ACT30

Rev 5, 05-Jun-09

High Performance Off-Line Controller
ActiveSwitcher^{TM} IC Family

Active-Semi

FEATURES

- **Lowest Total Cost Solution**
- **0.15W Standby Power**
- **Emitter Drive Allows Safe NPN Transistor Flyback Use**
- **Hiccup Mode Short Circuit**
- **Current Mode Operation**
- **Over-Current Protection**
- **Under-voltage Protection with Auto-Restart**
- **Proprietary Scalable Output Driver**
- **Flexible Packaging Options (Including TO-92)**
- **65kHz or 100kHz Switching Frequency**
- **Selectable 0.4A to 1.2A Current Limit**

APPLICATIONS

- **Battery Chargers**
- **Power Adaptors**
- **Standby Power Supplies**
- **Appliances**
- **Universal Off-Line Power Supplies**

GENERAL DESCRIPTION

The ACT30 is a high performance green-energy off-line power supply controller. It features a scalable driver for driving external NPN or MOSFET transistors for line voltage switching. This proprietary architecture enables many advanced features to be integrated into a small package (TO-92 or SOT23-B), resulting in lowest total cost solution.

The ACT30 design has six internal terminals and is a pulse frequency and width modulation IC with many flexible packaging options. One combination of internal terminals is packaged in the space-saving TO-92 package (A/B versions) for 65kHz or 100kHz switching frequency and with 400mA or 800mA current limit.

Consuming only 0.15W in standby, the IC features over-current, hiccup mode short circuit, and under-voltage protection mechanisms.

The ACT30 is ideal for use in high performance universal adaptors and chargers. For highest performance versus cost and smallest PCB area, use the ACT30 in combination with the ACT32 CV/CC Controller.
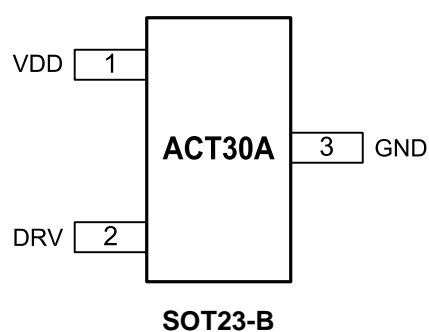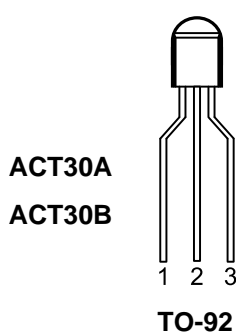
Figure 1:

Simplified Application Circuit
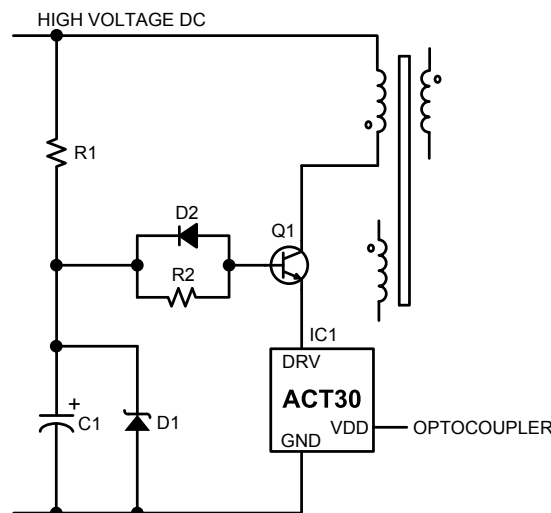
ORDERING INFORMATION

PART NUMBER	SWITCHING FREQUENCY	CURRENT LIMIT	JUNCTION TEMPERATURE	PACKAGE	PINS
ACT30AHT	65kHz	400mA	-40˚C to 150˚C	TO-92	3
ACT30BHT	65kHz	800mA	-40˚C to 150˚C	TO-92	3
ACT30AYT	65kHz	400mA	-40˚C to 150˚C	SOT23-B	3

PIN CONFIGURATION

ACT30A
ACT30B

1 2 3
TO-92

VDD 1
ACT30A
3 GND
DRV 2
SOT23-B

PIN DESCRIPTIONS

PIN		NAME	DESCRIPTION
TO-92	SOT23-B		
1	1	VDD	Power Supply Pin. Connect to optocoupler's emitter. Internally limited to 5.5V max. Bypass to GND with a proper compensation network.
2	3	GND	Ground.
3	2	DRV	Driver Output (TO-92 Only). Connect to emitter of the high voltage NPN or MOSFET. For ACT30A/C, DRV pin is internally connected to DRV1. For ACT30B/D, DRV pin is internally connected to both DRV1 and DRV2.

ABSOLUTE MAXIMUM RATINGS^{①}

| PARAMETER		VALUE	UNIT
VDD, FREQ to GND		-0.3 to 6	V
VDD Current		20	mA
DRV, DRV1, DRV2 to GND		-0.3 to 18	V
Continuous DRV, DRV1, DRV2 Current		Internally limited	A
Maximum Power Dissipation	TO-92	0.6	W
	SOT23-B	0.39	
Operating Junction Temperature		-40 to 150	˚C
Storage Temperature		-55 to 150	˚C
Lead Temperature (Soldering, 10 sec)		300	˚C

①: Do not exceed these limits to prevent damage to the device. Exposure to absolute maximum rating conditions for long periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{VDD} = 4V, T_J = 25°C, unless otherwise specified.)

PARAMETER	SYMBOL	TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{VDD} Start Voltage	V_{START}	Rising edge		4.75	5	5.25	V
DRV1 Start Voltage	V_{DRVST}	DRV1 must be higher than this voltage to start up.	ACT30A		8.6	10.5	V
			ACT30B		9.6	11.5	
DRV1 Short-Circuit Detect Threshold	V_{SCDRV}			6.35	6.8	7.25	V
V_{VDD} Under-Voltage Threshold	V_{UV}	Falling edge		3.17	3.35	3.63	V
V_{VDD} Clamp Voltage		10mA		5.15	5.45	5.95	V
Startup Supply Current	I_{DDST}	V_{VDD} = 4V before V_{UV}			0.23	0.45	mA
Supply Current	I_{DD}				0.7	1	mA
Switching Frequency	f_{SW}	ACT30A/B or FREQ = 0		50	65	80	kHz
Maximum Duty Cycle	D_{MAX}	ACT30A, V_{VDD} = 4V		67	75	83	%
		ACT30B, V_{VDD} = 4V		60			
Minimum Duty Cycle	D_{MIN}	V_{VDD} = 4.6V			3.5		%
Effective Current Limit	I_{LIM}	V_{VDD} = V_{UV} + 0.1V	ACT30A	340	400	480	mA
			ACT30B with DRV1 = DRV2	680	800	920	
V_{VDD} to DRV1 Current Coefficient	G_{GAIN}				-0.29		A/V
VDD Dynamic Impedance	R_{VDD}				9		kΩ
DRV1 or DRV2 Driver On-Resistance	R_{DRV1}, R_{DRV2}	I_{DRV1} = I_{DRV2} = 0.05A			3.6		Ω
DRV1 Rise Time		1nF load, 15Ω pull-up			30		ns
DRV1 Fall Time		1nF load, 15Ω pull-up			20		ns
DRV1 and DRV2 Switch Off Current		Driver off, V_{DRV1} = V_{DRV2} = 10V			12	30	µA

FUNCTIONAL BLOCK DIAGRAM
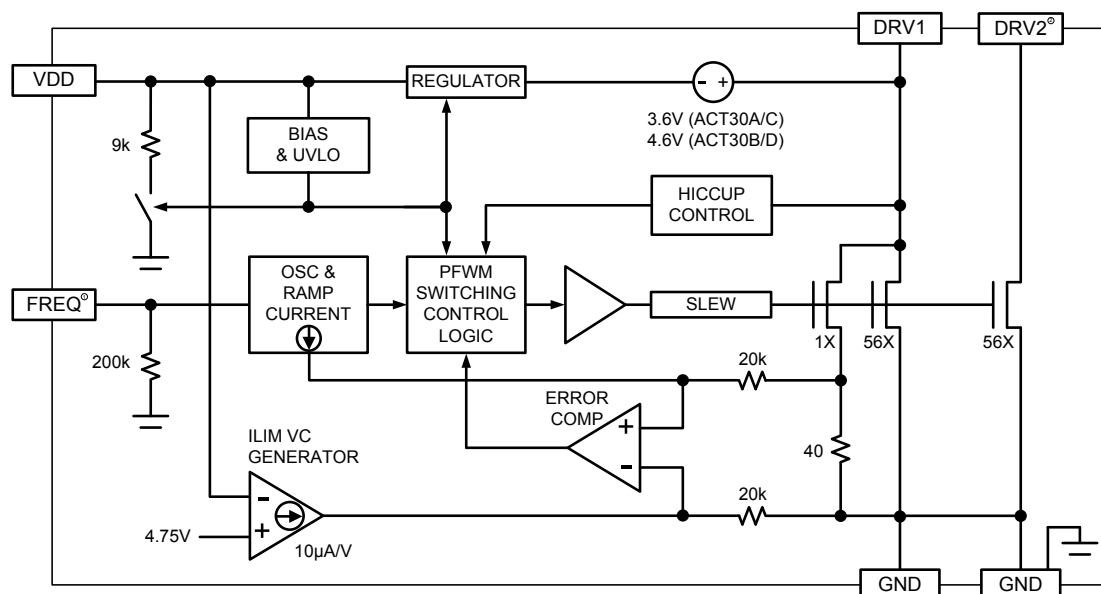
①: FREQ terminal wire-bonded to VDD in ACT30C/D (TO-92)

②: DRV2 terminal wire-bonded to DRV1 in ACT30B/D (TO-92)

FUNCTIONAL DESCRIPTION

As seen in the *Functional Block Diagram*, the main components include switching control logic, two on-chip medium-voltage power-MOSFETs with parallel current sensor, driver, oscillator and ramp generator, current limit VC generator, error comparator, hiccup control, bias and under voltage-lockout, and regulator circuitry.

As seen in the *Functional Block Diagram*, the design has six internal terminals. V_{VDD} is the power supply terminal. DRV1 and DRV2 are linear driver outputs that can drive the emitter of an external high voltage NPN transistor or N-channel MOSFET. This emitter-drive method takes advantage of the high V_{CBO} of the transistor, allowing a low cost transistor such as '13003 (V_{CBO} = 700V) or '13002 (V_{CBO} = 600V) to be used for a wide AC input range. The slew-rate limited driver coupled with the turn-off characteristics of an external NPN transitor result in lower EMI.

The driver peak current is designed to have a negative voltage coefficient with respect to supply voltage V_{VDD}, so that lower supply voltage automatically results in higher DRV1 peak current. This way, the optocoupler can control V_{VDD} directly to affect driver current.

Startup Sequence

Figure 1 shows a *Simplified Application Circuit* for the ACT30. Initially, the small current through resistor R1 charges up the capacitor C1, and the BJT acts as a follower to bring up the DRV1 voltage. An internal regulator generates a V_{VDD} voltage equal to V_{DRV1} – 3.6V for ACT30A (V_{DRV1} – 4.6V for ACT30B) but limits it to 5.5V max. As V_{VDD} crosses 5V, the regulator sourcing function stops and V_{VDD} begins to drop due to its current consumption. As V_{VDD} voltage decreases below 4.75V, the IC starts to operate with increasing driver current. When the output voltage reaches regulation point, the optocoupler feedback circuit stops V_{VDD} from decreasing further. The switching action also allows the auxiliary windings to take over in supplying the C1 capacitor. Figure 2 shows a typical startup sequence for the ACT30.

To limit the auxiliary voltage, use a 12V zener diode for ACT30A or a 13V zener diode for ACT30B (D1 diode in Figure 1).

Even though up to 2MΩ startup resistor (R1) can be used due to the very low startup current, the actual R1 value should be chosen as a compromise between standby power and startup time delay.

Figure 2:

Startup Waveforms
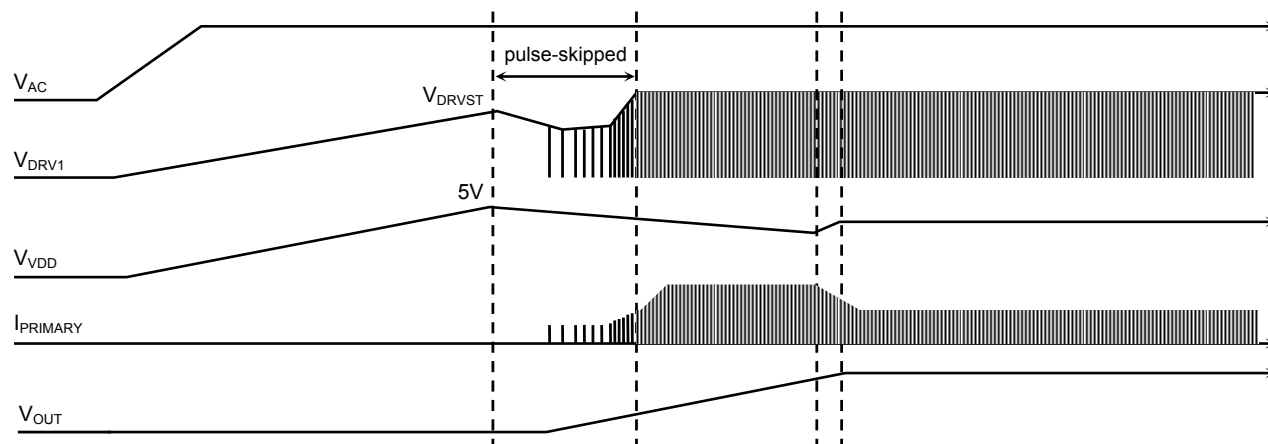
Normal Operation

In normal operation, the feedback signal from the secondary side is transmitted through the optocoupler as a current signal into V_{VDD} pin, which has dynamic impedance of 9kΩ. The resulting V_{VDD} voltage affects the switching of the IC. As seen in the *Functional Block Diagram*, the Current Limit VC Generator uses the V_{VDD} voltage difference with 4.75V to generate a proportional offset at the negative input of the Error Comparator.

The drivers turn on at the beginning of each switching cycle. The current sense resistor current, which is a fraction of the transformer primary current, increases with time as the primary current increases. When the voltage across this current sense resistor plus the oscillator ramp signal equals Error Comparator's negative input voltage, the drivers turn off. Thus, the peak DRV1 current has a negative voltage coefficient of -0.29A/V and can be calculated from the following:

$$I_{DRV1PEAK} = 0.29A/V \times \left(4.75V - V_{VDD}\right)$$

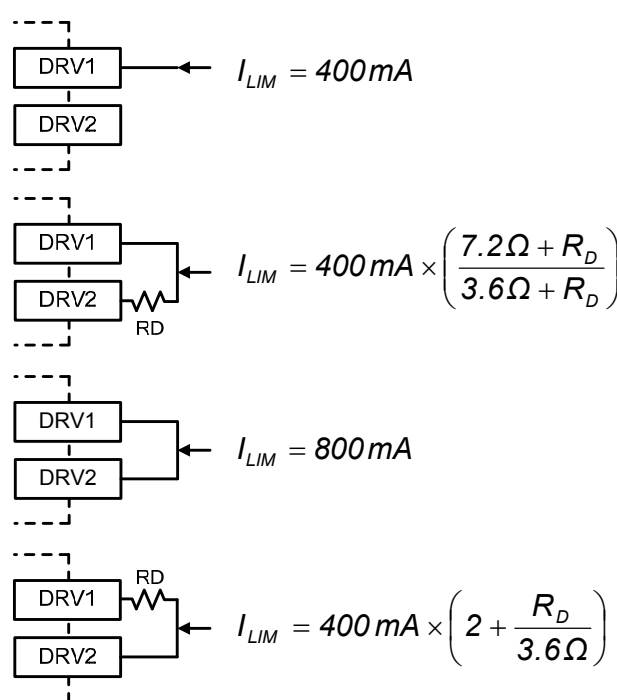for V_{VDD} < 4.75V and duty cycle < 50%.

When the output voltage is lower than regulation, the current into V_{VDD} pin is zero and V_{VDD} voltage decreases. At $V_{VDD} = V_{UV} = 3.35V$, the peak DRV1 current has maximum value of 400mA.

Current Limit Adjustment

The IC's proprietary driver arrangement allows the current limit to be easily adjusted between 400mA and 1.2A. To understand this, the drivers have to be utilized as linear resistive devices with typically

3.6Ω (rather than as digital output switches). The current limit can then be calculated through linear combination as shown in Figure 3. For TO-92 package, the ACT30A are preprogrammed to 400mA current limit and the ACT30B are preprogrammed to 800mA current limit, for SOT23-B package, the ACT30A are preprogrammed to 400mA current limit.

Figure 3:

Driver Output Configurations

$$I_{LIM} = 400\,mA$$

$$I_{LIM} = 400\,mA \times \left(\frac{7.2\Omega + R_D}{3.6\Omega + R_D}\right)$$

$$I_{LIM} = 800\,mA$$

$$I_{LIM} = 400\,mA \times \left(2 + \frac{R_D}{3.6\Omega}\right)$$

Pulse Modulation

The PFWM Switching Control Logic block operates in different modes depending on the output load current level. At light load, the V_{VDD} voltage is around 4.75V. The energy delivered by each switching cycle (with minimum on time of 500ns) to the output causes V_{VDD} to increase slightly above 4.75V. The FPWM Switching Control Logic block is able to detect this condition and prevents the IC from switching until V_{VDD} is below 4.75V again. This results in a pulse-modulation action with fixed pulse width and varying frequency, and low power consumption because the switching frequency is reduced. Typical system standby power consumption is 0.15W.

Short Circuit Hiccup

When the output is short circuited, the ACT30 enters hiccup mode operation. In this condition, the auxiliary supply voltage collapses. An on-chip detector compares DRV1 voltage during the off-time of each cycle to 6.8V. If DRV1 voltage is below 6.8V, the IC will not start the next cycle, causing both the auxiliary supply voltage and V_{VDD} to reduce further. The circuit enters startup mode when V_{VDD} drops below 3.35V. This hiccup behavior continues until the short circuit is removed. In this behavior, the effective duty cycle is very low resulting in very low short circuit current.

To make sure that the IC enters hiccup mode easily, the transformer should be constructed so that there is close coupling between secondary and auxiliary, so that the auxiliary voltage is low when the output is short-circuited. This can be achieved with the primary/auxiliary/secondary sequencing from the bobbin.

Active-Semi

APPLICATIONS INFORMATION

External Power Transistor

The ACT30 allows a low-cost high voltage power NPN transistor such as '13003 or '13002 to be used safely in a flyback configuration. The required collector voltage rating for V_{AC} = 265V with full output load is at least 600V to 700V. As seen in Figure 4, the breakdown voltage of an NPN is significantly improved when it is driven at its emitter. Thus, the ACT30 and '13002 or '13003 combination meet the necessary breakdown safety requirement even though RCC circuits using '13002 or '13003 do not. Table 1 lists the breakdown voltage of some transistors appropriate for use with the ACT30.

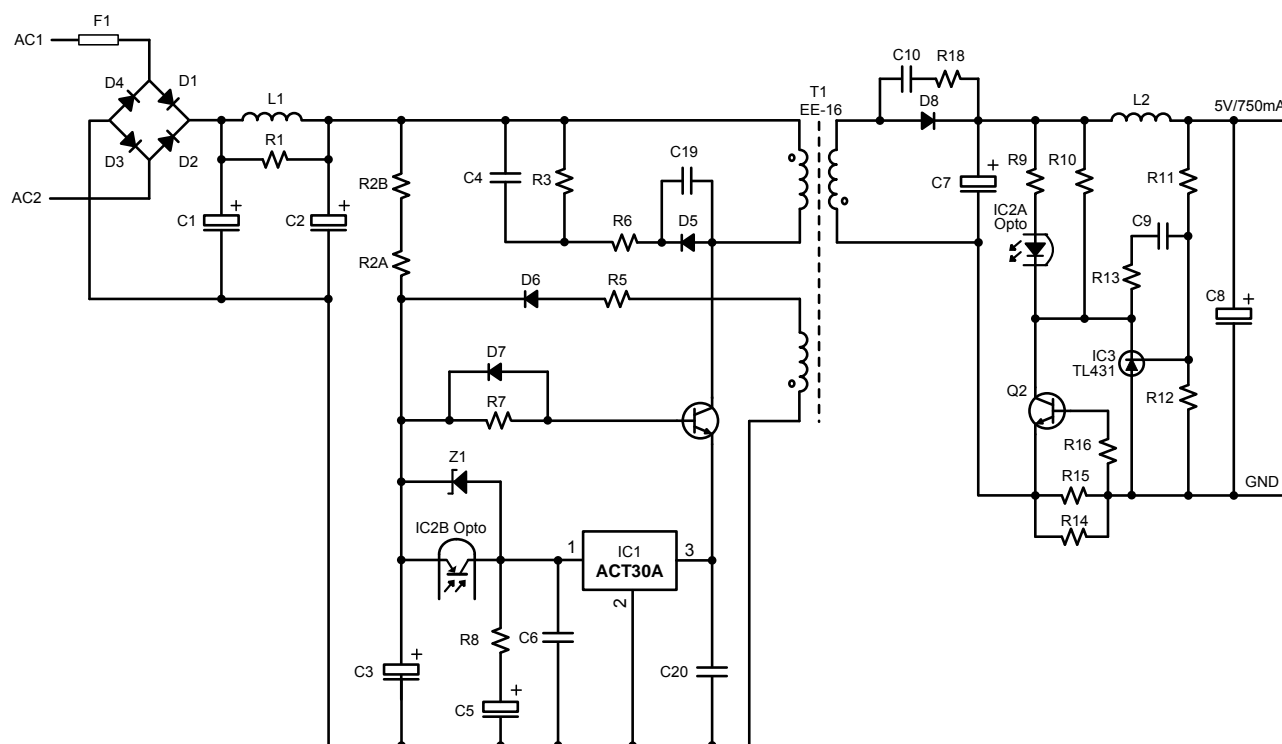**Table 1:**

Recommended Power Transistor List

DEVICE	V_{CBO}	V_{CEO}	I_C	h_{FEMIN}	PACKAGE
MJE13002	600V	300V	1.5A	8	TO-126
MJE13003, KSE13003	700V	400V	1.5A	8	TO-126
STX13003	700V	400V	1A	8	TO-92

Figure 4:

NPN Reverse Bias Safe Operation Area
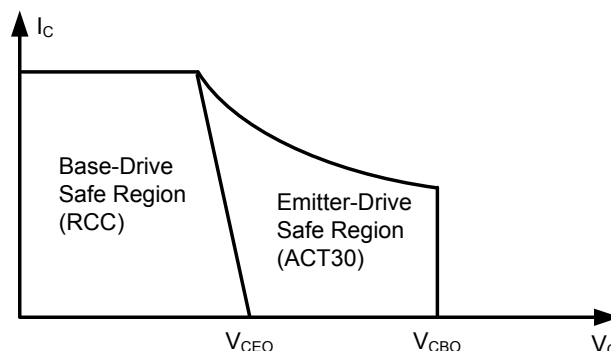
The power dissipated in the NPN transistor is equal to the collector current times the collector-emitter voltage. As a result, the transistor must always be in saturation when turned on to prevent excessive power dissipation. Select an NPN transistor with sufficiently high current gain (h_{FEMIN} > 8) and a base drive resistor (R2 in Figure 1) low enough to ensure that the transistor easily saturates.

Figure 5:

A 3.75W Charger Using ACT30A in Combination with TL431

Application Example

The application circuit in Figure 5 provides a 5V/0.75A constant voltage/constant current output. The performance of this circuit is summarized in Table 2.

Table 2:

System Performance of Circuit in Figure 5

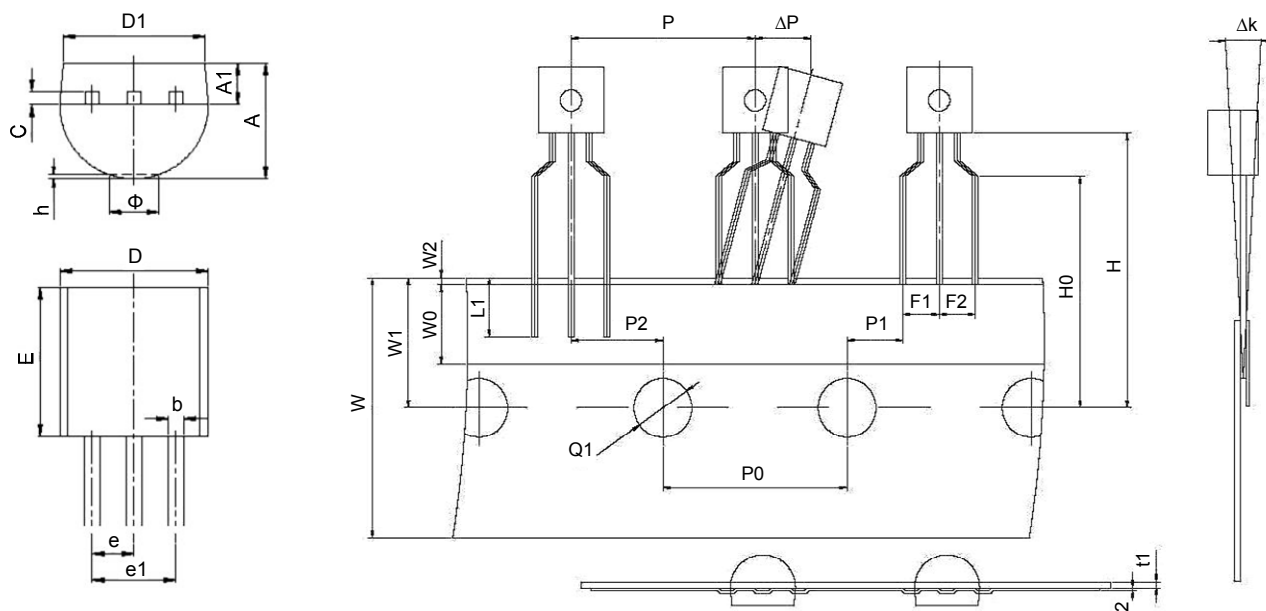|	110VAC	220VAC
Standby Power	0.09W	0.15W
Current Limit	0.75A	0.75A
Full Load Efficiency	65%	67%

Layout Considerations

The following should be observed when doing layout for the ACT30:

1) Use a "star point" connection at the GND pin of ACT30 for the VDD bypass components (C5 and C6 in Figure 5), the input filter capacitor (C2 in Figure 5) and other ground connections on the primary side.

2) Keep the loop across the input filter capacitor, the transformer primary windings, and the high voltage transistor, and the ACT30 as small as possible.

3) Keep ACT30 pins and the high voltage transistor pins as short as possible.

4) Keep the loop across the secondary windings, the output diode, and the output capacitors as small as possible.

5) Allow enough copper area under the high voltage transistor, output diode, and current shunt resistor for heat sink.

PACKAGE OUTLINE

TO-92 PACKAGE OUTLINE AND DIMENSIONS (AMMO TAPE PACKING)

SYMBOL	DIMENSION IN MILIMETERS		DIMENSION IN INCHES	
	MIN	MAX	MIN	MAX
A	3.300	3.700	0.130	0.146
A1	1.100	1.400	0.043	0.055
b	0.380	0.550	0.015	0.022
c	0.360	0.510	0.014	0.020
D	4.400	4.700	0.173	0.185
D1	3.430		0.135	
E	4.300	4.700	0.169	0.185
e	1.270 TYP		0.050 TYP	
e1	2.440	2.640	0.096	0.104
Φ		1.600		0.063
h	0.000	0.380	0.000	0.015

SYMBOL	DIMENSION IN MILIMETERS		DIMENSION IN INCHES	
	MIN	MAX	MIN	MAX
Δk	-1.000	1.000	-0.039	0.039
F1, F2	2.200	2.800	0.087	0.110
H	19.00	21.00	0.748	0.827
H0	15.50	16.50	0.610	0.650
L1	2.500		0.098	
P	12.40	13.00	0.488	0.512
ΔP	-1.000	1.000	-0.039	0.039
P0	12.50	12.90	0.492	0.508
P1	3.550	4.150	0.140	0.163
P2	6.050	6.650	0.238	0.262
Q1	3.800	4.200	0.150	0.165
t1	0.350	0.450	0.014	0.018
t2	0.150	0.250	0.006	0.010
W	17.50	19.00	0.689	0.748
W0	5.500	6.500	0.217	0.256
W1	8.500	9.500	0.335	0.374
W2		1.000		0.039

SOT23-B PACKAGE OUTLINE AND DIMENSIONS
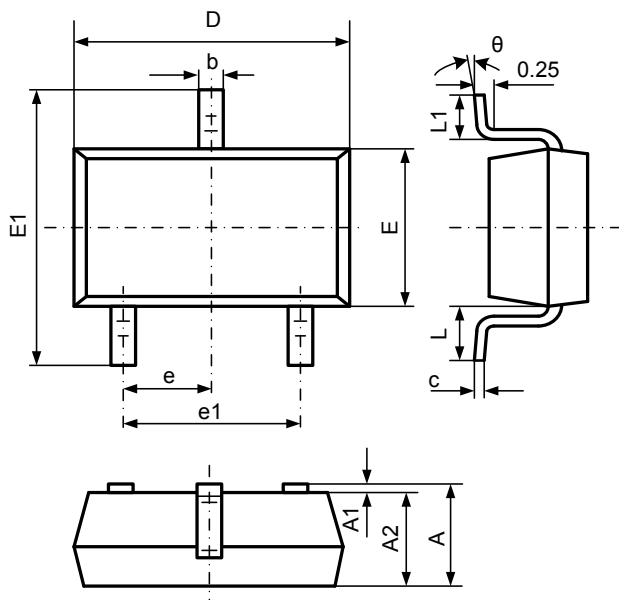
SYMBOL	DIMENSION IN MILLIMETERS		DIMENSION IN INCHES	
	MIN	MAX	MIN	MAX
A	1.900	1.150	0.035	0.045
A1	0.000	0.100	0.000	0.004
A2	0.900	1.050	0.035	0.041
b	0.300	0.500	0.012	0.020
c	0.080	0.150	0.003	0.006
D	2.800	3.000	0.110	0.118
E	1.200	1.400	0.047	0.055
E1	2.250	2.550	0.089	0.100
e	0.950 TYP		0.037 TYP	
e1	1.800	2.000	0.071	0.079
L	0.550 REF		0.022 REF	
L1	0.300	0.500	0.012	0.020
θ	0°	8°	0°	8°

Компания «ЭлектроПласт» предлагает заключение долгосрочных отношений при поставках импортных электронных компонентов на взаимовыгодных условиях!

Наши преимущества:
- Оперативные поставки широкого спектра электронных компонентов отечественного и импортного производства напрямую от производителей и с крупнейших мировых складов;
- Поставка более 17-ти миллионов наименований электронных компонентов;
- Поставка сложных, дефицитных, либо снятых с производства позиций;
- Оперативные сроки поставки под заказ (от 5 рабочих дней);
- Экспресс доставка в любую точку России;
- Техническая поддержка проекта, помощь в подборе аналогов, поставка прототипов;
- Система менеджмента качества сертифицирована по Международному стандарту ISO 9001;
- Лицензия ФСБ на осуществление работ с использованием сведений, составляющих государственную тайну;
- Поставка специализированных компонентов (Xilinx, Altera, Analog Devices, Intersil, Interpoint, Microsemi, Aeroflex, Peregrine, Syfer, Eurofarad, Texas Instrument, Miteq, Cobham, E2V, MA-COM, Hittite, Mini-Circuits,General Dynamics и др.);

Помимо этого, одним из направлений компании «ЭлектроПласт» является направление «Источники питания». Мы предлагаем Вам помощь Конструкторского отдела:
- Подбор оптимального решения, техническое обоснование при выборе компонента;
- Подбор аналогов;
- Консультации по применению компонента;
- Поставка образцов и прототипов;
- Техническая поддержка проекта;
- Защита от снятия компонента с производства.

Как с нами связаться

Телефон: 8 (812) 309 58 32 (многоканальный)
Факс: 8 (812) 320-02-42
Электронная почта: org@eplast1.ru
Адрес: 198099, г. Санкт-Петербург, ул. Калинина, дом 2, корпус 4, литера А.