

Stereo, Low Power, 96 kHz, 24-Bit Audio Codec with Integrated PLL

Data Sheet

FEATURES

24-bit stereo audio ADC and DAC: >98 dB SNR Sampling rates from 8 kHz to 96 kHz Low power: 17 mW record, 18 mW playback, 48 kHz 6 analog input pins, configurable for single-ended or differential inputs Flexible analog input/output mixers Stereo digital microphone input Analog outputs: 2 differential stereo, 2 single-ended stereo, 1 mono headphone output driver PLL supporting input clocks from 8 MHz to 27 MHz Analog automatic level control (ALC) **Microphone bias reference voltage** Analog and digital I/O: 3.3 V I²C and SPI control interfaces Digital audio serial data I/O: stereo and time-division multiplexing (TDM) modes Software-controllable clickless mute 32-lead, 5 mm × 5 mm LFCSP -40°C to +105°C operating temperature range **Qualified for automotive applications**

GENERAL DESCRIPTION

The ADAU1961 is a low power, stereo audio codec that supports stereo 48 kHz record and playback at 35 mW from a 3.3 V analog supply. The stereo audio ADCs and DACs support sample rates from 8 kHz to 96 kHz as well as a digital volume control.

ADAU1961

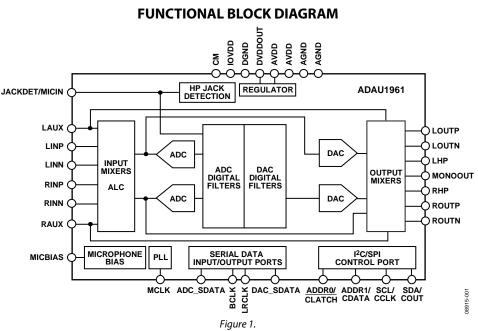
The record path includes an integrated microphone bias circuit and six inputs. The inputs can be mixed and muxed before the ADC, or they can be configured to bypass the ADC. The ADAU1961 includes a stereo digital microphone input.

The ADAU1961 includes five high power output drivers (two differential and three single-ended), supporting stereo headphones, an earpiece, or other output transducer. AC-coupled or capless configurations are supported. Individual fine level controls are supported on all analog outputs. The output mixer stage allows for flexible routing of audio.

The serial control bus supports the I²C and SPI protocols. The serial audio bus is programmable for I²S, left-/right-justified, and TDM modes. A programmable PLL supports flexible clock generation for all standard integer rates and fractional master clocks from 8 MHz to 27 MHz.

APPLICATIONS

Automotive head units Automotive amplifiers Navigation systems Rear-seat entertainment systems



Document Feedback

Rev. A

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REVISION HISTORY

4/13—Rev. 0 to Rev. A

Added Maximum Junction Temperature of 125°C	12
Updated Outline Dimensions	75

10/10—Revision 0: Initial Version

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SPECIFICATIONS

Supply voltage (AVDD) = 3.3 V, $T_A = 25^{\circ}$ C, master clock = 12.288 MHz (48 kHz fs, $256 \times \text{fs}$ mode), input sample rate = 48 kHz, measurement bandwidth = 20 Hz to 20 kHz, word width = 24 bits, C_{LOAD} (digital output) = 20 pF, I_{LOAD} (digital output) = 2 mA, $V_{\text{IH}} = 2 \text{ V}$, $V_{\text{IL}} = 0.8 \text{ V}$, unless otherwise noted. Performance of all channels is identical, exclusive of the interchannel gain mismatch and interchannel phase deviation specifications.

ANALOG PERFORMANCE SPECIFICATIONS, $T_{\text{A}} = 25^{\circ}\text{C}$

IOVDD = $3.3 \text{ V} \pm 10\%$.

Parameter	Test Conditions/Comments	Min	Тур	Мах	Unit
ANALOG-TO-DIGITAL CONVERTERS	ADC performance excludes mixers and PGA				
ADC Resolution	All ADCs		24		Bits
Digital Attenuation Step			0.375		dB
Digital Attenuation Range			95		dB
INPUT RESISTANCE					
Single-Ended Line Input	–12 dB gain		80.4		kΩ
	0 dB gain		21		kΩ
	6 dB gain		10.5		kΩ
PGA Inverting Inputs	–12 dB gain		84.5		kΩ
	0 dB gain		53		kΩ
	35.25 dB gain		1.7		kΩ
PGA Noninverting Inputs	All gains		105		kΩ
SINGLE-ENDED LINE INPUT					
Full-Scale Input Voltage (0 dB)			1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, –60 dB input				
With A-Weighted Filter (RMS)		83.5	99		dB
No Filter (RMS)		83	96		dB
Total Harmonic Distortion + Noise	–1 dBFS		-90	-71	dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)			99		dB
No Filter (RMS)			96		dB
Input Mixer Gain per Step	–12 dB to +6 dB range	2.89	3	3.07	dB
Mute Attenuation	LINPG[2:0], LINNG[2:0] = 000, RINPG[2:0], RINNG[2:0] = 000, MX1AUXG[2:0], MX2AUXG[2:0] = 000		-85.5	-77	dB
Interchannel Gain Mismatch		-0.3	+0.032	+0.3	dB
Offset Error		-5	0	+5	mV
Gain Error		-17	-12	-8	%
Interchannel Isolation			68		dB
Power Supply Rejection Ratio	CM capacitor = 20 μ F, 100 mV p-p @ 1 kHz		67		dB
PSEUDO-DIFFERENTIAL PGA INPUT					
Full-Scale Input Voltage (0 dB)			1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, –60 dB input				
With A-Weighted Filter (RMS)		94	98		dB
No Filter (RMS)		91	95		dB
Total Harmonic Distortion + Noise	–1 dBFS		-89	-83	dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)			98		dB
No Filter (RMS)			95		dB
PGA Boost Gain Error	20 dB gain setting (RDBOOST[1:0], LDBOOST[1:0] = 10)	-8	+0.4	+8	dB

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
Mute Attenuation	PGA muted				
	LDMUTE, RDMUTE $= 0$		-76	-73	dB
	RDBOOST[1:0], LDBOOST[1:0] = 00		-87	-82	dB
Interchannel Gain Mismatch		-0.6	-0.073	+0.6	dB
Offset Error		-6	0	+6	mV
Gain Error		-24	-14	-3	%
Interchannel Isolation			83	Ū.	dB
Common-Mode Rejection Ratio	100 mV rms, 1 kHz		-58		dB
	100 mV rms, 20 kHz	-52	-48	-44	dB
FULL DIFFERENTIAL PGA INPUT	Differential PGA inputs	52	10		45
Full-Scale Input Voltage (0 dB)	Differentiari Granputs		1.0 (2.83)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, –60 dB input		1.0 (2.05)		v mis (v p p)
With A-Weighted Filter (RMS)	20 Hz to 20 kHz, -00 db input	94	98		dB
-		94 91	98 95		dB
No Filter (RMS)		91		74	
Total Harmonic Distortion + Noise	-1 dBFS		-78	-74	dB
Signal-to-Noise Ratio					
With A-Weighted Filter (RMS)			98		dB
No Filter (RMS)			95		dB
PGA Boost Gain Error	20 dB gain setting (RDBOOST[1:0], LDBOOST[1:0] = 10)	-8	-0.15	+8	dB
Mute Attenuation	PGA muted				
	LDMUTE, $RDMUTE = 0$		-76	-73	dB
	RDBOOST[1:0], LDBOOST[1:0] = 00		-87	-82	dB
Interchannel Gain Mismatch		-0.3	-0.0005	+0.3	dB
Offset Error		-6	0	+6	mV
Gain Error		-17	-14	-9	%
Interchannel Isolation			83		dB
Common-Mode Rejection Ratio	100 mV rms, 1 kHz		-58		dB
	100 mV rms, 20 kHz	-52	-48	-44	dB
MICROPHONE BIAS	MBIEN = 1				
Bias Voltage					
$0.65 \times AVDD$	MBI = 1, $MPERF = 0$	2.00	2.145	2.19	V
	MBI = 1, $MPERF = 1$	2.04	2.13	2.21	V
$0.90 \times \text{AVDD}$	MBI = 0, $MPERF = 0$	2.89	2.97	3.04	V
	MBI = 0, $MPERF = 1$	2.89	2.99	3.11	V
Bias Current Source	MBI = 0, $MPERF = 1$			3	mA
Noise in the Signal Bandwidth	1 kHz to 20 kHz				
<u> </u>	MBI = 0, $MPERF = 0$		42		nV/√Hz
	MBI = 0, $MPERF = 1$		85		nV/√Hz
	MBI = 1, MPERF = 0		25		nV/√Hz
	MBI = 1, MPERF = 1	13	22	36	nV/√Hz
DIGITAL-TO-ANALOG CONVERTERS	DAC performance excludes mixers and headphone amplifier				
DAC Resolution	All DACs		24		Bits
Digital Attenuation Step			24 0.375		dB
Digital Attenuation Step					dB
			95		UD
DAC TO LINE OUTPUT					
Full-Scale Output Voltage (0 dB)			0.92 (2.60)		V rms (V p-p)
Dynamic Range	20 Hz to 20 kHz, –60 dBFS input, line				
With A Waighted Filter (DMC)	output mode	05	101		dP
With A-Weighted Filter (RMS)		95 03 5	101		dB
No Filter (RMS)		93.5	98		dB

Data Sheet

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
Total Harmonic Distortion + Noise	0 dBFS, 10 kΩ load				
Line Output Mode			-92	-77	dB
Headphone Output Mode			-89	-79	dB
Signal-to-Noise Ratio	Line output mode				
With A-Weighted Filter (RMS)			101		dB
No Filter (RMS)			98		dB
Mute Attenuation					
Mixer 3 and Mixer 4 Muted	MX3RM, MX3LM, MX4RM, MX4LM = 0, MX3AUXG[3:0], MX4AUXG[3:0] = 0000, MX3G1[3:0], MX3G2[3:0] = 0000, MX4G1[3:0], MX4G2[3:0] = 0000		-85	-78	dB
Mixer 5, Mixer 6, and Mixer 7 Muted	MX5G3[1:0], MX5G4[1:0], MX6G3[1:0], MX6G4[1:0], MX7[1:0] = 00		-89	-80	dB
All Volume Controls Muted	LOUTM, ROUTM = 0		-82	-74	dB
	MONOM, LHPM, RHPM = 0		-74	-69	dB
Interchannel Gain Mismatch		-0.3	-0.005	+0.3	dB
Offset Error		-22	0	+22	mV
Gain Error		-10	+3	+10	%
Interchannel Isolation	1 kHz, 0 dBFS input signal		100		dB
Power Supply Rejection Ratio	CM capacitor = 20 μ F, 100 mV p-p @ 1 kHz		70		dB
DAC TO HEADPHONE/EARPIECE OUTPUT	LOUTx, ROUTx, LHP, RHP in headphone output mode; P_0 = output power per channel				
Full-Scale Output Voltage (0 dB)	Scales linearly with AVDD		0.92 (2.60)		V rms (V p-p)
Total Harmonic Distortion + Noise	-4 dBFS, 16 Ω load, P ₀ = 21.1 mW		-82		dB
	-4 dBFS, 32 Ω load, P ₀ = 10.6 mW		-82		dB
Capless Headphone Mode	–2 dBFS, 16 Ω load		-78	-71	dB
	–2 dBFS, 32 Ω load		-75	-65	dB
Headphone Output Mode	0 dBFS, 10 kΩ load		-86	-77	dB
Interchannel Isolation	1 kHz, 0 dBFS input signal, 32 Ω load				
	Referred to GND		73		dB
	Referred to CM (capless headphone mode)		50		dB
Power Supply Rejection Ratio	CM capacitor = 20μ F, $100 \text{ mV p-p} @ 1 \text{ kHz}$		67		dB
REFERENCE					
Common-Mode Reference Output	CM pin	1.62	1.65	1.67	V

ANALOG PERFORMANCE SPECIFICATIONS, $-40^{\circ}C < T_A < +105^{\circ}C$

 $IOVDD = 3.3 V \pm 10\%.$

Table 2.

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
SINGLE-ENDED LINE INPUT					
Dynamic Range	20 Hz to 20 kHz, –60 dB input				
With A-Weighted Filter (RMS)		74			dB
No Filter (RMS)		71			dB
Total Harmonic Distortion + Noise	–1 dBFS			-67	dB
Input Mixer Gain per Step	–12 dB to +6 dB range	2.88		3.09	dB
Mute Attenuation	LINPG[2:0], LINNG[2:0] = 000, RINPG[2:0], RINNG[2:0] = 000, MX1AUXG[2:0], MX2AUXG[2:0] = 000			-77	dB
Interchannel Gain Mismatch		-0.5		+0.5	dB
Offset Error		-5		+5	mV
Gain Error		-22		-6	%

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
PSEUDO-DIFFERENTIAL PGA INPUT					
Dynamic Range	20 Hz to 20 kHz, –60 dB input				
With A-Weighted Filter (RMS)		94			dB
No Filter (RMS)		91			dB
Total Harmonic Distortion + Noise	–1 dBFS			-75	dB
PGA Boost Gain Error	20 dB gain setting (RDBOOST[1:0], LDBOOST[1:0] = 10)	-11		-7	dB
Mute Attenuation	PGA muted				
	LDMUTE, RDMUTE = 0			-73	dB
	RDBOOST[1:0], LDBOOST[1:0] = 00			-82	dB
Interchannel Gain Mismatch		-0.6		+0.6	dB
Offset Error		-6		+6	mV
Gain Error		-24		-3	%
Common-Mode Rejection Ratio	100 mV rms, 1 kHz	-64		-38	dB
	100 mV rms, 20 kHz	-53		-43	dB
FULL DIFFERENTIAL PGA INPUT	Differential PGA inputs				
Dynamic Range	20 Hz to 20 kHz, –60 dB input				
With A-Weighted Filter (RMS)		89			dB
No Filter (RMS)		86			dB
Total Harmonic Distortion + Noise	–1 dBFS			-70	dB
PGA Boost Gain Error	20 dB gain setting (RDBOOST[1:0], LDBOOST[1:0] = 10)	-11		-7	dB
Mute Attenuation	PGA muted				
	LDMUTE, RDMUTE = 0			-73	dB
	RDBOOST[1:0], LDBOOST[1:0] = 00			-82	dB
Interchannel Gain Mismatch		-0.4		+0.4	dB
Offset Error		-6		+6	mV
Gain Error		-21		-7	%
Common-Mode Rejection Ratio	100 mV rms, 1 kHz	-64		-38	dB
	100 mV rms, 20 kHz	-53		-43	dB
MICROPHONE BIAS Bias Voltage	MBIEN = 1				
0.65 × AVDD	MBI = 1, $MPERF = 0$	1.85		2.45	v
	MBI = 1, $MPERF = 1$	1.87		2.45	V
$0.90 \times AVDD$	MBI = 0, $MPERF = 0$	2.65		3.40	v
	MBI = 0, $MPERF = 1$	2.65		3.40	V
Noise in the Signal Bandwidth	1 kHz to 20 kHz	11		36	nV/√Hz
DAC TO LINE OUTPUT					
Dynamic Range	20 Hz to 20 kHz, –60 dB input, line output mode				
With A-Weighted Filter (RMS)		85			dB
No Filter (RMS)		78			dB
Total Harmonic Distortion + Noise Line Output Mode	0 dBFS, 10 kΩ load			-76	dB
Headphone Output Mode				-70 -78	dB
Mute Attenuation				-70	ub
Mute Attenuation Mixer 3 and Mixer 4 Muted	MX3RM, MX3LM, MX4RM, MX4LM = 0 ,			-77	dB
Mixer 5 and Mixer 4 Muteu	MX3RW, MX3LW, MX4RW, MX4LW = 0, MX3AUXG[3:0], MX4AUXG[3:0] = 0000, MX3G1[3:0], MX3G2[3:0] = 0000, MX4G1[3:0], MX4G2[3:0] = 0000			-//	UD
Mixer 5, Mixer 6, and Mixer 7 Muted	MX5G3[1:0], MX5G4[1:0], MX6G3[1:0], MX6G4[1:0], MX7[1:0] = 00			-77	dB
All Volume Controls Muted	LOUTM, ROUTM = 0			-74	dB
	MONOM, LHPM, RHPM = 0			-69	dB

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
Interchannel Gain Mismatch		-0.3		+0.3	dB
Offset Error		-22		+22	mV
Gain Error		-10		+10	%
DAC TO HEADPHONE/EARPIECE OUTPUT	LOUTx, ROUTx, LHP, RHP in headphone output mode; P ₀ = output power per channel				
Total Harmonic Distortion + Noise					
Capless Headphone Mode	–2 dBFS, 16 Ω load			-61	dB
	–2 dBFS, 32 Ω load			-63	dB
Headphone Output Mode	0 dBFS, 10 kΩ load			-76	dB
REFERENCE					
Common-Mode Reference Output	CM pin	1.47		1.83	V

POWER SUPPLY SPECIFICATIONS

 $\begin{array}{l} \text{Master clock} = 12.288 \text{ MHz, input sample rate} = 48 \text{ kHz, input tone} = 1 \text{ kHz, ADC input @ -1 dBFS, DAC input @ 0 dBFS, \\ -40^{\circ}\text{C} < T_{\text{A}} < +105^{\circ}\text{C}, \text{ IOVDD} = 3.3 \text{ V} \pm 10\%. \\ \end{array} \\ \begin{array}{l} \text{For total power consumption, add the IOVDD current listed in Table 3.} \end{array}$

Parameter	Test Conditions/Comments	Min	Тур	Max	Unit
SUPPLIES					
Voltage	DVDDOUT		1.56		V
	AVDD	2.97	3.3	3.65	V
	IOVDD	2.97	3.3	3.65	V
Digital I/O Current (IOVDD)	20 pF capacitive load on all digital pins				
Slave Mode	$f_s = 48 \text{ kHz}$		0.48		mA
	$f_s = 96 \text{ kHz}$		0.9		mA
	$f_s = 8 \text{ kHz}$		0.13		mA
Master Mode	$f_s = 48 \text{ kHz}$		1.51		mA
	$f_s = 96 \text{ kHz}$		3		mA
	$f_s = 8 \text{ kHz}$		0.27		mA
Analog Current (AVDD)					
Record Stereo Differential to ADC					
	PLL bypass		5.24		mA
	Integer PLL		6.57		mA
DAC Stereo Playback to Line Output	10 kΩ load				
	PLL bypass		5.55		mA
	Integer PLL		6.90		mA
DAC Stereo Playback to Headphone	32 Ω load				
	PLL bypass		30.9		mA
	Integer PLL		32.25		mA
DAC Stereo Playback to Capless Headphone	32 Ω load				
	PLL bypass		56.75		mA
	Integer PLL		58		mA

DIGITAL FILTERS

Table 4.

Parameter	Mode	Factor	Min	Тур	Max	Unit
ADC DECIMATION FILTER	All modes, typ @ 48 kHz					
Pass Band		0.4375 fs		21		kHz
Pass-Band Ripple				±0.015		dB
Transition Band		0.5 fs		24		kHz
Stop Band		0.5625 fs		27		kHz
Stop-Band Attenuation				67		dB
Group Delay		22.9844/fs		479		μs
DAC INTERPOLATION FILTER						
Pass Band	48 kHz mode, typ @ 48 kHz	0.4535 fs		22		kHz
	96 kHz mode, typ @ 96 kHz	0.3646 fs		35		kHz
Pass-Band Ripple	48 kHz mode, typ @ 48 kHz				±0.01	dB
	96 kHz mode, typ @ 96 kHz				±0.05	dB
Transition Band	48 kHz mode, typ @ 48 kHz	0.5 fs		24		kHz
	96 kHz mode, typ @ 96 kHz	0.5 fs		48		kHz
Stop Band	48 kHz mode, typ @ 48 kHz	0.5465 fs		26		kHz
	96 kHz mode, typ @ 96 kHz	0.6354 fs		61		kHz
Stop-Band Attenuation	48 kHz mode, typ @ 48 kHz			69		dB
	96 kHz mode, typ @ 96 kHz			68		dB
Group Delay	48 kHz mode, typ @ 48 kHz	25/f s		521		μs
	96 kHz mode, typ @ 96 kHz	11/fs		115		μs

DIGITAL INPUT/OUTPUT SPECIFICATIONS

 $-40^{\circ}C < T_A < +105^{\circ}C$, IOVDD = 3.3 V ± 10%.

Table 5.

Parameter	Test Conditions/Comments	Min Typ	Max	Unit
INPUT SPECIFICATIONS				
Input Voltage High (V _{II})		$0.7 \times IOVDD$		V
Input Voltage Low (V _{IL})			$0.3 \times IOVDD$	V
Input Leakage				
Pull-Ups/Pull-Downs Disabled	I _{IH} @ V _{IH} = 3.3 V	-0.17	+0.17	μA
	$I_{IL} @ V_{IL} = 0 V$	-0.17	+0.17	μA
	$I_{IL} @ V_{IL} = 0 V (MCLK pin)$	-13.5	-0.5	μA
Pull-Ups Enabled	I _{IH} @ V _{IH} = 3.3 V	-0.7	+0.7	μΑ
	$I_{IL} @ V_{IL} = 0 V$	-13.5	-0.5	μA
Pull-Downs Enabled	I _{IH} @ V _{IH} = 3.3 V	2.7	8.3	μΑ
	$I_{IL} @ V_{IL} = 0 V$	-0.18	+0.18	μΑ
Input Capacitance			5	рF
OUTPUT SPECIFICATIONS				
Output Voltage High (V _{он})	I _{OH} = 2 mA @ 3.3 V	$0.8 \times IOVDD$		V
Output Voltage Low (V _{OL})	I _{OL} = 2 mA @ 3.3 V		$0.1 \times IOVDD$	V

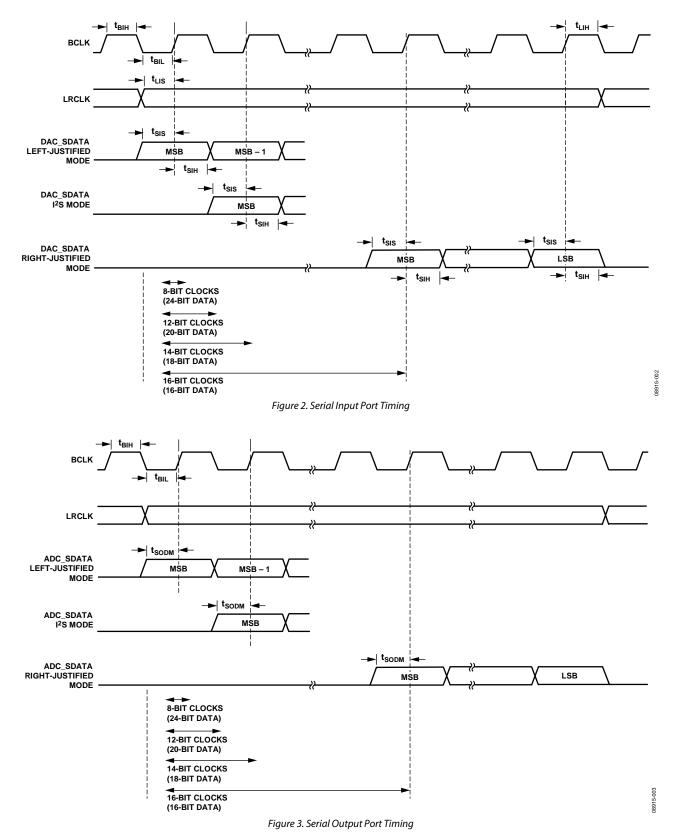
DIGITAL TIMING SPECIFICATIONS

 $-40^{o}C < T_{\rm A} < +105^{o}C, \mbox{ IOVDD} = 3.3 \ V \pm 10\%.$

Table 6. Digital Timing

	Limit				
Parameter	t _{MIN}	t _{MAX}	Unit	Description	
MASTER CLOCK					
t _{MP}	74	488	ns	MCLK period, $256 \times f_s$ mode.	
t _{мр}	37	244	ns	MCLK period, 512 × fs mode.	
t _{мр}	24.7	162.7	ns	MCLK period, 768 × fs mode.	
t _{MP}	18.5	122	ns	MCLK period, 1024 × fs mode.	
SERIAL PORT					
t _{BIL}	5		ns	BCLK pulse width low.	
t _{він}	5		ns	BCLK pulse width high.	
tlis	5		ns	LRCLK setup. Time to BCLK rising.	
tuн	5		ns	LRCLK hold. Time from BCLK rising.	
tsis	5		ns	DAC_SDATA setup. Time to BCLK rising.	
tsih	5		ns	DAC_SDATA hold. Time from BCLK rising.	
tsodm		50	ns	ADC_SDATA delay. Time from BCLK falling in master mode.	
SPI PORT					
f _{CCLK}		10	MHz	CCLK frequency.	
t _{CCPL}	10		ns	CCLK pulse width low.	
tссрн	10		ns	CCLK pulse width high.	
t _{cls}	5		ns	CLATCH setup. Time to CCLK rising.	
t _{clH}	10		ns	CLATCH hold. Time from CCLK rising.	
t _{CLPH}	10		ns	CLATCH pulse width high.	
t _{CDS}	5		ns	CDATA setup. Time to CCLK rising.	
t _{CDH}	5		ns	CDATA hold. Time from CCLK rising.	
t _{COD}	5	50	ns	COUT three-stated. Time from CLATCH rising.	
I ² C PORT		50	115	coor the stated. The non-certer rising.	
fscL		400	kHz	SCL frequency.	
	0.6	400		SCL high.	
t _{sclH}	1.3		μs	SCL low.	
tscll	0.6		μs	Setup time; relevant for repeated start condition.	
tscs	0.6		μs	Hold time. After this period, the first clock is generated.	
tscн +	100		μs	Data setup time.	
t _{Ds}	100	300	ns	SCL rise time.	
t _{SCR}			ns	SCL fall time.	
t _{SCF}		300 300	ns	SDA rise time.	
t _{sDR}			ns		
t _{sDF}	0.6	300	ns	SDA fall time.	
	0.6		μs	Bus-free time. Time between stop and start.	
DIGITAL MICROPHONE		10		$R_{LOAD} = 1 M\Omega$, $C_{LOAD} = 14 pF$.	
t _{DCF}		10	ns	Digital microphone clock fall time.	
t _{DCR}		10	ns	Digital microphone clock rise time.	
t _{DDV}	22	30	ns	Digital microphone delay time for valid data.	
t _{DDH}	0	12	ns	Digital microphone delay time for data three-stated.	

DIGITAL TIMING DIAGRAMS



Data Sheet

ADAU1961

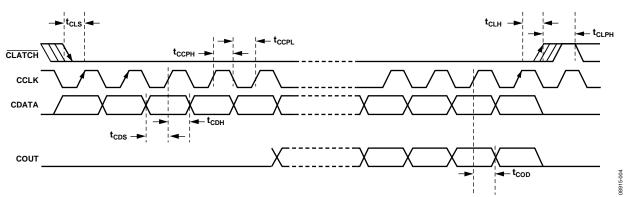


Figure 4. SPI Port Timing

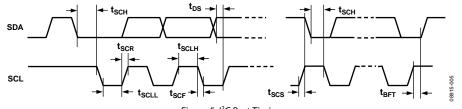


Figure 5. I²C Port Timing

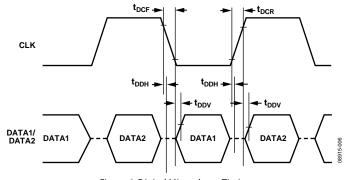


Figure 6. Digital Microphone Timing

ABSOLUTE MAXIMUM RATINGS

Table 7.

Parameter	Rating
Power Supply (AVDD)	–0.3 V to +3.65 V
Input Current (Except Supply Pins)	±20 mA
Analog Input Voltage (Signal Pins)	-0.3 V to AVDD + 0.3 V
Digital Input Voltage (Signal Pins)	-0.3 V to IOVDD + 0.3 V
Operating Temperature Range	-40°C to +105°C
Storage Temperature Range	–65°C to +150°C
Maximum Junction Temperature	125°C

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL RESISTANCE

 θ_{JA} represents thermal resistance, junction-to-ambient; θ_{JC} represents thermal resistance, junction-to-case. All characteristics are for a 4-layer board.

Table 8. Thermal Resistance

Package Type	θ _{JA}	οıc	Unit
32-Lead LFCSP	50.1	17	°C/W

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

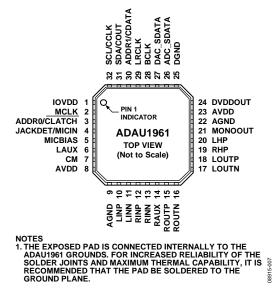


Figure 7. Pin Configuration

Table 9. Pin Function Descriptions

Pin No.	Mnemonic	Type ¹	Description
1	IOVDD	PWR	Supply for Digital Input and Output Pins. The digital output pins are supplied from IOVDD, which also sets the highest input voltage that should be seen on the digital input pins. IOVDD should be set to 3.3 V. The current draw of this pin is variable because it is dependent on the loads of the digital outputs. IOVDD should be decoupled to DGND with a 100 nF capacitor and a 10 μ F capacitor.
2	MCLK	D_IN	External Master Clock Input.
3	ADDR0/CLATCH	D_IN	I ² C Address Bit 0 (ADDR0). SPI Latch Signal (CLATCH). Must go low at the beginning of an SPI transaction and high at the end of a transaction. Each SPI transaction can take a different number of CCLKs to complete, depending on the address and read/write bit that are sent at the beginning of the SPI transaction.
4	JACKDET/MICIN	D_IN	Detect Insertion/Removal of Headphone Plug (JACKDET).
			Digital Microphone Stereo Input (MICIN).
5	MICBIAS	A_OUT	Bias Voltage for Electret Microphone.
6	LAUX	A_IN	Left Channel Single-Ended Auxiliary Input. Biased at AVDD/2.
7	СМ	A_OUT	AVDD/2 V Common-Mode Reference. A 10 μ F to 47 μ F standard decoupling capacitor should be connected between this pin and AGND to reduce crosstalk between the ADCs and DACs. This pin can be used to bias external analog circuits, as long as they are not drawing current from CM (for example, the noninverting input of an op amp).
8	AVDD	PWR	3.3 V Analog Supply for DAC and Microphone Bias. This pin should be decoupled locally to AGND with a 100 nF capacitor.
9	AGND	PWR	Analog Ground. The AGND and DGND pins can be tied together on a common ground plane. AGND should be decoupled locally to AVDD with a 100 nF capacitor.
10	LINP	A_IN	Left Channel Noninverting Input or Single-Ended Input 0. Biased at AVDD/2.
11	LINN	A_IN	Left Channel Inverting Input or Single-Ended Input 1. Biased at AVDD/2.
12	RINP	A_IN	Right Channel Noninverting Input or Single-Ended Input 2. Biased at AVDD/2.
13	RINN	A_IN	Right Channel Inverting Input or Single-Ended Input 3. Biased at AVDD/2.
14	RAUX	A_IN	Right Channel Single-Ended Auxiliary Input. Biased at AVDD/2.
15	ROUTP	A_OUT	Right Line Output, Positive. Biased at AVDD/2.
16	ROUTN	A_OUT	Right Line Output, Negative. Biased at AVDD/2.
17	LOUTN	A_OUT	Left Line Output, Negative. Biased at AVDD/2.
18	LOUTP	A_OUT	Left Line Output, Positive. Biased at AVDD/2.

Pin No.	Mnemonic	Type ¹	Description
19	RHP	A_OUT	Right Headphone Output. Biased at AVDD/2.
20	LHP	A_OUT	Left Headphone Output. Biased at AVDD/2.
20	MONOOUT	A OUT	Mono Output or Virtual Ground for Capless Headphone. Biased at AVDD/2 when set as mono
21	MONOCOT	A_001	output.
22	AGND	PWR	Analog Ground. The AGND and DGND pins can be tied together on a common ground plane. AGND should be decoupled locally to AVDD with a 100 nF capacitor.
23	AVDD	PWR	3.3 V Analog Supply for ADC, Output Driver, and Input to Digital Supply Regulator. This pin should be decoupled locally to AGND with a 100 nF capacitor.
24	DVDDOUT	PWR	Digital Core Supply Decoupling Point. The digital supply is generated from an on-board regulator and does not require an external supply. DVDDOUT should be decoupled to DGND with a 100 nF capacitor and a 10 μ F capacitor.
25	DGND	PWR	Digital Ground. The AGND and DGND pins can be tied together on a common ground plane. DGND should be decoupled to DVDDOUT and to IOVDD with 100 nF capacitors and 10 μ F capacitors.
26	ADC_SDATA	D_OUT	ADC Serial Output Data.
27	DAC_SDATA	D_IN	DAC Serial Input Data.
28	BCLK	D_IO	Serial Data Port Bit Clock.
29	LRCLK	D_IO	Serial Data Port Frame Clock.
30	ADDR1/CDATA	D_IN	I ² C Address Bit 1 (ADDR1).
			SPI Data Input (CDATA).
31	SDA/COUT	D_IO	I ² C Data (SDA). This pin is a bidirectional open-collector input/output. The line connected to this pin should have a 2 k Ω pull-up resistor.
			SPI Data Output (COUT). This pin is used for reading back registers and memory locations. It is three-state when an SPI read is not active.
32	SCL/CCLK	D_IN	I ² C Clock (SCL). This pin is always an open-collector input when in I ² C control mode. The line connected to this pin should have a 2 k Ω pull-up resistor.
			SPI Clock (CCLK). This pin can run continuously or be gated off between SPI transactions.
EP	Exposed Pad		Exposed Pad. The exposed pad is connected internally to the ADAU1961 grounds. For increased reliability of the solder joints and maximum thermal capability, it is recommended that the pad be soldered to the ground plane. See the Exposed Pad PCB Design section for more information.

¹ A_IN = analog input, A_OUT = analog output, D_IN = digital input, D_IO = digital input/output, D_OUT = digital output, PWR = power.

TYPICAL PERFORMANCE CHARACTERISTICS

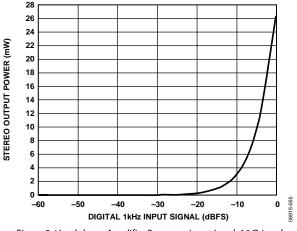


Figure 8. Headphone Amplifier Power vs. Input Level, 16 Ω Load

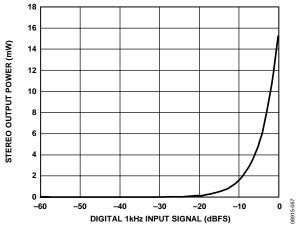


Figure 9. Headphone Amplifier Power vs. Input Level, 32Ω Load

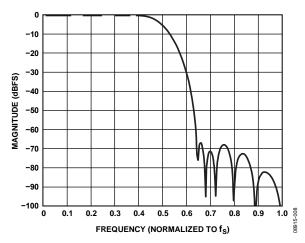


Figure 10. ADC Decimation Filter, $64 \times$ Oversampling, Normalized to f_s

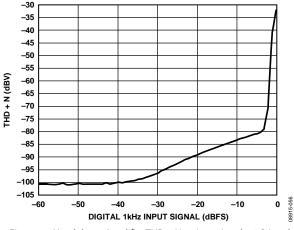


Figure 11. Headphone Amplifier THD + N vs. Input Level, 16 Ω Load

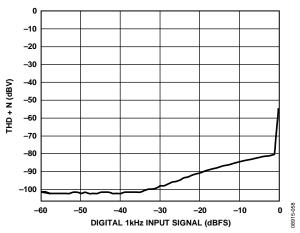


Figure 12. Headphone Amplifier THD + N vs. Input Level, 32 Ω Load

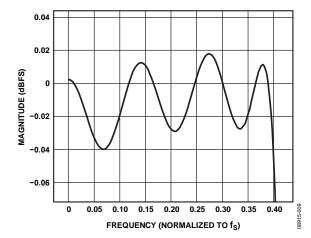


Figure 13. ADC Decimation Filter Pass-Band Ripple, 64× Oversampling, Normalized to fs

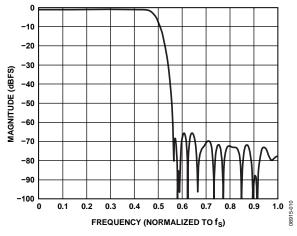


Figure 14. ADC Decimation Filter, 128× Oversampling, Normalized to fs

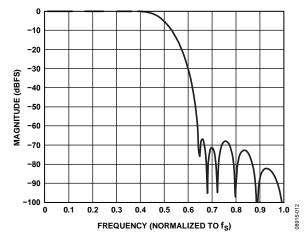


Figure 15. ADC Decimation Filter, 128× Oversampling, Double-Rate Mode, Normalized to fs

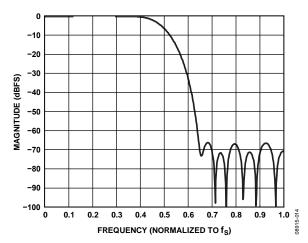


Figure 16. DAC Interpolation Filter, 64× Oversampling, Double-Rate Mode, Normalized to f_s

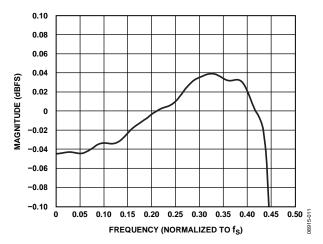


Figure 17. ADC Decimation Filter Pass-Band Ripple, 128× Oversampling, Normalized to fs

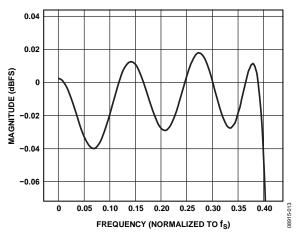


Figure 18. ADC Decimation Filter Pass-Band Ripple, 128× Oversampling, Double-Rate Mode, Normalized to fs

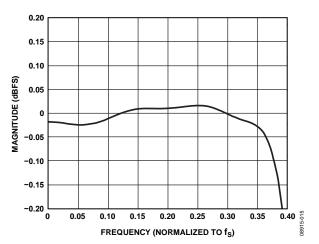


Figure 19. DAC Interpolation Filter Pass-Band Ripple, 64× Oversampling, Double-Rate Mode, Normalized to f_s

Data Sheet

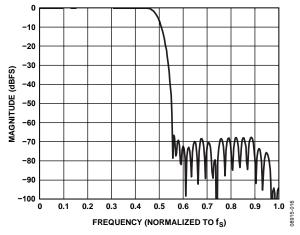


Figure 20. DAC Interpolation Filter, 128× Oversampling, Normalized to fs

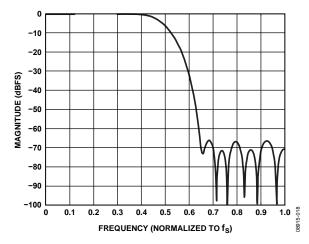


Figure 21. DAC Interpolation Filter, 128× Oversampling, Double-Rate Mode, Normalized to fs

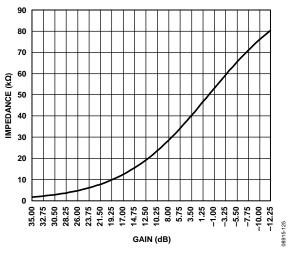


Figure 22. Input Impedance vs. Gain for Analog Inputs

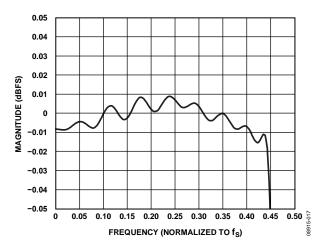


Figure 23. DAC Interpolation Filter Pass-Band Ripple, 128× Oversampling, Normalized to fs

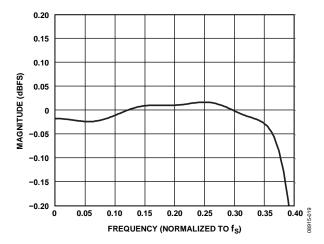
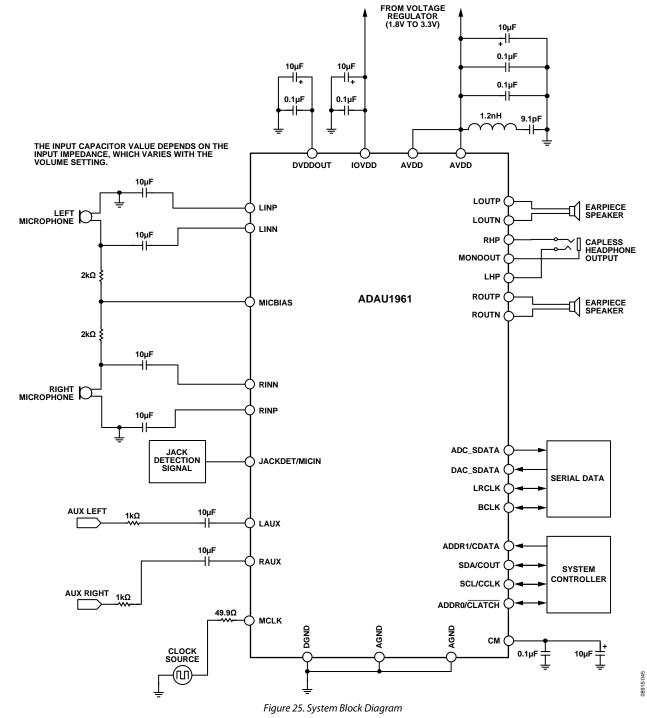


Figure 24. DAC Interpolation Filter Pass-Band Ripple, 128× Oversampling, Double-Rate Mode, Normalized to fs

ADAU1961

SYSTEM BLOCK DIAGRAMS



Data Sheet

ADAU1961

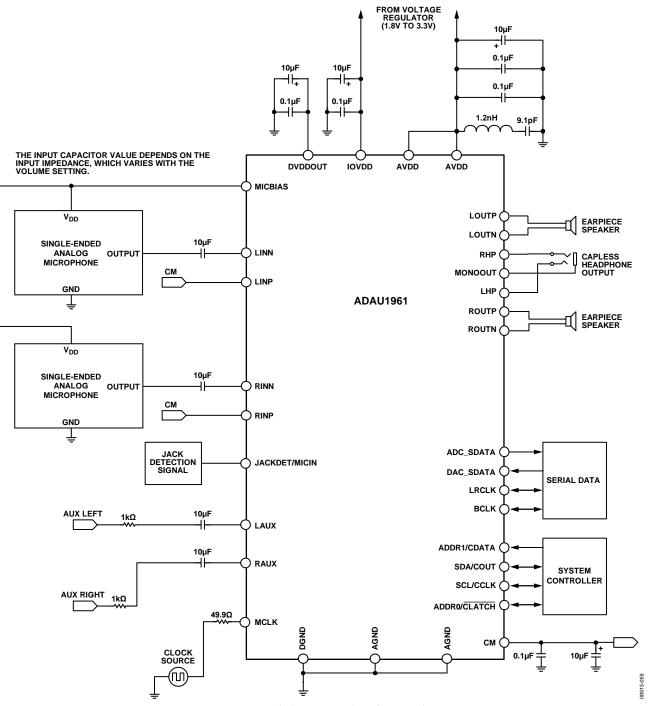


Figure 26. System Block Diagram with Analog Microphones

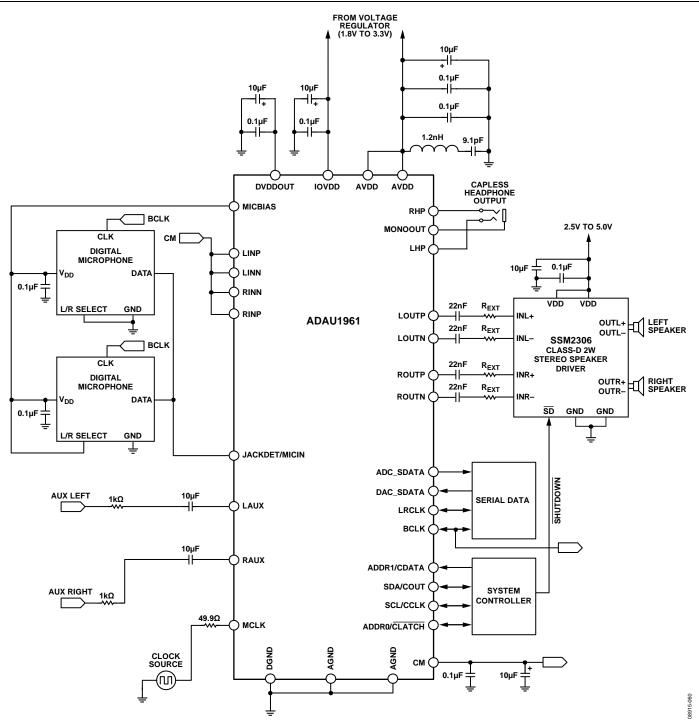


Figure 27. System Block Diagram with Digital Microphones and SSM2306 Class-D Speaker Driver

THEORY OF OPERATION

The ADAU1961 is a low power audio codec that offers high quality audio, low power, small size, and many advanced features. The stereo ADC and stereo DAC each have an SNR of at least +98 dB and a THD + N of at least -90 dB. The serial data port is compatible with I²S, left-justified, right-justified, and TDM modes for interfacing to digital audio data. The operating voltage is 3.3 V, with an on-board regulator generating the internal digital supply voltage.

The record signal path includes very flexible input configurations that can accept differential and single-ended analog microphone inputs as well as a digital microphone input. A microphone bias pin provides seamless interfacing to electret microphones. Input configurations can accept up to six single-ended analog signals or variations of stereo differential or stereo single-ended signals with two additional auxiliary single-ended inputs. Each input signal has its own programmable gain amplifier (PGA) for volume adjustment and can be routed directly to the playback path output mixers, bypassing the ADCs. An automatic level control (ALC) can also be implemented to keep the recording volume constant. The ADCs and DACs are high quality, 24-bit Σ - Δ converters that operate at selectable 64× or 128× oversampling ratios. The base sampling rate of the converters is set by the input clock rate and can be further scaled with the converter control register settings. The converters can operate at sampling frequencies from 8 kHz to 96 kHz. The ADCs and DACs also include very fine-step digital volume controls.

The playback path allows input signals and DAC outputs to be mixed into various output configurations. Headphone drivers are available for a stereo headphone output, and the other output pins are capable of differentially driving an earpiece speaker. Capless headphone outputs are possible with the use of the mono output as a virtual ground connection. The stereo line outputs can be used as either single-ended or differential outputs and as an optional mix-down mono output.

The ADAU1961 can generate its internal clocks from a wide range of input clocks by using the on-board fractional PLL. The PLL accepts inputs from 8 MHz to 27 MHz.

The ADAU1961 is provided in a small, 32-lead, 5 mm \times 5 mm LFCSP with an exposed bottom pad.

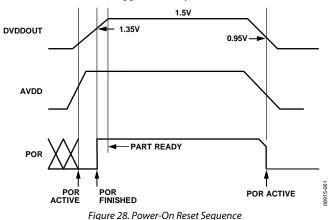
STARTUP, INITIALIZATION, AND POWER

This section describes the procedure for properly starting up the ADAU1961. The following sequence provides a high level approach to the proper initiation of the system.

- 1. Apply power to the ADAU1961.
- 2. Lock the PLL to the input clock (if using the PLL).
- 3. Enable the core clock.
- 4. Load the register settings.

POWER-UP SEQUENCE

The ADAU1961 uses a power-on reset (POR) circuit to reset the registers upon power-up. The POR monitors the DVDDOUT pin and generates a reset signal whenever power is applied to the chip. During the reset, the ADAU1961 is set to the default values documented in the register map (see the Control Registers section). Typically, with a 10 μ F capacitor on AVDD, the POR takes approximately 14 ms.



The PLL lock time is dependent on the MCLK rate. Typical lock times are provided in Table 10.

Table 10. PLL Lock Times

Tuble To: T LE Lock Times				
PLL Mode	MCLK Frequency	Lock Time (Typical)		
Fractional	8 MHz	3.5 ms		
Fractional	12 MHz	3.0 ms		
Integer	12.288 MHz	2.96 ms		
Fractional	13 MHz	2.4 ms		
Fractional	14.4 MHz	2.4 ms		
Fractional	19.2 MHz	2.98 ms		
Fractional	19.68 MHz	2.98 ms		
Fractional	19.8 MHz	2.98 ms		
Fractional	24 MHz	2.95 ms		
Integer	24.576 MHz	2.96 ms		
Fractional	26 MHz	2.4 ms		
Fractional	27 MHz	2.4 ms		

POWER REDUCTION MODES

Sections of the ADAU1961 chip can be turned on and off as needed to reduce power consumption. These include the ADCs, the DACs, and the PLL.

The digital filters of the ADCs and DACs can each be set to oversampling ratios of $64 \times$ or $128 \times$ (default). Setting the oversampling ratios to $64 \times$ for these filters lowers power consumption with a minimal impact on performance. See the Digital Filters section for specifications; see the Typical Performance Characteristics section for graphs of these filters.

DIGITAL POWER SUPPLY

The digital power supply for the ADAU1961 is generated from an internal regulator. This regulator generates a 1.5 V supply internally. The only external connection to this regulator is the DVDDOUT bypassing point. A 100 nF capacitor and a 10 μ F capacitor should be connected between this pin and DGND.

INPUT/OUTPUT POWER SUPPLY

The power for the digital output pins is supplied from IOVDD, and this pin also sets the highest input voltage that should be seen on the digital input pins. IOVDD should be set to 3.3 V; no digital input signal should be at a voltage level higher than the one on IOVDD. The current draw of this pin is variable because it depends on the loads of the digital outputs. IOVDD should be decoupled to DGND with a 100 nF capacitor and a 10 μ F capacitor.

CLOCK GENERATION AND MANAGEMENT

The ADAU1961 uses a flexible clocking scheme that enables the use of many different input clock rates. The PLL can be bypassed or used, resulting in two different approaches to clock management. For more information about clocking schemes, PLL configuration, and sampling rates, see the Clocking and Sampling Rates section.

Case 1: PLL Is Bypassed

If the PLL is bypassed, the core clock is derived directly from the MCLK input. The rate of this clock must be set properly in Register R0 (clock control register, Address 0x4000) using the INFREQ[1:0] bits. When the PLL is bypassed, supported external clock rates are $256 \times f_s$, $512 \times f_s$, $768 \times f_s$, and $1024 \times f_s$, where f_s is the base sampling rate. The core clock of the chip is off until the core clock enable bit (COREN) is asserted.

Case 2: PLL Is Used

The core clock to the entire chip is off during the PLL lock acquisition period. The user can poll the lock bit to determine when the PLL has locked. After lock is acquired, the ADAU1961 can be started by asserting the core clock enable bit (COREN) in Register R0 (clock control register, Address 0x4000). This bit enables the core clock to all the internal blocks of the ADAU1961.

PLL Lock Acquisition

During the lock acquisition period, only Register R0 (Address 0x4000) and Register R1 (Address 0x4002) are accessible through the control port. Because all other registers require a valid master clock for reading and writing, do not attempt to access any other register. Any read or write is prohibited until the core clock enable bit (COREN) and the lock bit are both asserted.

To program the PLL during initialization or reconfiguration of the clock setting, the following procedure must be followed:

- 1. Power down the PLL.
- 2. Reset the PLL control register.
- 3. Start the PLL.
- 4. Poll the lock bit.
- 5. Assert the core clock enable bit after the PLL lock is acquired.

The PLL control register (Register R1, Address 0x4002) is a 48-bit register where all bits must be written with a single continuous write to the control port.

CLOCKING AND SAMPLING RATES

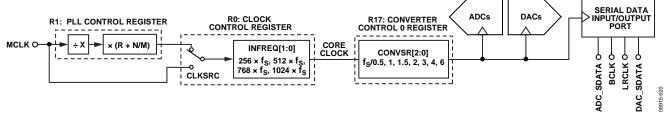


Figure 29. Clock Tree Diagram

CORE CLOCK

Clocks for the converters and the serial ports are derived from the core clock. The core clock can be derived directly from MCLK or it can be generated by the PLL. The CLKSRC bit (Bit 3 in Register R0, Address 0x4000) determines the clock source.

The INFREQ[1:0] bits should be set according to the expected input clock rate selected by CLKSRC; this value also determines the core clock rate and the base sampling frequency, fs.

For example, if the input to CLKSRC = 49.152 MHz (from PLL), then

 $INFREQ[1:0] = 1024 \times f_s$

 $f_s = 49.152 \text{ MHz}/1024 = 48 \text{ kHz}$

The PLL output clock rate is always $1024 \times f_s$, and the clock control register automatically sets the INFREQ[1:0] bits to $1024 \times f_s$ when using the PLL. When using a direct clock, the INFREQ[1:0] frequency should be set according to the MCLK pin clock rate and the desired base sampling frequency.

Table 11. Clock Control Register (Register R0, Address 0x4000)

Bits	Bit Name	Settings
3	CLKSRC	0: Direct from MCLK pin (default) 1: PLL clock
[2:1]	INFREQ[1:0]	00: $256 \times f_s$ (default) 01: $512 \times f_s$ 10: $768 \times f_s$ 11: $1024 \times f_s$
0	COREN	0: Core clock disabled (default) 1: Core clock enabled

SAMPLING RATES

The ADCs, DACs, and serial port share a common sampling rate that is set in Register R17 (Converter Control 0 register, Address 0x4017). The CONVSR[2:0] bits set the sampling rate as a ratio of the base sampling frequency.

Table 12 and Table 13 list the sampling rate divisions for common base sampling rates.

Table 12. 48 kHz Base	Sampling Rate Divisions
-----------------------	-------------------------

Base Sampling Frequency	Sampling Rate Scaling	Sampling Rate
$f_s = 48 \text{ kHz}$	f _s /1	48 kHz
	fs/6	8 kHz
	f _s /4	12 kHz
	fs/3	16 kHz
	fs/2	24 kHz
	fs/1.5	32 kHz
	fs/0.5	96 kHz

Base Sampling Frequency	Sampling Rate Scaling	Sampling Rate
fs = 44.1 kHz	fs/1	44.1 kHz
	fs/6	7.35 kHz
	fs/4	11.025 kHz
	fs/3	14.7 kHz
	fs/2	22.05 kHz
	fs/1.5	29.4 kHz
	f _s /0.5	88.2 kHz

PLL

The PLL uses the MCLK as a reference to generate the core clock. PLL settings are set in Register R1 (PLL control register, Address 0x4002). Depending on the MCLK frequency, the PLL must be set for either integer or fractional mode. The PLL can accept input frequencies in the range of 8 MHz to 27 MHz.

All six bytes in the PLL control register must be written with a single continuous write to the control port.

$$\mathsf{MCLK} \circ \longrightarrow figure 30. PLL Block Diagram$$

Integer Mode

Integer mode is used when the MCLK is an integer (R) multiple of the PLL output $(1024 \times f_s)$.

For example, if MCLK = 12.288 MHz and f_s = 48 kHz, then

PLL required output = 1024×48 kHz = 49.152 MHz

R = 49.152 MHz/12.288 MHz = 4

In integer mode, the values set for N and M are ignored.

• •

Fractional Mode

Fractional mode is used when the MCLK is a fractional (R + (N/M)) multiple of the PLL output.

For example, if MCLK = 12 MHz and f_s = 48 kHz, then

PLL required output = 1024×48 kHz = 49.152 MHz

R + (N/M) = 49.152 MHz/12 MHz = 4 + (12/125)

Common fractional PLL parameter settings for 44.1 kHz and 48 kHz sampling rates can be found in Table 15 and Table 16.

The PLL outputs a clock in the range of 41 MHz to 54 MHz, which should be taken into account when calculating PLL values and MCLK frequencies.

Bits	Bit Name	Description
[47:32]	M[15:0]	Denominator of the fractional PLL: 16-bit binary number
		0x00FD: M = 253 (default)
[31:16]	N[15:0]	Numerator of the fractional PLL: 16-bit binary number
		0x000C: N = 12 (default)
[14:11]	R[3:0]	Integer part of PLL: four bits, only values 2 to 8 are valid
		0010: R = 2 (default)
		0011: R = 3
		0100: R = 4
		0101: R = 5
		0110: R = 6 0111: R = 7
		1000: R = 8
[10:9]	X[1:0]	PLL input clock divider
[10.5]	X[1.0]	00: X = 1 (default)
		01: X = 2
		10: X = 3
		11: X = 4
8	Туре	PLL operation mode
		0: Integer (default)
		1: Fractional
1	Lock	PLL lock (read-only bit)
		0: PLL unlocked (default)
		1: PLL locked
0	PLLEN	PLL enable
		0: PLL disabled (default)
		1: PLL enabled

Table 14. PLL Co	ontrol Register (R	egister	R1,	Address 0x4002)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)
8	1	5	625	403	0x0271 0193 2901
12	1	3	625	477	0x0271 01DD 1901
13	1	3	8125	3849	0x1FBD 0F09 1901
14.4	2	6	125	34	0x007D 0022 3301
19.2	2	4	125	88	0x007D 0058 2301
19.68	2	4	1025	604	0x0401 025C 2301
19.8	2	4	1375	772	0x055F 0304 2301
24	2	3	625	477	0x0271 01DD 1B01
26	2	3	8125	3849	0x1FBD 0F09 1B01
27	2	3	1875	647	0x0753 0287 1B01

Table 15. Fractional PLL Parameter Settings for $f_s = 44.1$ kHz (PLL Output = 45.1584 MHz = $1024 \times f_s$)

Table 16. Fractional PLL Parameter Settings for $f_s = 48$ kHz (PLL Output = 49.152 MHz = $1024 \times f_s$)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex)
8	1	6	125	18	0x007D 0012 3101
12	1	4	125	12	0x007D 000C 2101
13	1	3	1625	1269	0x0659 04F5 1901
14.4	2	6	75	62	0x004B 003E 3301
19.2	2	5	25	3	0x0019 0003 2B01
19.68	2	4	205	204	0x00CD 00CC 2301
19.8	2	4	825	796	0x0339 031C 2301
24	2	4	125	12	0x007D 000C 2301
26	2	3	1625	1269	0x0659 04F5 1B01
27	2	3	1125	721	0x0465 02D1 1B01

Table 17. Integer PLL Parameter Settings for $f_s = 48$ kHz (PLL Output = 49.152 MHz = $1024 \times f_s$)

MCLK Input (MHz)	Input Divider (X)	Integer (R)	Denominator (M)	Numerator (N)	R2: PLL Control Setting (Hex) ¹			
12.288	1	4	Don't care	Don't care	0xXXXX XXXX 2001			
24.576	1	2	Don't care	Don't care	0xXXXX XXXX 1001			

¹ X = don't care.

RECORD SIGNAL PATH

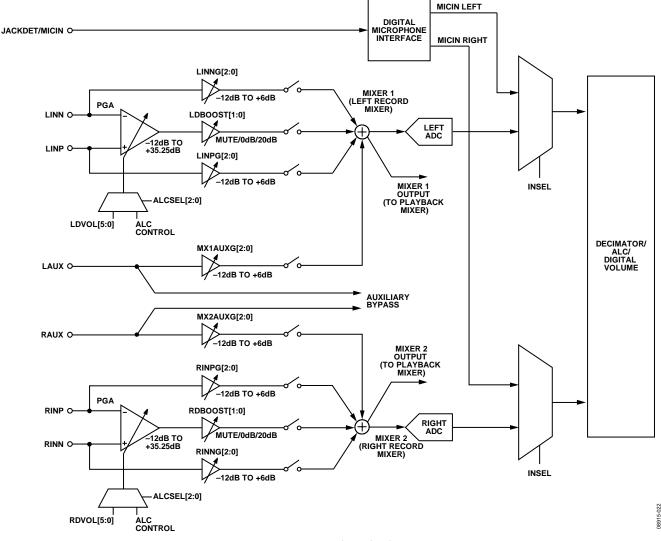


Figure 31. Record Signal Path

INPUT SIGNAL PATHS

The ADAU1961 can accept both line level and microphone inputs. The analog inputs can be configured in a single-ended or differential configuration. There is also an input for a digital microphone. The analog inputs are biased at AVDD/2. Unused input pins should be connected to CM.

Each of the six analog inputs has individual gain controls (boost or cut). The input signals are mixed and routed to an ADC. The mixed input signals can also bypass the ADCs and be routed directly to the playback mixers. Left channel inputs are mixed before the left ADC; however, it is possible to route the mixed analog signal around the ADC and output it into a left or right output channel. The same capabilities apply to the right channel and the right ADC. Signals are inverted through the PGAs and the mixers. The result of this inversion is that differential signals input through the PGA are output from the ADCs at the same polarity as they are input. Single-ended inputs that pass through the mixer but not through the PGA are inverted. The ADCs are noninverting.

The input impedance of the analog inputs varies with the gain of the PGA. This impedance ranges from 1.7 k Ω at the 35.25 dB gain setting to 80.4 k Ω at the –12 dB setting. This range is shown in Figure 22.

Analog Microphone Inputs

For microphone inputs, configure the part in either stereo pseudo-differential mode or stereo full differential mode.

The LINN and LINP pins are the inverting and noninverting inputs for the left channel, respectively. The RINN and RINP pins are the inverting and noninverting inputs for the right channel, respectively.

For a differential microphone input, connect the positive signal to the noninverting input of the PGA and the negative signal to the inverting input of the PGA, as shown in Figure 32. The PGA settings are controlled with Register R8 (left differential input volume control register, Address 0x400E) and Register R9 (right differential input volume control register, Address 0x400F). The PGA must first be enabled by setting the RDEN and LDEN bits.

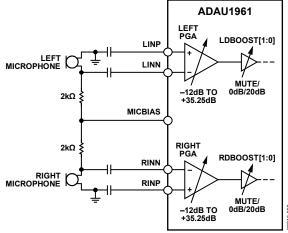


Figure 32. Stereo Differential Microphone Configuration

The PGA can also be used for single-ended microphone inputs. Connect LINP and/or RINP to the CM pin. In this configuration, the signal connects to the inverting input of the PGA, LINN and/or RINN, as shown in Figure 33.

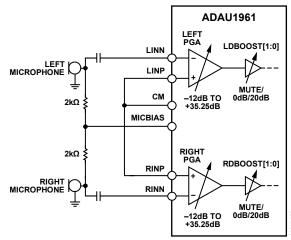


Figure 33. Stereo Single-Ended Microphone Configuration

Analog Line Inputs

Line input signals can be accepted by any analog input. It is possible to route signals on the RINN, RINP, LINN, and LINP pins around the differential amplifier to their own amplifier and to use these pins as single-ended line inputs by disabling the LDEN and RDEN bits (Bit 0 in Register R8, Address 0x400E, and Bit 0 in Register R9, Address 0x400F). Figure 34 depicts a stereo single-ended line input using the RINN and LINN pins.

The LAUX and RAUX pins are single-ended line inputs. They can be used together as a stereo single-ended auxiliary input, as shown in Figure 34. These inputs can bypass the input gain control, mixers, and ADCs to directly connect to the output playback mixers (see auxiliary bypass in Figure 31).

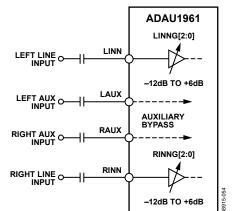


Figure 34. Stereo Single-Ended Line Input with Stereo Auxiliary Bypass

Digital Microphone Input

When using a digital microphone connected to the JACKDET/ MICIN pin, the JDFUNC[1:0] bits in Register R2 (Address 0x4008) must be set to 10 to enable the microphone input and disable the jack detection function. The ADAU1961 must operate in master mode and source BCLK to the input clock of the digital microphone.

The digital microphone signal bypasses record path mixers and ADCs and is routed directly into the decimation filters. The digital microphone and ADCs share decimation filters and, therefore, both cannot be used simultaneously. The digital microphone input select bit, INSEL, can be set in Register R19 (ADC control register, Address 0x4019). Figure 35 depicts the digital microphone interface and signal routing.

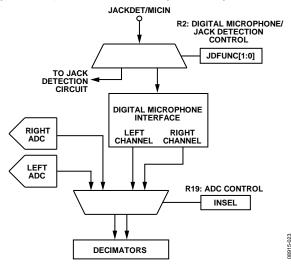


Figure 35. Digital Microphone Interface Block Diagram

Microphone Bias

The MICBIAS pin provides a voltage reference for electret analog microphones. The MICBIAS voltage is set in Register R10 (record microphone bias control register, Address 0x4010). In this register, the MICBIAS output can be enabled or disabled. Additional options include high performance operation and a gain boost. The gain boost provides two different voltage biases: $0.65 \times \text{AVDD}$ or $0.90 \times \text{AVDD}$. When enabled, the high performance bit increases supply current to the microphone bias circuit to decrease rms input noise.

The MICBIAS pin can also be used to cleanly supply voltage to digital microphones or analog microphones with separate power supply pins.

ANALOG-TO-DIGITAL CONVERTERS

The ADAU1961 uses two 24-bit Σ - Δ analog-to-digital converters (ADCs) with selectable oversampling ratios of 64× or 128× (selected by Bit 3 in Register R17, Address 0x4017).

ADC Full-Scale Level

The full-scale input to the ADCs (0 dBFS) is 1.0 V rms with AVDD = 3.3 V. This full-scale analog input will output a digital signal at -1.38 dBFS. This gain offset is built into the ADAU1961 to prevent clipping. The full-scale input level scales linearly with the level of AVDD.

For single-ended and pseudo-differential signals, the full-scale value corresponds to the signal level at the pins, 0 dBFS.

The full differential full-scale input level is measured after the differential amplifier, which corresponds to -6 dBFS at each pin.

Signal levels above the full-scale value cause the ADCs to clip.

Digital ADC Volume Control

The digital ADC volume can be attenuated using Register R20 (left input digital volume register, Address 0x401A) and Register R21 (right input digital volume register, Address 0x401B).

High-Pass Filter

By default, a high-pass filter is used in the ADC path to remove dc offsets; this filter can be enabled or disabled in Register R19 (ADC control register, Address 0x4019). At $f_s = 48$ kHz, the corner frequency of this high-pass filter is 2 Hz.

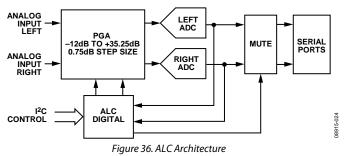
AUTOMATIC LEVEL CONTROL (ALC)

The ADAU1961 contains a hardware automatic level control (ALC). The ALC is designed to continuously adjust the PGA gain to keep the recording volume constant as the input level varies.

For optimal noise performance, the ALC uses the analog PGA to adjust the gain instead of using a digital method. This ensures that the ADC noise is not amplified at low signal levels. Extremely small gain step sizes are used to ensure high audio quality during gain changes.

To use the ALC function, the inputs must be applied either differentially or pseudo-differentially to input pins LINN and LINP, for the left channel, and RINN and RINP, for the right channel. The ALC function is not available for the auxiliary line input pins, LAUX and RAUX.

A block diagram of the ALC block is shown in Figure 36. The ALC logic receives the ADC output signals and analyzes these digital signals to set the PGA gain. The ALC control registers are used to control the time constants and output levels, as described in this section.



ALC PARAMETERS

The ALC function is controlled with the ALC control registers (Address 0x4011 through Address 0x4014) using the following parameters:

- ALCSEL[2:0]: The ALC select bits are used to enable the ALC and set the mode to left only, right only, or stereo. In stereo mode, the greater of the left or right inputs is used to calculate the gain, and the same gain is then applied to both the left and right channels.
- ALCTARG[3:0]: The ALC target is the desired input recording level that the ALC attempts to achieve.

- ALCATCK[3:0]: The ALC attack time sets how fast the ALC starts attenuating after a sudden increase in input level above the ALC target. Although it may seem that the attack time should be set as fast as possible to avoid clipping on transients, using a moderate value results in better overall sound quality. If the value is too fast, the ALC overreacts to very short transients, causing audible gain-pumping effects, which sounds worse than using a moderate value that allows brief periods of clipping on transients. A typical setting for music recording is 384 ms. A typical setting for voice recording is 24 ms.
- ALCHOLD[3:0]: These bits set the ALC hold time. When the output signal falls below the target output level, the gain is not increased unless the output remains below the target level for the period of time set by the hold time bits. The hold time is used to prevent the gain from modulating on a steady low frequency sine wave signal, which would cause distortion.
- ALCDEC[3:0]: The ALC decay time sets how fast the ALC increases the PGA gain after a sudden decrease in input level below the ALC target. A very slow setting can be used if the main function of the ALC is to set a music recording level. A faster setting can be used if the function of the ALC is to compress the dynamic range of a voice recording. Using a very fast decay time can cause audible artifacts such as noise pumping or distortion. A typical setting for music recording is 24.58 sec. A typical setting for voice recording is 1.54 sec.
- ALCMAX[2:0]: The maximum ALC gain bits are used to limit the maximum gain that can be programmed into the ALC. This can be used to prevent excessive noise in the recording for small input signals. Note that setting this register to a low value may prevent the ALC from reaching its target output level, but this behavior is often desirable to achieve the best overall sound.

Figure 37 shows the dynamic behavior of the PGA gain for a tone-burst input. The target output is achieved for three different input levels, with the effect of attack, hold, and decay shown in the figure. Note that for very small signals, the maximum PGA gain may prevent the ALC from achieving its target level; in the same way, for very large inputs, the minimum PGA gain may prevent the ALC from achieving its target level (assuming that the target output level is set to a very low value). The effects of the PGA gain limit are shown in the input/output graph of Figure 38.

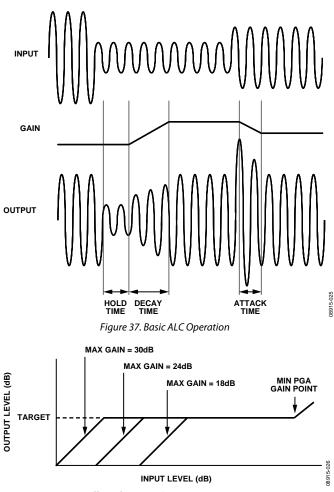


Figure 38. Effect of Varying the Maximum Gain Parameter

NOISE GATE FUNCTION

When using the ALC, one potential problem is that for small input signals, the PGA gain can become very large. A side effect of this is that the noise is amplified along with the signal of interest. To avoid this situation, the ADAU1961 noise gate can be used. The noise gate cuts off the ADC output when its signal level is below a set threshold. The noise gate is controlled using the following parameters in the ALC Control 3 register (Address 0x4014):

- NGTYP[1:0]: The noise gate type is set to one of four modes by writing to the NGTYP[1:0] bits.
- NGEN: The noise gate function is enabled by writing to the NGEN bit.
- NGTHR[4:0]: The threshold for muting the output is set by writing to the NGTHR[4:0] bits.

One common problem with noise gate functions is chatter, where a small signal that is close to the noise gate threshold varies in amplitude, causing the noise gate function to open and close rapidly. This causes an unpleasant sound.

To reduce this effect, the noise gate in the ADAU1961 uses a combination of a timeout period and hysteresis. The timeout period is set to 250 ms, so the signal must consistently be below

the threshold for 250 ms before the noise gate operates. Hysteresis is used so that the threshold for coming out of the mute state is 6 dB higher than the threshold for going into the mute state. There are four operating modes for the noise gate.

Noise Gate Mode 0 (see Figure 39) is selected by setting the NGTYP[1:0] bits to 00. In this mode, the current state of the PGA gain is held at its current state when the noise gate logic is activated. This prevents a large increase in background noise during periods of silence. When using this mode, it is advisable to use a relatively slow decay time. This is because the noise gate takes at least 250 ms to activate, and if the PGA gain has already increased to a large value during this time, the value at which the gain is held will be large.

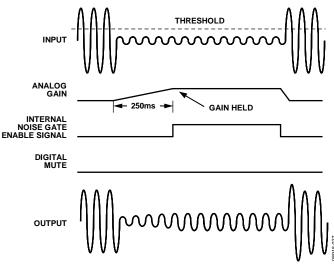


Figure 39. Noise Gate Mode 0 (PGA Gain Hold)

Noise Gate Mode 1 (see Figure 40) is selected by setting the NGTYP[1:0] bits to 01. In this mode, the ADAU1961 does a simple digital mute of the ADC output. Although this mode completely eliminates any background noise, the effect of an abrupt mute may not be pleasant to the ear.

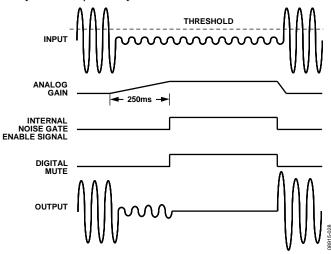


Figure 40. Noise Gate Mode 1 (Digital Mute)

Noise Gate Mode 2 (see Figure 41) is selected by setting the NGTYP[1:0] bits to 10. In this mode, the ADAU1961 improves the sound of the noise gate operation by first fading the PGA gain over a period of about 100 ms to the minimum PGA gain value. The ADAU1961 does not do a hard mute after the fade is complete, so some small background noise will still exist.

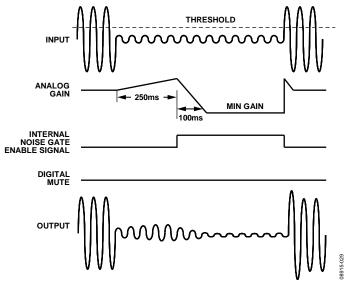


Figure 41. Noise Gate Mode 2 (Analog Fade)

Noise Gate Mode 3 (see Figure 42) is selected by setting the NGTYP[1:0] bits to 11. This mode is the same as Mode 2 except that at the end of the PGA fade gain interval, a digital mute is performed. In general, this mode is the best-sounding mode, because the audible effect of the digital hard mute is reduced by the fact that the gain has already faded to a low level before the mute occurs.

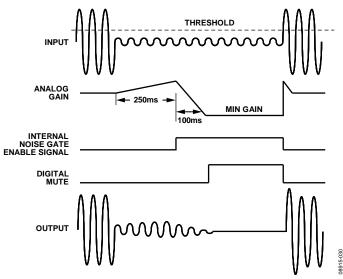
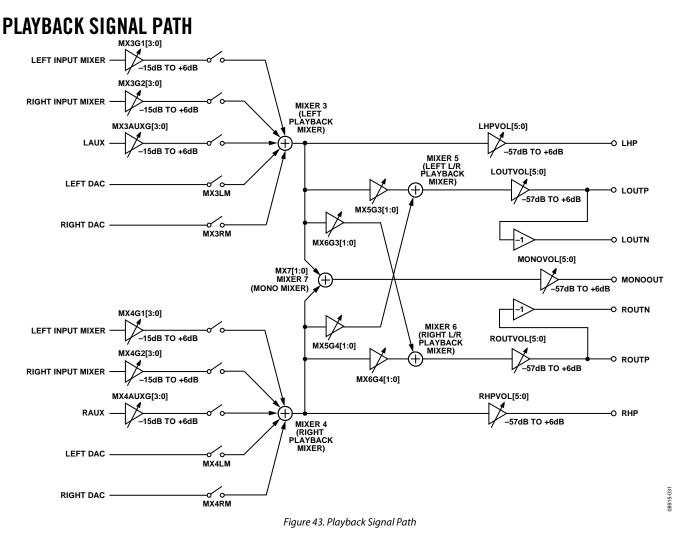


Figure 42. Noise Gate Mode 3 (Analog Fade/Digital Mute)



OUTPUT SIGNAL PATHS

The outputs of the ADAU1961 can be configured as a variety of differential or single-ended outputs. All analog output pins are capable of driving headphone or earpiece speakers. There are selectable output paths for stereo signals or a downmixed mono output. The line outputs can drive a load of at least 10 k Ω or can be put into HP mode to drive headphones or earpiece speakers. The analog output pins are biased at AVDD/2.

With a 0 dBFS digital input and AVDD = 3.3 V, the full-scale output level is 920 mV rms.

Signals are inverted through the mixers and volume controls. The result of this inversion is that the polarity of the differential outputs and the headphone outputs is preserved. The singleended mono output is inverted. The DACs are noninverting.

Routing Flexibility

The playback path contains five mixers (Mixer 3 to Mixer 7) that perform the following functions:

- Mix signals from the record path and the DACs.
- Mix or swap the left and right channels.
- Mix a mono signal or generate a common-mode output.

Mixer 3 and Mixer 4 are dedicated to mixing signals from the record path and the DACs. Each of these two mixers can accept signals from the left and right DACs, the left and right input mixers, and the dedicated channel auxiliary input. Signals coming from the record path can be boosted or cut before the playback mixer.

For example, the MX4G2[3:0] bits set the gain from the output of Mixer 2 (right record channel) to the input of Mixer 4, hence the naming convention.

Signals coming from the DACs have digital volume attenuation controls set in Register R20 (left input digital volume register, Address 0x401A) and Register R21 (right input digital volume register, Address 0x401B).

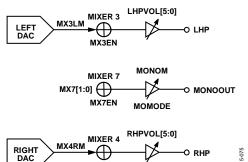
HEADPHONE OUTPUT

The LHP and RHP pins can be driven by either a line output driver or a headphone driver by setting the HPMODE bit in Register R30 (playback headphone right volume control register, Address 0x4024). The headphone outputs can drive a load of at least 16 Ω .

Separate volume controls for the left and right channels range from -57 dB to +6 dB. Slew can be applied to all the playback volume controls using the ASLEW[1:0] bits in Register R34 (playback pop/click suppression register, Address 0x4028).

Capless Headphone Configuration

The headphone outputs can be configured in a capless output configuration with the MONOOUT pin used as a dc virtual ground reference. Figure 44 depicts a typical playback path in a capless headphone configuration. Table 18 lists the register settings for this configuration. As shown in this table, the MONOOUT pin outputs common mode (AVDD/2), which is used as the virtual headphone reference.



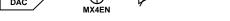


Figure 44. Capless Headphone Configuration Diagram

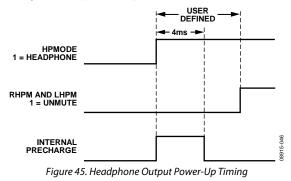
Table 18. Capless Headphone Register Settings

Register	Bit Name	Setting
R36	DACEN[1:0]	11 = both DACs on
R22	MX3EN	1 = enable Mixer 3
	MX3LM	1 = unmute left DAC input
R24	MX4EN	1 = enable Mixer 4
	MX4RM	1 = unmute right DAC input
R28	MX7EN	1 = enable Mixer 7
	MX7[1:0]	00 = common-mode output
R33	MONOM	1 = unmute mono output
	MOMODE	1 = headphone output
R29	LHPVOL[5:0]	Desired volume for LHP output
	LHPM	1 = unmute left headphone output
R30	HPMODE	1 = headphone output
	RHPVOL[5:0]	Desired volume for RHP output
	RHPM	1 = unmute right headphone output

Headphone Output Power-Up/Power-Down Sequencing

To prevent pops when turning on the headphone outputs, the user must wait at least 4 ms to unmute these outputs after enabling the headphone output with the HPMODE bit. This is because of an internal capacitor that must charge before these outputs can be used. Figure 45 and Figure 46 illustrate the headphone power-up/power-down sequencing.

For capless headphones, configure the MONOOUT pin before unmuting the headphone outputs.



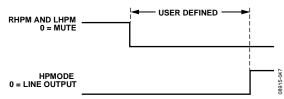


Figure 46. Headphone Output Power-Down Timing

Ground-Centered Headphone Configuration

The headphone outputs can also be configured as groundcentered outputs by placing coupling capacitors on the LHP and RHP pins. Ground-centered headphones should use the AGND pin as the ground reference.

When the headphone outputs are configured in this manner, the capacitors create a high-pass filter on the outputs. The corner frequency of this filter, at which point its attenuation is 3 dB, is calculated by the following formula:

$f_{3dB} =$	$1/(2\pi \times R \times C)$

where:

C is the capacitor value. *R* is the impedance of the headphones.

For a typical headphone impedance of 16 Ω and a 47 μF capacitor, the corner frequency is 211 Hz.

Jack Detection

When the JACKDET/MICIN pin is set to the jack detect function, a flag on this pin can be used to mute the line outputs when headphones are plugged into the jack. This pin can be configured in Register R2 (digital microphone/jack detection control register, Address 0x4008). The JDFUNC[1:0] bits set the functionality of the JACKDET/MICIN pin.

Additional settings for jack detection include debounce time (JDDB[1:0] bits) and detection polarity (JDPOL bit). Because the jack detection and digital microphone share a pin, both functions cannot be used simultaneously.

POP-AND-CLICK SUPPRESSION

Upon power-up, precharge circuitry is enabled to suppress pops and clicks. After power-up, the precharge circuitry can be put into a low power mode using the POPMODE bit in Register R34 (playback pop/click suppression register, Address 0x4028).

The precharge time depends on the capacitor value on the CM pin and the RC time constant of the load. For a typical line output load, the precharge time is between 2 ms and 3 ms. After this precharge time, the POPMODE bit can be set to low power mode.

Changing any register settings that affect the signal path can cause pops and clicks on the analog outputs. To avoid these pops and clicks, mute the appropriate outputs using Register R29 to Register R32 (Address 0x4023 to Address 0x4026). Unmute the analog outputs after the changes are made.

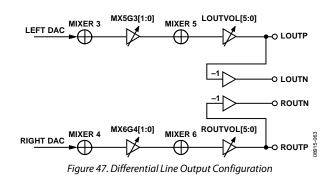
LINE OUTPUTS

The line output pins (LOUTP, LOUTN, ROUTP, and ROUTN) can be used to drive both differential and single-ended loads. In their default settings, these pins can drive typical line loads of 10 k Ω or greater, but they can also be put into headphone mode by setting the LOMODE bit in Register R31 (playback line output left volume control register, Address 0x4025) and the ROMODE bit in Register R32 (playback line output right volume control register, Address 0x4026). In headphone mode, the line output pins are capable of driving headphone and earpiece speakers of 16 Ω or greater. The output impedance of the line outputs is approximately 1 k Ω .

When the line output pins are used in single-ended mode, LOUTP and ROUTP should be used to output the signals, and LOUTN and ROUTN should be left unconnected.

The volume controls for these outputs range from -57 dB to +6 dB. Slew can be applied to all the playback volume controls using the ASLEW[1:0] bits in Register R34 (playback pop/click suppression register, Address 0x4028).

The MX5G4[1:0], MX5G3[1:0], MX6G3[1:0], and MX6G4[1:0] bits can all provide a 6 dB gain boost to the line outputs. This gain boost allows single-ended output signals to achieve 0 dBV (1.0 V rms) and differential output signals to achieve up to 6 dBV (2.0 V rms). For more information, see Register R26 (playback L/R mixer left (Mixer 5) line output control register, Address 0x4020) and Register R27 (playback L/R mixer right (Mixer 6) line output control register, Address 0x4021).



CONTROL PORTS

The ADAU1961 can operate in one of two control modes:

- I²C control
- SPI control

The ADAU1961 has both a 4-wire SPI control port and a 2-wire I²C bus control port. Both ports can be used to set the registers. The part defaults to I²C mode, but it can be put into SPI control mode by pulling the CLATCH pin low three times.

The control port is capable of full read/write operation for all addressable registers. The ADAU1961 must have a valid master clock in order to write to all registers except for Register R0 (Address 0x4000) and Register R1 (Address 0x4002).

All addresses can be accessed in both a single-address mode or a burst mode. The first byte (Byte 0) of a control port write contains the 7-bit chip address plus the R/W bit. The next two bytes (Byte 1 and Byte 2) together form the subaddress of the register location within the ADAU1961. This subaddress must be two bytes long because the memory locations within the ADAU1961 are directly addressable and their sizes exceed the range of single-byte addressing. All subsequent bytes (starting with Byte 3) contain the data. The number of bytes per word depends on the type of data that is being written.

The control port pins are multifunctional, depending on the mode in which the part is operating. Table 19 describes these multiple functions.

Table 19. Control Port Pin Functions

Pin Name	I ² C Mode	SPI Mode				
SCL/CCLK	SCL: input clock	CCLK: input clock				
SDA/COUT	SDA: open-collector input/output	COUT: output				
ADDR1/CDATA	I ² C Address Bit 1: input	CDATA: input				
ADDR0/CLATCH	I ² C Address Bit 0: input	CLATCH: input				

BURST MODE WRITING AND READING

Burst mode addressing, where the subaddresses are automatically incremented at word boundaries, can be used for writing large amounts of data to contiguous registers. This increment happens automatically after a single-word write or read unless a stop condition is encountered (I²C) or CLATCH is brought high (SPI). A burst write starts like a single-word write, but following the first data-word, the data-word for the next immediate address can be written immediately without sending its two-byte address.

The registers in the ADAU1961 are one byte wide with the exception of the PLL control register, which is six bytes wide. The autoincrement feature knows the word length at each subaddress, so the subaddress does not need to be specified manually for each address in a burst write.

The subaddresses are autoincremented by 1 following each read or write of a data-word, regardless of whether there is a valid register word at that address. Address holes in the register map can be written to or read from without consequence. In the ADAU1961, these address holes exist at Address 0x4001, Address 0x4003 to Address 0x4007, Address 0x402E, and Address 0x4032 to Address 0x4035. A single-byte write to these registers is ignored by the ADAU1961, and a read returns a single byte 0x00.

I²C PORT

The ADAU1961 supports a 2-wire serial (I²C-compatible) microprocessor bus driving multiple peripherals. Two pins, serial data (SDA) and serial clock (SCL), carry information between the ADAU1961 and the system I²C master controller. In I²C mode, the ADAU1961 is always a slave on the bus, meaning that it cannot initiate a data transfer. Each slave device is recognized by a unique address. The address and R/W byte format is shown in Table 20. The address resides in the first seven bits of the I²C write. Bits[5:6] of the I²C address for the ADAU1961 are set by the levels on the ADDR1 and ADDR0 pins. The LSB of the address—the R/W bit—specifies either a read or write operation. Logic Level 1 corresponds to a read operation, and Logic Level 0 corresponds to a write operation.

Table 20. ADAU1961 I²C Address and Read/Write Byte Format

Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7
0	1	1	1	0	ADDR1	ADDR0	R/W

The SDA and SCL pins should each have a 2 k Ω pull-up resistor on the line connected to it. The voltage on these signal lines should not be higher than IOVDD (3.3 V).

Addressing

Initially, each device on the I²C bus is in an idle state and monitors the SDA and SCL lines for a start condition and the proper address. The I²C master initiates a data transfer by establishing a start condition, defined by a high-to-low transition on SDA while SCL remains high. This indicates that an address/ data stream follows. All devices on the bus respond to the start condition and shift the next eight bits (the 7-bit address plus the R/W bit) MSB first. The device that recognizes the transmitted address responds by pulling the data line low during the ninth clock pulse. This ninth bit is known as an acknowledge bit. All other devices withdraw from the bus at this point and return to the idle condition.

Data Sheet

ADAU1961

The R/W bit determines the direction of the data. A Logic 0 on the LSB of the first byte means that the master will write information to the peripheral, whereas a Logic 1 means that the master will read information from the peripheral after writing the subaddress and repeating the start address. A data transfer takes place until a stop condition is encountered. A stop condition occurs when SDA transitions from low to high while SCL is held high. Figure 48 shows the timing of an I²C write, and Figure 49 shows an I²C read.

Stop and start conditions can be detected at any stage during the data transfer. If these conditions are asserted out of sequence with normal read and write operations, the ADAU1961 immediately jumps to the idle condition. During a given SCL high period,

the user should only issue one start condition, one stop condition, or a single stop condition followed by a single start condition. If an invalid subaddress is issued by the user, the ADAU1961 does not issue an acknowledge and returns to the idle condition.

If the user exceeds the highest subaddress while in autoincrement mode, one of two actions is taken. In read mode, the ADAU1961 outputs the highest subaddress register contents until the master device issues a no acknowledge, indicating the end of a read. A no acknowledge condition is where the SDA line is not pulled low on the ninth clock pulse on SCL. If the highest subaddress location is reached while in write mode, the data for the invalid byte is not loaded into any subaddress register, a no acknowledge is issued by the ADAU1961, and the part returns to the idle condition.

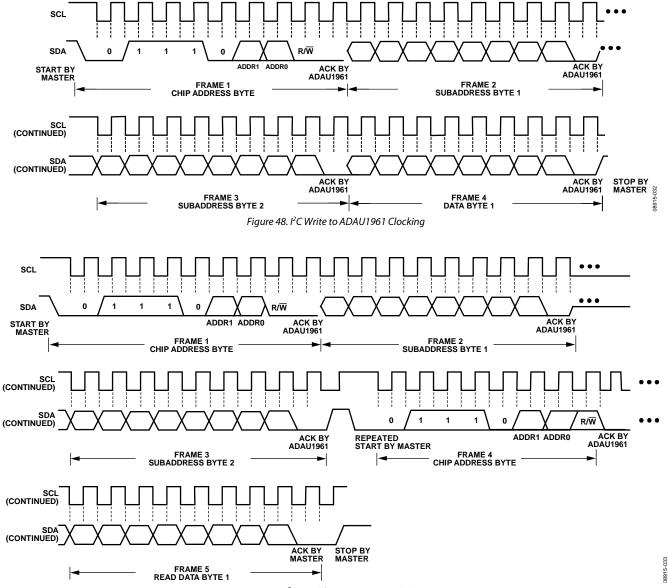


Figure 49. I²C Read from ADAU1961 Clocking

I²C Read and Write Operations

Figure 50 shows the format of a single-word write operation. Every ninth clock pulse, the ADAU1961 issues an acknowledge by pulling SDA low.

Figure 51 shows the format of a burst mode write sequence. This figure shows an example of a write to sequential single-byte registers. The ADAU1961 increments its subaddress register after every byte because the requested subaddress corresponds to a register or memory area with a 1-byte word length.

Figure 52 shows the format of a single-word read operation. Note that the first R/\overline{W} bit is 0, indicating a write operation. This is because the subaddress still needs to be written to set up the internal address. After the ADAU1961 acknowledges the receipt of the subaddress, the master must issue a repeated start command followed by the chip address byte with the R/\overline{W} bit set to 1 (read).

This causes the ADAU1961 SDA to reverse and begin driving data back to the master. The master then responds every ninth pulse with an acknowledge pulse to the ADAU1961.

Figure 53 shows the format of a burst mode read sequence. This figure shows an example of a read from sequential single-byte registers. The ADAU1961 increments its subaddress register after every byte because the requested subaddress corresponds to a register or memory area with a 1-byte word length. The ADAU1961 always decodes the subaddress and sets the auto-increment circuit so that the address increments after the appropriate number of bytes.

Figure 50 to Figure 53 use the following abbreviations: S =start bit

P = stop bit

AM = acknowledge by master

AS = acknowledge by slave

	S	Chip address, R/W = 0	AS	Subaddress high byte	AS	Subaddress low byte	AS	Data Byte 1	Р
--	---	--------------------------	----	----------------------	----	---------------------	----	-------------	---

Figure 50. Single-Word I²C Write Format

	baddress AS Subaddress of byte	AS Data AS Byte 1	Data AS Byte 2	Data AS Byte 3	Data AS Byte 4	Р
--	--------------------------------	----------------------	-------------------	-------------------	-------------------	---

Figure 51. Burst Mode I²C Write Format

S	Chip address, R/W = 0	AS	Subaddress high byte	AS	Subaddress low byte	AS	S	Chip address, R/W = 1	AS	Data Byte 1	Р
---	--------------------------	----	-------------------------	----	------------------------	----	---	--------------------------	----	----------------	---

Figure 52. Single-Word I²C Read Format

S	Ch <u>ip</u> address, R/W = 0	AS	Subaddress high byte	AS	Subaddress low byte	AS	S	Chip address, R/W = 1	AS	Data Byte 1	AM	Data Byte 2	AM		Ρ
---	----------------------------------	----	-------------------------	----	------------------------	----	---	--------------------------	----	----------------	----	----------------	----	--	---

Figure 53. Burst Mode I²C Read Format

SPI PORT

By default, the ADAU1961 is in I²C mode, but it can be put into SPI control mode by pulling CLATCH low three times. This is done by performing three dummy writes to the SPI port (the ADAU1961 does not acknowledge these three writes). Beginning with the fourth SPI write, data can be written to or read from the IC. The ADAU1961 can be taken out of SPI mode only by a full reset initiated by power-cycling the IC.

The SPI port uses a 4-wire interface, consisting of the CLATCH, CCLK, CDATA, and COUT signals, and it is always a slave port. The CLATCH signal should go low at the beginning of a transaction and high at the end of a transaction. The CCLK signal latches CDATA on a low-to-high transition. COUT data is shifted out of the ADAU1961 on the falling edge of CCLK and should be clocked into a receiving device, such as a microcontroller, on the CCLK rising edge. The CDATA signal carries the serial input data, and the COUT signal carries the serial output data. The COUT signal remains three-state until a read operation is requested. This allows other SPI-compatible peripherals to share the same readback line. All SPI transactions have the same basic format shown in Table 22. A timing diagram is shown in Figure 4. All data should be written MSB first.

Chip Address R/W

The LSB of the first byte of an SPI transaction is a R/\overline{W} bit. This bit determines whether the communication is a read (Logic Level 1) or a write (Logic Level 0). This format is shown in Table 21.

Table 21. ADAU1961 SPI Address and Read/Write By	4 . T
I ADIE 21. ADAU 1961 SPI Address and Read/ write BV	te Format
1 4010 210112 110 19 01 01 1114 41 000 4114 10044, ··· ==·· 2 }	

Bit 0	Bit 1	Bit 2	Bit 3	Bit 4	Bit 5	Bit 6	Bit 7
0	0	0	0	0	0	0	R/W

Subaddress

The 16-bit subaddress word is decoded into a location in one of the registers. This subaddress is the location of the appropriate register. The MSBs of the subaddress are zero-padded to bring the word to a full 2-byte length.

Data Bytes

The number of data bytes varies according to the register being accessed. During a burst mode write, an initial subaddress is written followed by a continuous sequence of data for consecutive register locations.

A sample timing diagram for a single-word SPI write operation to a register is shown in Figure 54. A sample timing diagram of a single-word SPI read operation is shown in Figure 55. The COUT pin goes from being three-state to being driven at the beginning of Byte 3. In this example, Byte 0 to Byte 2 contain the addresses and R/\overline{W} bit, and subsequent bytes carry the data.

Table 22. Generic Control Word Format

Byte 0	Byte 1	Byte 2	Byte 3	Byte 4 ¹
chip_adr[6:0], R/W	subaddr[15:8]	subaddr[7:0]	data	data

¹ Continues to end of data.

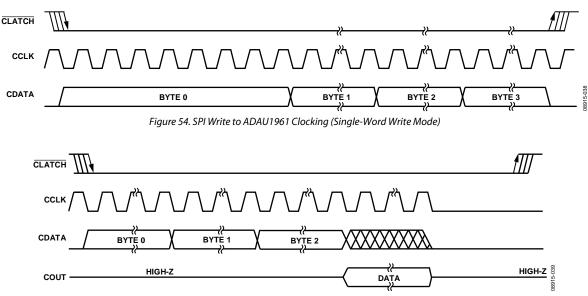


Figure 55. SPI Read from ADAU1961 Clocking (Single-Word Read Mode)

SERIAL DATA INPUT/OUTPUT PORTS

The flexible serial data input and output ports of the ADAU1961 can be set to accept or transmit data in 2-channel format or in a 4-channel TDM stream to interface to external ADCs or DACs. Data is processed in twos complement, MSB first format. The left channel data field always precedes the right channel data field in 2-channel streams. In TDM mode, Slot 0 and Slot 1 are in the first half of the audio frame, and Slot 2 and Slot 3 are in the second half of the frame. The serial modes and the position of the data in the frame are set in Register R15 to Register R18 (serial port and converter control registers, Address 0x4015 to Address 0x4018).

If the PLL of the ADAU1961 is not used, the serial data clocks must be synchronous with the ADAU1961 master clock input. The LRCLK and BCLK pins are used to clock both the serial input and output ports. The ADAU1961 can be set as the master or the slave in a system. Because there is only one set of serial data clocks, the input and output ports must always be both master or both slave.

Register R15 and Register R16 (serial port control registers, Address 0x4015 and Address 0x4016) allow control of clock polarity and data input modes. The valid data formats are I²S, left-justified, right-justified (24-/20-/18-/16-bit), and TDM. In all modes except for the right-justified modes, the serial port inputs an arbitrary number of bits up to a limit of 24. Extra bits do not cause an error, but they are truncated internally. The serial port can operate with an arbitrary number of BCLK transitions in each LRCLK frame. The LRCLK in TDM mode can be input to the ADAU1961 either as a 50% duty cycle clock or as a bit-wide pulse.

When the LRCLK is set as a pulse, a 47 pF capacitor should be connected between the LRCLK pin and ground (see Figure 56). This capacitor is necessary in both master and slave modes to properly align the LRCLK signal to the serial data stream.

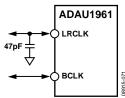


Figure 56. LRCLK Capacitor Alignment, TDM Pulse Mode

In TDM mode, the ADAU1961 can be a master for $f_{\rm S}$ up to 48 kHz. Table 23 lists the modes in which the serial output port can function.

fs	2-Channel Modes (I ² S, Left- Justified, Right-Justified)	4-Channel TDM
48 kHz	Master and slave	Master and slave
96 kHz	Master and slave	Slave

Table 24 describes the proper configurations for standard audio data formats.

Table 24. Data Format Configurations

Format	LRCLK Polarity (LRPOL)	LRCLK Mode (LRMOD)	BCLK Polarity (BPOL)	BCLK Cycles/Audio Frame (BPF[2:0])	Data Delay from LRCLK Edge (LRDEL[1:0])
I ² S (see Figure 57)	Frame begins on falling edge	50% duty cycle	Data changes on falling edge	32 to 64	Delayed from LRCLK edge by 1 BCLK
Left-Justified (see Figure 58)	Frame begins on rising edge	50% duty cycle	Data changes on falling edge	32 to 64	Aligned with LRCLK edge
Right-Justified (see Figure 59)	Frame begins on rising edge	50% duty cycle	Data changes on falling edge	32 to 64	Delayed from LRCLK edge by 8 or 16 BCLKs
TDM with Clock (see Figure 60)	Frame begins on falling edge	50% duty cycle	Data changes on falling edge	64 to 128	Delayed from start of word clock by 1 BCLK
TDM with Pulse (see Figure 61)	Frame begins on rising edge	Pulse	Data changes on falling edge	64 to 128	Delayed from start of word clock by 1 BCLK

ADAU1961 Data Sheet LEFT CHANNEL LRCLK RIGHT CHANNEL BCLK SDATA MSB MSB LSB LSB 1/f_S Figure 57. I²S Mode—16 Bits to 24 Bits per Channel RIGHT CHANNEL LRCLK BCLK LEFT CHANNEL MSB LSB MSB LSB SDATA 1/f_s Figure 58. Left-Justified Mode—16 Bits to 24 Bits per Channel RIGHT CHANNEL LRCLK BCLK LEFT CHANNEL SDATA MSB LSB MSB LSE 89 15 -1/f_S Figure 59. Right-Justified Mode—16 Bits to 24 Bits per Channel LRCLK 128 BCLKs ហហ ஹ BCLK 32 BCLKs SDATA SLOT 0 SLOT 3 SLOT 1 SLOT 2 LRCLK BCLK 08915-043 MSB MSB – 1 MSB – 2 SDATA Figure 60. TDM 4 Mode LRCLK BCLK ł ł 11 1i MSB TDM СН SDATA SLOT 0 SLOT 1 SLOT 2 SLOT 3 8915-044 32 BCLKs -



APPLICATIONS INFORMATION POWER SUPPLY BYPASS CAPACITORS

Each analog and digital power supply pin should be bypassed to its nearest appropriate ground pin with a single 100 nF capacitor. The connections to each side of the capacitor should be as short as possible, and the trace should stay on a single layer with no vias. For maximum effectiveness, locate the capacitor equidistant from the power and ground pins or, when equidistant placement is not possible, slightly closer to the power pin. Thermal connections to the ground planes should be made on the far side of the capacitor.

Each supply signal on the board should also be by passed with a single bulk capacitor (10 μF to 47 $\mu F).$

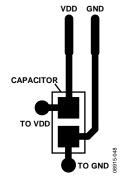


Figure 62. Recommended Power Supply Bypass Capacitor Layout

GSM NOISE FILTER

In mobile phone applications, excessive 217 Hz GSM noise on the analog supply pins can degrade the audio quality. To avoid this problem, it is recommended that an L-C filter be used in series with the bypass capacitors for the AVDD pins. This filter should consist of a 1.2 nH inductor and a 9.1 pF capacitor in series between AVDD and ground, as shown in Figure 63.

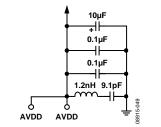


Figure 63. GSM Filter on the Analog Supply Pins

GROUNDING

A single ground plane should be used in the application layout. Components in an analog signal path should be placed away from digital signals.

EXPOSED PAD PCB DESIGN

The ADAU1961 has an exposed pad on the underside of the LFCSP. This pad is used to couple the package to the PCB for heat dissipation when using the outputs to drive earpiece or headphone loads. When designing a board for the ADAU1961, special consideration should be given to the following:

- A copper layer equal in size to the exposed pad should be on all layers of the board, from top to bottom, and should connect somewhere to a dedicated copper board layer (see Figure 64).
- Vias should be placed to connect all layers of copper, allowing for efficient heat and energy conductivity. For an example, see Figure 65, which has nine vias arranged in a 3 inch × 3 inch grid in the pad area.

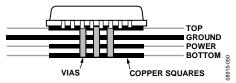
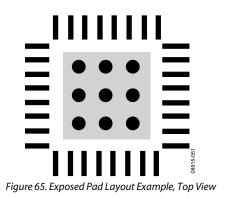


Figure 64. Exposed Pad Layout Example, Side View



CONTROL REGISTERS

Table 25. Register Map

Reg	Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default
R0	0x4000	Clock control		Rese	erved		CLKSRC	INFR	EQ[1:0]	COREN	00000000
R1	0x4002	PLL control				M[[*]	15:8]				00000000
						M[7:0]				11111101
						N [1	5:8]				00000000
						N[7:0]				00001100
			Reserved			[3:0]		X	[1:0]	Туре	00010000
						erved			Lock	PLLEN	00000000
R2	0x4008	Dig mic/jack detect	JDDE	3[1:0]	JDFU	NC[1:0]		Reserved		JDPOL	00000000
R3	0x4009	Reserved				Rese	erved				00000000
R4	0x400A	Rec Mixer Left 0	Reserved		LINPG[2:0]			LINNG[2:0]		MX1EN	00000000
R5	0x400B	Rec Mixer Left 1		Reserved			DST[1:0]		MX1AUXG[2:	_	00000000
R6	0x400C	Rec Mixer Right 0	Reserved		RINPG[2:0]			RINNG[2:0]		MX2EN	00000000
R7	0x400D	Rec Mixer Right 1		Reserved			DST[1:0]		MX2AUXG[2:	-	00000000
R8	0x400E	Left diff input vol				OL[5:0]			LDMUTE	LDEN	00000000
R9	0x400F	Right diff input vol				OL[5:0]			RDMUTE	RDEN	00000000
R10	0x4010	Record mic bias			erved		MPERF	MBI	Reserved	MBIEN	00000000
R11	0x4011	ALC 0	PGASL			ALCMAX[2:0]		ALCSEL[2:0]		00000000
R12	0x4012	ALC 1			DLD[3:0]				ARG[3:0]		00000000
R13	0x4013	ALC 2			CK[3:0]	1			EC[3:0]		00000000
R14	0x4014	ALC 3	NGTY		NGEN			NGTHR[4:0]			00000000
R15	0x4015	Serial Port 0	DITHEN	Reserved	LRMOD	BPOL	LRPOL	-	PF[1:0]	MS	00000000
R16	0x4016	Serial Port 1		BPF[2:0]		ADTDM	DATDM	MSBP		EL[1:0]	00000000
R17	0x4017	Converter 0	Reserved	DAPA	JR[1:0]	DAOSR	ADOSR		CONVSR[2:0]	00000000
R18	0x4018	Converter 1		T		erved			ADPA	\IR[1:0]	00000000
R19	0x4019	ADC control	Reserved	ADCPOL	HPF	DMPOL	DMSW	INSEL	ADCE	EN[1:0]	00010000
R20	0x401A	Left digital vol					OL[7:0]				00000000
R21	0x401B	Right digital vol		1	1	RADV	OL[7:0]			1	00000000
R22	0x401C	Play Mixer Left 0	Reserved	MX3RM	MX3LM		MX3AU			MX3EN	00000000
R23	0x401D	Play Mixer Left 1		1	52[3:0]	1			G1[3:0]	1	00000000
R24	0x401E	Play Mixer Right 0	Reserved	MX4RM	MX4LM		MX4AU			MX4EN	00000000
R25	0x401F	Play Mixer Right 1			52[3:0]			1	G1[3:0]		0000000
R26	0x4020	Play L/R mixer left		Reserved			64[1:0]		G3[1:0]	MX5EN	00000000
R27	0x4021	Play L/R mixer right		Reserved		MX60	64[1:0]		G3[1:0]	MX6EN	0000000
R28	0x4022	Play L/R mixer mono			Reserved			MX	7[1:0]	MX7EN	0000000
R29	0x4023	Play HP left vol				OL[5:0]			LHPM	HPEN	00000010
R30	0x4024	Play HP right vol				OL[5:0]			RHPM	HPMODE	00000010
R31	0x4025	Line output left vol				VOL[5:0]			LOUTM	LOMODE	00000010
R32	0x4026	Line output right vol				VOL[5:0]			ROUTM	ROMODE	00000010
R33	0x4027	Play mono output			MONO	VOL[5:0]			MONOM	MOMODE	00000010
R34	0x4028	Pop/click suppress		Reserved		POPMODE	POPLESS	ASLE	W[1:0]	Reserved	00000000
R35	0x4029	Play power mgmt				erved			PREN	PLEN	00000000
R36	0x402A	DAC Control 0	DACMC	NO[1:0]	DACPOL		erved	DEMPH	DACE	EN[1:0]	00000000
R37	0x402B	DAC Control 1					OL[7:0]				00000000
R38	0x402C	DAC Control 2			- •		OL[7:0]	(014.01		(014.01	00000000
R39	0x402D	Serial port pad		DP[1:0]		DP[1:0]	LRCLK			(P[1:0]	10101010
R40	0x402F	Control Port Pad 0	CDAT	P[1:0]	CLCF	HP[1:0]	SCLP	'[1:0]	SDA	P[1:0]	10101010
R41	0x4030	Control Port Pad 1			10.075	Reserved			-	SDASTR	00000000
R42	0x4031	Jack detect pin	Rese	rved	JDSTR	Reserved	JDP	[1:0]	Rese	erved	00001000
R67	0x4036	Dejitter control				DEJI	T[7:0]				00000011

CONTROL REGISTER DETAILS

All registers except for the PLL control register are 1-byte write and read registers.

R0: Clock Control, 16,384 (0x4000)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved				CLKSRC	INFRE	O[1:0]	COREN

Table 2	6. Clock Control	Register		
Bits	Bit Name	Description		
3	CLKSRC	Clock source select. 0 = direct from MCLK pin (default). 1 = PLL clock.		
[2:1]	INFREQ[1:0]	Input clock frequency. Sets th automatically set to 1024 × fs	ne core clock rate that generates the core clock. If the PLL is used, this value is	
		Setting	Input Clock Frequency	
		00	$256 \times f_s$ (default)	
		01	512 × fs	
		10	$768 \times f_s$	
		11	1024 × fs	
0	COREN		Core clock enable. Only the R0 and R1 registers can be accessed when this bit is set to 0 (core clock disabled). 0 = core clock disabled (default).	

R1: PLL Control, 16,386 (0x4002)

Byte	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0		M[15:8]							
1		M[7:0]							
2	N[15:8]								
3					N[7:0]				
4	Reserved	ed R[3:0] X[X[1:0]	Туре	
5	Reserved Lock PLLEN								

Table 27. PLL Control Register

Byte	Bits	Bit Name	Description					
0	[7:0]	M[15:8]	PLL denominator MSB. Thi	PLL denominator MSB. This value is concatenated with M[7:0] to make up a 16-bit number.				
1	[7:0]	M[7:0]	PLL denominator LSB. This	value is concatenated with	M[15:8] to make up a 16-bit number.			
			M[15:8] (MSB)	M[7:0] (LSB)	Value of M			
			0000000	0000000	0			
			0000000	11111101	253 (default)			
			11111111	11111111	65,535			
2	[7:0]	N[15:8]	PLL numerator MSB. This v	alue is concatenated with N	ا[7:0] to make up a 16-bit number.			
3	[7:0]	N[7:0]	PLL numerator LSB. This va	alue is concatenated with N	[15:8] to make up a 16-bit number.			
			N[15:8] (MSB)	N[7:0] (LSB)	Value of N			
			0000000	0000000	0			
			0000000	00001100	12 (default)			
			11111111	11111111	65,535			

Data Sheet

Byte	Bits	Bit Name	Description			
4	[6:3]	R[3:0]	PLL integer setting.			
			Setting	Value of R		
			0010	2 (default)		
			0011	3		
			0100	4		
			0101	5		
			0110	6		
			0111	7		
			1000	8		
4	[2:1]	X[1:0]	PLL input clock divider.			
			Setting	Value of X		
			00	1 (default)		
			01	2		
			10	3		
			11	4		
4	0	Туре		iteger mode, the values of M and N are ignored.		
			0 = integer (default).			
	_		1 = fractional.			
5	1	Lock	PLL lock. This read-only bit is flagged when the PLL has finished locking.			
			0 = PLL unlocked (default). 1 = PLL locked.			
5	0	PLLEN	PLL enable.			
J	U	FLLEIN	0 = PLL disabled (default).			
			1 = PLL enabled.			

R2: Digital Microphone/Jack Detection Control, 16,392 (0x4008)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
JDDE	3[1:0]	JDFUN	NC[1:0]		Reserved		JDPOL

Table 28. Digital Microphone/Jack Detection Control Register

Bits	Bit Name	Description				
[7:6]	JDDB[1:0]	Jack detect debounce time.				
		Setting	Debounce Time			
		00	5 ms (default)			
		01	10 ms			
		10	20 ms			
		11	40 ms			
[5:4]	JDFUNC[1:0]	JACKDET/MICIN pin microphone input.	function. Enables or disables the jack detect function or configures the pin for a digital			
		Setting	Pin Function			
		00	Jack detect off (default)			
		01	Jack detect on			
		10	Digital microphone input			
		11	Reserved			
0	JDPOL	Jack detect polarity.	. Detects high or low signal.			
		0 = detect high sign				
		1 = detect low signa	al.			

R4: Record Mixer Left (Mixer 1) Control 0, 16,394 (0x400A)

This register controls the gain of single-ended inputs for the left channel record path. The left channel record mixer is referred to as Mixer 1.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved		LINPG[2:0]			LINNG[2:0]		MX1EN

Table 29. Record Mixer Left (Mixer 1) Control 0 Register

Bits	Bit Name	Description			
[6:4]	LINPG[2:0]	Gain for a left channel sing	gle-ended input from the LINP pin, input to Mixer 1.		
		Setting	Gain		
		000	Mute (default)		
		001	-12 dB		
		010	-9 dB		
		011	-6 dB		
		100	-3 dB		
		101	0 dB		
		110	3 dB		
		111	6 dB		
[3:1]	LINNG[2:0]	Gain for a left channel single-ended input from the LINN pin, input to Mixer 1.			
		Setting	Gain		
		000	Mute (default)		
		001	-12 dB		
		010	-9 dB		
		011	-6 dB		
		100	-3 dB		
		101	0 dB		
		110	3 dB		
		111	6 dB		
0	MX1EN	Left channel mixer enable	in the record path. Referred to as Mixer 1.		
		0 = mixer disabled (defaul	t).		
		1 = mixer enabled.			

R5: Record Mixer Left (Mixer 1) Control 1, 16,395 (0x400B)

This register controls the gain boost of the left channel differential PGA input and the gain for the left channel auxiliary input in the record path. The left channel record mixer is referred to as Mixer 1.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved		LDBOC	DST[1:0]		MX1AUXG[2:0]	

Table 30. Record Mixer Left (Mixer 1) Control 1 Register

Bits	Bit Name	Description				
[4:3]	LDBOOST[1:0]	Left channel differential PGA input gain boost, input to Mixer 1. The left differential input uses the LINP (positive signal) and LINN (negative signal) pins.				
		Setting	Gain Boost			
		00	Mute (default)			
		01	0 dB			
		10	20 dB			
		11	Reserved			
[2:0]	MX1AUXG[2:0]	Left single-ended auxiliary input gain from the LAUX pin in the record path, input to Mixer 1.				
		Setting	Auxiliary Input Gain			
		000	Mute (default)			
		001	-12 dB			
		010	-9 dB			
		011	-6 dB			
		100	-3 dB			
		101	0 dB			
		110	3 dB			
		111	6 dB			

R6: Record Mixer Right (Mixer 2) Control 0, 16,396 (0x400C)

This register controls the gain of single-ended inputs for the right channel record path. The right channel record mixer is referred to as Mixer 2.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved		RINPG[2:0]			RINNG[2:0]		MX2EN

Table 31. Record Mixer Right (Mixer 2) Control 0 Register

Bits	Bit Name	Description			
[6:4]	RINPG[2:0]	Gain for a right channe	el single-ended input from the RINP pin, input to Mixer 2.		
		Setting	Gain		
		000	Mute (default)		
		001	-12 dB		
		010	-9 dB		
		011	-6 dB		
		100	-3 dB		
		101	0 dB		
		110	3 dB		
		111	6 dB		
[3:1]	RINNG[2:0]	Gain for a right channel single-ended input from the RINN pin, input to Mixer 2.			
		Setting	Gain		
		000	Mute (default)		
		001	-12 dB		
		010	-9 dB		
		011	-6 dB		
		100	-3 dB		
		101	0 dB		
		110	3 dB		
		111	6 dB		
0	MX2EN	Right channel mixer e	nable in the record path. Referred to as Mixer 2.		
		0 = mixer disabled (de	fault).		
		1 = mixer enabled.			

R7: Record Mixer Right (Mixer 2) Control 1, 16,397 (0x400D)

This register controls the gain boost of the right channel differential PGA input and the gain for the right channel auxiliary input in the record path. The right channel record mixer is referred to as Mixer 2.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved		RDBOC	DST[1:0]		MX2AUXG[2:0]	

Bits	Bit Name	Description			
[4:3]	RDBOOST[1:0]	Right channel differential PG (positive signal) and RINN (ne	A input gain boost, input to Mixer 2. The right differential input uses the RINP egative signal) pins.		
		Setting	Gain Boost		
		00	Mute (default)		
		01	0 dB		
		10	20 dB		
		11	Reserved		
[2:0]	MX2AUXG[2:0]	Right single-ended auxiliary i	input gain from the RAUX pin in the record path, input to Mixer 2.		
		Setting	Auxiliary Input Gain		
		000	Mute (default)		
		001	-12 dB		
		010	-9 dB		
		011	-6 dB		
		100	-3 dB		
		101	0 dB		
		110	3 dB		
		111	6 dB		

Table 32. Record Mixer Right (Mixer 2) Control 1 Register

R8: Left Differential Input Volume Control, 16,398 (0x400E)

This register enables the differential path and sets the volume control for the left differential PGA input.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		LDVO				LDMUTE	LDEN

Table 33. Left Differential Input Volume Control Register

Bits	Bit Name	Description				
[7:2]	LDVOL[5:0]	Left channel differential PGA input volume control. The left differential input uses the LINP (positive signal) and LINN (negative signal) pins. Each step corresponds to a 0.75 dB increase in gain. See Table 69 for a complete list of the volume settings.				
		Setting	Volume			
		000000	–12 dB (default)			
		000001	-11.25 dB			
		010000	0 dB			
		111110	34.5 dB			
		111111	35.25 dB			
1	LDMUTE	Left differential input mute co	ontrol.			
		0 = mute (default).				
		1 = unmute.				
0	LDEN	Left differential PGA enable. When enabled, the LINP and LINN pins are used as a full differential pair. When				
			onfigured as two single-ended inputs with the signals routed around the PGA.			
		0 = disabled (default).				
		1 = enabled.				

R9: Right Differential Input Volume Control, 16,399 (0x400F)

This register enables the differential path and sets the volume control for the right differential PGA input.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	RDVOL[5:0]						RDEN

Table 34. Right Differential Input Volume Control Register

Bits	Bit Name	Description				
[7:2]	7:2] RDVOL[5:0]	Right channel differential PGA input volume control. The right differential input uses the RINP (positive signal) and RINN (negative signal) pins. Each step corresponds to a 0.75 dB increase in gain. See Table 69 for a complete list of the volume settings.				
		Setting	Volume			
		000000	-12 dB (default)			
		000001	-11.25 dB			
		010000	0 dB			
		111110	34.5 dB			
		111111	35.25 dB			
1	RDMUTE	Right differential input mute	control.			
		0 = mute (default).				
		1 = unmute.				
0	RDEN	5	When enabled, the RINP and RINN pins are used as a full differential pair. When			
		<i>i</i>	configured as two single-ended inputs with the signals routed around the PGA.			
		0 = disabled (default).				
		1 = enabled.				

R10: Record Microphone Bias Control, 16,400 (0x4010)

This register controls the MICBIAS pin settings for biasing electret type analog microphones.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved				MBI	Reserved	MBIEN

Table 35. Record Microphone Bias Control Register

Bits	Bit Name	Description
3	MPERF	 Microphone bias is enabled for high performance or normal operation. High performance operation sources more current to the microphone. 0 = normal operation (default). 1 = high performance.
2	MBI	Microphone voltage bias as a fraction of AVDD. $0 = 0.90 \times \text{AVDD}$ (default). $1 = 0.65 \times \text{AVDD}$.
0	MBIEN	Enables the MICBIAS output. 0 = disabled (default). 1 = enabled.

R11: ALC Control 0, 16,401 (0x4011)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PGASLEW[1:0] ALCM		ALCMAX[2:0]			ALCSEL[2:0]		

Table 36. ALC Control 0 Register

Bits	Bit Name	Description	
[7:6]	PGASLEW[1:0]		the ALC is off. The slew time is the period of time that a volume increase or decrease wn to the target volume set in Register R8 (left differential input volume control) ntial input volume control).
		Setting	Slew Time
		00	24 ms (default)
		01	48 ms
		10	96 ms
		11	Off
[5:3]	ALCMAX[2:0]	The maximum ALC gain sets protects small signals from ex	a limit to the amount of gain that the ALC can provide to the input signal. This xcessive amplification.
		Setting	Maximum ALC Gain
		000	-12 dB (default)
		001	-6 dB
		010	0 dB
		011	6 dB
		100	12 dB
		101	18 dB
		110	24 dB
		111	30 dB
[2:0]	ALCSEL[2:0]	only to the right channel input ALC responds only to the left stereo, the ALC responds to t	channels that are controlled by the ALC. When set to right only, the ALC responds ut and controls the gain of the right PGA amplifier only. When set to left only, the channel input and controls the gain of the left PGA amplifier only. When set to he greater of the left or right channel and controls the gain of both the left and its must be off if manual control of the volume is desired.
		Setting	Channels
		000	Off (default)
		001	Right only
		010	Left only
		011	Stereo
		100	Reserved
		101	Reserved
		110	Reserved
		111	Reserved

R12: ALC Control 1, 16,402 (0x4012)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	ALCHO	LD[3:0]			ALCTA	RG[3:0]	

Table 37. ALC Control 1 Register

Bits	Bit Name	Description	
[7:4]	ALCHOLD[3:0]	increasing the gain to achiev	time is the amount of time that the ALC waits after a decrease in input level before e the target level. The recommended minimum setting is 21 ms (0011) to prevent ignals. The hold time doubles with every 1-bit increase.
		Setting	Hold Time
		0000	2.67 ms (default)
		0001	5.34 ms
		0010	10.68 ms
		0011	21.36 ms
		0100	42.72 ms
		0101	85.44 ms
		0110	170.88 ms
		0111	341.76 ms
		1000	683.52 ms
		1001	1.367 sec
		1010	2.7341 sec
		1011	5.4682 sec
		1100	10.936 sec
		1101	21.873 sec
		1110	43.745 sec
		1111	87.491 sec
[3:0]	ALCTARG[3:0]		ts the desired ADC input level. The PGA gain is adjusted by the ALC to reach this ed target level is between -16 dB and -10 dB to accommodate transients without
		Setting	ALC Target
		0000	–28.5 dB (default)
		0001	–27 dB
		0010	–25.5 dB
		0011	-24 dB
		0100	–22.5 dB
		0101	-21 dB
		0110	–19.5 dB
		0111	-18 dB
		1000	–16.5 dB
		1001	–15 dB
		1010	–13.5 dB
		1011	-12 dB
		1100	–10.5 dB
		1101	-9 dB
		1110	–7.5 dB
		1111	-6 dB

R13: ALC Control 2, 16,403 (0x4013)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	ALCATCK[3:0]				ALCDE		

Table 38. ALC Control 2 Register

Bits	Bit Name	Description	
[7:4]	ALCATCK[3:0]		attack time sets how fast the ALC starts attenuating after an increase in input level above setting for music recording is 384 ms, and a typical setting for voice recording is 24 ms.
		Setting	Attack Time
		0000	6 ms (default)
		0001	12 ms
		0010	24 ms
		0011	48 ms
		0100	96 ms
		0101	192 ms
		0110	384 ms
		0111	768 ms
		1000	1.54 sec
		1001	3.07 sec
		1010	6.14 sec
		1011	12.29 sec
		1100	24.58 sec
		1101	49.15 sec
		1110	98.30 sec
		1111	196.61 sec
[3:0]	ALCDEC[3:0]		decay time sets how fast the ALC increases the PGA gain after a decrease in input level ypical setting for music recording is 24.58 seconds, and a typical setting for voice recording
		Setting	Decay Time
		0000	24 ms
		0001	48 ms
		0010	96 ms
		0011	192 ms
		0100	384 ms
		0101	768 ms
		0110	1.54 sec
		0111	3.07 sec
		1000	6.14 sec
		1001	12.29 sec
		1010	24.58 sec
		1011	49.15 sec
		1100	98.30 sec
		1101	196.61 sec
		1110	393.22 sec
		1111	786.43 sec

R14: ALC Control 3, 16,404 (0x4014)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
NGTY	NGTYP[1:0]				NGTHR[4:0]		

Table 39. ALC Control 3 Register

Bits	Bit Name	Description				
[7:6]	NGTYP[1:0]	Noise gate type. When the input signal falls below the threshold for 250 ms, the noise gate can hold a c PGA gain, mute the ADC output, fade the PGA gain to the minimum gain value, or fade then mute.				
		Setting	Noise Gate			
		00	Hold PGA constant (default)			
		01	Mute ADC output (digital mute)			
		10	Fade to PGA minimum value (analog fade)			
		11	Fade then mute (analog fade/digital mute)			
5	NGEN	Noise gate enable. 0 = disabled (default). 1 = enabled.	0 = disabled (default).			
[4:0]	NGTHR[4:0]	5	the input signal falls below the threshold for 250 ms, the noise gate is activated. Is to a -1.5 dB change. See Table 70 for a complete list of the threshold settings.			
		Setting	Threshold			
		00000	–76.5 dB (default)			
		00001	–75 dB			
		11110	–31.5 dB			
		11111	-30 dB			

R15: Serial Port Control 0, 16,405 (0x4015)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
DITHEN	Reserved	LRMOD	BPOL	LRPOL	CHPI	[1:0]	MS

Table 40. Serial Port Control 0 Register

Bits	Bit Name	Description					
7	DITHEN	Dither enable is applicable only for 16-bit data width modes. 0 = disabled (default). 1 = enabled.					
5	LRMOD	LRCLK mode sets the LRCLK for either a 50% duty cycle or a pulse. The pulse mode should be at least 1 BCLK wide. 0 = 50% duty cycle (default). 1 = pulse mode.					
4	BPOL	BCLK polarity sets the BCLK edge that triggers a change in audio data. This can be set for the falling or rising edge of the BCLK. 0 = falling edge (default). 1 = rising edge.					
3	LRPOL	 LRCLK polarity sets the LRCLK edge that triggers the beginning of the left channel audio frame. This can be set for the falling or rising edge of the LRCLK. 0 = falling edge (default). 1 = rising edge. 					
[2:1]	CHPF[1:0]	Channels per frame sets the number of channels per LRCLK frame.					
		Setting	Channels per LRCLK Frame				
		00	Stereo (default)				
		01	TDM 4				
		10	Reserved				
		Reserved					
0	MS	Serial data port bus mode. Both LRCLK and BCLK are master of the serial port when set in master mode a serial port slave in slave mode. 0 = slave mode (default).					
		1 = master mode.					

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
BPF[2:0]		ADTDM	DATDM	MSBP	LRDE	L[1:0]	

Table 41. Serial Port Control 1 Register

Bits	Bit Name	Description				
[7:5]	BPF[2:0]	Number of bit clock cycles pe	er LRCLK audio frame.			
		Setting	Bit Clock Cycles			
		000	64 (default)			
		001	32			
		010	48			
		011	128			
		100	256			
		101	Reserved			
		110	Reserved			
		111	Reserved			
4	ADTDM	ADC serial audio data channel position in TDM mode.				
		0 = left first (default).				
		1 = right first.				
3	DATDM	DAC serial audio data channe	el position in TDM mode.			
		0 = left first (default).				
		1 = right first.				
2	MSBP	MSB position in the LRCLK fra	ame.			
		0 = MSB first (default). 1 = LSB first.				
[1:0]	LRDEL[1:0]	Data delay from LRCLK edge	(in BCLK units)			
[1.0]		Setting	Delay (Bit Clock Cycles)			
		00	1 (default)			
		01	0			
		10	8			
		11	16			

R17: Converter Control 0, 16,407 (0x4017)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	DAPA	R[1:0]	DAOSR	ADOSR		CONVSR[2:0]	

Table 42. Converter Control 0 Register

Bits	Bit Name	Description						
[6:5]	DAPAIR[1:0]	On-chip DAC serial data sele	ction in TDM mode.					
		Setting	ting Pair					
		00	First pair (default)					
		01	Second pair					
		10	Third pair					
		11	Fourth pair					
4	DAOSR	DAC oversampling ratio. This bit cannot be set for $64 \times$ when CONVSR[2:0] is set to 96 kHz. 0 = $128 \times$ (default). 1 = $64 \times$.						
3	ADOSR	ADC oversampling ratio. This bit cannot be set for $64 \times$ when CONVSR[2:0] is set to 96 kHz. 0 = $128 \times$ (default). 1 = $64 \times$.						
[2:0]	CONVSR[2:0]	Converter sampling rate. The ADCs and DACs operate at the sampling rate set in this register. The converter rate selected is a ratio of the base sampling rate, f _s . The base sampling rate is determined by the operating frequency of the core clock. The serial port mirrors the converter sampling rates set in this register.						
		Setting	Sampling Rate	Base Sampling Rate (fs = 48 kHz)				
		000	fs	48 kHz, base (default)				
		001	fs/6	8 kHz				
		010	f _s /4	12 kHz				
		011	fs/3	16 kHz				
		100	f _s /2	24 kHz				
		101	fs/1.5	32 kHz				
		110	f _s /0.5	96 kHz				
		111	Reserved					

R18: Converter Control 1, 16,408 (0x4018)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		Rese	rved			ADPA	

Table 43. Converter Control 1 Register

Bits	Bit Name	Description			
[1:0]	ADPAIR[1:0]	On-chip ADC serial data selection in TDM mode.			
	Setting Pair		Pair		
		00	First pair (default)		
		01	Second pair		
		10	Third pair		
		11	Fourth pair		

R19: ADC Control, 16,409 (0x4019)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	ADCPOL	HPF	DMPOL	DMSW	INSEL	ADCE	N[1:0]

Table 44. ADC Control Register

Bits	Bit Name	Description					
6	ADCPOL	Invert input polarity.					
		0 = normal (default).					
		1 = inverted.					
5	HPF	ADC high-pass filter select. At 4	$18 \text{ kHz}, f_{3dB} = 2 \text{ Hz}.$				
		0 = off (default).					
		1 = on.					
4	DMPOL	Digital microphone data polarit	ty swap.				
		0 = invert polarity.					
2	DMGM	1 = normal (default).					
3	DMSW	Digital microphone channel swap. Normal operation sends the left channel on the rising edge of the clock and the right channel on the falling edge of the clock.					
		0 = normal (default).					
		1 = swap left and right channel	S.				
2	INSEL		t. When asserted, the on-chip ADCs are off, BCLK is master at $128 \times f_s$, and				
	-	ADC_SDATA is expected to have left and right channels interleaved.					
		0 = digital microphone inputs off, ADCs enabled (default).					
		1 = digital microphone inputs e	enabled, ADCs off.				
[1:0]	ADCEN[1:0]	ADC enable.					
		Setting	ADCs Enabled				
		00	Both off (default)				
	01 Left on		Left on				
		10	Right on				
		11 Both on					

R20: Left Input Digital Volume, 16,410 (0x401A)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
			LADVO				

Table 45. Left Input Digital Volume Register

Bits	Bit Name	Description	
[7:0]	LADVOL[7:0]	3	enuation for left channel inputs from either the left ADC or the left digital micro- nds to a 0.375 dB step with slewing between settings. See Table 71 for a complete
		Setting	Volume Attenuation
		0000000	0 dB (default)
		0000001	–0.375 dB
		0000010	–0.75 dB
		1111110	–95.25 dB
		11111111	–95.625 dB

R21: Right Input Digital Volume, 16,411 (0x401B)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
			RADVO	DL[7:0]			

Table 46. Right Input Digital Volume Register

Bits	Bit Name	Description			
[7:0]	RADVOL[7:0]		volume attenuation for right channel inputs from either the right ADC or the right digital Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 71 for a volume settings.		
		Setting	Volume Attenuation		
	0000000	0 dB (default)			
		0000001	–0.375 dB		
		0000010	–0.75 dB		
		11111110	-95.25 dB		
		11111111	–95.625 dB		

R22: Playback Mixer Left (Mixer 3) Control 0, 16,412 (0x401C)

-							
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	MX3RM	MX3LM		MX3AU	IXG[3:0]		MX3EN

Table 47. Playback Mixer Left (Mixer 3) Control 0 Register

Bits	Bit Name	Description			
6	MX3RM	Mixer input mute. Mutes the r 0 = muted (default). 1 = unmuted.	ight DAC input to the left channel playback mixer (Mixer 3).		
5	MX3LM	Mixer input mute. Mutes the left DAC input to the left channel playback mixer (Mixer 3). 0 = muted (default). 1 = unmuted.			
[4:1]	MX3AUXG[3:0]	Mixer input gain. Controls the	left channel auxiliary input gain to the left channel playback mixer (Mixer 3).		
		Setting	Gain		
		0000	Mute (default)		
		0001	–15 dB		
		0010	-12 dB		
		0011	-9 dB		
		0100	-6 dB		
		0101	-3 dB		
		0110	0 dB		
		0111	3 dB		
		1000	6 dB		
0	MX3EN	Mixer 3 enable. 0 = disabled (default). 1 = enabled.			

R23: Playback Mixer Left (Mixer 3) Control 1, 16,413 (0x401D)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	MX3G	i2[3:0]			MX3G	i1[3:0]	

Table 48. Playback Mixer Left (Mixer 3) Control 1 Register

Bits	Bit Name	Description			
[7:4]	MX3G2[3:0]	Bypass gain control. The sign can be applied before the le	nal from the right channel record mixer (Mixer 2) bypasses the converters and gain ft playback mixer (Mixer 3).		
		Setting	Gain		
		0000	Mute (default)		
		0001	–15 dB		
		0010	–12 dB		
		0011	-9 dB		
		0100	-6 dB		
		0101	-3 dB		
		0110	0 dB		
		0111	3 dB		
		1000	6 dB		
[3:0]	MX3G1[3:0]	Bypass gain control. The signal from the left channel record mixer (Mixer 1) bypasses the converters and gain can be applied before the left playback mixer (Mixer 3).			
		Setting	Gain		
		0000	Mute (default)		
		0001	–15 dB		
		0010	–12 dB		
		0011	-9 dB		
		0100	-6 dB		
		0101	-3 dB		
		0110	0 dB		
		0111	3 dB		
		1000	6 dB		

R24: Playback Mixer Right (Mixer 4) Control 0, 16,414 (0x401E)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Reserved	MX4RM	MX4LM		MX4AU	IXG[3:0]		MX4EN

Table 49. Playback Mixer Right (Mixer 4) Control 0 Register

Bits	Bit Name	Description				
6	MX4RM	Mixer input mute. Mutes the 0 = muted (default). 1 = unmuted.				
5	MX4LM	Mixer input mute. Mutes the left DAC input to the right channel playback mixer (Mixer 4). 0 = muted (default). 1 = unmuted.				
[4:1]	MX4AUXG[3:0]	Mixer input gain. Controls th	e right channel auxiliary input gain to the right channel playback mixer (Mixer 4).			
		Setting	Gain			
		0000	Mute (default)			
		0001	–15 dB			
		0010	-12 dB			
		0011	-9 dB			
		0100	-6 dB			
		0101	-3 dB			
		0110	0 dB			
		0111	3 dB			
		1000	6 dB			
0	MX4EN	Mixer 4 enable. 0 = disabled (default). 1 = enabled.				

R25: Playback Mixer Right (Mixer 4) Control 1, 16,415 (0x401F)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	MX4G	i2[3:0]			MX40	i1[3:0]	

Table 50. Playback Mixer Right (Mixer 4) Control 1 Register

Bits	Bit Name	Description			
[7:4]	MX4G2[3:0]		ignal from the right channel record mixer (Mixer 2) bypasses the converters and gain right playback mixer (Mixer 4).		
		Setting	Gain		
		0000	Mute (default)		
		0001	–15 dB		
		0010	-12 dB		
		0011	-9 dB		
		0100	6 dB		
		0101	-3 dB		
		0110	0 dB		
		0111	3 dB		
		1000	6 dB		
[3:0]	MX4G1[3:0]	Bypass gain control. The signal from the left channel record mixer (Mixer 1) bypasses the converters and gain can be applied before the right playback mixer (Mixer 4).			
		Setting	Gain		
		0000	Mute (default)		
		0001	–15 dB		
		0010	-12 dB		
		0011	-9 dB		
		0100	6 dB		
		0101	-3 dB		
		0110	0 dB		
		0111	3 dB		
		1000	6 dB		

R26: Playback L/R Mixer Left (Mixer 5) Line Output Control, 16,416 (0x4020)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved		MX5G	i4[1:0]	MX5G	i3[1:0]	MX5EN

Table 51. Playback L/R Mixer Left (Mixer 5) Line Output Control Register

Bits	Bit Name	Description			
[4:3]	MX5G4[1:0]	[1:0] Mixer input gain boost. The signal from the right channel playback mixer (Mixer 4) can be enabled and in the playback L/R mixer left (Mixer 5).			
		Setting	Gain Boost		
		00	Mute (default)		
		01	0 dB output (–6 dB gain on each of the two inputs)		
		10	6 dB output (0 dB gain on each of the two inputs)		
		11	Reserved		
[2:1]	MX5G3[1:0]	Mixer input gain boost. The the playback L/R mixer left	e signal from the left channel playback mixer (Mixer 3) can be enabled and boosted in (Mixer 5).		
		Setting	Gain Boost		
		00	Mute (default)		
		01	0 dB output (–6 dB gain on each of the two inputs)		
		10	6 dB output (0 dB gain on each of the two inputs)		
		11	Reserved		
0	MX5EN	Mixer 5 enable.	- ·		
		0 = disabled (default).			
		1 = enabled.			

R27: Playback L/R Mixer Right (Mixer 6) Line Output Control, 16,417 (0x4021)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved		MX6G	i4[1:0]	MX6G	3[1:0]	MX6EN

Table 52. Playback L/R Mixer Right (Mixer 6) Line Output Control Register

Bits	Bit Name	Description	
[4:3]	MX6G4[1:0]	Mixer input gain boost. The in the playback L/R mixer ri	signal from the right channel playback mixer (Mixer 4) can be enabled and boosted ght (Mixer 6).
		Setting	Gain Boost
		00	Mute (default)
		01	0 dB output (–6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
[2:1]	MX6G3[1:0]	Mixer input gain boost. The the playback L/R mixer righ	signal from the left channel playback mixer (Mixer 3) can be enabled and boosted in t (Mixer 6).
		Setting	Gain Boost
		00	Mute (default)
		01	0 dB output (–6 dB gain on each of the two inputs)
		10	6 dB output (0 dB gain on each of the two inputs)
		11	Reserved
0	MX6EN	Mixer 6 enable.	
		0 = disabled (default).	
		1 = enabled.	

R28: Playback L/R Mixer Mono Output (Mixer 7) Control, 16,418 (0x4022)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		Reserved			MX7	[1:0]	MX7EN

Table 53. Playback L/R Mixer Mono Output (Mixer 7) Control Register

Bits	Bit Name	Description				
[2:1]	MX7[1:0]	L/R mono playback mixer (Mixer 7). Mixes the left and right playback mixers (Mixer 3 and Mixer 4) with eithe 0 dB or 6 dB gain boost. Additionally, this mixer can operate as a common-mode output, which is used as th virtual ground in a capless headphone configuration.				
		Setting	Gain Boost			
		00	Common-mode output (default)			
		01	0 dB output (–6 dB gain on each of the two inputs)			
		10	6 dB output (0 dB gain on each of the two inputs)			
		11	Reserved			
0	MX7EN	Mixer 7 enable.	· ·			
		0 = disabled (default). 1 = enabled.				

R29: Playback Headphone Left Volume Control, 16,419 (0x4023)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	LHPVOL[5:0]					LHPM	HPEN

Table 54. Playback Headphone Left Volume Control Register

Bits	Bit Name	Description				
[7:2]	LHPVOL[5:0]	Headphone volume control for left channel, LHP output. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 72 for a complete list of the volume settings.				
		Setting	Volume			
		000000	–57 dB (default)			
		111001	0 dB			
		111111	6 dB			
1	LHPM	Headphone mute for left char	nnel, LHP output (active low).			
		0 = mute.				
		1 = unmute (default).				
0	HPEN		nable. Logical OR with the HPMODE bit in Register R30. If either the HPEN bit or e headphone output is enabled.			

R30: Playback Headphone Right Volume Control, 16,420 (0x4024)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		RHPVOL[5:0] R					

Table 55. Playback Headphone Right Volume Control Register

Bits	Bit Name	Description					
[7:2]	RHPVOL[5:0]		Headphone volume control for right channel, RHP output. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 72 for a complete list of the volume settings.				
		Setting	Volume				
		000000	-57 dB (default)				
		111001	0 dB				
		111111	6 dB				
1	RHPM	Headphone mute for right channel, RHP output (active low).					
		0 = mute.					
		1 = unmute (default).					
0	HPMODE	OR with the HPEN bit in Regi is enabled.	These pins can be configured for either line outputs or headphone outputs. Logical ster R29. If either the HPMODE bit or the HPEN bit is set to 1, the headphone output				
		0 = enable line output (defa	,				
		1 = enable headphone outp	ut.				

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		LOUTV	OL[5:0]			LOUTM	LOMODE

Table 56. Playback Line Output Left Volume Control Register

Bits	Bit Name	Description	
[7:2]	LOUTVOL[5:0]	Line output volume control for left channel, LOUTN and LOUTP outputs. Each 1-bit step correspondence in volume. See Table 72 for a complete list of the volume settings.	
		Setting	Volume
		000000	–57 dB (default)
		111001	0 dB
		111111	6 dB
1	LOUTM	Line output mute for left chan	nel, LOUTN and LOUTP outputs (active low).
		0 = mute.	
		1 = unmute (default).	
0	LOMODE	•	nnel, LOUTN and LOUTP outputs. These pins can be configured for either line ts. To drive earpiece speakers, set this bit to 1 (headphone output).

R32: Playback Line Output Right Volume Control, 16,422 (0x4026)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		'OL[5:0]			ROUTM	ROMODE	

Table 57. Playback Line Output Right Volume Control Register

Bits	Bit Name	Description				
[7:2]	[7:2] ROUTVOL[5:0]	Line output volume control for right channel, ROUTN and ROUTP outputs. Each 1-bit step corresponds to a 1 dB increase in volume. See Table 72 for a complete list of the volume settings.				
		Setting	Volume			
		000000	–57 dB (default)			
		111001	0 dB			
		111111	6 dB			
1	ROUTM	Line output mute for right channel, ROUTN and ROUTP outputs (active low).				
		0 = mute.				
		1 = unmute (default).				
0	ROMODE	Line output mode for right channel, ROUTN and ROUTP outputs. These pins can be configured for either line outputs or headphone outputs. To drive earpiece speakers, set this bit to 1 (headphone output).				
		0 = line output (default).				
		1 = headphone output.				

R33: Playback Mono Output Control, 16,423 (0x4027)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	MONOVOL[5:0]					MONOM	MOMODE

Table 58. Playback Mono Output Control Register

Bits	Bit Name	Description				
[7:2]	MONOVOL[5:0]	Mono output volume control. Each 1-bit step corresponds to a 1 dB increase in volume. If MX7[1:0] in Register R28 is set for common-mode output, volume control is disabled. See Table 72 for a complete list of the volume settings.				
		Setting	Volume			
		000000	-57 dB (default)			
		111001	0 dB			
		111111	6 dB			
1	MONOM	Mono output mute (active low	<i>ı</i>).			
		0 = mute.				
		1 = unmute (default).				
0	MOMODE	Headphone mode enable. If MX7[1:0] in Register R28 is set for common-mode output for a capless headphone configuration, this bit should be set to 1 (headphone output). 0 = line output (default). 1 = headphone output.				

R34: Playback Pop/Click Suppression, 16,424 (0x4028)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved		POPMODE	POPLESS	ASLE\	W[1:0]	Reserved

Table 59. Playback Pop/Click Suppression Register

Bits	Bit Name	Description			
4	POPMODE	Pop suppression circuit power saving mode. The pop suppression circuits charge faster in normal operation; however, after they are charged, they can be put into low power operation. 0 = normal (default). 1 = low power.			
3	POPLESS	Pop suppression disable. The pop suppression circuits are enabled by default. They can be disabled to save power; however, disabling the circuits increases the risk of pops and clicks. 0 = enabled (default). 1 = disabled.			
[2:1]	ASLEW[1:0]	Analog volume slew rate for p	layback volume controls.		
		Setting	Slew Rate		
		00	21.25 ms (default)		
		01	42.5 ms		
		10	85 ms		
		11	Off		

R35: Playback Power Management, 16,425 (0x4029)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved						PLEN

Table 60. Playback Power Management Register

Bits	Bit Name	Description
1	PREN	Playback right channel enable.
		0 = disabled (default).
		1 = enabled.
0	PLEN	Playback left channel enable.
		0 = disabled (default).
		1 = enabled.

R36: DAC Control 0, 16,426 (0x402A)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
DACMC			Rese		DEMPH	DACE	N[1:0]

Table 61. DAC Control 0 Register

Bits	Bit Name	Description					
[7:6]	DACMONO[1:0]		DAC mono mode. The DAC channels can be set to mono mode within the DAC and output on the left channel, the right channel, or both channels.				
		Setting	Mono Mode				
		00	Stereo (default)				
		01	Left channel in mono mode				
		10	Right channel in mono mode				
		11	Both channels in mono mode				
5	DACPOL	Invert input polarity of the DAG 0 = normal (default). 1 = inverted.					
2	DEMPH	DAC de-emphasis filter enable. 0 = disabled (default). 1 = enabled.	The de-emphasis filter is designed for use with a sampling rate of 44.1 kHz only.				
[1:0]	DACEN[1:0]	DAC enable.					
		Setting	DACs Enabled				
		00	Both off (default)				
		01	Left on				
		10	Right on				
		11	Both on				

R37: DAC Control 1, 16,427 (0x402B)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
	LDAVOL[7:0]							

Table 62. DAC Control 1 Register

Bits	Bit Name	Description				
[7:0]	7:0] LDAVOL[7:0]	Controls the digital volume attenuation for left channel inputs from the left DAC. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 71 for a complete list of the volume settings.				
		Setting	Volume Attenuation			
		0000000	0 dB (default)			
		0000001	–0.375 dB			
		0000010	–0.75 dB			
		1111110	–95.25 dB			
		11111111	-95.625 dB			

R38: DAC Control 2, 16,428 (0x402C)

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	RDAVOL[7:0]						

Table 63. DAC Control 2 Register

Bits	Bit Name	Description					
[7:0]	RDAVOL[7:0]	Controls the digital volume attenuation for right channel inputs from the right DAC. Each bit corresponds to a 0.375 dB step with slewing between settings. See Table 71 for a complete list of the volume settings.					
		Setting	Volume Attenuation				
		0000000	0 dB (default)				
		0000001	–0.375 dB				
		0000010	–0.75 dB				
		1111110	–95.25 dB				
		1111111	–95.625 dB				

R39: Serial Port Pad Control, 16,429 (0x402D)

The optional pull-up/pull-down resistors are nominally 250 k Ω . When enabled, these pull-up/pull-down resistors set the serial port signals to a defined state when the signal source becomes three-state.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ADCSI	OP[1:0]	DACSE	DP[1:0]	LRCLK	(P[1:0]	BCLK	P[1:0]

Table 64. Serial Port Pad Control Register

Bits	Bit Name	Description				
[7:6]	ADCSDP[1:0]	ADC_SDATA pad pull-up/pu	ll-down configuration.			
		Setting	Configuration			
		00	Pull-up			
		01	Reserved			
		10	None (default)			
		11	Pull-down			
[5:4]	DACSDP[1:0]	DAC_SDATA pad pull-up/pu	II-down configuration.			
		Setting	Configuration			
		00	Pull-up			
		01	Reserved			
		10	None (default)			
		11	Pull-down			
[3:2]	LRCLKP[1:0]	LRCLK pad pull-up/pull-down configuration.				
		Setting	Configuration			
		00	Pull-up			
		01	Reserved			
		10	None (default)			
		11	Pull-down			
[1:0]	BCLKP[1:0]	BCLK pad pull-up/pull-dowr	n configuration.			
		Setting	Configuration			
		00	Pull-up			
		01	Reserved			
		10	None (default)			
		11	Pull-down			

R40: Control Port Pad Control 0, 16,431 (0x402F)

The optional pull-up/pull-down resistors are nominally 250 k Ω . When enabled, these pull-up/pull-down resistors set the control port signals to a defined state when the signal source becomes three-state.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CDAT	CDATP[1:0] CLCHP[1:0]		P[1:0]	SCLP	P[1:0]	SDAF	P[1:0]

Bits	Bit Name	Description			
[7:6]	CDATP[1:0]	CDATA pad pull-up/pu	Ill-down configuration.		
		Setting	Configuration		
		00	Pull-up		
		01	Reserved		
		10	None (default)		
		11	Pull-down		
[5:4]	CLCHP[1:0]	CLATCH pad pull-up/p	ull-down configuration.		
		Setting	Configuration		
		00	Pull-up		
		01	Reserved		
		10	None (default)		
		11	Pull-down		
[3:2]	SCLP[1:0]	SCL/CCLK pad pull-up/pull-down configuration.			
		Setting	Configuration		
		00	Pull-up		
		01	Reserved		
		10	None (default)		
		11	Pull-down		
[1:0]	SDAP[1:0]	SDA/COUT pad pull-up	o/pull-down configuration.		
		Setting	Configuration		
		00	Pull-up		
		01	Reserved		
		10	None (default)		
		11	Pull-down		

Table 65. Control Port Pad Control 0 Register

R41: Control Port Pad Control 1, 16,432 (0x4030)

With IOVDD set to 3.3 V, the low and high drive strengths of the SDA/COUT pin are approximately 2.0 mA and 4.0 mA, respectively. The high drive strength mode may be useful for generating a stronger ACK pulse in I²C mode, if needed.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	Reserved						SDASTR

Table 66. Control Port Pad Control 1 Register

Bits	Bit Name	Description
0	SDASTR	SDA/COUT pin drive strength. 0 = low (default). 1 = high.

R42: Jack Detect Pin Control, 16,433 (0x4031)

With IOVDD set to 3.3 V, the low and high drive strengths of the JACKDET/MICIN pin are approximately 2.0 mA and 4.0 mA, respectively. The optional pull-up/pull-down resistors are nominally 250 k Ω . When enabled, these pull-up/pull-down resistors set the input signals to a defined state when the signal source becomes three-state.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Rese	rved	JDSTR	Reserved	JDP[[1:0]	Rese	rved

Bits	Bit Name	Description				
5	JDSTR	JACKDET/MICIN pin drive stree	JACKDET/MICIN pin drive strength.			
		0 = low (default).	0 = low (default).			
		1 = high.	1 = high.			
[3:2]	JDP[1:0]	JACKDET/MICIN pad pull-up/p	oull-down configuration.			
		Setting	Configuration			
		00	Pull-up			
		01	Reserved			
		10	None (default)			
		11	Pull-down			

Table 67. Jack Detect Pin Control Register

R67: Dejitter Control, 16,438 (0x4036)

The dejitter control register allows the size of the dejitter window to be set, and also allows all dejitter circuits in the device to be activated or bypassed. Dejitter circuits protect against duplicate samples or skipped samples due to jitter from the serial ports in slave mode. Disabling and reenabling certain subsystems in the device—that is, the ADCs, serial ports, and DACs—during operation can cause the associated dejitter circuits to fail. As a result, audio data fails to be output to the next subsystem in the device.

When the serial ports are in master mode, the dejitter circuit can be bypassed by setting the dejitter window to 0. When the serial ports are in slave mode, the dejitter circuit can be reinitialized prior to outputting audio from the device, guaranteeing that audio is output to the next subsystem in the device. Any time that audio must pass through the ADCs, serial port, or DACs, the dejitter circuit can be bypassed and reset by setting the dejitter window size to 0. In this way, the dejitter circuit can be immediately reactivated, without a wait period, by setting the dejitter window size to the default value of 3.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
DEJIT[7:0]							

Table 68. Dejitter Control Register

Bits	Bit Name	Description	
[7:0]	DEJIT[7:0]	Dejitter window size.	
		Window Size	Core Clock Cycles
		0000000	0
			 3 (default)
		00000011	
		 00000101	5

Binary Value	R9 Volume Settings Volume Setting (dB)	Binary Value	Volume Setting (dB)
000000	-12	110011	26.25
000001	-11.25	110100	27
000010	-10.5	110100	27.75
000010	-9.75	110110	28.5
000100	-9	110111	29.25
			30
000101	-8.25	111000	
000110	-7.5	111001	30.75
000111	-6.75	111010	31.5
001000	-6	111011	32.25
001001	-5.25	111100	33
001010	-4.5	111101	33.75
001011	-3.75	111110	34.5
001100	-3	111111	35.25
001101	-2.25		
001110	-1.5		ise Gate Threshold
001111	-0.75	Binary Value	Noise Gate Threshold (dB)
010000	0	00000	-76.5
010001	0.75	00001	-75
010010	1.5	00010	-73.5
010011	2.25	00011	-72
010100	3	00100	-70.5
010101	3.75	00101	-69
010110	4.5		
010111	5.25	00110	-67.5
011000	6	00111	-66
011001	6.75	01000	-64.5
011010	7.5	01001	-63
011011	8.25	01010	-61.5
011100	9	01011	-60
011101	9.75	01100	-58.5
011110	10.5	01101	-57
011111	11.25	01110	-55.5
100000	12	01111	-54
100001	12.75	10000	-52.5
100010	13.5	10001	-51
100011	14.25	10010	-49.5
100100	15	10010	-48
100101	15.75	10100	-46.5
100110	16.5	10101	-45
100111	17.25	10110	-43.5
101000	18	10111	-42
101001	18.75	11000	-40.5
101010	19.5	11001	-39
101011	20.25	11010	-37.5
101100	21	11011	-36
101101	21.75	11100	-34.5
101110	22.5	11101	-33
101111	23.25	11110	-31.5
110000	24		-30
110001	24.75	11111	-50
110010	25.5		

25.5

110010

Table 71	. R20, R2	l, R37, and	l R38 Volum	e Settings
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	Volume Attenuation (dP)	Binary Value	Volume Attenuation (dB)
Binary Value 00000000	Volume Attenuation (dB)	00110000	Volume Attenuation (dB)
	-0.375		
00000001		00110001	-18.375
00000010	-0.75	00110010	-18.75
00000011	-1.125	00110011	-19.125
00000100	-1.5	00110100	-19.5
00000101	-1.875	00110101	-19.875
00000110	-2.25	00110110	-20.25
00000111	-2.625	00110111	-20.625
00001000	-3	00111000	-21
00001001	-3.375	00111001	-21.375
00001010	-3.75	00111010	-21.75
00001011	-4.125	00111011	-22.125
00001100	-4.5	00111100	-22.5
00001101	-4.875	00111101	-22.875
00001110	-5.25	00111110	-23.25
00001111	-5.625	00111111	-23.625
00010000	-6	0100000	-24
00010001	-6.375	01000001	-24.375
00010010	-6.75	01000010	-24.75
00010011	-7.125	01000011	-25.125
00010100	-7.5	01000100	-25.5
00010101	-7.875	01000101	-25.875
00010110	-8.25	01000110	-26.25
00010111	-8.625	01000111	-26.625
00011000	-9	01001000	-27
00011001	-9.375	01001001	-27.375
00011010	-9.75	01001010	-27.75
00011011	-10.125	01001011	-28.125
00011100	-10.5	01001100	-28.5
00011101	-10.875	01001101	-28.875
00011110	-11.25	01001110	-29.25
00011111	-11.625	01001111	-29.625
00100000	-12	01010000	-30
00100001	-12.375	01010001	-30.375
00100010	-12.75	01010010	-30.75
00100011	-13.125	01010010	-31.125
00100100	-13.5	01010100	-31.5
00100101	-13.875	01010101	-31.875
00100110	-14.25	01010110	-32.25
00100111	-14.625	01010111	-32.625
00101000	-15	01011000	-33
	-15.375		
00101001		01011001	-33.375
00101010	-15.75	01011010	-33.75
00101011	-16.125	01011011	-34.125
00101100	-16.5	01011100	-34.5
00101101	-16.875	01011101	-34.875
00101110	-17.25	01011110	-35.25
00101111	-17.625	01011111	-35.625

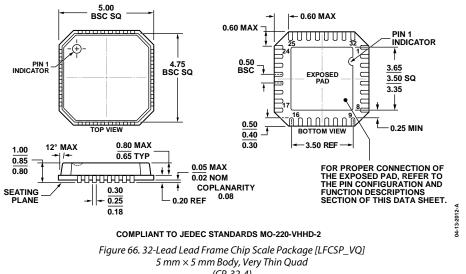
Binary Value	Volume Attenuation (dB)	Bi	nary Value	Volume Attenuation (dB)
01100000	-36		010001	-54.375
01100001	-36.375		010010	-54.75
01100010	-36.75		010011	-55.125
01100011	-37.125		010100	-55.5
01100100	-37.5		010101	-55.875
01100101	-37.875		010110	-56.25
01100110	-38.25		010111	-56.625
01100111	-38.625		011000	-57
01101000	-39		011001	-57.375
01101001	-39.375		011010	-57.75
01101010	-39.75		011011	-58.125
01101011	-40.125		011100	-58.5
01101100	-40.5		011101	-58.875
01101101	-40.875		011110	-59.25
01101110	-41.25		011111	-59.625
01101111	-41.625		100000	-60
01110000	-42		100001	-60.375
01110001	-42.375		100010	-60.75
01110010	-42.75		100011	-61.125
01110011	-43.125		100100	-61.5
01110100	-43.5		0100101	-61.875
01110101	-43.875	10	0100110	-62.25
01110110	-44.25	10	0100111	-62.625
01110111	-44.625	10	101000	-63
01111000	-45	10	101001	-63.375
01111001	-45.375	10	101010	-63.75
01111010	-45.75	10	0101011	-64.125
01111011	-46.125	10	101100	-64.5
01111100	-46.5	10	0101101	-64.875
01111101	-46.875	10	0101110	-65.25
01111110	-47.25	10	0101111	-65.625
01111111	-47.625	10	110000	-66
1000000	-48	10	110001	-66.375
1000001	-48.375	10	110010	-66.75
10000010	-48.75	10	110011	-67.125
10000011	-49.125	10	110100	-67.5
10000100	-49.5	10	110101	-67.875
10000101	-49.875	10	0110110	-68.25
10000110	-50.25	10	0110111	-68.625
10000111	-50.625	10	111000	-69
10001000	-51		0111001	-69.375
10001001	-51.375		0111010	-69.75
10001010	-51.75		0111011	-70.125
10001011	-52.125		0111100	-70.5
10001100	-52.5		0111101	-70.875
10001101	-52.875		0111110	-71.25
10001110	-53.25		0111111	-71.625
10001111	-53.625		000000	-72
10010000	-54		000001	-72.375
				, 2.313

Data Sheet

Binary Value	Volume Attenuation (dB)	Binary Value	Volume Attenuation (dB)
11000010	-72.75	11110011	-91.125
11000011	-73.125	11110100	-91.5
11000100	-73.5	11110101	-91.875
11000101	-73.875	11110110	-92.25
11000110	-74.25	11110111	-92.625
11000111	-74.625	11111000	-93
11001000	-75	11111001	-93.375
11001001	-75.375	11111010	-93.75
11001010	-75.75	11111011	-94.125
11001011	-76.125	1111100	-94.5
11001100	-76.5	1111101	-94.875
11001101	-76.875	1111110	-95.25
11001110	-77.25	11111111	-95.625
11001111	-77.625		
11010000	-78	Table 72. R29 throu	ugh R33 Volume Settings
11010001	-78.375	Binary Value	Volume Setting (dB)
11010010	-78.75	000000	-57
11010011	-79.125	000001	-56
11010100	-79.5	000010	-55
11010101	-79.875	000011	-54
11010110	-80.25	000100	-53
11010111	-80.625	000101	-52
11011000	-81	000110	-51
11011001	-81.375	000111	-50
11011010	-81.75	001000	-49
11011011	-82.125	001001	-48
11011100	-82.5	001010	-47
11011101	-82.875	001010	-46
11011110	-83.25	001100	-45
11011111		001100	-44
	-83.625	001101	
11100000	-84		-43
11100001	-84.375	001111	-42
11100010	-84.75	010000	-41
11100011	-85.125	010001	-40
11100100	-85.5	010010	-39
11100101	-85.875	010011	-38
11100110	-86.25	010100	-37
11100111	-86.625	010101	-36
11101000	-87	010110	-35
11101001	-87.375	010111	-34
11101010	-87.75	011000	-33
11101011	-88.125	011001	-32
11101100	-88.5	011010	-31
11101101	-88.875	011011	-30
11101110	-89.25	011100	-29
11101111	-89.625	011101	-28
11110000	-90	011110	-27
11110001	-90.375	011111	-26
11110010	-90.75	100000	-25

Binary Value	Volume Setting (dB)
100001	-24
100010	-23
100011	-22
100100	-21
100101	-20
100110	–19
100111	-18
101000	-17
101001	-16
101010	–15
101011	-14
101100	-13
101101	-12
101110	-11
101111	-10
110000	-9
110001	-8
110010	-7
110011	-6
110100	-5
110101	-4
110110	-3
110111	-2
111000	-1
111001	0
111010	1
111011	2
111100	3
111101	4
111110	5
111111	6

OUTLINE DIMENSIONS



(CP-32-4) Dimensions shown in millimeters

ORDERING GUIDE

Model ^{1, 2}	Temperature Range	Package Description	Package Option
ADAU1961WBCPZ	-40°C to +105°C	32-Lead Lead Frame Chip Scale Package [LFCSP_VQ]	CP-32-4
ADAU1961WBCPZ-R7	-40°C to +105°C	32-Lead Lead Frame Chip Scale Package [LFCSP_VQ], 7"Tape and Reel	CP-32-4
ADAU1961WBCPZ-RL	-40°C to +105°C	32-Lead Lead Frame Chip Scale Package [LFCSP_VQ], 13" Tape and Reel	CP-32-4

¹ Z = RoHS Compliant Part.

 2 W = Qualified for Automotive Applications.

AUTOMOTIVE PRODUCTS

The ADAU1961W models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

NOTES

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- Консультации по применению компонента;
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- Техническая поддержка проекта;
- Защита от снятия компонента с производства.



Как с нами связаться

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