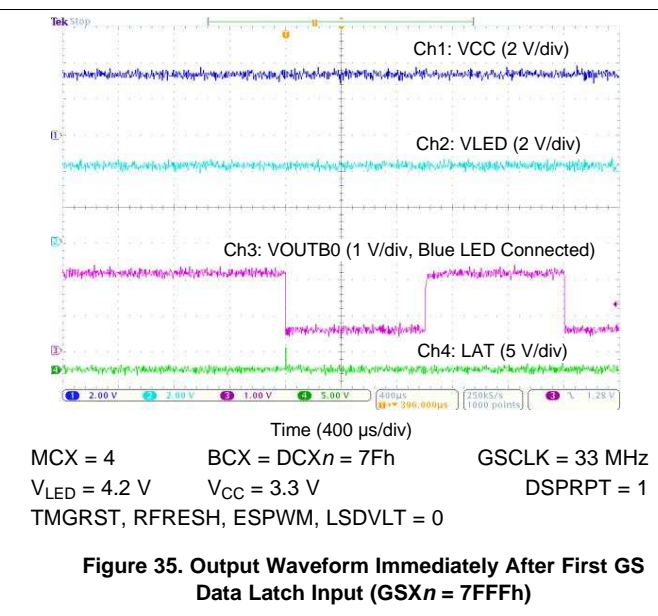
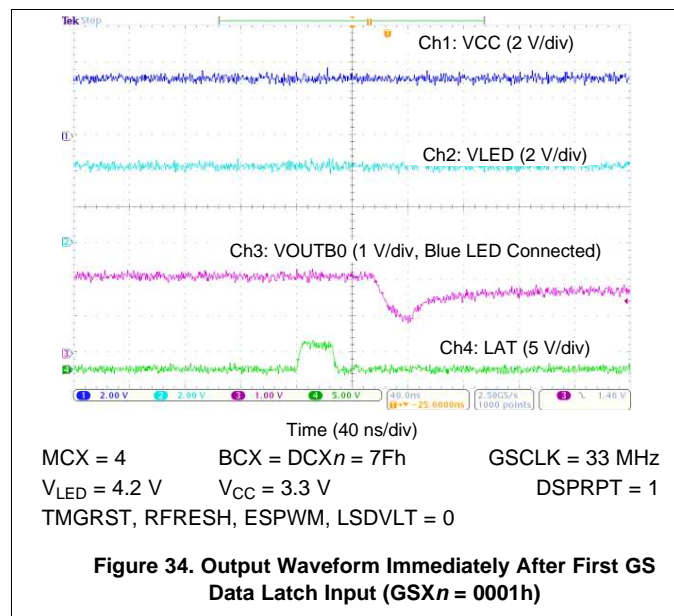
There are a total of 48 sets of 16-bit GS data for the PWM control of each output. Select the GS data of the LED ramp intensity and color control and write the data with other GS data.

9.2.1.2.6 Other Control Data

There are five bits control data to set the function mode for the auto display repeat, display timing reset, auto data refresh, ES-PWM, and LSD functions explained in the [Device Functional Modes](#) section. Write the 5-bit control data for the appropriate operation of the display system with MC, BC, and DC data as the control data.

9.2.1.3 Application Curves

One LED connected to each output.



10 Power Supply Recommendations

The V_{CC} power-supply voltage should be well regulated. An electrolytic capacitor must be used to reduce the voltage ripple to less than 5% of the input voltage. Furthermore, the V_{LED} voltage should be set to the voltage calculated by [Equation 3](#):

$$V_{LED} \geq LED\ V_F \times \text{Number of LED Lamps Connected in Series} + 0.3\ V\ (20\ mA\ \text{for Constant-Current Example})$$

where:

- V_F = Forward voltage (3)

Because the total current of the constant-current output is large, some electrolytic capacitors must be used to prevent the $OUTXn$ terminal voltage from dropping lower than the calculated voltage from [Equation 3](#).

11 Layout

11.1 Layout Guidelines

1. Place the decoupling capacitor near the VCC and GND terminals.
2. Route the GND pattern as widely as possible for large GND currents. Maximum GND current is approximately 1.53 A.
3. Routing between the LED cathode side and the device $OUTXn$ should be as short and straight as possible to reduce wire inductance.
4. The PowerPAD must be connected to the GND layer because the pad is not internally connected to GND and should be connected to a heat sink layer to reduce device temperature.

11.2 Layout Example

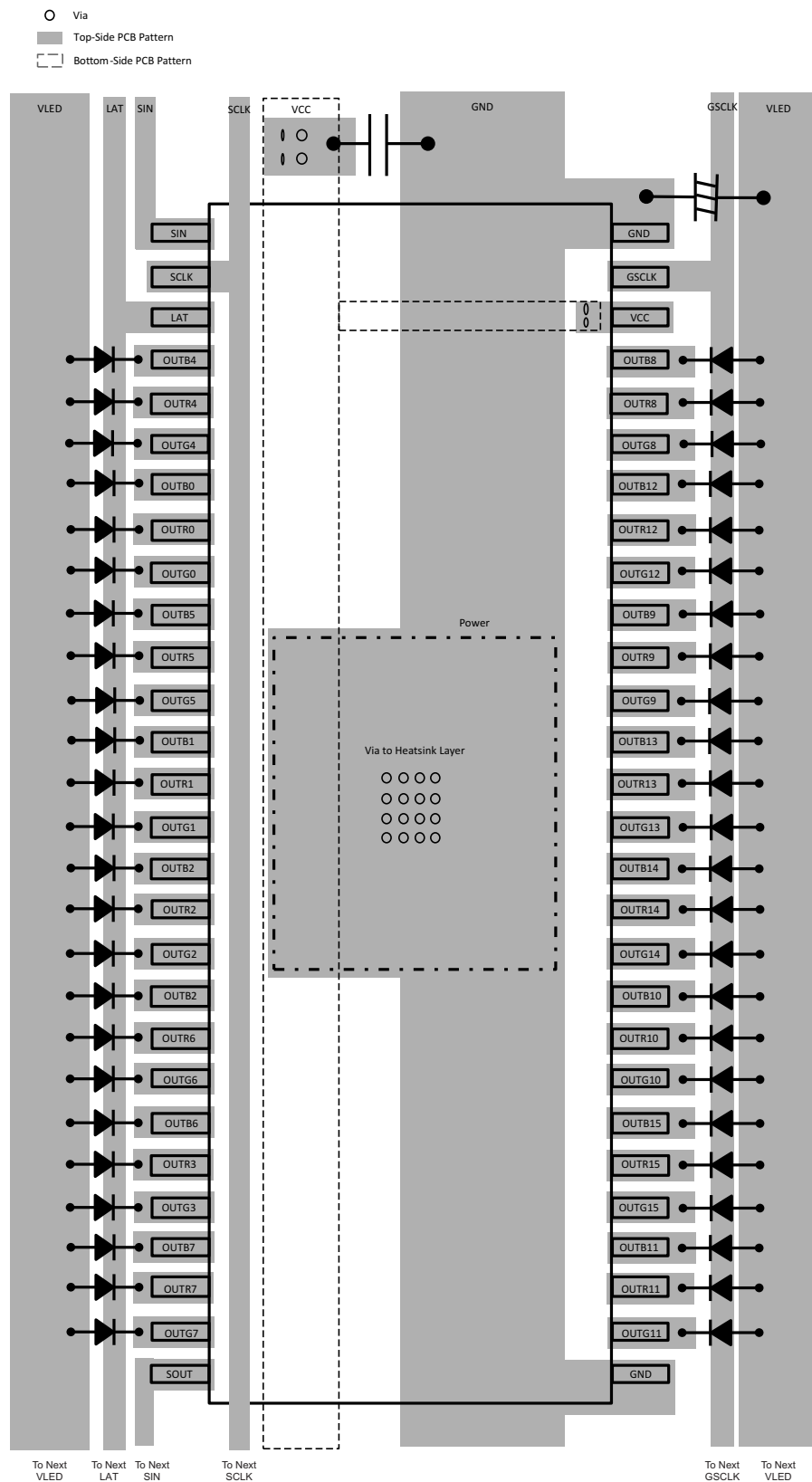


Figure 36. Layout Example

12 Device and Documentation Support

12.1 Device Support

12.1.1 Development Support

For the LED driver solution, go to www.ti.com/solution/lighting_signage.

12.2 Documentation Support

12.2.1 Related Documentation

For related documentation see the following:

- PowerPAD™ Thermally Enhanced Package Application Report, [SLMA002](#)

12.3 Trademarks

All trademarks are the property of their respective owners.

12.4 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.5 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC5955DCA	ACTIVE	HTSSOP	DCA	56	35	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	TLC5955	Samples
TLC5955DCAR	ACTIVE	HTSSOP	DCA	56	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	TLC5955	Samples
TLC5955RTQR	ACTIVE	QFN	RTQ	56	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	TLC5955	Samples
TLC5955RTQT	ACTIVE	QFN	RTQ	56	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	TLC5955	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

TAPE AND REEL INFORMATION


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC5955DCAR	HTSSOP	DCA	56	2000	330.0	24.4	8.6	15.6	1.8	12.0	24.0	Q1
TLC5955RTQR	QFN	RTQ	56	2000	330.0	16.4	8.3	8.3	2.25	12.0	16.0	Q2
TLC5955RTQT	QFN	RTQ	56	250	180.0	16.4	8.3	8.3	2.25	12.0	16.0	Q2

TAPE AND REEL BOX DIMENSIONS

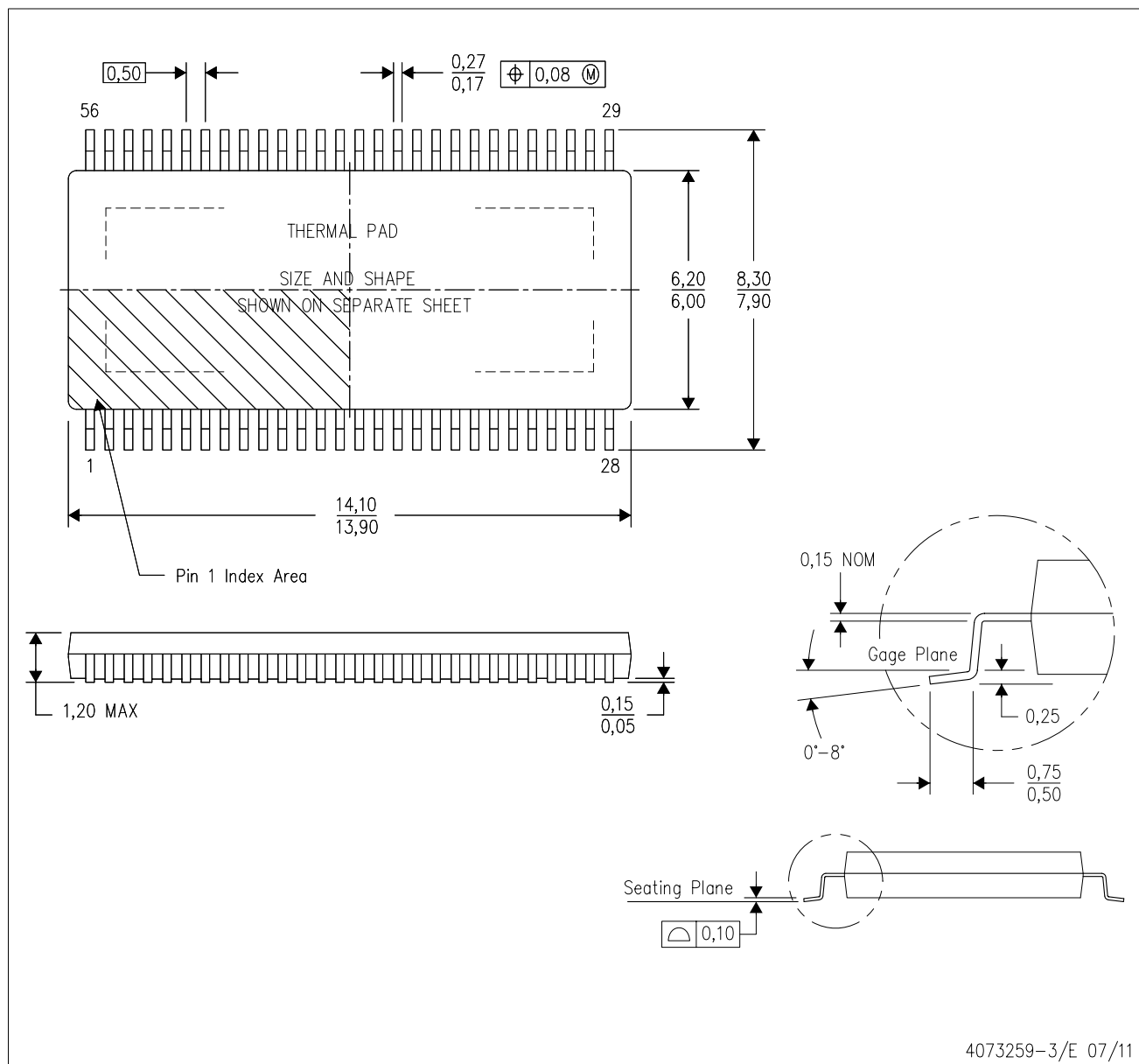


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC5955DCAR	HTSSOP	DCA	56	2000	367.0	367.0	45.0
TLC5955RTQR	QFN	RTQ	56	2000	367.0	367.0	38.0
TLC5955RTQT	QFN	RTQ	56	250	210.0	185.0	35.0

DCA (R-PDSO-G56)

PowerPAD™ PLASTIC SMALL-OUTLINE



- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusion not to exceed 0.15.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - Falls within JEDEC MO-153

PowerPAD is a trademark of Texas Instruments.

DCA (R-PDSO-G56)

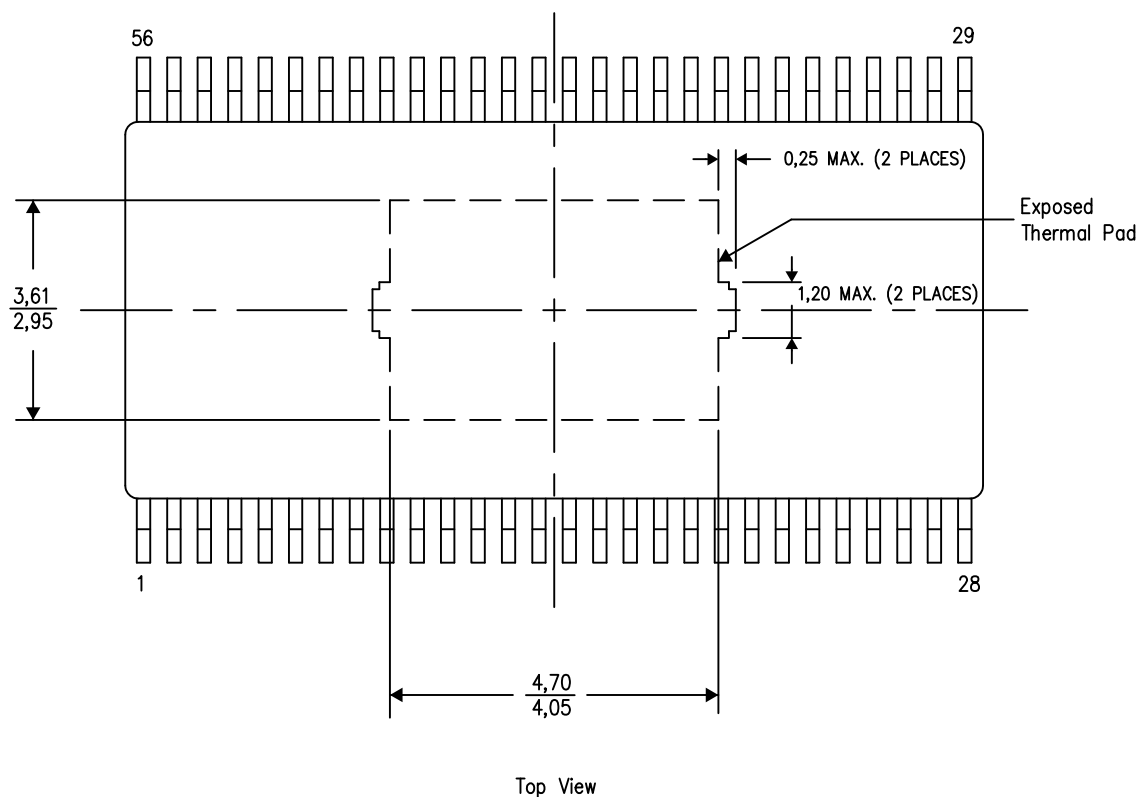
PowerPAD™ PLASTIC SMALL OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

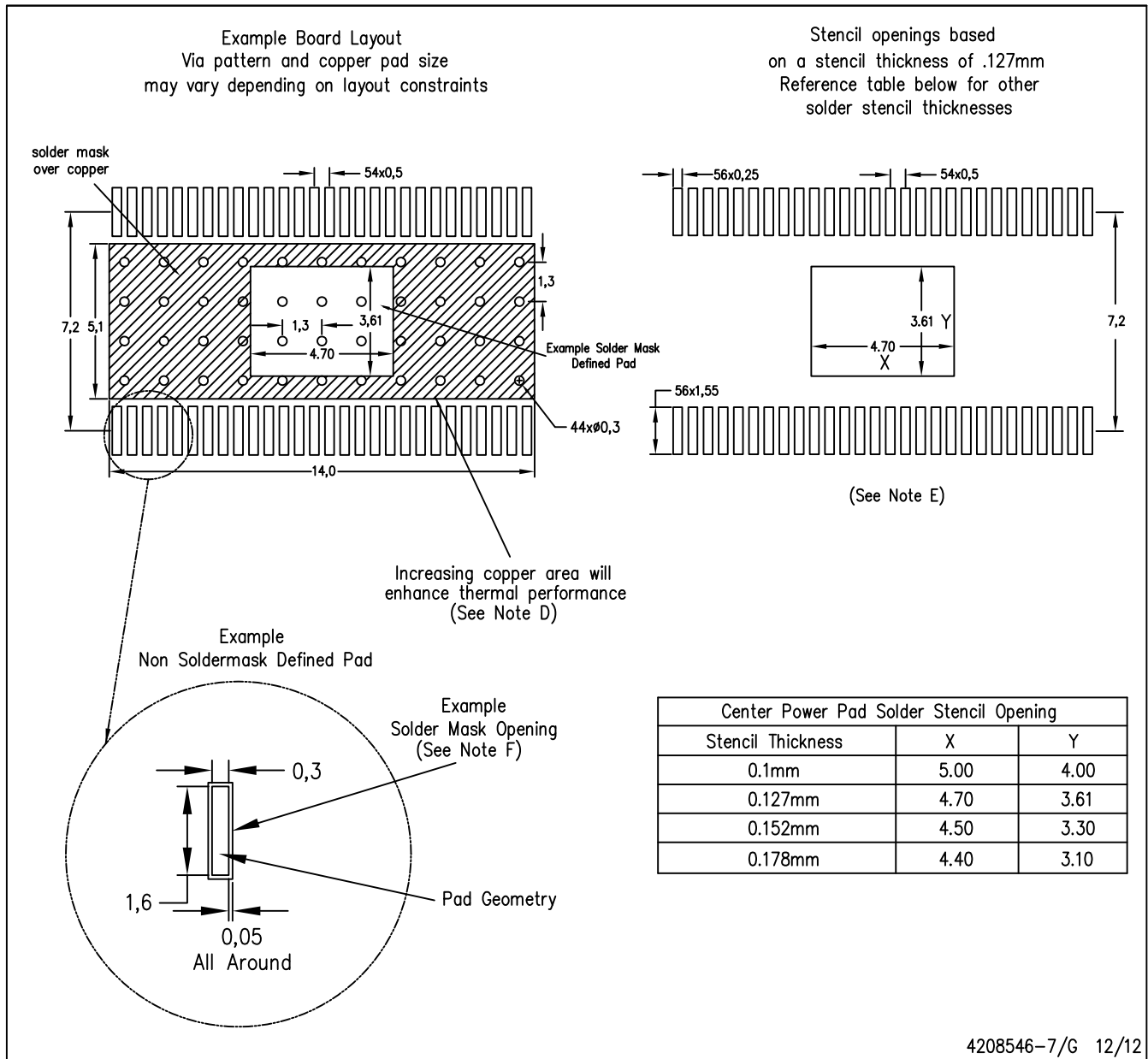


Exposed Thermal Pad Dimensions

4206320-9/R 03/13

- NOTE:
- A. All linear dimensions are in millimeters
 - B. Keep-out features are identified to prevent board routing interference.
- These exposed metal features may vary within the identified area or completely absent on some devices.

PowerPAD is a trademark of Texas Instruments.

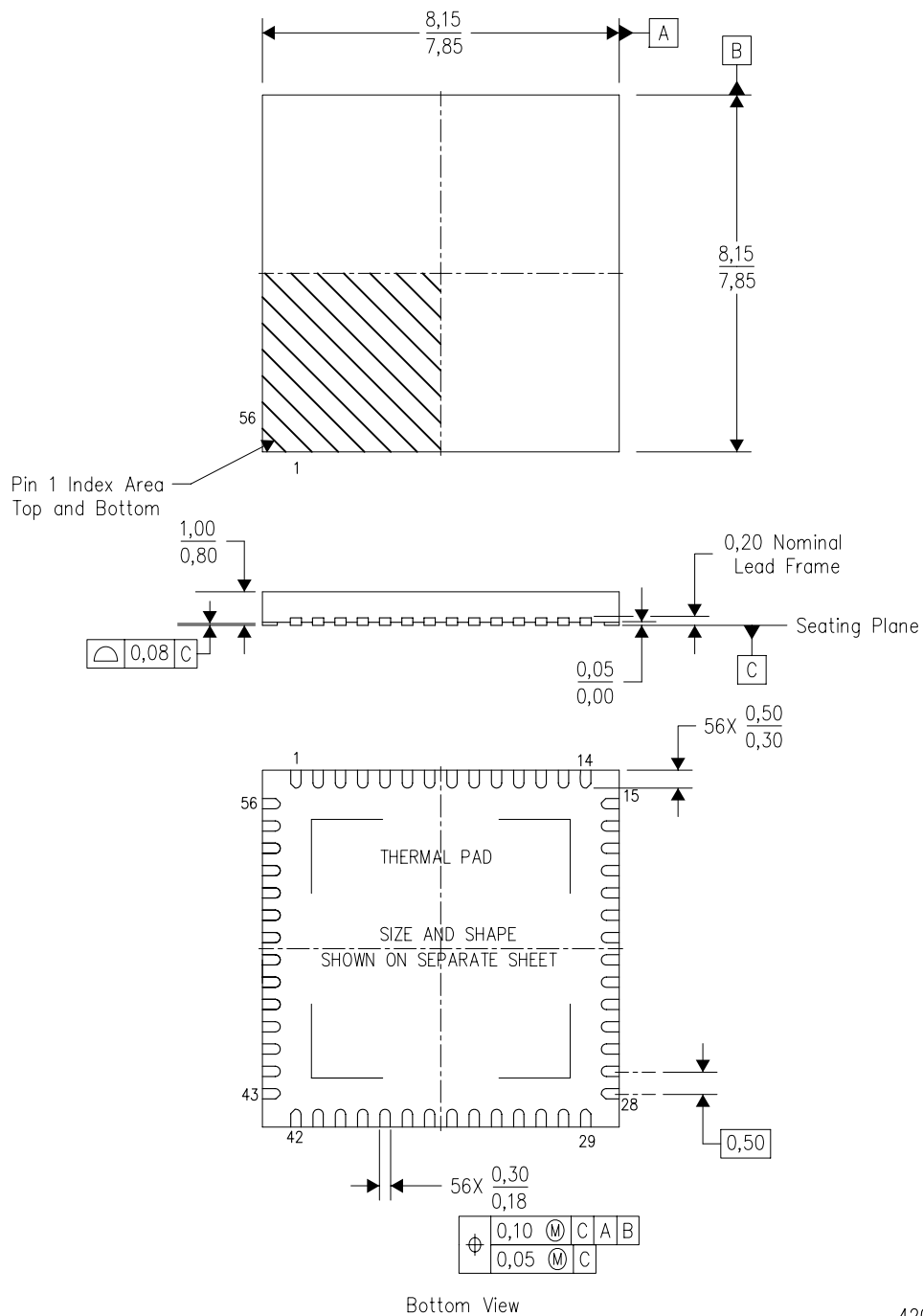


NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

RTQ (S-PVQFN-N56)

PLASTIC QUAD FLATPACK NO-LEAD



4206089/D 08/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5–1994.
 - B. This drawing is subject to change without notice.
 - C. QFN (Quad Flatpack No-Lead) Package configuration.
 - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
 - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - F. Package complies to JEDEC MO-220.

RTQ (S-PVQFN-N56)

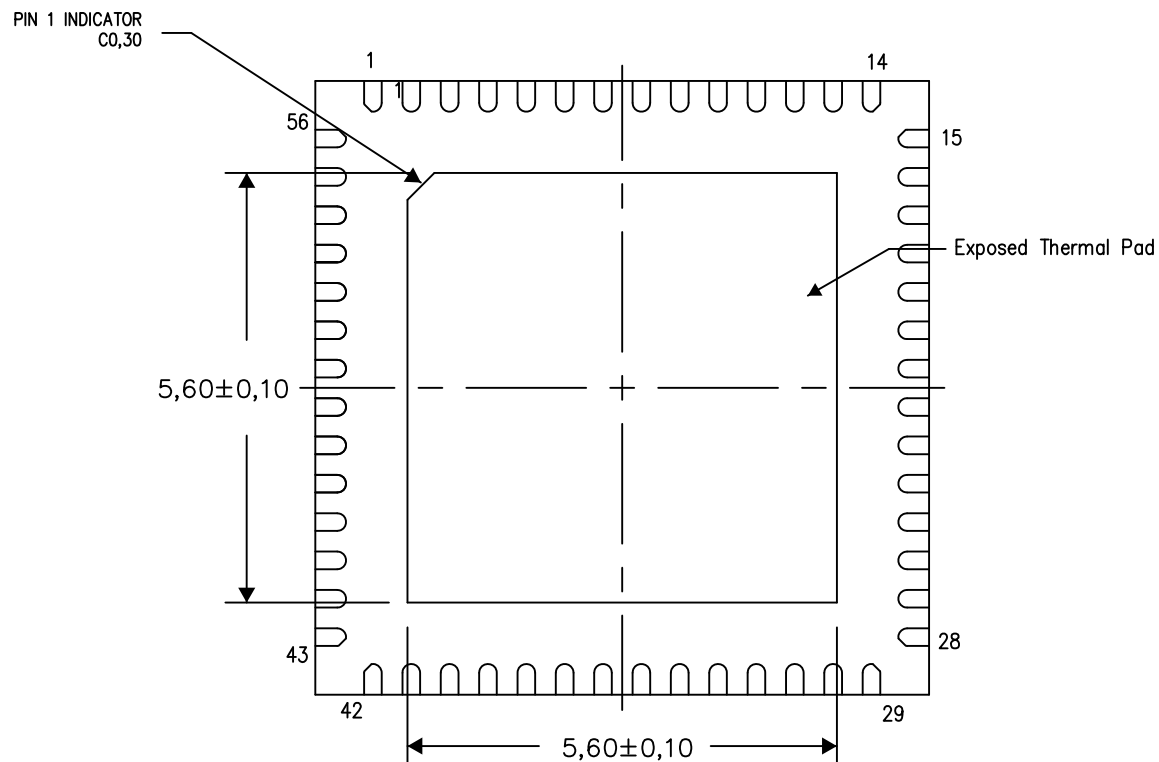
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

4206252-8/P 10/13

NOTE: All linear dimensions are in millimeters

IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

Products

Audio	www.ti.com/audio
Amplifiers	amplifier.ti.com
Data Converters	dataconverter.ti.com
DLP® Products	www.dlp.com
DSP	dsp.ti.com
Clocks and Timers	www.ti.com/clocks
Interface	interface.ti.com
Logic	logic.ti.com
Power Mgmt	power.ti.com
Microcontrollers	microcontroller.ti.com
RFID	www.ti-rfid.com
OMAP Applications Processors	www.ti.com/omap
Wireless Connectivity	www.ti.com/wirelessconnectivity

Applications

Automotive and Transportation	www.ti.com/automotive
Communications and Telecom	www.ti.com/communications
Computers and Peripherals	www.ti.com/computers
Consumer Electronics	www.ti.com/consumer-apps
Energy and Lighting	www.ti.com/energy
Industrial	www.ti.com/industrial
Medical	www.ti.com/medical
Security	www.ti.com/security
Space, Avionics and Defense	www.ti.com/space-avionics-defense
Video and Imaging	www.ti.com/video

TI E2E Community

e2e.ti.com



Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:

- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)

Факс: 8 (812) 320-02-42

Электронная почта: org@eplast1.ru

Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.