# LVDS Frequency-Programmable Crystal Oscillator

DATA SHEET

### **General Description**

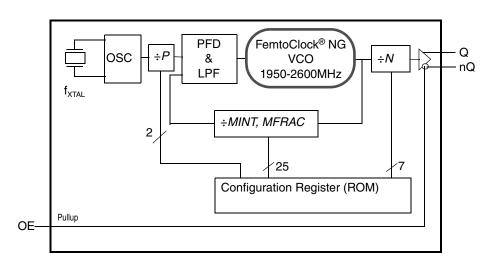
The IDT8N4S271 is a Factory Frequency-Programmable Crystal Oscillator with very flexible frequency programming capabilities. The device uses IDT's fourth generation FemtoClock® NG technology for an optimum of high clock frequency and low phase noise performance. The device accepts 2.5V or 3.3V supply and is packaged in a small, lead-free (RoHS 6) 6-lead ceramic 5mm x 7mm x 1.55mm package.

The device can be factory programmed to any in the range from 15.476MHz to 866.67MHz and from 975MHz to 1,300MHz and supports a very high degree of frequency precision of 218Hz or better. The extended temperature range supports wireless infrastructure, telecommunication and networking end equipment requirements.

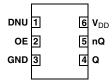
### **Features**

- Fourth generation FemtoClock<sup>®</sup> NG technology
- Factory-programmable clock output frequency from 15.476MHz to 866.67MHz and from 975MHz to 1,300MHz
- Frequency programming resolution is 218Hz and better
- One 2.5V, 3.3V LVDS clock output
- Output enable control (positive polarity), LVCMOS/LVTTL compatible
- RMS phase jitter @ 231.25MHz (12kHz 20MHz): 0.48ps (typical), integer PLL feedback configuration
- RMS phase jitter @ 231.25MHz (1kHz 40MHz): 0.50ps (typical), integer PLL feedback configuration
- 2.5V or 3.3V supply
- -40°C to 85°C ambient operating temperature
- Available in a lead-free (RoHS 6) 6-pin ceramic package

# **Block Diagram**



# **Pin Assignment**



**IDT8N4S271** 6-lead ceramic 5mm x 7mm x 1.55mm package body **CD Package Top View** 

# **Pin Description and Characteristic Tables**

**Table 1. Pin Descriptions** 

Number	Name	Туре		Description
1	DNU			Do not use (factory use only).
2	OE	Input	Pullup	Output enable pin. See Table 3A for function. LVCMOS/LVTTL interface levels.
3	GND	Power		Power supply ground.
4, 5	Q, nQ	Output		Differential clock output pair. LVDS interface levels.
6	V <sub>DD</sub>	Power		Power supply pin.

NOTE: Pullup refers to an internal input resistor. See Table 2, Pin Characteristics, for typical values.

#### **Table 2. Pin Characteristics**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			5.5		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			50		kΩ

### **Function Tables**

### **Table 3A. OE Configuration**

Input	
OE	Output Enable
0	Outputs Q, nQ are in high-impedance state
1 (default)	Outputs are enabled

NOTE: OE is an asynchronous control.

### Table 3B. Output Frequency Range

15.476MHz to 866.67MHz
975MHz to 1,300MHz

NOTE: Supported output frequency range. The output frequency can be programmed to any frequency in this range and to a precision of 218Hz or better.

# **Principles of Operation**

The block diagram consists of the internal 3rd overtone crystal and oscillator which provide the reference clock f<sub>XTAL</sub> of either 114.285MHz or 100MHz. The PLL includes the FemtoClock NG VCO along with the Pre-divider (*P*), the feedback divider (*M*) and the post divider (*N*). The *P*, *M*, and *N* dividers determine the output frequency based on the f<sub>XTAL</sub> reference. The feedback divider is fractional supporting a huge number of output frequencies. The configuration of the feedback divider to integer-only values results in an improved output phase noise characteristics at the expense of the range of output frequencies. Internal registers are used to hold one factory pre-set *P*, *M*, and *N* configuration setting. The *P*, *M*, and *N* frequency configuration supports an output frequency range from 15.476MHz to 866.67MHz and from 975MHz to 1,300MHz.

The devices use the fractional feedback divider with a delta-sigma modulator for noise shaping and robust frequency synthesis capability. The relatively high reference frequency minimizes phase noise generated by frequency multiplication and allows more efficient shaping of noise by the delta-sigma modulator.

The output frequency is determined by the 2-bit pre-divider (P), the feedback divider (M) and the 7-bit post divider (N). The feedback divider (M) consists of both a 7-bit integer portion (MINT) and an

18-bit fractional portion (MFRAC) and provides the means for high-resolution frequency generation. The output frequency  $f_{OUT}$  is calculated by:

$$f_{OUT} = f_{XTAL} \cdot \frac{1}{P \cdot N} \cdot \left[ MINT + \frac{MFRAC + 0.5}{2^{18}} \right]$$

### **Frequency Configuration**

An order code is assigned to each frequency configuration programmed by the factory (default frequencies). For more information on the available default frequencies and order codes, please see the Ordering Information section in this document. For available order codes, see the FemtoClock NG Ceramic-Package XO and VCXO Ordering Product Information document.

For more information on programming capabilities of the device for custom frequency and pull-range configurations, see the *FemtoClock NG Ceramic 5x7 Module Programming Guide*.

# **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC* 

Characteristics or AC Characteristics is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating	
Supply Voltage, V <sub>DD</sub>	3.63V	
Inputs, V <sub>I</sub>	-0.5V to V <sub>DD</sub> + 0.5V	
Outputs, I <sub>O</sub> (LVDS)		
Continuous Current	10mA	
Surge Current	15mA	
Package Thermal Impedance, $\theta_{JA}$	49.4°C/W (0 mps)	
Storage Temperature, T <sub>STG</sub>	-65°C to 150°C	

### **DC Electrical Characteristics**

Table 4A. Power Supply DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Power Supply Voltage		3.135	3.3	3.465	V
I <sub>DD</sub>	Power Supply Current			134	160	mA

### Table 4B. Power Supply DC Characteristics, $V_{DD} = 2.5V \pm 5\%$ , $T_A = -40$ °C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Power Supply Voltage		2.375	2.5	2.625	V
I <sub>DD</sub>	Power Supply Current			129	155	mA

### Table 4C. LVCMOS/LVTTL DC Characteristic, $V_{DD}$ = 3.3V ± 5% or 2.5V ± 5%, $T_A$ = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
.,	Input High Voltage	OE	$V_{DD} = 3.3V$	2		V <sub>DD</sub> + 0.3	V
V <sub>IH</sub>	Input High Voltage		$V_{DD} = 2.5V$	1.7		$V_{DD} + 0.3$	V
V	Input Low Voltage	OE	$V_{DD} = 3.3V$	-0.3		0.8	V
V <sub>IL</sub>		OE	V <sub>DD</sub> = 2.5V	-0.3		0.7	V
I <sub>IH</sub>	Input High Current	OE	$V_{DD} = V_{IN} = 3.465V \text{ or } 2.625V$			10	μΑ
I <sub>IL</sub>	Input Low Current	OE	$V_{DD} = 3.465V \text{ or } 2.625V, V_{IN} = 0V$	-150			μΑ

Table 4D. LVDS DC Characteristics,  $V_{DD}$  = 3.3V ± 5%,  $T_A$  = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OD</sub>	Differential Output Voltage		247	370	454	mV
$\Delta V_{OD}$	V <sub>OD</sub> Magnitude Change				50	mV
Vos	Offset Voltage		1.125	1.22	1.375	V
$\Delta V_{\sf OS}$	V <sub>os</sub> Magnitude Change				50	mV

### Table 4E. LVDS DC Characteristics, $V_{\text{DD}}$ = 2.5V ± 5%, $T_{\text{A}}$ = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OD</sub>	Differential Output Voltage		247	360	454	mV
$\Delta V_{OD}$	V <sub>OD</sub> Magnitude Change				50	mV
Vos	Offset Voltage		1.125	1.21	1.375	V
$\Delta V_{OS}$	V <sub>os</sub> Magnitude Change				50	mV

# **AC Electrical Characteristics**

Table 5. AC Characteristics,  $V_{DD}$  = 3.3V  $\pm$  5% or 2.5V  $\pm$  5%,  $T_A$  = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
	Output Fraguanay		15.476		866.67	MHz
fout	Output Frequency		975		1,300	MHz
f <sub>l</sub>	Initial Accuracy	Measured @ 25°C			±10	ppm
		Option code = A or B			±100	ppm
$f_S$	Temperature Stability	Option code = E or F			±50	ppm
		Option code = K or L			±20	ppm
	Aging	Frequency drift over 10 year life			±3	ppm
f <sub>A</sub>		Frequency drift over 15 year life			±5	ppm
		Option code A, B (10 year life)			±113	ppm
$f_{T}$	Total Stability	Option code E, F (10 year life)			±63	ppm
		Option code K, L (10 year life)			±33	ppm
tjit(cc)	Cycle-to-Cycle Jitter; NOTE 1				20	ps
tjit(per)	RMS Period Jitter; NOTE 1			3	5	ps

Table 5 (continued). AC Characteristics,  $V_{DD}$  = 3.3V ± 5% or 2.5V ± 5%,  $T_A$  = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
	RMS Phase Jitter (Random); Fractional PLL feedback and f <sub>XTAL</sub> =100.000MHz (2xxx order codes), NOTES 2, 3, 4	17MHz ≤ f <sub>OUT</sub> ≤ 1300MHz, Integration range: 12kHz-20MHz		0.497	0.882	ps
		$500 MHz \le f_{OUT} \le 1300 MHz$ , Integration range: $12 kHz$ - $20 MHz$		0.232	0.322	ps
		125MHz ≤ f <sub>OUT</sub> < 500MHz, Integration range: 12kHz-20MHz		0.250	0.450	ps
	RMS Phase Jitter (Random);	$17MHz \le f_{OUT} < 125MHz$ , Integration range: $12kHz$ - $20MHz$		0.275	0.405	ps
fjit(Ø)	Integer PLL feedback and f <sub>XTAL</sub> =100.00MHz (1xxx order codes),	f <sub>OUT</sub> = 156.25MHz, Integration range: 12kHz-20MHz		0.242	0.311	ps
	NOTES 2, 3, 5	f <sub>OUT</sub> = 231.25MHz, Integration range: 12kHz-20MHz		0.476	0.680	ps
		f <sub>OUT</sub> = 156.25MHz, Integration range: 12kHz-20MHz		0.275	0.359	ps
		f <sub>OUT</sub> = 231.25MHz, Integration range: 12kHz-20MHz		0.504	0.700	ps
	RMS Phase Jitter (Random) Fractional PLL feedback and f <sub>XTAL</sub> =114.285MHz (0xxx order codes), NOTES 2, 3, 6	17MHz ≤ f <sub>OUT</sub> ≤ 1300MHz, Integration range: 12kHz-20MHz		0.474	0.986	ps
Φ <sub>N</sub> (100)	Single-side Band Phase Noise, 100Hz from Carrier	f <sub>OUT</sub> = 231.25MHz		-88		dBc/Hz
Φ <sub>N</sub> (1k)	Single-side Band Phase Noise, 1kHz from Carrier	f <sub>OUT</sub> = 231.25MHz		-110		dBc/Hz
Φ <sub>N</sub> (10k)	Single-side Band Phase Noise, 10kHz from Carrier	f <sub>OUT</sub> = 231.25MHz		-123		dBc/Hz
Φ <sub>N</sub> (100k)	Single-side Band Phase Noise, 100kHz from Carrier	f <sub>OUT</sub> = 231.25MHz		-125		dBc/Hz
Φ <sub>N</sub> (1M)	Single-side Band Phase Noise, 1MHz from Carrier	f <sub>OUT</sub> = 231.25MHz		-137		dBc/Hz
Φ <sub>N</sub> (10M)	Single-side band phase noise, 10MHz from Carrier	f <sub>OUT</sub> = 231.25MHz		-141		dBc/Hz
t <sub>R</sub> / t <sub>F</sub>	Output Rise/Fall Time	20% to 80%	50		450	ps
odc	Output Duty Cycle		47		53	%
t <sub>STARTUP</sub>	Device Startup Time After Power Up				20	ms

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE: XTAL parameters (initial accuracy, temperature stability, aging and total stability) are guaranteed by manufacturing.

NOTE 1: This parameter is defined in accordance with JEDEC standard 65.

NOTE 2: Refer to the phase noise plot.

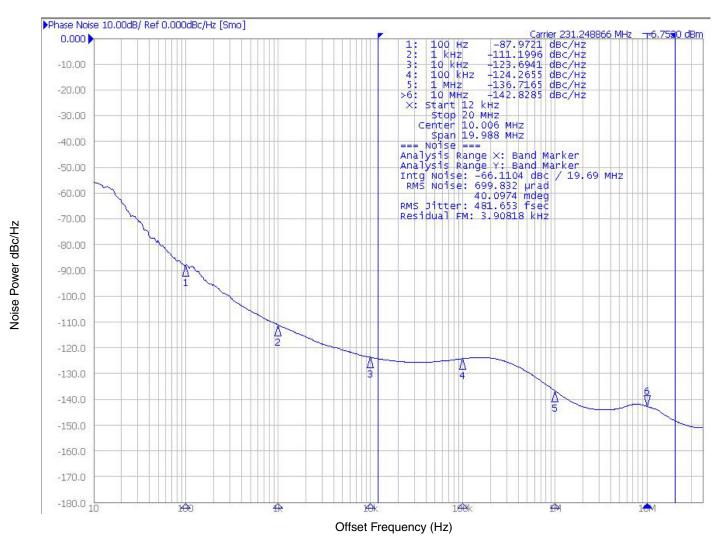
NOTE 3: See the FemtoClock NG Ceramic 5x7 Modules Programming guide for more information on PLL feedback modes and the optimum configuration for phase noise. Integer PLL feedback is the default operation for the dddd = 1xxx order codes.

NOTE 4: Applies to output frequencies: 81MHz, 122.88MHz, 231.25MHz, 622.08MHz, 866.67MHz and 1124MHz.

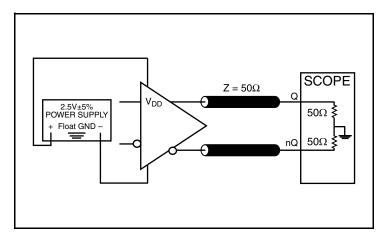
NOTE 5: Applies to output frequencies: 75MHz, 100MHz, 106.25MHz, 125MHz, 156.25MHz, 425MHz, 500MHz, 625MHz, 975MHz and 1300MHz.

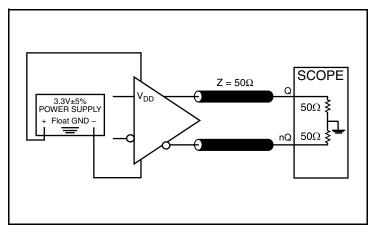
NOTE 6: Applies to output frequencies: 15.4762MHz, 38.88MHz, 114.285MHz, 496MHz, 669.32MHz and 658MHz.

# Typical Phase Noise at 231.25MHz (12kHz - 20MHz)



### **Parameter Measurement Information**





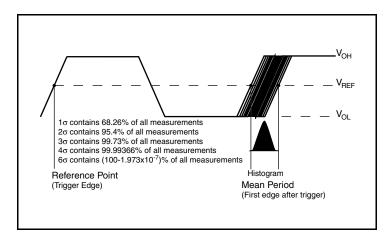
2.5V LVDS Output Load Test Circuit

Phase Noise Plot

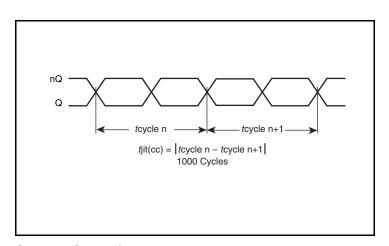
Phase Noise Plot

Offset Frequency  $f_1$ Offset Frequency  $f_2$ RMS Phase Jitter =  $\frac{1}{\sqrt{2} * \pi * f} * \sqrt{\text{Area Under Curve Defined by the Offset Frequency Markers}}$ 

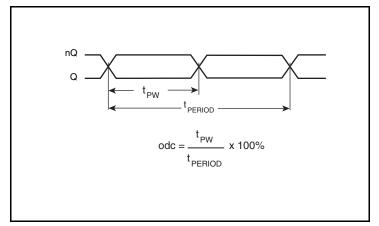
3.3V LVDS Output Load Test Circuit



**RMS Phase Jitter** 



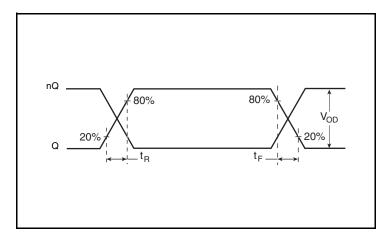
**RMS Period Jitter** 

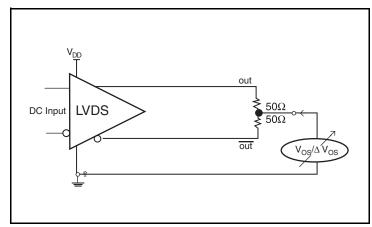


**Cycle-to-Cycle Jitter** 

**Output Duty Cycle/Pulse Width/Period** 

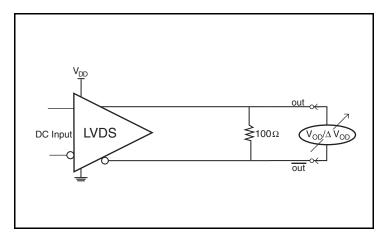
# **Parameter Measurement Information, continued**





**Output Rise/Fall Time** 

Offset Voltage Setup



**Differential Output Voltage Setup** 

# **Applications Information**

#### LVDS Driver Termination

For a general LVDS interface, the recommended value for the termination impedance  $(Z_T)$  is between  $90\Omega$  and  $132\Omega$ . The actual value should be selected to match the differential impedance  $(Z_0)$  of your transmission line. A typical point-to-point LVDS design uses a  $100\Omega$  parallel resistor at the receiver and a  $100\Omega$  differential transmission-line environment. In order to avoid any transmission-line reflection issues, the components should be surface mounted and must be placed as close to the receiver as possible. IDT offers a full line of LVDS compliant devices with two types of output structures: current source and voltage source. The

standard termination schematic as shown in *Figure 1A* can be used with either type of output structure. *Figure 1B*, which can also be used with both output types, is an optional termination with center tap capacitance to help filter common mode noise. The capacitor value should be approximately 50pF. If using a non-standard termination, it is recommended to contact IDT and confirm if the output structure is current source or voltage source type. In addition, since these outputs are LVDS compatible, the input receiver's amplitude and common-mode input range should be verified for compatibility with the output.

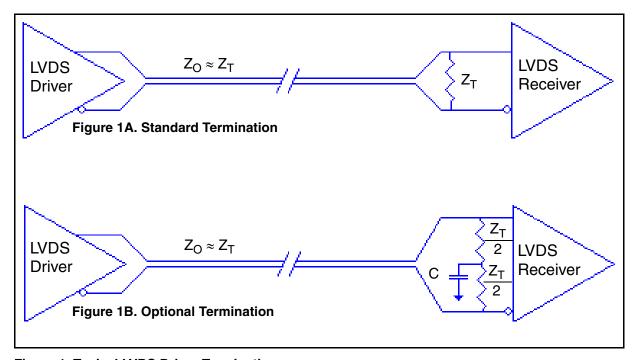


Figure 1. Typical LVDS Driver Termination

### **Schematic Layout**

Figure 2 shows an example IDT8N4S271 application schematic. The schematic example focuses on functional connections and is intended as an example only and may not represent the exact user configuration. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set. For example OE can be configured from an FPGA instead of set with pull up and pull down resistors as shown.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise, so to achieve optimum jitter performance isolation of the  $V_{DD}$  pin from power supply is required. In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB as close to the power pins as possible. If space is limited, the  $0.1\mu F$ 

capacitor on the  $V_{DD}$  pin must be placed on the device side with direct return to the ground plane though vias. The remaining filter components can be on the opposite side of the PCB.

Power supply filter component recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for a wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

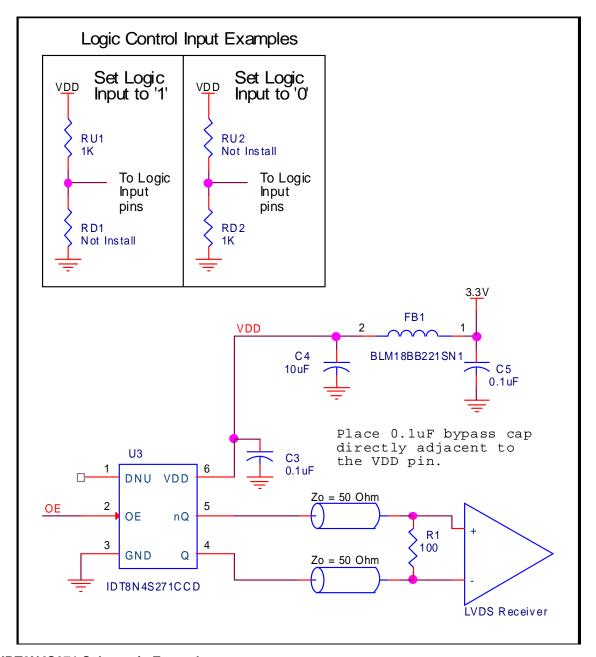


Figure 2. IDT8N4S271 Schematic Example

### **Power Considerations**

This section provides information on power dissipation and junction temperature for the IDT8N4S271. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the IDT8N4S271 is the sum of the core power plus the output power dissipated due to the loading. The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

Power (core)<sub>MAX</sub> = V<sub>DD MAX</sub> \* I<sub>DD MAX</sub> = 3.465V \* 160mA = 554.4mW

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{1A}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is  $49.4^{\circ}$ C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.554\text{W} * 49.4^{\circ}\text{C/W} = 112.4^{\circ}\text{C}$ . This is below the limit of  $125^{\circ}\text{C}$ .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 6. Thermal Resistance  $\theta_{JA}$  for 6 Lead Ceramic 5mm x 7mm Package, Forced Convection

$\theta_{JA}$ by Velocity						
Meters per Second	0	1	2.5			
Multi-Layer PCB, JEDEC Standard Test Boards	49.4°C/W	44.2°C/W	42.1°C/W			

# **Reliability Information**

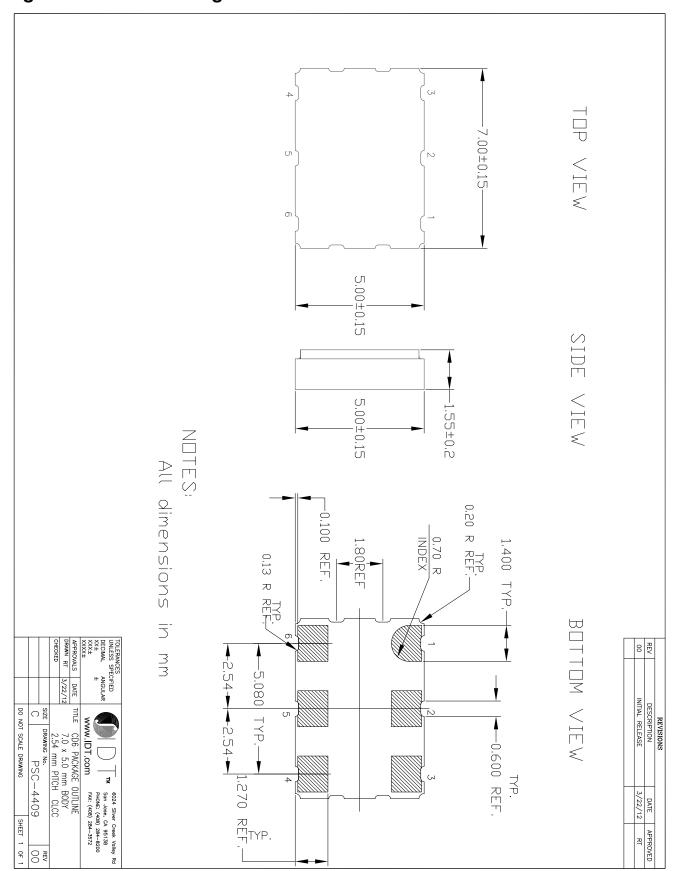
# Table 7. $\theta_{\text{JA}}$ vs. Air Flow Table for a 6-lead Ceramic 5mm x 7mm Package

$\theta_{\sf JA}$ vs. Air Flow						
Meters per Second	0	1	2.5			
Multi-Layer PCB, JEDEC Standard Test Boards	49.4°C/W	44.2°C/W	42.1°C/W			

### **Transistor Count**

The transistor count for IDT8N4S271 is: 47,511

# **Package Outline and Package Dimensions**

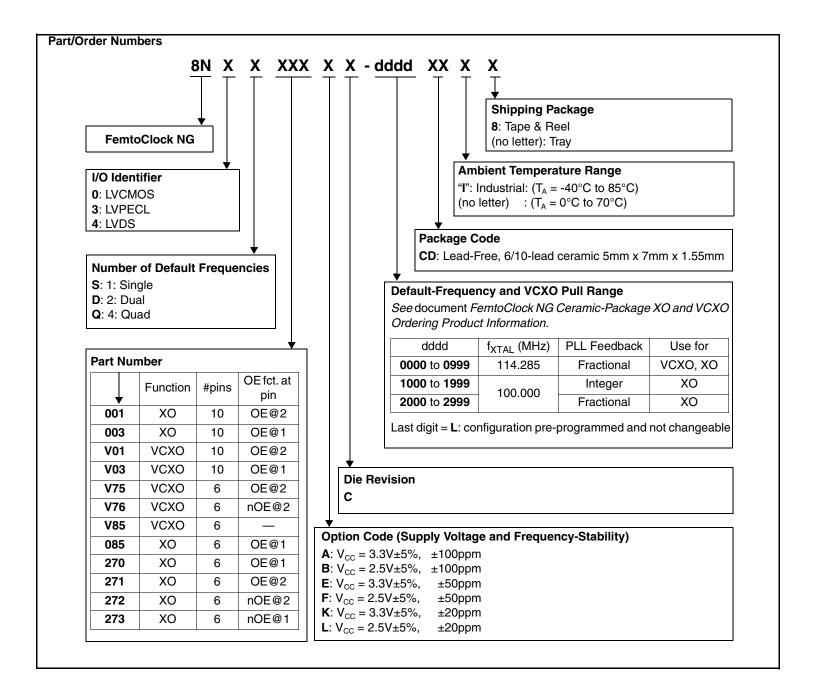


# Ordering Information for FemtoClock NG Ceramic-Package XO and VCXO Products

The programmable VCXO and XO devices support a variety of devices options such as the output type, number of default frequencies, internal crystal frequency, power supply voltage, ambient temperature range and the frequency accuracy. The device options, default frequencies and default VCXO pull range must be specified at the time of order and are programmed by IDT before the shipment. The table below specifies the available order codes, including the device options and default frequency configurations. Example part number: the order code 8N3QV01FG-0001CDI specifies a programmable, quad default-frequency VCXO with a voltage supply of 2.5V, a LVPECL output, a  $\pm 50$ ppm crystal frequency accuracy,

contains a 114.285MHz internal crystal as frequency source, industrial temperature range, a lead-free (6/6 RoHS) 6-lead ceramic 5mm x 7mm x 1.55mm package and is factory-programmed to the default frequencies of 100MHz, 122.88MHz, 125MHz and 156.25MHz and to the VCXO pull range of min.  $\pm 100$ ppm.

Other default frequencies and order codes are available from IDT on request. For more information on available default frequencies, see the FemtoClock NG Ceramic-Package XO and VCXO Ordering Product Information document.



# **Table 9. Device Marking**

	Industrial Temperature Range (T <sub>A</sub> = -40°C to 85°C)	Commercial Temperature Range (T <sub>A</sub> = 0°C to 70°C)	
Moulsing	IDT8N4S271 <b>y</b> C-	IDT8N4S271 <b>y</b> C-	
Marking	<b>dddd</b> CDI	<b>dddd</b> CD	
	y = Option Code, dddd=Default-Frequency and VCXO Pull Range		

# **Revision History Sheet**

Rev	Table	Page	Description of Change	Date
Α	Т9	15 16	Ordering Information Table - corrected Die Revision from "G" to "C".  Marking Table - corrected marking.	11/29/2012

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- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001:
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits, General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:

- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



#### Как с нами связаться

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